# IEEE Standard Verilog ${ }^{\oplus}$ Hardware Description Language 

## IEEE Computer Society

Sponsored by the
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# IEEE Standard Verilog ${ }^{\circledR}$ Hardware Description Language 

Sponsor

Design Automation Standards Committee<br>of the<br>IEEE Computer Society

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#### Abstract

The Verilog ${ }^{\alpha}$ Hardware Description Language (HDL) is defined in this standard. Verilog HDL is a formal notation intended for use in all phases of the creation of electronic systems. Because it is both machine readable and human readable, it supports the development, verification, synthesis, and testing of hardware designs; the communication of hardware design data; and the maintenance, modification, and procurement of hardware. The primary audiences for this standard are the implementors of tools supporting the language and advanced users of the language. Keywords: computer, computer languages, digital systems, electronic systems, hardware, hardware description languages, hardware design, HDL, PLI, programming language interface, Verilog HDL, Verilog PLI, Verilog ${ }^{\text {a }}$


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## Introduction

(This introduction is not part of IEEE Std 1364-2001, IEEE Standard Verilog ${ }^{\circledR}$ Hardware Description Language.)

The Verilog ${ }^{\alpha}$ Hardware Description Language (Verilog HDL) became an IEEE standard in 1995 as IEEE Std 1364-1995. It was designed to be simple, intuitive, and effective at multiple levels of abstraction in a standard textual format for a variety of design tools, including verification simulation, timing analysis, test analysis, and synthesis. It is because of these rich features that Verilog has been accepted to be the language of choice by an overwhelming number of IC designers.

Verilog contains a rich set of built-in primitives, including logic gates, user-definable primitives, switches, and wired logic. It also has device pin-to-pin delays and timing checks. The mixing of abstract levels is essentially provided by the semantics of two data types: nets and variables. Continuous assignments, in which expressions of both variables and nets can continuously drive values onto nets, provide the basic structural construct. Procedural assignments, in which the results of calculations involving variable and net values can be stored into variables, provide the basic behavioral construct. A design consists of a set of modules, each of which has an I/O interface, and a description of its function, which can be structural, behavioral, or a mix. These modules are formed into a hierarchy and are interconnected with nets.

The Verilog language is extensible via the Programming Language Interface (PLI) and the Verilog Procedural Interface (VPI) routines. The PLI/VPI is a collection of routines that allows foreign functions to access information contained in a Verilog HDL description of the design and facilitates dynamic interaction with simulation. Applications of PLI/VPI include connecting to a Verilog HDL simulator with other simulation and CAD systems, customized debugging tasks, delay calculators, and annotators.

The language that influenced Verilog HDL the most was HILO-2, which was developed at Brunel University in England under a contract to produce a test generation system for the British Ministry of Defense. HILO-2 successfully combined the gate and register transfer levels of abstraction and supported verification simulation, timing analysis, fault simulation, and test generation.

In 1990, Cadence Design Systems placed the Verilog HDL into the public domain and the independent Open Verilog International (OVI) was formed to manage and promote Verilog HDL. In 1992, the Board of Directors of OVI began an effort to establish Verilog HDL as an IEEE standard. In 1993, the first IEEE Working Group was formed and after 18 months of focused efforts Verilog became an IEEE standard as IEEE Std 1364-1995.

After the standardization process was complete the 1364 Working Group started looking for feedback from 1364 users worldwide so the standard could be enhanced and modified accordingly. This led to a five year effort to get a much better Verilog standard in IEEE Std 1364-2001.

## Objective of the IEEE Std 1364-2001 effort

The starting point for the IEEE 1364 Working Group for this standard was the feedback received from the IEEE Std 1364-1995 users worldwide. It was clear from the feedback that users wanted improvements in all aspects of the language. Users at the higher levels wanted to expand and improve the language at the RTL and behavioral levels, while users at the lower levels wanted improved capability for ASIC designs and signoff. It was for this reason that the 1364 Working Group was organized into three task forces: Behavioral, ASIC, and PLI.

The clear directive from the users for these three task forces was to start by solving some of the following problems:

Consolidate existing IEEE Std 1364-1995
Verilog Generate statement
Multi-dimensional arrays
Enhanced Verilog file I/O
Re-entrant tasks
Standardize Verilog configurations
Enhance timing representation
Enhance the VPI routines

## Achievements

Over a period of four years the 1364 Verilog Standards Group (VSG) has produced five drafts of the LRM. The three task forces went through the IEEE Std 1364-1995 LRM very thoroughly and in the process of consolidating the existing LRM have been able to provide nearly three hundred clarifications and errata for the Behavioral, ASIC, and PLI sections. In addition, the VSG has also been able to agree on all the enhancements that were requested (including the ones stated above).

Three new sections have been added. Clause 13, Configuring the contents of a design, deals with configuration management and has been added to facilitate both the sharing of Verilog designs between designers and/or design groups and the repeatability of the exact contents of a given simulation session. Clause 15 , Timing checks, has been broken out of Clause 17, System tasks and functions, and details more fully how timing checks are used in specify blocks. Clause 16, Backannotation using the Standard Delay Format (SDF), addresses using back annotation (IEEE Std 1497-1999) within IEEE Std 1364-2001.

Extreme care has been taken to enhance the VPI routines to handle all the enhancements in the Behavioral and other areas of the LRM. Minimum work has been done on the PLI routines and most of the work has been concentrated on the VPI routines. Some of the enhancements in the VPI are the save and restart, simulation control, work area access, error handling, assign/deassign and support for array of instances, generate, and file I/O.

Work on this standard would not have been possible without funding from the CAS society of the IEEE and Open Verilog International.

## The IEEE Std 1364-2001 Verilog Standards Group organization

Many individuals from many different organizations participated directly or indirectly in the standardization process. The main body of the IEEE Std 1364-2001 working group is located in the United States, with a subgroup in Japan (EIAJ/1364HDL).

The members of the IEEE Std 1364-2001 working group had voting privileges and all motions had to be approved by this group to be implemented. The three task forces focused on their specific areas and their recommendations were eventually voted on by the IEEE Std 1364-2001 working group.

At the time this document was approved, the IEEE Std 1364-2001 working group had the following membership:

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| Adam Krolnik | Karen Pieper |
| :--- | :--- |
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The IEEE 1364 Japan subgroup (EIAJ/1364HDL) consisted of the following members:
Kasumi Hamaguchi, Vice Chair (Japan)

Yokozeki Atsushi<br>Yasuaki Hatta

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# IEEE Standard Verilog ${ }^{\circledR}$ Hardware Description Language 

## 1. Overview

### 1.1 Objectives of this standard

The intent of this standard is to serve as a complete specification of the Verilog ${ }^{d}$ Hardware Description Language (HDL). This document contains

- The formal syntax and semantics of all Verilog HDL constructs
- The formal syntax and semantics of Standard Delay Format (SDF) constructs
- Simulation system tasks and functions, such as text output display commands
- Compiler directives, such as text substitution macros and simulation time scaling
- The Programming Language Interface (PLI) binding mechanism
- The formal syntax and semantics of access routines, task/function routines, and Verilog procedural interface routines
- Informative usage examples
- Informative delay model for SDF
- Listings of header files for PLI


### 1.2 Conventions used in this standard

This standard is organized into clauses, each of which focuses on a specific area of the language. There are subclauses within each clause to discuss individual constructs and concepts. The discussion begins with an introduction and an optional rationale for the construct or the concept, followed by syntax and semantic descriptions, followed by some examples and notes.

The term shall is used through out this standard to indicate mandatory requirements, whereas the term can is used to indicate optional features. These terms denote different meanings to different readers of this standard:
a) To the developers of tools that process the Verilog HDL, the term shall denotes a requirement that the standard imposes. The resulting implementation is required to enforce the requirements and to issue an error if the requirement is not met by the input.
b) To the Verilog HDL model developer, the term shall denotes that the characteristics of the Verilog HDL are natural consequences of the language definition. The model developer is required to adhere to the constraint implied by the characteristic. The term can denotes optional features that the model
developer can exercise at discretion. If used, however, the model developer is required to follow the requirements set forth by the language definition.
c) To the Verilog HDL model user, the term shall denotes that the characteristics of the models are natural consequences of the language definition. The model user can depend on the characteristics of the model implied by its Verilog HDL source text.

### 1.3 Syntactic description

The formal syntax of the Verilog HDL is described using Backus-Naur Form (BNF). The following conventions are used:
a) Lowercase words, some containing embedded underscores, are used to denote syntactic categories. For example:
module_declaration
b) Boldface words are used to denote reserved keywords, operators, and punctuation marks as a required part of the syntax. These words appear in a larger font for distinction. For example:
module $=>$;
c) A vertical bar separates alternative items unless it appears in boldface, in which case it stands for itself. For example:

```
unary_operator ::=
```


d) Square brackets enclose optional items. For example:

```
input_declaration ::= input [range] list_of_variables;
```

e) Braces enclose a repeated item unless it appears in boldface, in which case it stands for itself. The item may appear zero or more times; the repetitions occur from left to right as with an equivalent left-recursive rule. Thus, the following two rules are equivalent:

```
list_of_param_assignments ::= param_assignment {, param_assignment }
list_of_param_assignments ::=
            param_assignment
            | list_of_param_assignment, param_assignment
```

f) If the name of any category starts with an italicized part, it is equivalent to the category name without the italicized part. The italicized part is intended to convey some semantic information. For example, $m s b$ _constant_expression and $l s b$ _constant_expression are equivalent to constant_expression.

The main text uses italicized font when a term is being defined, and constant-width font for examples, file names, and while referring to constants, especially $0,1, x$, and $z$ values.

### 1.4 Contents of this standard

A synopsis of the clauses and annexes is presented as a quick reference. There are 27 clauses and 8 annexes. All clauses, as well as Annex A, Annex B, Annex E, Annex F, and Annex G, are normative parts of this standard. Annex C, Annex D, and Annex H are included for informative purposes only.

Clause 1-Overview: This clause discusses the conventions used in this standard and its contents.

Clause 2-This clause describes the lexical tokens used in Verilog HDL source text and their conventions.: This clause describes how to specify and interpret the lexical tokens.

Clause 3-Data types: This clause describes net and variable data types. This clause also discusses the parameter data type for constant values and describes drive and charge strength of the values on nets.

Clause 4-Expressions: This clause describes the operators and operands that can be used in expressions.
Clause 5-Scheduling semantics: This clause describes the scheduling semantics of the Verilog HDL.
Clause 6-Assignments: This clause compares the two main types of assignment statements in the Verilog HDL-continuous assignments and procedural assignments. It describes the continuous assignment statement that drives values onto nets.

Clause 7-Gate and switch level modeling: This clause describes the gate and switch level primitives and logic strength modeling.

Clause 8-User-defined primitives (UDPs): This clause describes how a primitive can be defined in the Verilog HDL and how these primitives are included in Verilog HDL models.

Clause 9-Behavioral modeling: This clause describes procedural assignments, procedural continuous assignments, and behavioral language statements.

Clause 10-Tasks and functions: This clause describes tasks and functions-procedures that can be called from more than one place in a behavioral model. It describes how tasks can be used like subroutines and how functions can be used to define new operators.

Clause 11-Disabling of named blocks and tasks: This clause describes how to disable the execution of a task and a block of statements that has a specified name.

Clause 12-Hierarchical structures: This clause describes how hierarchies are created in the Verilog HDL and how parameter values declared in a module can be overridden. It describes how generated instantiations can be used to do conditional or multiple instantiations in a design.

Clause 13-Configuring the contents of a design: This clause describes how to configure the contents of a design.

Clause 14-Specify blocks: This clause describes how to specify timing relationships between input and output ports of a module.

Clause 15-Timing checks: This clause describes how timing checks are used in specify blocks to determine if signals obey the timing constraints.

Clause 16-Backannotation using the Standard Delay Format (SDF): This clause describes syntax and semantics of Standard Delay Format (SDF) constructs.

Clause 17-System tasks and functions: This clause describes the system tasks and functions.
Clause 18-Value change dump (VCD) files: This clause describes the system tasks associated with Value Change Dump (VCD) file, and the format of the file.

Clause 19-Compiler directives: This clause describes the compiler directives.

Clause 20-PLI overview: This clause previews the C language procedural interface standard (Programming Language Interface or PLI) and interface mechanisms that are part of the Verilog HDL.

## Clause 21-PLI TF and ACC interface mechanism

This clause describes the interface mechanism that provides a means for users to link PLI task/function (TF) routine and access (ACC) routine applications to Verilog software tools.

Clause 22-Using ACC routines: This clause describes the ACC routines in general, including how and why to use them.

Clause 23-ACC routine definitions: This clause describes the specific ACC routines, explaining their function, syntax, and usage.

Clause 24-Using TF routines: This clause provides an overview of the types of operations that are done with the TF routines.

Clause 25-TF routine definitions: This clause describes the specific TF routines, explaining their function, syntax, and usage.

Clause 26-Using VPI routines: This clause provides an overview of the types of operations that are done with the Verilog Programming Interface (VPI) routines.

Clause 27-VPI routine definitions: This clause describes the VPI routines.

Annex A-Formal syntax definition: This normative annex describes, using BNF, the syntax of the Verilog HDL.

Annex B—List of keywords: This normative annex lists the Verilog HDL keywords.
Annex C-System tasks and functions: This informative annex describes system tasks and functions that are frequently used, but that are not part of the standard.

Annex D-Compiler directives: This informative annex describes compiler directives that are frequently used, but that are not part of the standard.

Annex E-acc_user.h: This normative annex provides a listing of the contents of the acc_user.h file.
Annex F-veriuser.h: This normative annex provides a listing of the contents of the vpi_user. h file.

Annex G—vpi_user.h: This normative annex provides a listing of the contents of the veriuser.h file.
Annex H-Bibliography: This informative annex contains bibliographic entries pertaining to this standard.

### 1.5 Header file listings

The header file listings included in the annexes E, F, and G for acc_user.h, veriuser.h, and vpi_user. h are a normative part of this standard. All compliant software tools should use the same function declarations, constant definitions, and structure definitions contained in these header file listings.

### 1.6 Examples

Several small examples in the Verilog HDL and the C programming language are shown throughout this standard. These examples are informative - they are intended to illustrate the usage of Verilog HDL constructs and PLI functions in a simple context and do not define the full syntax.

### 1.7 Prerequisites

Clauses 20 through 27 and Annexes E through G presuppose a working knowledge of the C programming language.

## 2. Lexical conventions

This clause describes the lexical tokens used in Verilog HDL source text and their conventions.

### 2.1 Lexical tokens

Verilog HDL source text files shall be a stream of lexical tokens. A lexical token shall consist of one or more characters. The layout of tokens in a source file shall be free format-that is, spaces and newlines shall not be syntactically significant other than being token separators, except for escaped identifiers (see 2.7.1).

The types of lexical tokens in the language are as follows:

- White space
- Comment
- Operator
- Number
- String
- Identifier
- Keyword


### 2.2 White space

White space shall contain the characters for spaces, tabs, newlines, and formfeeds. These characters shall be ignored except when they serve to separate other lexical tokens. However, blanks and tabs shall be considered significant characters in strings (see 2.6).

### 2.3 Comments

The Verilog HDL has two forms to introduce comments. A one-line comment shall start with the two characters / / and end with a new line. A block comment shall start with /* and end with * /. Block comments shall not be nested. The one-line comment token / / shall not have any special meaning in a block comment.

### 2.4 Operators

Operators are single-, double-, or triple-character sequences and are used in expressions. Clause 4 discusses

```
number ::= (From Annex A-A.8.7)
        decimal_number
        octal_number
        binary_number
        hex_number
        real_number
real_number \({ }^{*}::=\)
            unsigned_number . unsigned_number
            | unsigned_number [ . unsigned_number ] exp [ sign ] unsigned_number
\(\exp ::=\mathbf{e} \mid \mathbf{E}\)
decimal_number ::=
    unsigned_number
    | [ size ] decimal_base unsigned_number
    [ [ size ] decimal_base x_digit \(\left\{\begin{array}{l}- \\ \}\end{array}\right.\)
    | [ size ] decimal_base z_digit \(\{\) _ \(\}\)
binary_number \(::=\)
    [ size ] binary_base binary_value
octal_number ::=
    [ size ] octal_base octal_value
hex_number ::=
    [ size ] hex_base hex_value
sign \(::=+\mid-\)
size \(::=\) non_zero_unsigned_number
non_zero_unsigned_number \({ }^{*}::=\) non_zero_decimal_digit \(\{\) _ \(\mid\) decimal_digit \(\}\)
unsigned_number \({ }^{*}::=\) decimal_digit \(\{\) _ \(\mid\) decimal_digit \(\}\)
binary_value \({ }^{*}::=\) binary_digit \(\{\) _ \(\mid\) binary_digit \(\}\)
octal_value* \(::=\) octal_digit \(\{\) _ \(\mid\) octal_digit \(\}\)
hex_value \({ }^{*}::=\) hex_digit \(\{\) _ \(\mid\) hex_digit \(\}\)
decimal_base \({ }^{*}::=\) ' \([\mathbf{s} \mid \mathbf{S}] \mathbf{d} \mid{ }^{\prime}[\mathbf{s} \mid \mathbf{S}] \mathbf{D}\)
binary_base \({ }^{*}::={ }^{\prime}[\mathbf{s} \mid \mathbf{S}] \mathbf{b} \mid{ }^{\prime}[\mathbf{s} \mid \mathbf{S}] \mathbf{B}\)
octal_base \({ }^{*}::={ }^{\prime}[\mathbf{s} \mid \mathbf{S}] \mathbf{0} \mid,{ }^{\prime}[\mathbf{s} \mid \mathbf{S}] \mathbf{O}\)
hex_base \({ }^{*}::={ }^{\prime}[\mathbf{s} \mid \mathbf{S}] \mathbf{h} \mid{ }^{\prime}[\mathbf{s} \mid \mathbf{S}] \mathbf{H}\)
non_zero_decimal_digit ::=1|2|3|4|5|6|7|8|9
decimal_digit ::= 0 \(|\mathbf{1}| \mathbf{2}|\mathbf{3}| \mathbf{4}|\mathbf{5}| \mathbf{6}|\mathbf{7}| \mathbf{8} \mid \mathbf{9}\)
binary_digit ::=x_digit \(\mid\) z_digit \(|\mathbf{0}| \mathbf{1}\)
octal_digit ::= x_digit \(\mid\) z_digit \(|\mathbf{0}| \mathbf{1}|\mathbf{2}| \mathbf{3}|\mathbf{4}| \mathbf{5}|\mathbf{6}| \mathbf{7}\)
hex_digit ::=
            x_digit \(\mid\) z_digit \(|\mathbf{0}| \mathbf{1}|\mathbf{2}| \mathbf{3}|\mathbf{4}| \mathbf{5}|\mathbf{6}| \mathbf{7}|\mathbf{8}| \mathbf{9}\)
            \(|\mathbf{a}| \mathbf{b}|\mathbf{c}| \mathbf{d}|\mathbf{e}| \mathbf{f}|\mathbf{A}| \mathbf{B}|\mathbf{C}| \mathbf{D}|\mathbf{E}| \mathbf{F}\)
x_digit \(::=\mathbf{x} \mid \mathbf{X}\)
z_digit ::= \(\mathbf{z}|\mathbf{Z}|\) ?
```

*Embedded spaces are illegal.

## Syntax 2-1-Syntax for integer and real numbers

### 2.5.1 Integer constants

Integer constants can be specified in decimal, hexadecimal, octal, or binary format.

There are two forms to express integer constants. The first form is a simple decimal number, which shall be specified as a sequence of digits 0 through 9 , optionally starting with a plus or minus unary operator. The second form specifies a size constant, which shall be composed of up to three tokens-an optional size constant, a single quote followed by a base format character, and the digits representing the value of the number.

The first token, a size constant, shall specify the size of the constant in terms of its exact number of bits. It shall be specified as a non-zero unsigned decimal number. For example, the size specification for two hexadecimal digits is 8 , because one hexadecimal digit requires 4 bits. Unsized unsigned constants where the high order bit is unknown ( X or x ) or three-state ( Z or z ) are extended to the size of the expression containing the constant.

NOTE-In IEEE Std 1364-1995, unsized constants where the high order bit is unknown or three-state, the x or z was only extended to 32 bits.

The second token, a base_format, shall consist of a case-insensitive letter specifying the base for the number, optionally preceded by the single character $s$ (or $S$ ) to indicate a signed quantity, preceded by the single quote character ( ${ }^{\prime}$ ). Legal base specifications are $d, D, h, H, O, O, b$, or $B$, for the bases decimal, hexadecimal, octal, and binary respectively.

The use of x and z in defining the value of a number is case insensitive.

The single quote and the base format character shall not be separated by any white space.
The third token, an unsigned number, shall consist of digits that are legal for the specified base format. The unsigned number token shall immediately follow the base format, optionally preceded by white space. The hexadecimal digits a to $f$ shall be case insensitive.

Simple decimal numbers without the size and the base format shall be treated as signed integers, whereas the numbers specified with the base format shall be treated as signed integers if the $s$ designator is included or as unsigned integers if the base format only is used. The $s$ designator does not affect the bit pattern specified, only its interpretation.

A plus or minus operator preceding the size constant is a unary plus or minus operator. A plus or minus operator between the base format and the number is an illegal syntax.

Negative numbers shall be represented in 2 s complement form.
An x represents the unknown value in hexadecimal, octal, and binary constants. A z represents the highimpedance value. See 3.1 for a discussion of the Verilog HDL value set. An x shall set 4 bits to unknown in the hexadecimal base, 3 bits in the octal base, and 1 bit in the binary base. Similarly, a z shall set 4 bits, 3 bits, and 1 bit, respectively, to the high-impedance value.

If the size of the unsigned number is smaller than the size specified for the constant, the unsigned number shall be padded to the left with zeros. If the leftmost bit in the unsigned number is an x or a z , then an x or a z shall be used to pad to the left respectively.

When used in a number, the question-mark (? ) character is a Verilog HDL alternative for the $z$ character. It sets 4 bits to the high-impedance value in hexadecimal numbers, 3 bits in octal, and 1 bit in binary. The question mark can be used to enhance readability in cases where the high-impedance value is a don t-care condition. See the discussion of casez and casex in 9.5.1. The question-mark character is also used in userdefined primitive state table. See 8.1.6, Table 8-1.

The underscore character (_) shall be legal anywhere in a number except as the first character. The underscore character is ignored. This feature can be used to break up long numbers for readability purposes.

## Examples:

Example 1-Unsized constant numbers

```
659 // is a decimal number
'h 837FF // is a hexadecimal number
'o7460 // is an octal number
4af // is illegal (hexadecimal format requires 'h)
```

Example 2-Sized constant numbers

```
4' b1001 // is a 4-bit binary number
5'D 3 // is a 5-bit decimal number
3' b01x // is a 3-bit number with the least
            // significant bit unknown
12' hx // is a 12-bit unknown number
16' hz // is a 16-bit high-impedance number
```

Example 3-Using sign with constant numbers

```
8'd -6 // this is illegal syntax
-8 'd 6 // this defines the two's complement of 6,
    // held in 8 bits-equivalent to -(8' d 6)
4'shf // this denotes the 4-bit number '1111', to
    // be interpreted as a 2's complement number,
    // or '-1'. This is equivalent to -4' h 1
-4 'sd15 // this is equivalent to -(-4' d 1), or '0001'.
```

Example 4-Automatic left padding

```
reg [ 11:0] a, b, c, d;
initial begin
            a ='h x; // yields xxx
    b = 'h 3x; // yields 03x
    c ='h z3; // yields zz3
    d = 'h Oz3; // yields Oz3
end
reg [ 84:0] e, f, g;
```

```
    e = 'h5; // yields { 82{1'b0},3'b101}
```

    e = 'h5; // yields { 82{1'b0},3'b101}
    f = 'hx; // yields { 85{ 1'hx}}
    f = 'hx; // yields { 85{ 1'hx}}
    g = 'hz; // yields { 85{1'hz}}
    ```
    g = 'hz; // yields { 85{1'hz}}
```

Example 5-Using underscore character in numbers

```
27_195_000
16' b0011_0101_0001_1111
32'h 12ab f001
```


## NOTES:

1) Sized negative constant numbers and sized signed constant numbers are sign-extended when assigned to a reg data type, regardless of whether the reg itself is signed or not.
2) Each of the three tokens for specifying a number may be macro substituted.
3) The number of bits that make up an unsized number (which is a simple decimal number or a number without the size specification) shall be at least 32 .

### 2.5.2 Real constants

The real constant numbers shall be represented as described by IEEE Std 754-1985 [B1], ${ }^{1}$ an IEEE standard for double-precision floating-point numbers.

Real numbers can be specified in either decimal notation (for example, 14.72) or in scientific notation (for example, 39 e 8 , which indicates 39 multiplied by 10 to the eighth power). Real numbers expressed with a decimal point shall have at least one digit on each side of the decimal point.

## Examples:

```
1.2
0.1
2394.26331
1.2E12 (the exponent symbol can be e or E)
1.30e-2
0.1e-0
23E10
29E-2
236.123_763_e-12 (underscores are ignored)
```

The following are invalid forms of real numbers because they do not have at least one digit on each side of the decimal point:

```
. }1
```

9. 

4.E3
. $2 \mathrm{e}-7$

### 2.5.3 Conversion

Real numbers shall be converted to integers by rounding the real number to the nearest integer, rather than by truncating it. Implicit conversion shall take place when a real number is assigned to an integer. The ties shall be rounded away from zero. For example:

The real numbers 35.7 and 35.5 both become 36 when converted to an integer and 35.2 becomes 35 . Converting -1.5 to integer yields -2 , converting 1.5 to integer yields 2 .

### 2.6 Strings

A string is a sequence of characters enclosed by double quotes (" ") and contained on a single line. Strings used as operands in expressions and assignments shall be treated as unsigned integer constants represented by a sequence of 8 -bit ASCII values, with one 8 -bit ASCII value representing one character.

[^1]
### 2.6.1 String variable declaration

String variables are variables of reg type (see 3.2) with width equal to the number of characters in the string multiplied by 8 .

## Example:

To store the twelve-character string "Hello world!" requires a reg $8 * 12$, or 96 bits wide

```
reg[ 8* 12:1] stringvar;
initial begin
    stringvar = "Hello world!";
end
```


### 2.6.2 String manipulation

Strings can be manipulated using the Verilog HDL operators. The value being manipulated by the operator is the sequence of 8 -bit ASCII values.

## Example:

```
module string test;
reg [ 8*14:1] stringvar;
initial begin
    stringvar = "Hello world";
    \$display("\%s is stored as \%h", stringvar,stringvar);
    stringvar = \{stringvar,"!!!"\} ;
    \$display("\%s is stored as \%h", stringvar,stringvar);
end
endmodule
```

The output is:

```
Hello world is stored as 00000048656c6c6f20776f726c64
Hello world!!! is stored as 48656c6c6f20776f726c64212121
```

NOTE - When a variable is larger than required to hold a value being assigned, the contents on the left are padded with zeros after the assignment. This is consistent with the padding that occurs during assignment of nonstring values. If a string is larger than the destination string variable, the string is truncated to the left, and the leftmost characters will be lost.

### 2.6.3 Special characters in strings

Certain characters can only be used in strings when preceded by an introductory character called an escape character. Table 1 lists these characters in the right-hand column, with the escape sequence that represents the character in the left-hand column.

## Table 1-Specifying special characters in string

| Escape <br> string | Character produced by <br> escape string |
| :--- | :--- |
| $\backslash n$ | New line character |
| $\backslash t$ | Tab character |
| $\backslash \backslash$ | $\backslash$ character |
| $\backslash "$ | " character |
| $\backslash d d d$ | A character specified in $1-3$ octal digits <br> $(0 \leq \mathrm{d} \leq 7)$ |

### 2.7 Identifiers, keywords, and system names

An identifier is used to give an object a unique name so it can be referenced. An identifier is either a simple identifier or an escaped identifier (see 2.7.1). A simple identifier shall be any sequence of letters, digits, dollar signs (\$), and underscore characters (_).

The first character of a simple identifier shall not be a digit or $\$$; it can be a letter or an underscore. Identifiers shall be case sensitive.

## Example:

```
shiftreg_a
busa indēx
error_condition
merge_ab
_bus3
n$657
```

NOTE-Implementations may set a limit on the maximum length of identifiers, but they shall at least be 1024 characters. If an identifier exceeds the implementation-specified length limit, an error shall be reported.

### 2.7.1 Escaped identifiers

Escaped identifiers shall start with the backslash character ( $\backslash$ ) and end with white space (space, tab, newline). They provide a means of including any of the printable ASCII characters in an identifier (the decimal values 33 through 126, or 21 through 7 E in hexadecimal).

Neither the leading backslash character nor the terminating white space is considered to be part of the identifier. Therefore, an escaped identifier $\backslash$ cpu 3 is treated the same as a nonescaped identifier cpu3.

## Example:

```
\busa+index
\-clock
\***error-condition***
\net1/\ net2
\{ a,b}
\a* (b+c)
```


### 2.7.2 Generated identifiers

Generated identifiers are created by generate loops (see 12.1.3.2); and are a special case of identifiers in that they can be used in hierarchical names (see 12.4). A generated identifier is the named generate block identifier terminated with a ([digit(s)]) string. This identifier is used as a node name in hierarchical names (see 12.4).

### 2.7.3 Keywords

Keywords are predefined nonescaped identifiers that are used to define the language constructs. A Verilog HDL keyword preceded by an escape character is not interpreted as a keyword.

All keywords are defined in lowercase only. Annex B gives a list of all defined keywords.

### 2.7.4 System tasks and functions

The $\$$ character introduces a language construct that enables development of user-defined tasks and functions. System constructs are not design semantics, but refer to simulator functionality. A name following the \$ is interpreted as a system task or a system function.

The syntax for a system task or function is given in Syntax 2-2.

```
system_task_enable ::= (From Annex A - A.6.9)
    system_task_identifier [ ( expression \{, expression \} ) ];
system_function_call ::= (From Annex \(A\) - A.8.2)
    system_function_identifier [ ( expression \{ , expression \} )]
system_function_identifier \({ }^{*}::=(\) (From Annex \(A-\) A.9.3) \()\)
    \$[ a-zA-Z0-9_\$ \(\{\) [ a-zA-Z0-9_\$] \}
system_task_identifier \({ }^{*}::=\)
    \$[ a-zA-Z0-9_\$]\{[ a-zA-Z0-9_\$] \}
```

*The \$ character in a system_function_identifier or system_task_identifier shall not be followed by white space. A system_function_identifier or system_task_identifier shall not be escaped.

Syntax 2-2-Syntax for system tasks and functions

The \$identifier system task or function can be defined in three places
A standard set of \$identifier system tasks and functions, as defined in Clauses 17 and 19. Additional \$identifier system tasks and functions defined using the PLI, as described in Clause 20. Additional \$identifier system tasks and functions defined by software implementations.

Any valid identifier, including keywords already in use in contexts other than this construct, can be used as a system task or function name. The system tasks and functions described in 17. are part of this standard. Additional system tasks and functions with the \$identifier construct are not part of this standard.

## Example:

```
$display ("display a message");
$finish;
```


### 2.7.5 Compiler directives

The ' character (the ASCII value 60, called open quote or accent grave) introduces a language construct used to implement compiler directives. The compiler behavior dictated by a compiler directive shall take effect as soon as the compiler reads the directive. The directive shall remain in effect for the rest of the compilation unless a different compiler directive specifies otherwise. A compiler directive in one description file can therefore control compilation behavior in multiple description files.

The 'identifier compiler directive construct can be defined in two places
A standard set of 'identifier compiler directives defined in Clause 19.
Additional 'identifier compiler directives defined by software implementations.
Any valid identifier, including keywords already in use in contexts other than this construct, can be used as a compiler directive name. The compiler directives described in Clause 19 are part of this standard. Additional compiler directives with the 'identifier construct are not part of this standard.

## Example:

'define wordsize 8

### 2.8 Attributes

With the proliferation of tools other than simulators that use Verilog HDL as their source, a mechanism is included for specifying properties about objects, statements and groups of statements in the HDL source that may be used by various tools, including simulators, to control the operation or behavior of the tool. These properties shall be referred to as "attributes". This subclause specifies the syntactic mechanism that shall be used for specifying attributes, without standardizing on any particular attributes.

The syntax for specifying an attribute is shown in Syntax 2-3.

```
attribute_instance ::= (From Annex A - A.9.1)
    (* attr_spec {, attr_spec } *)
attr_spec ::=
        attr_name = constant_expression
        | attr_name
attr_name ::=
    identifier
```


## Syntax 2-3-Syntax for attributes

An attribute instance can appear in the Verilog description as a prefix attached to a declaration, a module item, a statement, or a port connection. It can appear as a suffix to an operator or a Verilog function name in an expression.

If a value is not specifically assigned to the attribute, then its value shall be 1 . If the same attribute name is defined more than once for the same language element, the last attribute value shall be used and a tool can give a warning that a duplicate attribute specification has occurred.

### 2.8.1 Examples

Example 1-The following example shows how to attach attributes to a case statement:

```
(* full_case, parallel_case *)
case (foo)
<rest_of_case_statement>
```

or

```
(* full_case=1, parallel_case=1 *)
case (foo)
<rest_of_case_statement>
```

or

```
(* full_case, // no value assigned
    parallel_case=1 *)
case (foo)
<rest_of_case_statement>
```

Example 2-To attach the full_case attribute, but NOT the parallel_case attribute:

```
(* full_case *) // parallel_case not specified
case (foo)
<rest_of_case_statement>
```

or

```
(* full_case=1, parallel_case = 0 *)
case (foo)
<rest_of_case_statement>
```

Example 3-To attach an attribute to a module definition:

```
(* optimize power *)
module mod1 (<port_list>);
```

or

```
(* optimize_power=1 *)
module mod1 (<port_list>);
```

Example 4-To attach an attribute to a module instantiation:

```
(* optimize_power=0 *)
mod1 synth1 (<port_list>);
```

Example 5-To attach an attribute to a reg declaration:

```
(* fsm_state *) reg [ 7:0] state1;
(* fsm_state=1 *) reg [ 3:0] state2, state3;
reg [ 3:0] reg1; // this reg does NOT have fsm_state set
(* fsm_state=0 *) reg [ 3:0] reg2; // nor does this one
```

Example 6-To attach an attribute to an operator:
$\mathrm{a}=\mathrm{b}+(* \operatorname{mode}=" \mathrm{cla} \mathrm{*}$ *) c ;
This sets the value for the attribute mode to be the string cla.
Example 7-To attach an attribute to a Verilog function call:
$\mathrm{a}=$ add (* mode = "cla" *) (b, c);
Example 8-To attach an attribute to a conditional operator:

```
a = b ? (* no_glitch *) c : d;
```


### 2.8.2 Syntax

The syntax for legal statements with attributes is shown in Syntax 2-4-Syntax 2-11.
The syntax for module declaration attributes is given in Syntax 2-4.
module_declaration $::=$ (From Annex $A-$ A.1.3)
\{ attribute_instance \} module_keyword module_identifier [ module_parameter_port_list ] [ list_of_ports ] ; \{ module_item \} endmodule
| \{ attribute_instance \} module_keyword module_identifier [ module_parameter_port_list ] [ list_of_port_declarations ] ; \{ non_port_module_item \} endmodule

## Syntax 2-4-Syntax for module declaration attributes

The syntax for port declaration attributes is given in Syntax 2-5.
port_declaration $::=$ (From Annex $A$ - A.1.4)
\{attribute_instance\} inout_declaration
| \{attribute_instance\} input_declaration
| \{attribute_instance\} output_declaration

Syntax 2-5—Syntax for port declaration attributes

The syntax for module item attributes is given in Syntax 2-6.

```
module_item ::= (From Annex A - A.1.5)
    module_or_generate_item
    port_declaration ;
    | \{ attribute_instance \} generated_instantiation
    | \{ attribute_instance \} local_parameter_declaration
    | \{ attribute_instance \} parameter_declaration
    | \{ attribute_instance \} specify_block
    | \{ attribute_instance \} specparam_declaration
module_or_generate_item ::=
            \{ attribute_instance \} module_or_generate_item_declaration
    \{ attribute_instance \} parameter_override
    \{ attribute_instance \} continuous_assign
    \{ attribute_instance \} gate_instantiation
    \{ attribute_instance \} udp_instantiation
    | \{ attribute_instance \} module_instantiation
    | \{ attribute_instance \} initial_construct
    | \{ attribute_instance \} always_construct
non_port_module_item ::=
    \{ attribute_instance \} generated_instantiation
    | \{ attribute_instance \} local_parameter_declaration
    | \{ attribute_instance \} module_or_generate_item
    \{ attribute_instance \} parameter_declaration
    | \{ attribute_instance \} specify_block
    | \{attribute_instance \} specparam_declaration
```

Syntax 2-6-Syntax for module item attributes

The syntax for function port, task, and block attributes is given in Syntax 2-7.

```
function_port_list \(::=(\) From Annex \(A-A .2 .6)\)
    \{attribute_instance\} input_declaration \{, \{attribute_instance \} input_declaration\}
task_item_declaration \(::=\) (From Annex \(A-\) A.2.7)
        block_item_declaration
        | \{ attribute_instance \} input_declaration ;
        | \{ attribute_instance \} output_declaration ;
        | \{ attribute_instance \} inout_declaration;
task_port_item ::=
        \{ attribute_instance \} input_declaration
        | \{ attribute_instance \} output_declaration
        | \{ attribute_instance \} inout_declaration
block_item_declaration ::= (From Annex A-A.2.8)
        \{ attribute_instance \} block_reg_declaration
        | \{ attribute_instance \} event_declaration
        \{ attribute_instance \} integer_declaration
        | \{ attribute_instance \} local_parameter_declaration
        \{ attribute_instance \} parameter_declaration
        \{ attribute_instance \} real_declaration
        | \{ attribute_instance \} realtime_declaration
        | \{ attribute_instance \} time_declaration
```

    Syntax 2-7-Syntax for function port, task, and block attributes
    The syntax for port connection attributes is given in Syntax 2-8.

```
ordered_port_connection \(::=\) (From Annex A - A.4.1)
    \{ attribute_instance \} [ expression ]
named_port_connection ::=
    \{ attribute_instance \} .port_identifier ([ expression ] )
```

Syntax 2-8-Syntax for port connection attributes

The syntax for udp attributes is given in Syntax 2-9.

```
udp_declaration ::= (From Annex A - A.5.1)
    { attribute_instance } primitive udp_identifier ( udp_port_list );
        udp_port_declaration { udp_port_declaration }
        udp_body
    endprimitive
    | { attribute_instance } primitive udp_identifier (udp_declaration_port_list ) ;
        udp_body
    endprimitive
udp_output_declaration ::= (From Annex A - A.5.2)
        { attribute_instance } output port_identifier
        | { attribute_instance } output reg port_identifier [ = constant_expression ]
udp_input_declaration ::=
        { attribute_instance } input list_of_port_identifiers
udp_reg_declaration ::=
    {attribute_instance } reg variable_identifier
```

Syntax 2-9—Syntax for udp attributes

## 3. Data types

The set of Verilog HDL data types is designed to represent the data storage and transmission elements found in digital hardware.

### 3.1 Value set

The Verilog HDL value set consists of four basic values:

```
0 - represents a logic zero, or a false condition
1 - represents a logic one, or a true condition
\(x\) - represents an unknown logic value
z - represents a high-impedance state
```

The values 0 and 1 are logical complements of one another.
When the $z$ value is present at the input of a gate, or when it is encountered in an expression, the effect is usually the same as an $x$ value. Notable exceptions are the metal-oxide semiconductor (MOS) primitives, which can pass the $z$ value.

Almost all of the data types in the Verilog HDL store all four basic values. The exception is the event type (see 9.7.3), which has no storage. All bits of vectors can be independently set to one of the four basic values.

The language includes strength information in addition to the basic value information for net variables. This is described in detail in 7 ..

### 3.2 Nets and variables

There are two main groups of data types: the variable data types and the net data types. These two groups differ in the way that they are assigned and hold values. They also represent different hardware structures.

### 3.2.1 Net declarations

The net data types shall represent physical connections between structural entities, such as gates. A net shall not store a value (except for the trireg net). Instead, its value shall be determined by the values of its drivers, such as a continuous assignment or a gate. See Section 6 and 7. for definitions of these constructs. If no driver is connected to a net, its value shall be high-impedance ( $z$ ) unless the net is a trireg, in which case it shall hold the previously driven value. It is illegal to redeclare a name already declared by a net, parameter, or variable declaration (see 3.12).

The syntax for net declarations is given in Syntax 3-1.

```
net_declaration ::= (From Annex A - A.2.1.3)
        net_type [ signed ]
        [ delay3 ] list_of_net_identifiers;
    | net_type [drive_strength ] [ signed ]
        [ delay3 ] list_of_net_decl_assignments ;
    | net_type [ vectored | scalared ] [ signed ]
        range [ delay3 ] list_of_net_identifiers ;
    | net_type [ drive_strength ] [ vectored | scalared ] [ signed ]
        range [ delay3 ] list_of_net_decl_assignments ;
    | trireg [ charge_strength ] [ signed ]
        [ delay3 ] list_of_net_identifiers;
        trireg [drive_strength ] [ signed ]
        [ delay3 ] list_of_net_decl_assignments ;
    | trireg [ charge_strength ] [ vectored | scalared ] [ signed ]
        range [ delay3 ] list_of_net_identifiers ;
    | trireg [ drive_strength ] [ vectored | scalared ] [ signed ]
        range [ delay3 ] list_of_net_decl_assignments;
net_type ::= (From Annex A - A.2.2.1)
        supply0 | supply1
        tri | triand | trior | tri0 | tri1
        wire |wand | wor
drive_strength ::= (From Annex A - A.2.2.2)
        ( strength0, strength1 )
        ( strength1, strength0)
        ( (strength0, highz1 )
        ( strength1, highz0 )
        ( highz0 , strength1 )
        ( highz1 , strength0 )
strength0 ::= supply0 | strong0 | pull0 | weak0
strength1 ::= supply1 | strong1 | pull1 | weak1
charge_strength ::= ( small ) | ( medium ) | ( large )
delay3 ::= (From Annex A - A.2.2.3)
    # delay_value|# ( delay_value [, delay_value [, delay_value ] ])
delay2 ::=
    # delay_value | # (delay_value [, delay_value ])
delay_value ::=
        unsigned_number
        parameter_identifier
        | specparam_identifier
        | mintypmax_expression
list_of_net_decl_assignments ::= (From Annex A - A.2.3)
    net_decl_assignment {, net_decl_assignment }
list_of_net_identifiers ::=
    net_identifier [ dimension { dimension }]
        {, net_identifier [ dimension { dimension }] }
net_decl_assignment ::= (From Annex A - A.2.4)
    net_identifier = expression
dimension ::= (From Annex A -A.2.5)
    [ dimension_constant_expression : dimension_constant_expression ]
range ::=
    [ msb_constant_expression : lsb_constant_expression ]
```

The first two forms of net declaration are described in this section. The third form, called net assignment, is described in Section 6.

### 3.2.2 Variable declarations

A variable is an abstraction of a data storage element. A variable shall store a value from one assignment to the next. An assignment statement in a procedure acts as a trigger that changes the value in the data storage element. The initialization value for reg, time, and integer data types shall be the unknown value, $x$. The default initialization value for real and realtime variable datatypes shall be 0.0 . If a variable declaration assignment is used (see 6.2.1), the variable shall take this value as if the assignment occurred in a blocking assignment in an initial construct. It is illegal to redeclare a name already declared by a net, parameter, or variable declaration.

NOTE In previous versions of the Verilog standard, the term register was used to encompass both the reg, integer, time, real and realtime types; but that the term is no longer used as a Verilog data type.

The syntax for variable declarations is given in Syntax 3-2.

```
integer_declaration ::= (From Annex A-A.2.1.3)
        integer list_of_variable_identifiers ;
real_declaration ::=
        real list_of_real_identifiers ;
realtime_declaration ::=
        realtime list_of_real_identifiers ;
reg_declaration ::=
    reg [ signed ] [ range ] list_of_variable_identifiers ;
time_declaration ::=
        time list_of_variable_identifiers;
real_type \(::=(\) From Annex \(A-\) A.2.2.1)
        real_identifier [ = constant_expression ]
        \(\mid\) real_identifier dimension \(\{\) dimension \}
variable_type \(::=\)
        variable_identifier [ = constant_expression ]
        | variable_identifier dimension \{ dimension \}
list_of_real_identifiers ::= (From Annex A - A.2.3)
            real_type \(\{\), real_type \}
list_of_variable_identifiers ::=
            variable_type \(\{\), variable_type \}
dimension \(::=(\) From Annex \(A-A .2 .5)\)
    [ dimension_constant_expression : dimension_constant_expression ]
range ::=
        [ msb_constant_expression : 1sb_constant_expression ]
```

Syntax 3-2 Syntax for variable declaration

If a set of nets or variables share the same characteristics, they can be declared in the same declaration statement.

## CAUTI N

Variables can be assigned negative values, but only signed regs, integer, real, and realtime variables shall retain the significance of the sign. The unsigned reg and time variables shall treat the value assigned to them as an unsigned value. Refer to 4.1.6 for a description of how signed and unsigned variables are treated by certain Verilog operators.

### 3.3 Vectors

A net or reg declaration without a range specification shall be considered 1 bit wide and is known as a scalar. Multiple bit net and reg data types shall be declared by specifying a range, which is known as a vector.

### 3.3.1 Specifying vectors

The range specification gives addresses to the individual bits in a multibit net or reg. The most significant bit specified by the $m s b$ constant expression is the left-hand value in the range and the least significant bit specified by the $l s b$ constant expression is the righthand value in the range.

Both msb constant expression and lsb constant expression shall be constant expressions. The msb and lsb constant expressions can be any value-positive, negative, or zero. The lsb constant expression can be a greater, equal, or lesser value than msb constant expression.

Vector nets and regs shall obey laws of arithmetic modulo 2 to the power $n\left(2^{n}\right)$, where $n$ is the number of bits in the vector. Vector nets and regs shall be treated as unsigned quantities, unless the net or reg is declared to be signed or is connected to a port that is declared to be signed (see 12.2.3).

Examples:

| wand $\mathrm{w} ;$ | // a scalar net of type wand |
| :--- | :--- |
| tri $[15: 0]$ busa; | // a three-state 16-bit bus |
| trireg (small) storeit; | // a charge storage node of strength small |
| reg a; | // a scalar reg |
| reg[3:0] v; | // a 4-bit vector reg made up of (from most to |
|  | // least significant) v[3], v[2], v[1], and v[0] |
| reg signed [3:0] signed_reg; $/ /$ a 4-bit vector in range -8 to 7 |  |
| reg [-1:4] b; | // a 6-bit vector reg |
| wire $\mathrm{wl}, \mathrm{w} 2 ;$ | // declares two wires |
| reg [4:0] $\mathrm{x}, \mathrm{y}, \mathrm{z} ;$ | // declares three 5-bit regs |

NOTES:

1) Implementations may set a limit on the maximum length of a vector, but they will at least be $65536\left(2^{16}\right)$ bits.
2) Implementations do not have to detect overflow of integer operations.

### 3.3.2 Vector net accessibility

Vectored and scalared shall be optional advisory keywords to be used in vector net or reg declaration. If these keywords are implemented, certain operations on vectors may be restricted. If the keyword vectored is used, bit and part selects and strength specifications may not be permitted, and the PLI may consider the object unexpanded. If the keyword scalared is used, bit and part selects of the object shall be permitted, and the PLI shall consider the object expanded.

Examples:

```
tri1 scalared [ 63:0] bus64; //a bus that will be expanded
tri vectored [ 31:0] data; //a bus that may or may not be expanded
```


### 3.4 Strengths

There are two types of strengths that can be specified in a net declaration. They are as follows:
charge strength shall only be used when declaring a net of type trireg
drive strength shall only be used when placing a continuous assignment on a net in the same statement that declares the net

Gate declarations can also specify a drive strength. See 7. for more information on gates and for information on strengths.

### 3.4.1 Charge strength

The charge strength specification shall be used only with trireg nets. A trireg net shall be used to model charge storage; charge strength shall specify the relative size of the capacitance indicated by one of the following keywords:

## small

medium
large

The default charge strength of a trireg net shall be medium.
A trireg net can model a charge storage node whose charge decays over time. The simulation time of a charge decay shall be specified in the delay specification for the trireg net (see 7.14.2).

## Examples:

```
trireg a; // a trireg net of charge strength medium
trireg (large) #(0,0,50) cap1 ; // a trireg net of charge strength
large
    //with charge decay time 50 time units
trireg (small) signed [ 3:0] cap2 ; // a signed 4-bit trireg vector of
    // charge strength small
```


### 3.4.2 Drive strength

The drive strength specification allows a continuous assignment to be placed on a net in the same statement that declares that net. See Section 6 for more details. Net strength properties are described in detail in Clause 7.

### 3.5 Implicit declarations

The syntax shown in 3.2 shall be used to declare nets and variables explicitly. In the absence of an explicit declaration, an implicit net of default net type shall be assumed in the following circumstances:

If an identifier is used in a port expression declaration, then an implicit net of type wire shall be assumed, with the vector width of the port expression declaration. See 12.3.3 for a discussion of port expression declarations.
If an identifier is used in the terminal list of a primitive instance or a module instance, and that identifier has not been explicitly declared previously in one of the declaration statements of the instantiating module, then an implicit scalar net of default net type shall be assumed. See Section 19 for a discussion of control of the type for implicitly declared nets with the 'default_nettype compiler directive.

### 3.6 Net initialization

The default initialization value for a net shall be the value $z$. Nets with drivers shall assume the output value of their drivers. The trireg net is an exception. The trireg net shall default to the value $x$, with the strength specified in the net declaration (small, medium, or large).

### 3.7 Net types

There are several distinct types of nets, as shown in Table 2.
Table 2-Net types

| wire | tri | tri0 | supply0 |
| :--- | :--- | :--- | :--- |
| wand | triand | tri1 | supply1 |
| wor | trior | trireg |  |

### 3.7.1 Wire and tri nets

The wire and tri nets connect elements. The net types wire and tri shall be identical in their syntax and functions; two names are provided so that the name of a net can indicate the purpose of the net in that model. A wire net can be used for nets that are driven by a single gate or continuous assignment. The tri net type can be used where multiple drivers drive a net.

Logical conflicts from multiple sources of the same strength on a wire or a tri net result in x (unknown) values.

Table 3 is a truth table for resolving multiple drivers on wire and tri nets. Note that it assumes equal strengths for both drivers. Please refer to 7.9 for a discussion of logic strength modeling.

Table 3-Truth table for wire and tri nets

| wire/ <br> tri | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{x}$ | $\mathbf{z}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\mathbf{0}$ | 0 | x | x | 0 |
| $\mathbf{1}$ | x | 1 | x | 1 |
| $\mathbf{x}$ | x | x | x | x |
| $\mathbf{z}$ | 0 | 1 | x | z |

### 3.7.2 Wired nets

Wired nets are of type wor, wand, trior, and triand, and are used to model wired logic configurations. Wired nets use different truth tables to resolve the conflicts that result when multiple drivers drive the same net. The wor and trior nets shall create wired or configurations, such that when any of the drivers is 1 , the resulting value of the net is 1 . The wand and triand nets shall create wired and configurations, such that if any driver is 0 , the value of the net is 0 .

The net types wor and trior shall be identical in their syntax and functionality. The net types wand and triand shall be identical in their syntax and functionality. Table 4 and Table 5 give the truth tables for wired nets. Note that they assume equal strengths for both drivers. See 7.9 for a discussion of logic strength modeling.

Table 4-Truth table for wand and triand nets

| wand/ <br> triand | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{x}$ | $\mathbf{z}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\mathbf{0}$ | 0 | 0 | 0 | 0 |
| $\mathbf{1}$ | 0 | 1 | x | 1 |
| $\mathbf{x}$ | 0 | x | x | x |
| $\mathbf{z}$ | 0 | 1 | x | z |

Table 5-Truth table for wor and trior nets

| wor/ <br> trior | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{x}$ | $\mathbf{z}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\mathbf{0}$ | 0 | 1 | x | 0 |
| $\mathbf{1}$ | 1 | 1 | 1 | 1 |
| $\mathbf{x}$ | x | 1 | x | x |
| $\mathbf{z}$ | 0 | 1 | x | z |

### 3.7.3 Trireg net

The trireg net stores a value and is used to model charge storage nodes. A trireg net can be in one of two states:
driven state
capacitive state

When at least one driver of a trireg net has a value of 1,0 , or $x$, the resolved value propagates into the trireg net and is the driven value of the trireg net.

When all the drivers of a trireg net are at the high-impedance value ( $z$ ), the trireg net retains its last driven value; the high-impedance value does not propagate from the driver to the trireg.

The strength of the value on the trireg net in the capacitive state can be small, medium, or large, depending on the size specified in the declaration of the trireg net. The strength of a trireg net in the driven state can be supply, strong, pull, or weak, depending on the strength of the driver.

## Examples:

Figure 1 shows a schematic that includes a trireg net whose size is medium, its driver, and the simulation results.


Figure 1-Simulation values of a trireg and its driver
a) At simulation time 0 , wire a and wire b have a value of 1 . A value of 1 with a strong strength propagates from the and gate through the nmos switches connected to each other by wire $c$ into trireg net d .
b) At simulation time 10 , wire a changes value to 0 , disconnecting wire $c$ from the and gate. When wire c is no longer connected to the and gate, the value of wire c changes to HiZ . The value of wire b remains 1 so wire c remains connected to trireg net $d$ through the nmos 2 switch. The HiZ value does not propagate from wire $c$ into trireg net $d$. Instead, trireg net $d$ enters the capacitive state, storing its last driven value of 1 . It stores the 1 with a medium strength.

### 3.7.3.1 Capacitive networks

A capacitive network is a connection between two or more trireg nets. In a capacitive network whose trireg nets are in the capacitive state, logic and strength values can propagate between trireg nets.

## Examples:

Figure 2 shows a capacitive network in which the logic value of some trireg nets change the logic value of other trireg nets of equal or smaller size.


Figure 2-Simulation results of a capacitive network
In Figure 2, the capacitive strength of trireg_la net is large, trireg_me1 and trireg_me2 are medium, and trireg_sm is small. Simulation reports the following sequence of events:
a) At simulation time 0 , wire a and wire $b$ have a value of 1 . The wire $c$ drives a value of 1 into trireg_la and trireg_sm; wire d drives a value of 1 into trireg_me1 and trireg_me2.
b) At simulation time 10 , the value of wire b changes to 0 , disconnecting trireg_sm and trireg_me2 from their drivers. These trireg nets enter the capacitive state and store the value 1 , their last $\bar{d}$ riven value.
c) At simulation time 20 , wire c drives a value of 0 into trireg_la.
d) At simulation time 30 , wire $d$ drives a value of 0 into trireg_me1.
e) At simulation time 40 , the value of wire a changes to 0 , disconnecting trireg_la and trireg_me1 from their drivers. These trireg nets enter the capacitive state and store the value 0 .
f) At simulation time 50 , the value of wire b changes to 1 .

This change of value in wire b connects trireg_sm to trireg_la; these trireg nets have different sizes and stored different values. This connection causes the smaller trireg net to store the value of the larger trireg net, and trireg_sm now stores a value of 0 .

This change of value in wire $b$ also connects trireg_me1 to trireg_me2; these trireg nets have the same size and stored different values. The connection causes both trireg_me1 and trireg_me2 to change value to x .

In a capacitive network, charge strengths propagate from a larger trireg net to a smaller trireg net. Figure 3 shows a capacitive network and its simulation results.


Figure 3-Simulation results of charge sharing
In Figure 3, the capacitive strength of trireg_la is large and the capacitive strength of trireg_sm is small. Simulation reports the following results:
a) At simulation time 0 , the values of wire a, wire $b$, and wire $c$ are 1 , and wire a drives a strong 1 into trireg_la and trireg_sm.
b) At simulation time 10 , the value of wire b changes to 0 , disconnecting trireg_la and trireg_sm from wire a. The trireg_la and trireg_sm nets enter the capacitive state. Both trireg nets share the large charge of trireg_la because they remain connected through tranif1_2.
c) At simulation time 20 , the value of wire c changes to 0 , disconnecting trireg_sm from trireg_la. The trireg_sm no longer shares large charge of trireg_la and now stores a small charge.
d) At simulation time 30, the value of wire $c$ changes to 1 , connecting the two trireg nets. These trireg nets now share the same charge.
e) At simulation time 40, the value of wire c changes again to 0 , disconnecting trireg_sm from trireg_la. Once again, trireg_sm no longer shares the large charge of trireg_la and now stores a small charge.

### 3.7.3.2 Ideal capacitive state and charge decay

A trireg net can retain its value indefinitely or its charge can decay over time. The simulation time of charge decay is specified in the delay specification of the trireg net. See 7.14.2 for charge decay explanation.

### 3.7.4 Tri0 and tri1 nets

The tri0 and tril nets model nets with resistive pulldown and resistive pullup devices on them. When no driver drives a tri0 net, its value is 0 . When no driver drives a tril net, its value is 1 . The strength of this value is pull. See Clause 7. for a description of strength modeling.

A tri0 net is equivalent to a wire net with a continuous 0 value of pull strength driving it. A tril net is equivalent to a wire net with a continuous 1 value of pull strength driving it.

A truth table for tri0 is shown in Table 6. A truth table for tri1 is shown in Table 7.
Table 6-Truth table for tri0 net

| $\operatorname{tri0}$ | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{x}$ | $\mathbf{z}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\mathbf{0}$ | 0 | x | x | 0 |
| $\mathbf{1}$ | x | 1 | x | 1 |
| $\mathbf{x}$ | x | x | x | x |
| $\mathbf{z}$ | 0 | 1 | x | 0 |

Table 7-Truth table for tri1 net

| tri1 | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{x}$ | $\mathbf{z}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\mathbf{0}$ | 0 | x | x | 0 |
| $\mathbf{1}$ | x | 1 | x | 1 |
| $\mathbf{x}$ | x | x | x | x |
| $\mathbf{z}$ | 0 | 1 | x | 1 |

### 3.7.5 Supply nets

The supply0 and supplyl nets may be used to model the power supplies in a circuit. These nets shall have supply strengths.

## 3.8 regs

Assignments to a reg are made by procedural assignments (see 6.2 and 9.2). Since the reg holds a value between assignments, it can be used to model hardware registers. Edge-sensitive (i.e., flip-flops) and level sensitive (i.e., RS and transparent latches) storage elements can be modeled. A reg needs not represent a hardware storage element since it can also be used to represent combinatorial logic.

### 3.9 Integers, reals, times, and realtimes

In addition to modeling hardware, there are other uses for variables in an HDL model. Although reg variables can be used for general purposes such as counting the number of times a particular net changes value, the integer and time variable data types are provided for convenience and to make the description more selfdocumenting.

The syntax for declaring integer, time, real, and realtime variables is given in Syntax 3-3 (from Syntax 3-2).

```
integer_declaration ::= (From Annex A-A.2.1.3)
    integer list_of_variable_identifiers ;
real_declaration ::=
    real list_of_real_identifiers ;
realtime_declaration ::=
    realtime list_of_real_identifiers ;
time_declaration ::=
    time list_of_variable_identifiers ;
real_type ::= (From Annex A-A.2.2.1)
            real_identifier [ = constant_expression ]
        | real_identifier dimension \{ dimension \}
variable_type ::=
            variable_identifier [ = constant_expression ]
            | variable_identifier dimension \{ dimension \}
list_of_real_identifiers ::= (From Annex A- A.2.3)
            real_type \{, real_type \}
list_of_variable_identifiers ::=
            variable_type \{, variable_type \}
dimension ::= (From Annex A-A.2.5)
    [ dimension_constant_expression : dimension_constant_expression ]
```

Syntax 3-3 Syntax for integer, time, real, and realtime declarations

The syntax for list of reg variables is defined in 3.2.2.
An integer is a general-purpose variable used for manipulating quantities that are not regarded as hardware registers.

A time variable is used for storing and manipulating simulation time quantities in situations where timing checks are required and for diagnostics and debugging purposes. This data type is typically used in conjunction with the \$time system function (see 17.).

The integer and time variables shall be assigned values in the same manner as reg. Procedural assignments shall be used to trigger their value changes.

The time variables shall behave the same as a reg of at least 64 bits, with the least significant bit being bit 0 . They shall be unsigned quantities, and unsigned arithmetic shall be performed on them. In contrast, integer variables shall be treated as signed regs with the least significant bit being zero. Arithmetic operations performed on integer variables shall produce 2 s complement results.

NOTE Implementations may limit the maximum size of an integer variable, but they shall at least be 32 bits.
The Verilog HDL supports real number constants and real variable data types in addition to integer and time variable data types. Except for the following restrictions, variables declared as real can be used in the same places that integer and time variables are used:

Not all Verilog HDL operators can be used with real number values. See Table 17 for lists of valid and invalid operators for real numbers and real variables.
Real variables shall not use range in the declaration
Real variables shall default to an initial value of zero.

The realtime declarations shall be treated synonymously with real declarations and can be used interchangeably.

## Examples:

```
integer a; // integer value
time last_chng; // time value
real float ; // a variable to store real value
realtime rtime ; // a variable to store time as a real
```

value

### 3.9.1 Operators and real numbers

The result of using logical or relational operators on real numbers and real variables is a single-bit scalar value. Not all Verilog HDL operators can be used with expressions involving real numbers and real variables. Table 4-9 lists the valid operators for use with real numbers and real variables. Real number constants and real variables are also prohibited in the following cases:

Edge descriptors ( posedge, negedge) applied to real variables
Bit-select or part-select references of variables declared as real
Real number index expressions of bit-select or part-select references of vectors

### 3.9.2 Conversion

Real numbers shall be converted to integers by rounding the real number to the nearest integer, rather than by truncating it. Implicit conversion shall take place when a real number is assigned to an integer. If the fractional part of the real number is exactly 0.5 , it shall be rounded away from zero.

Implicit conversion shall take place when an expression is assigned to a real. Individual bits that are x or z in the net or the variable shall be treated as zero upon conversion.

See Clause 17 for a discussion of system tasks that perform explicit conversion.

### 3.10 Arrays

An array declaration for a net or a variable declares an element type which is either scalar or vector (see 3.3). For example:

| Declaration | Element Type |
| :--- | :--- |
| $\operatorname{reg} \mathrm{x}[11: 0] ;$ | scalar reg |
| wire $[0: 7] \mathrm{y}[5: 0] ;$ | eight-bit-wide vector wire indexed from 0 to 7 |
| $\operatorname{reg}[31: 0] \times[127: 0] ;$ | thirty-two-bit-wide reg |

NOTE-Array size does not affect the element size..
Arrays can be used to group elements of the declared element type into multi-dimensional objects. Arrays shall be declared by specifying the element address range(s) after the declared identifier. Each dimension shall be represented by an address range. See 3.2.1 and 3.2.2 for net and variable declarations. The expression(s) that specify the indices of the array shall be constant expressions. The value of the constant expression can be a positive integer, a negative integer, or zero.

One declaration statement can be used for declaring both arrays and elements of the declared data type. This ability makes it convenient to declare both arrays and elements that match the element vector width in the same declaration statement.

An element can be assigned a value in a single assignment, but complete or partial array dimensions cannot. Nor can complete or partial array dimensions be used to provide a value to an expression. To assign a value to an element of an array, an index for every dimension shall be specified. The index can be an expression. This option provides a mechanism to reference different array elements depending on the value of other variables and nets in the circuit. For example, a program counter reg can be used to index into a RAM.

Implementations may limit the maximum size of an array, but they shall at least be $16777216\left(2^{24}\right)$.

### 3.10.1 Net arrays

Arrays of nets can be used to connect ports of generated instances. Each element of the array can be used in the same fashion as a scalar or vector net.

### 3.10.2 reg and variable arrays

Arrays for all variables types (reg, integer, time, real, realtime) shall be possible.

### 3.10.3 Memories

A one dimensional array with elements of type reg is also called a memory. These memories can be used to model read-only memories (ROMs), random access memories (RAMs), and reg files. Each reg in the array is known as an element or word and is addressed by a single array index.

An $n$-bit reg can be assigned a value in a single assignment, but a complete memory cannot. To assign a value to a memory word, an index shall be specified. The index can be an expression. This option provides a mechanism to reference different memory words, depending on the value of other variables and nets in the circuit. For example, a program counter reg could be used to index into a RAM.

### 3.10.3.1 Array examples

### 3.10.3.1.1 Array declarations

```
reg [ 7:0] mema[ 0:255]; // declares a memory mema of 256 8-bit
    // registers. The indices are 0 to 255
reg arrayb[ 7:0][ 0:255] ; // declare a two dimensional array of
    // one bit registers
wire w_array[ 7:0][ 5:0] ; // declare array of wires
integer inta[ 1:64]; // an array of 64 integer values
time chng_hist[ 1:1000] // an array of 1000 time values
integer t_index;
```


### 3.10.3.1.2 Assignment to array elements

The assignment statements in this section assume the presence of the declarations in 3.10.3.1.1.

```
mema = 0; // Illegal syntax- Attempt to write to entire array
arrayb[ 1] = 0; // Illegal Syntax - Attempt to write to elements
    // [ 1][ 0] ..[ 1][ 255]
arrayb[ 1][ 12:31] = 0; // Illegal Syntax - Attempt to write to
    // elements [ 1][ 12]..[ 1][ 31]
mema[ 1] = 0; //Assigns 0 to the second element of mema
arrayb[ 1][ 0] = 0; // Assigns 0 to the bit referenced by indices
    // [ 1][ 0]
inta[ 4] = 33559; // Assign decimal number to integer in array
chng_hist[ t_index] = $time; // Assign current simulation time to
    // element addressed by integer index
```


### 3.10.3.1.3 Memory differences

A memory of $n$ 1-bit regs is different from an $n$-bit vector reg

```
reg [ 1:n] rega; // An n-bit register is not the same
reg mema [ 1:n] ; // as a memory of n 1-bit registers
```


### 3.11 Parameters

Verilog HDL parameters do not belong to either the variable or the net group. Parameters are not variables, they are constants. There are two types of parameters: module parameters and specify parameters. It is illegal to redeclare a name already declared by a net, parameter or variable declaration.

Both types of parameters accept a range specification. By default, parameters and specparams shall be as wide as necessary to contain the value of the constant, except when a range specification is present.

### 3.11.1 Module parameters

The syntax for module parameter declarations is given in Syntax 3-4.

```
local_parameter_declaration ::= (From Annex A-A.2.1.1)
    localparam [ signed ] [ range ] list_of_param_assignments ;
    localparam integer list_of_param_assignments ;
    localparam real list_of_param_assignments ;
    localparam realtime list_of_param_assignments ;
    localparam time list_of_param_assignments ;
parameter_declaration ::=
        parameter [ signed ] [ range ] list_of_param_assignments ;
        parameter integer list_of_param_assignments ;
        parameter real list_of_param_assignments ;
        parameter realtime list_of_param_assignments ;
        parameter time list_of_param_assignments ;
list_of_param_assignments ::= (From Annex A - A.2.3)
        param_assignment \{, param_assignment \}
param_assignment ::= (From Annex A - A.2.4)
    parameter_identifier = constant_expression
range ::= (From Annex A-A.2.5)
    [ msb_constant_expression : lsb_constant_expression ]
```


## Syntax 3-4 Syntax for module parameter declaration

The list_of_param_assignments shall be a comma-separated list of assignments, where the right hand side of the assignment shall be a constant expression; that is, an expression containing only constant numbers and previously defined parameters. (See 4.)

The list_of_param_assignments can appear in a module as a set of module_items or in the module declaration in the module_parameter_port_list. (See 12.1). If any param_assignments appear in a module_parameter_port_list, then any param_assignments that appear in the module become local parameters and shall not be overridden by any method.

Parameters represent constants; hence, it is illegal to modify their value at runtime. However, module parameters can be modified at compilation time to have values that are different from those specified in the declaration assignment. This allows customization of module instances. A parameter can be modified with the defparam statement or in the module instance statement. Typical uses of parameters are to specify delays and width of variables. See Section 12 for details on parameter value assignment.

A module parameter can have a type specification and a range specification. The type and range of module parameters shall be in accordance with the following rules:

A parameter declaration with no type or range specification shall default to the type and range of the final value assigned to the parameter, after any value overrides have been applied.
A parameter with a range specification, but with no type specification, shall be the range of the parameter declaration and shall be unsigned. The sign and range shall not be affected by value overrides.
A parameter with a type specification, but with no range specification, shall be of the type specified. A signed parameter shall default to the range of the final value assigned to the parameter, after any value overrides have been applied.

A parameter with a signed type specification and with a range specification shall be signed, and shall be the range of its declaration. The sign and range shall not be affected by value overrides.
A parameter with no range specification, and with either a signed type specification or no type specification, shall have an implied range with an $l s b$ equal to 0 and an $m s b$ equal to one less than the size of the final value assigned to the parameter.
A parameter with no range specification, and with either a signed type specification or no type specification, and for which the final value assigned to it is unsized, shall have an implied range with an $l s b$ equal to 0 and an $m s b$ equal to an implementation-dependent value of at least 31 .

The conversion rules between real and integer values described in 3.9.2 apply to parameters as well.

## Examples:

```
    parameter msb = 7; // defines msb as a constant value 7
    parameter e = 25, f = 9; // defines two constant numbers
    parameter r = 5.7; // declares r as a real parameter
    parameter byte_size = 8,
    byte mask = byte size - 1;
    parameter averäge_delay = (r + f) / 2;
    parameter signed [ 3:0] mux_selector = 0;
    parameter real r1 = 3.5e17;
    parameter p1 = 13' h7e;
    parameter [ 31:0] dec const = 1'b1; // value converted to 32 bits
    parameter newconst = '3' h4; // implied range of [ 2:0]
    parameter newconst = 4; // implied range of at least
[ 31:0]
```


### 3.11.2 Local parameters - localparam

Verilog HDL localparam - local parameter(s) are identical to parameters except that they can not directly be modified with the defparam statement or by the ordered or named parameter value assignment. Local parameters can be assigned to a constant expression containing a parameter which can be modified with the defparam statement or by the ordered or named parameter value assignment. See 12.1.3 for details.

The syntax for local parameter declarations is given in Syntax 3-4.

### 3.11.3 Specify parameters

The syntax for declaring specify parameters is shown in Syntax 3-5.

```
specparam_declaration ::= (From Annex A-A.2.2.1)
    specparam [ range ] list_of_specparam_assignments ;
list_of_specparam_assignments \(::=\) (From Annex A-A.2.3)
    specparam_assignment \{ , specparam_assignment \}
specparam_assignment \(::=\) (From Annex A-A.2.4)
        specparam_identifier = constant_mintypmax_expression
        | pulse_control_specparam
pulse_control_specparam ::=
    PATHPULSE \(=(\) reject_limit_value [ , error_limit_value ] ) ;
    PATHPULSE\$specify_input_terminal_descriptor\$specify_output_terminal_descriptor
        = ( reject_limit_value [, error_limit_value ]);
error_limit_value ::=
    limit_value
reject_limit_value ::=
    limit_value
limit_value ::=
    constant_mintypmax_expression
range \(::=\) (From Annex \(A\) - A.2.5)
    [ msb_constant_expression : lsb_constant_expression ]
```


## Syntax 3-5 Syntax of the specparam declaration

The keyword specparam declares a special type of parameter which is intended only for providing timing and delay values, but can appear in any expression that is not assigned to a parameter and is not part of the range specification of a declaration. Originally permitted only in specify blocks (see Section 14), with this revision specify parameters (also called specparams) are now permitted both within the specify block and in the main module body.

A specify parameter declared outside a specify block shall be declared before it is referenced. The value assigned to a specify parameter can be any constant expression. A specify parameter can be used as part of a constant expression for a subsequent specify parameter declaration. Unlike a module parameter, a specify parameter cannot be modified from within the language, but it may be modified through SDF annotation (see Clause 16).

The specify parameters and module parameters shall not be interchangeable. In addition, module parameters shall not be assigned a constant expression that includes any specify parameters. Table 8 summarizes the differences between the two types of parameter declarations.

Table 8-Differences between specparams and parameters

| Specparams <br> (specify parameter) | Parameters <br> (module parameter) |
| :--- | :--- |
| Use keyword specparam | Use keyword parameter |
| Shall be declared inside a module or specify block | Shall be declared outside specify blocks |
| May only be used inside a module or specify block | May not be used inside specify blocks |
| May be assigned specparams and parameters | May not be assigned specparams |
| Use SDF annotation to override values | Use defparam or instance declaration <br> parameter value passing to override values |

A specify parameter can have a range specification. The range of specify parameters shall be in accordance with the following rules:

A specparam declaration with no range specification shall default to the range of the final value assigned to the parameter, after any value overrides have been applied.
A specparam with a range specification shall be the range of the parameter declaration. The range shall not be affected by value overrides.

Examples:

```
specify
    specparam tRise_clk_q = 150, tFall_clk_q = 200;
    specparam tRise_control = 40, tFall_control = 50;
endspecify
```

The lines between the keywords specify and endspecify declare four specify parameters. The first line declares specify parameters called tRise_clk_q and tFall_clk_q with values 150 and 200 respectively; the second line declares tRise_control and tFall_control specify parameters with values 40 and 50 respectively.

Examples:

```
module RAM16GEN (DOUT, DIN, ADR, WE, CE)
specparam dhold = 1.0;
specparam ddly = 1.0;
parameter width = 1;
parameter regsize = dhold + 1.0; // Illegal - can't assign
    // specparams to parameters
endmodule
```


### 3.12 Name spaces

In Verilog HDL, there are seven name spaces; two are global and five are local. The global name spaces are definitions and text macros. The definitions name space unifies all the module (see 12.1), macromodule (see 12.1), and primitive (see 8.1) definitions. Once a name is used to define a module, macromodule, or primitive, the name shall not be used again to declare another module, macromodule, or primitive.

The text macro name space is global. Since text macro names are introduced and used with a leading ' character, they remain unambiguous with any other name space (see 19.3). The text macro names are defined in the linear order of appearance in the set of input files that make up the description of the design unit. Subsequent definitions of the same name override the previous definitions for the balance of the input files.

There are five local name spaces: block, module, port, specify block, and attribute. Once a name is defined within one of the five name spaces, it shall not be defined again in that space (with the same or a different type).

The block name space is introduced by the named block (see 9.8), function (see 10.3), and task (see 10.2) constructs. It unifies the definitions of the named blocks, functions, tasks, parameters, named events and the variable type of declaration (see 3.2.2). The variable type of declaration includes the reg, integer, time, real, and realtime declarations.

The module name space is introduced by the module, macromodule, and primitive constructs. It unifies the definition of functions, tasks, named blocks, instance names, parameters, named events, net type of declaration, and variable type of declaration. The net type of declaration includes wire, wor, wand, tri, trior, triand, tri0, tri1, trireg, supply0, and supply1 (see 3.7).

The port name space is introduced by the module, macromodule, primitive, function, and task constructs. It provides a means of structurally defining connections between two objects that are in two different name spaces. The connection can be unidirectional (either input or output) or bidirectional (inout). The port name space overlaps the module and the block name spaces. Essentially, the port name space specifies the type of connection between names in different name spaces. The port type of declarations include input, output, and inout (see 12.3). A port name introduced in the port name space may be reintroduced in the module name space by declaring a variable or a wire with the same name as the port name.

The specify block name space is introduced by the specify construct (see 14.2).
The attribute name space is enclosed by the (* and *) constructs attached to a language element (see 2.8). An attribute name can be defined and used only in the attribute name space. Any other type of name cannot be defined in this name space.

## 4. Expressions

This clause describes the operators and operands available in the Verilog HDL and how to use them to form expressions.

An expression is a construct that combines operands with operators to produce a result that is a function of the values of the operands and the semantic meaning of the operator. Any legal operand, such as a net bitselect, without any operator is considered an expression. Wherever a value is needed in a Verilog HDL statement, an expression can be used.

Some statement constructs require an expression to be a constant expression. The operands of a constant expression consist of constant numbers, parameter names, constant bit-selects of parameters, constant partselects of parameters, and constant function calls (see 10.3.5) only, but they can use any of the operators defined in Table 9.

A scalar expression is an expression that evaluates to a scalar (single-bit) result. If the expression evaluates to a vector (multibit) result, then the least significant bit of the result is used as the scalar result.

The data types reg, integer, time, real, and realtime are all variable data types. Descriptions pertaining to variable usage apply to all of these data types.

An operand can be one of the following:
Constant number (including real)
Net
Variables of type reg, integer, time, real, and realtime
Net bit-select
Bit-select of type reg, integer, and time
Net part-select
Part-select of type reg, integer, and time
Array element
A call to a user-defined function or system-defined function that returns any of the above

### 4.1 Operators

The symbols for the Verilog HDL operators are similar to those in the C programming language. Table 9 lists these operators.

Table 9-Operators in the Verilog HDL

| $\}\{\}\}$ | Concatenation, replication |
| :--- | :--- |
| $+-^{* / * *}$ | Arithmetic |
| $\%$ | Modulus |
| $\gg=\ll=$ | Relational |
| $!$ | Logical negation |
| $\& \&$ | Logical and |
| $\\|$ | Logical or |
| $==$ | Logical equality |
| $!=$ | Logical inequality |

Table 9-Operators in the Verilog HDL (continued)

| $=$ | Case equality |
| :---: | :---: |
| ! $==$ | Case inequality |
| $\sim$ | Bit-wise negation |
| \& | Bit-wise and |
| \| | Bit-wise inclusive or |
| $\wedge$ | Bit-wise exclusive or |
| $\wedge \sim$ or ~^ | Bit-wise equivalence |
| \& | Reduction and |
| $\sim$ \& | Reduction nand |
| 1 | Reduction or |
| $\sim$ | Reduction nor |
| $\wedge$ | Reduction xor |
| $\sim \wedge$ or $\wedge \sim$ | Reduction xnor |
| << | Logical left shift |
| >> | Logical right shift |
| <<< | Arithmetic left shift |
| >>> | Arithmetic right shift |
| ? : | Conditional |
| or | Event or |

### 4.1.1 Operators with real operands

The operators shown in Table 10 shall be legal when applied to real operands. All other operators shall be considered illegal when used with real operands.

Table 10-Legal operators for use in real expressions

| unary + unary - | Unary operators |
| :--- | :--- |
| $+\quad * / * *$ | Arithmetic |
| $\gg=\ll=$ | Relational |
| $!\& \& \\|$ | Logical |
| $==\quad!=$ | Logical equality |
| $?$ or | Conditional |
| or | Event or |

The result of using logical or relational operators on real numbers is a single-bit scalar value.

Table 11 lists operators that shall not be used to operate on real numbers.
Table 11-Operators not allowed for real expressions

| \{\} $\{$ \{ \} $\}$ | Concatenate, replicate |
| :---: | :---: |
| \% | Modulus |
| === ! $=$ | Case equality |
| $\left.\sim_{\wedge}^{\sim} \&\right\|_{\sim \wedge}$ | Bit-wise |
| ^ ^~~~ $\& \sim \&\|\sim\|$ | Reduction |
| << >> <<< >>> | Shift |

See 3.9.1 for more information on use of real numbers.

### 4.1.2 Binary operator precedence

The precedence order of binary operators and the conditional operator (?:) is shown in Table 12. The Verilog HDL has two equality operators. They are discussed in 4.1.8.

Table 12-Precedence rules for operators

| +- ! ( unary) | Highest precedence |
| :---: | :---: |
| ** |  |
| * / \% |  |
| + - (binary) |  |
| << >> <<< >>> |  |
| \ll $=\gg=$ |  |
| $=$ ! $=$ === ! $==$ |  |
| \& ~\& |  |
| $\wedge \sim \sim \sim$ |  |
| \| ~ |  |
| \&\& |  |
| \|| |  |
| ?: (conditional operator) | Lowest precedence |

Operators shown on the same row in Table 12 shall have the same precedence. Rows are arranged in order of decreasing precedence for the operators. For example, *, $/$, and $\%$ all have the same precedence, which is higher than that of the binary + and - operators.

All operators shall associate left to right with the exception of the conditional operator, which shall associate right to left. Associativity refers to the order in which the operators having the same precedence are evaluated. Thus, in the following example $B$ is added to $A$ and then $C$ is subtracted from the result of $A+B$.

$$
A+B-C
$$

When operators differ in precedence, the operators with higher precedence shall associate first. In the following example, $B$ is divided by $C$ (division has higher precedence than addition) and then the result is added to A.

$$
A+B / C
$$

Parentheses can be used to change the operator precedence.

```
(A + B) / C // not the same as A + B / C
```


### 4.1.3 Using integer numbers in expressions

Integer numbers can be used as operands in expressions. An integer number can be expressed as
An unsized, unbased integer (e.g., 12)
An unsized, based integer (e.g., ' d12, ' sd12)
A sized, based integer (e.g., $16^{\prime} \mathrm{d} 12,16^{\prime} \mathrm{sd} 12$ )

A negative value for an integer with no base specifier shall be interpreted differently than for an integer with a base specifier. An integer with no base specifier shall be interpreted as a signed value in 2 s complement form. An integer with an unsigned base specifier shall be interpreted as an unsigned value.

## Example:

This example shows four ways to write the expression minus 12 divided by 3 . Note that -12 and $-^{\prime} d 12$ both evaluate to the same 2 s complement bit pattern, but, in an expression, the $-^{\prime} \mathrm{d} 12$ loses its identity as a signed negative number.

```
integer IntA;
IntA = -12 / 3; // The result is -4.
IntA = _' d 12 / 3; // The result is 1431655761.
IntA = -' sd 12 / 3; // The result is -4.
IntA = -4'sd 12 / 3; // -4'sd12 is the negative of the 4-bit
    // quantity 1100, which is -4. -(-4) = 4.
    // The result is 1.
```


### 4.1.4 Expression evaluation order

The operators shall follow the associativity rules while evaluating an expression as described in 4.1.2. However, if the final result of an expression can be determined early, the entire expression need not be evaluated. This is called short-circuiting an expression evaluation.

## Example:

```
reg regA, regB, regC, result ;
result = regA & (regB | regC) ;
```

If regA is known to be zero, the result of the expression can be determined as zero without evaluating the sub-expression regB | regC.

### 4.1.5 Arithmetic operators

The binary arithmetic operators are given in Table 13.
Table 13-Arithmetic operators defined

| $a+b$ | $a$ plus $b$ |
| :--- | :--- |
| $a-b$ | a minus b |
| $a * b$ | a multiplied by b <br> (or a times b) |
| $a / b$ | a divided by b |
| $a \% b$ | a modulo $b$ |
| $a^{* *}$ b | a to the power of $b$ |

The integer division shall truncate any fractional part toward zero. For the division or modulus operators, if the second operand is a zero, then the entire result value shall be x . The modulus operator, for example $y \%$ $z$, gives the remainder when the first operand is divided by the second, and thus is zero when $z$ divides $y$ exactly. The result of a modulus operation shall take the sign of the first operand.

The result of the power operator shall be real if either operand is a real, integer, or signed. If both operands are unsigned then the result shall be unsigned. The result of the power operator is unspecified if the first operand is zero and the second operand is non-positive, or if the first operand is negative and the second operand is not an integral value.

The unary arithmetic operators shall take precedence over the binary operators. The unary operators are given in Table 14.

Table 14-Unary operators defined

| +m | Unary plus m (same as m ) |
| :--- | :--- |
| -m | Unary minus m |

For the arithmetic operators, if any operand bit value is the unknown value x or the high-impedance value z , then the entire result value shall be x .

## Example:

Table 15 gives examples of modulus operations.

Table 15-Examples of modulus operators

| Modulus expression | Result | Comments |
| :--- | :--- | :--- |
| $10 \% 3$ | 1 | $10 / 3$ yields a remainder of 1 |
| $11 \% 3$ | 2 | $11 / 3$ yields a remainder of 2 |
| $12 \% 3$ | 0 | $12 / 3$ yields no remainder |
| $-10 \% 3$ | -1 | The result takes the sign of the first operand |
| $11 \%-3$ | 2 | The result takes the sign of the first operand |
| $-4 \mathrm{~d} 12 \% 3$ | 1 | -4 d12 is seen as a large, positive number that leaves a remain- <br> der of 1 when divided by 3 |

### 4.1.6 Arithmetic expressions with regs and integers

A reg data type shall be treated as an unsigned value unless explicitly declared to be signed. An integer variable shall be treated as signed. Signed values shall use a 2's complement representation. Conversions between signed and unsigned values shall keep the same bit representation; only the interpretation changes.

Table 16 lists how arithmetic operators interpret each data type.

## Table 16-Data type interpretation by arithmetic operators

| Data type | Interpretation |
| :--- | :--- |
| unsigned net | Unsigned |
| signed net | Signed, 2 s complement |
| unsigned reg | Unsigned |
| signed reg | Signed, 2 s complement |
| integer | Signed, 2 s complement |
| time | Unsigned |
| real, realtime | Signed, floating point |

## Example:

The following example shows various ways to divide minus twelve by three using integer and reg data types in expressions.

```
integer intA;
reg [ 15:0] regA;
reg signed [ 15:0] regS;
intA = -4' d12;
regA = intA / 3; // expression result is -4,
    // intA is an integer data type, regA is 65532
regA = -4' d12; // regA is 65524
intA = regA / 3; // expression result is 21841,
    // regA is a reg data type
intA = -4' d12 / 3;// expression result is 1431655761.
    // -4' d12 is effectively a 32-bit reg data type
regA = -12 / 3; // expression result is -4, -12 is effectively
    // an integer data type. regA is 65532
regS = -12 / 3; // expression result is -4. regS is a signed
    // reg
regS = -4' sd12 / 3;// expression result is 1. -4' sol2 is actually
    // 4. The rules for integer division yield 4/3==1
```


### 4.1.7 Relational operators

Table 17 lists and defines the relational operators.
Table 17-Definitions of the relational operators

| $\mathrm{a}<\mathrm{b}$ | a less than b |
| :--- | :--- |
| $\mathrm{a}>\mathrm{b}$ | a greater than b |
| $\mathrm{a}<=\mathrm{b}$ | a less than or equal to b |
| $\mathrm{a}>=\mathrm{b}$ | a greater than or equal to b |

An expression using these relational operators shall yield the scalar value 0 if the specified relation is false or the value 1 if it is true. If either operand of a relational operator contains an unknown ( x ) or high impedance ( z ) value, then the result shall be a 1-bit unknown value ( x ).

When two operands of unequal bit lengths are used, the smaller operand shall be zero filled on the most significant bit side to extend to the size of the larger operand.

All the relational operators shall have the same precedence. Relational operators shall have lower precedence than arithmetic operators.

## Examples:

The following examples illustrate the implications of this precedence rule:

```
a < foo - 1 // this expression is the same as
a < (foo - 1) // this expression, but . . .
foo - (1 < a) // this one is not the same as
foo - < a // this expression
```

When foo - ( $1<a)$ evaluates, the relational expression evaluates first and then either zero or one is subtracted from foo. When foo - $1<$ a evaluates, the value of foo operand is reduced by one and then compared with a.

When both operands of a relational expression are signed integral operands (an integer, or a unsized, unbased integer) then the expression shall be interpreted as a comparison between signed values. When either operand of a relational expression is a real operand then the other operand shall be converted to an equivalent real value, and the expression shall be interpreted as a comparison between two real values.

Otherwise the expression shall be interpreted as a comparison between unsigned values.

### 4.1.8 Equality operators

The equality operators shall rank lower in precedence than the relational operators. Table 18 lists and defines the equality operators.

## Table 18-Definitions of the equality operators

| $\mathrm{a}===\mathrm{b}$ | a equal to b, including x and z |
| :--- | :--- |
| $\mathrm{a}!==\mathrm{b}$ | a not equal to b, including x and z |
| $\mathrm{a}==\mathrm{b}$ | a equal to $b$, result may be unknown |
| $\mathrm{a}!=\mathrm{b}$ | a not equal to b, result may be unknown |

All four equality operators shall have the same precedence. These four operators compare operands bit for bit, with zero filling if the two operands are of unequal bit length. As with the relational operators, the result shall be 0 if comparison fails, 1 if it succeeds.

For the logical equality and logical inequality operators (== and $!=$ ), if, due to unknown or high-impedance bits in the operands, the relation is ambiguous, then the result shall be a one bit unknown value ( x ).

For the case equality and case inequality operators ( $===$ and $!==$ ), the comparison shall be done just as it is in the procedural case statement (see 9.5). Bits that are x or z shall be included in the comparison and shall match for the result to be considered equal. The result of these operators shall always be a known value, either 1 or 0 .

### 4.1.9 Logical operators

The operators logical and $(\& \&)$ and logical or $(\|)$ are logical connectives. The result of the evaluation of a logical comparison shall be 1 (defined as true), 0 (defined as false), or, if the result is ambiguous, the unknown value (x). The precedence of $\& \&$ is greater than that of \|।, and both are lower than relational and equality operators.

A third logical operator is the unary logical negation operator (!). The negation operator converts a nonzero or true operand into 0 and a zero or false operand into 1 . An ambiguous truth value remains as $x$.

## Examples:

Example 1 If reg alpha holds the integer value 237 and beta holds the value zero, then the following examples perform as described:

```
regA = alpha && beta; // regA is set to 0
regB = alpha || beta; // regB is set to 1
```

Example 2 The following expression performs a logical and of three subexpressions without needing any parentheses:

```
a < size-1 && b != c && index != lastone
```

However, it is recommended for readability purposes that parentheses be used to show very clearly the precedence intended, as in the following rewrite of this example:

```
(a < size-1) && (b != c) && (index != lastone)
```

Example 3 A common use of! is in constructions like the following:
if (! inword)

In some cases, the preceding construct makes more sense to someone reading the code than this equivalent construct:

```
if (inword == 0)
```


### 4.1.10 Bit-wise operators

The bit-wise operators shall perform bit-wise manipulations on the operands that is, the operator shall combine a bit in one operand with its corresponding bit in the other operand to calculate one bit for the result. Logic Tables 19 through 23 show the results for each possible calculation.

Table 19-Bit-wise binary and operator

| $\boldsymbol{\&}$ | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{x}$ | $\mathbf{z}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\mathbf{0}$ | 0 | 0 | 0 | 0 |
| $\mathbf{1}$ | 0 | 1 | x | x |
| $\mathbf{x}$ | 0 | x | x | x |
| $\mathbf{z}$ | 0 | x | x | x |

Table 20-Bit-wise binary or operator

| $\mathbf{I}$ | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{x}$ | $\mathbf{z}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\mathbf{0}$ | 0 | 1 | x | x |
| $\mathbf{1}$ | 1 | 1 | 1 | 1 |
| $\mathbf{x}$ | x | 1 | x | x |
| $\mathbf{z}$ | x | 1 | x | x |

Table 21-Bit-wise binary exclusive or operator

| $\boldsymbol{\wedge}$ | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{x}$ | $\mathbf{z}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\mathbf{0}$ | 0 | 1 | x | x |
| $\mathbf{1}$ | 1 | 0 | x | x |
| $\mathbf{x}$ | x | x | x | x |
| $\mathbf{z}$ | x | x | x | x |

Table 22-Bit-wise binary exclusive nor operator

| $\sim \sim$ <br> $\sim$ | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{x}$ | $\mathbf{z}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\mathbf{0}$ | 1 | 0 | x | x |
| $\mathbf{1}$ | 0 | 1 | x | x |
| $\mathbf{x}$ | x | x | x | x |
| $\mathbf{z}$ | x | x | x | x |

Table 23-Bit-wise unary negation operator

| $\boldsymbol{\sim}$ |  |
| :---: | :---: |
| $\mathbf{0}$ | 1 |
| $\mathbf{1}$ | 0 |
| $\mathbf{x}$ | x |
| $\mathbf{z}$ | x |

When the operands are of unequal bit length, the shorter operand is zero-filled in the most significant bit positions.

### 4.1.11 Reduction operators

The unary reduction operators shall perform a bit-wise operation on a single operand to produce a single bit result. For reduction and, reduction or, and reduction xor operators, the first step of the operation shall apply the operator between the first bit of the operand and the second using logic Tables 24 through 26. The second and subsequent steps shall apply the operator between the 1-bit result of the prior step and the next bit of the operand using the same logic table. For reduction nand, reduction nor, and reduction xnor operators, the result shall be computed by inverting the result of the reduction and, reduction or, and reduction xor operation respectively.

Table 24-Reduction unary and operator

| $\boldsymbol{\&}$ | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{x}$ | $\mathbf{z}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\mathbf{0}$ | 0 | 0 | 0 | 0 |
| $\mathbf{1}$ | 0 | 1 | x | x |
| $\mathbf{x}$ | 0 | x | x | x |
| $\mathbf{z}$ | 0 | x | x | x |

Table 25-Reduction unary or operator

| $\mid$ | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{x}$ | $\mathbf{z}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\mathbf{0}$ | 0 | 1 | x | x |
| $\mathbf{1}$ | 1 | 1 | 1 | 1 |
| $\mathbf{x}$ | x | 1 | x | x |
| $\mathbf{z}$ | x | 1 | x | x |

Table 26-Reduction unary exclusive or operator

| $\wedge$ | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{x}$ | $\mathbf{z}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\mathbf{0}$ | 0 | 1 | x | x |
| $\mathbf{1}$ | 1 | 0 | x | x |
| $\mathbf{x}$ | x | x | x | x |
| $\mathbf{z}$ | x | x | x | x |

## Example:

Table 27 shows the results of applying reduction operators on different operands.
Table 27-Results of unary reduction operations

| Operand | $\boldsymbol{\&}$ | $\sim \boldsymbol{\sim}$ |  | $\sim$ | $\wedge$ | $\sim \wedge$ | Comments |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 4 b0000 | 0 | 1 | 0 | 1 | 0 | 1 | No bits set |
| 4 b1111 | 1 | 0 | 1 | 0 | 0 | 1 | All bits set |
| 4 b0110 | 0 | 1 | 1 | 0 | 0 | 1 | Even number of bits <br> set |
| 4 b1000 | 0 | 1 | 1 | 0 | 1 | 0 | Odd number of bits set |

### 4.1.12 Shift operators

There are two types of shift operators, the logical shift operators, $\ll$ and $\gg$, and the arithmetic shift operators, $\lll$ and $\ggg$. The left shift operators, $\ll$ and $\lll$, shall shift their left operand to the left by the number by the number of bit positions given by the right operand. In both cases, the vacated bit positions shall be filled with zeroes. The right shift operators, >> and >>>, shall shift their left operand to the right by the number of bit positions given by the right operand. The logical right shift shall fill the vacated bit positions with
zeroes. The arithmetic right shift shall fill the vacated bit positions with zeroes if the result type is unsigned. It shall fill the vacated bit positions with the value of the most-significant (i.e., sign) bit of the left operand if the result type is signed. If the right operand has an unknown or high impedence value, then the result shall be unknown. The right operand is always treated as an unsigned number and has no effect on the signedness of the result. The result signedness is determined by the left-hand operand and the remainder of the expression, as outlined in 4.5.1.

## Examples:

Example 1 In this example, the reg result is assigned the binary value 0100 , which is 0001 shifted to the left two positions and zero-filled.

```
module shift;
reg [ 3:0] start, result;
initial begin
    start = 1;
    result = (start << 2);
end
endmodule
```

Example 2 In this example, the reg result is assigned the binary value 1110 , which is 1000 shifted to the right two positions and sign-filled.

```
module ashift;
reg signed [ 3:0] start, result;
initial begin
        start = 4' b1000;
    result = (start >>> 2);
end
endmodule
```


### 4.1.13 Conditional operator

The conditional operator, also known as ternary operator, shall be right associative and shall be constructed using three operands separated by two operators in the format given in Syntax 4-1.

```
conditional_expression ::= (From Annex A - A.8.3)
        expression1 ? { attribute_instance } expression2: expression3
expression1 ::=
        expression
expression2 ::=
        expression
expression3 ::=
        expression
```


## Syntax 4-1 Syntax for conditional operator

The evaluation of a conditional operator shall begin with the evaluation of expression1. If expression1 evaluates to false (0), then expression3 shall be evaluated and used as the result of the conditional expression. If expression 1 evaluates to true (known value other than 0 ), then expression 2 is evaluated and used as the result. If expression1 evaluates to ambiguous value ( x or z ), then both expression 2 and expression 3 shall be evaluated and their results shall be combined, bit by bit, using Table 28 to calculate the final result unless
expression 2 or expression 3 is real, in which case the result shall be 0 . If the lengths of expression 2 and expression 3 are different, the shorter operand shall be lengthened to match the longer and zero-filled from the left (the high-order end).

Table 28-Ambiguous condition results for conditional operator

| $\mathbf{? :}$ | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{x}$ | z |
| :---: | :---: | :---: | :---: | :---: |
| $\mathbf{0}$ | 0 | x | x | x |
| $\mathbf{1}$ | x | 1 | x | x |
| $\mathbf{x}$ | x | x | x | x |
| $\mathbf{z}$ | x | x | x | x |

## Example:

The following example of a three-state output bus illustrates a common use of the conditional operator.

```
wire [ 15:0] busa = drive_busa ? data : 16' bz;
```

The bus called data is driven onto busa when drive_busa is 1 . If drive_busa is unknown, then an unknown value is driven onto busa. Otherwise, busa is not driven.

### 4.1.14 Concatenations

A concatenation is the joining together of bits resulting from two or more expressions. The concatenation shall be expressed using the brace characters $\{$ and $\}$, with commas separating the expressions within.

Unsized constant numbers shall not be allowed in concatenations. This is because the size of each operand in the concatenation is needed to calculate the complete size of the concatenation.

## Examples:

This example concatenates four expressions:

$$
\left\{\mathrm{a}, \mathrm{~b}[3: 0], \mathrm{w}, 3^{\prime} \mathrm{b} 101\right\}
$$

and it is equivalent to the following example:

$$
\left\{\mathrm{a}, \mathrm{~b}[3], \mathrm{b}[2], \mathrm{b}[1], \mathrm{b}[0], \mathrm{w}, 1^{\prime} \mathrm{b} 1,1^{\prime} \mathrm{b} 0,1^{\prime} \mathrm{b} 1\right\}
$$

Another form of concatenation is the replication operation. The first expression shall be a non-zero, non-X and non-Z constant expression, the second expression follows the rules for concatenations. This example replicates "w" 4 times.

```
{4{w}} // This is equivalent to {w, w, w, w}
a[ 31:0] = { 1' b1, { O{ 1' b0}} } ; //illegal. RHS becomes { 1' b1,;
a[ 31:0] = { 1'b1, { 1'bz{ 1'b0}} } ; //illegal. RHS becomes { 1'b1,;
a[ 31:0] = { 1'b1, { 1'bx{ 1'b0}} }; //illegal. RHS becomes { 1' b1,;
```

If the replication operator is used on a function call operand, the function need not be evaluated multiple times. For example:

```
result = {4{func(w)}}
```

may be computed as

```
result = { func(w), func(w), func(w), func(w)}
```

or

```
y = func(w) ;
result = {y, y, y, y}
```

This is another form of expression evaluation short-circuiting.
The next example illustrates nested concatenations:

```
{b, { {{a, b}}} // This is equivalent to {b, a, b, a, b, a, b}
```


### 4.1.15 Event or

The event or operator shall perform an or of events. The , operator does the same thing. See 9.7 for events and triggering of events.

## Example:

The following example shows both ways to make an assignment to rega when an event (change) occurs on trig or enable.

```
@(trig or enable) rega = regb ;
@(trig , enable) rega = regb ;
```


## 4.2 perands

There are several types of operands that can be specified in expressions. The simplest type is a reference to a net or variable in its complete form that is, just the name of the net or variable is given. In this case, all of the bits making up the net or variable value shall be used as the operand.

If a single bit of a vector net, reg variable, integer variable, or time variable is required, then a bit-select operand shall be used. A part-select operand shall be used to reference a group of adjacent bits in a vector net, vector reg, integer variable, or time variable.

A memory word can be referenced as an operand. A concatenation of other operands (including nested concatenations) can be specified as an operand. A function call is an operand.

### 4.2.1 Vector bit-select and part-select addressing

Bit-selects extract a particular bit from a vector net, vector reg, integer variable, or time variable. The bit can be addressed using an expression. If the bit-select is out of the address bounds or the bit-select is x or z , then the value returned by the reference shall be x . The bit-select or part-select of a variable declared as real or realtime shall be considered illegal.

Several contiguous bits in a vector net, vector reg, integer variable, or time variable can be addressed and are known as part-selects. There are two types of part-selects, a constant part-select and an indexed part-select.

A constant part-select of a vector reg or net is given with the following syntax:

```
vect[ msb_expr:lsb_expr]
```

Both expressions shall be constant expressions. The first expression has to address a more significant bit than the second expression. If the part-select is out of the address bounds or the part-select is x or z , then the value returned by the reference shall be x .

An indexed part select of a vector net, vector reg, integer variable, or time variable is given with the following syntax:

```
reg [ 15:0] big_vect;
reg [ 0:15] little_vect;
    big_vect[ lsb_base_expr +: width_expr]
little_vect[ msb_base_expr +: width_expr]
    big_vect[ msb_base_expr -: width_expr]
little_vect[ lsb__base_expr -: width_expr]
```

The width_expr shall be a constant expression. It also shall not be affected by run-time parameter assignments. The lsb_base_expr and msb_base_expr can vary at run-time. The first two examples select bits starting at the base and ascending the bit range. The number of bits selected is equal to the width expression. The second two examples select bits starting at the base and descending the bit range. Part-selects that address a range of bits that are completely out of the address bounds of the net, reg, integer, or time, or when the part-select is $x$ or $z$, shall yield the value $x$ when read, and shall have no effect on the data stored when written. Part-selects that are partially out of range shall when read return $x$ for the bits that are out of range, and when written shall only affect the bits that are in range.

## Examples:

```
reg [ 31:0] big_vect;
reg [ 0:31] little vect;
reg [ 63:0] dword;
integer sel;
```

The first four if statements show the identity between the two part select constructs. The last one shows an indexable nature.

```
initial begin
    if (big_vect[ 0 +:8] == big_vect[7 : 0] ) begin end
    if (little_vect[ 0 +:8] == little_vect[ 0 : 7]) begin end
    if (big_vect[ 15 -: 8] == big_vect[ 15 : 8]) begin end
    if (little_vect[ 15 -: 8] == little_vect[ 8 :15]) begin end
    if (sel >0 && sel < 8)
            dword[ 8* sel +:8] = big_vect[ 7:0] ; // Replace the byte
selected.
```


## Examples:

Example 1 The following example specifies the single bit of acc vector that is addressed by the operand index.
acc[ index]

The actual bit that is accessed by an address is, in part, determined by the declaration of acc. For instance, each of the declarations of acc shown in the next example causes a particular value of index to access a different bit:

```
reg [ 15:0] acc;
reg [ 2:17] acc
```

Example 2 The next example and the bullet items that follow it illustrate the principles of bit addressing. The code declares an 8-bit reg called vect and initializes it to a value of 4 . The list describes how the separate bits of that vector can be addressed.

```
reg [ 7:0] vect;
vect = 4;// fills vect with the pattern 00000100
    // msb is bit 7, lsb is bit 0
```

If the value of addr is 2, then vect[ addr] returns 1 .
If the value of addr is out of bounds, then vect[ addr] returns $x$.
If addr is 0,1 , or 3 through 7 , vect[ addr] returns 0 .
vect[ 3:0] returns the bits 0100 .
vect[ 5:1] returns the bits 00010 .
vect $[$ expression that returns $x$ ] returns x .
vect [ expression that returns $z$ ] returns x .
If any bit of addr is $x$ or $z$, then the value of addr is $x$.

## NOTES:

1) Part-select indices that evaluate to $x$ or $z$ may be flagged as a compile time error.
2) Bit-select or part-select indices that are outside of the declared range may be flagged as a compile time error.

### 4.2.2 Array and memory addressing

Declaration of arrays and memories (one dimensional arrays of reg) are discussed in 3.10. This subclause discusses array addressing.

## Examples:

The next example declares a memory of 1024 8-bit words:

```
reg [ 7:0] mem_name[ 0:1023];
```

The syntax for a memory address shall consist of the name of the memory and an expression for the address, specified with the following format:

```
mem_name[ addr_expr]
```

The addr_expr can be any expression; therefore, memory indirections can be specified in a single expression. The next example illustrates memory indirection:

```
mem_name[ mem_name[ 3] ]
```

In this example, mem name[ 3] addresses word three of the memory called mem name. The value at word three is the index into mem_name that is used by the memory address mem_name[ mem_name[3] ]. As with bit-selects, the address bounds given in the declaration of the memory determine the effect of the address expression. If the index is out of the address bounds or if any bit in the address is $x$ or $z$, then the value of the reference shall be $x$.

## Examples:

The next example declares an array of 256 by 256 8-bit elements and an array 256 by 256 by 8 1-bit elements:

```
reg [ 7:0] twod_array[ 0:255][0:255] ;
wire threed_array[ 0:255][0:255][0:7] ;
```

The syntax for access to the array shall consist of the name of the memory or array and an expression for each addressed dimension:

```
twod_array[ addr_expr][ addr_expr]
three\overline{d_array[ ad\overline{d}r_expr][ ad\overline{d}r_expr][ addr_expr]}]
```

As before, the addr_expr can be any expression. The array twod_array accesses a whole 8-bit vector, while the array threed_array accesses a single bit of the three dimensional array.

To express bit selects or part selects of array elements, the desired word shall first be selected by supplying an address for each dimension. Once selected, bit and part selects shall be addressed in the same manner as net and reg bit and part selects (see 4.2.1).

## Examples:

```
twod array[ 14][ 1][ 3:0] // access lower 4 bits of word
twod_array[ 1][ 3][ 6] // access bit 6 of word
twod_array[ 1][ 3][ sel] // use variable bit select
threed_array[ 14][ 1][ 3:0] // Illegal
```


### 4.2.3 Strings

String operands shall be treated as constant numbers consisting of a sequence of 8-bit ASCII codes, one per character. Any Verilog HDL operator can manipulate string operands. The operator shall behave as though the entire string were a single numeric value.

When a variable is larger than required to hold the value being assigned, the contents after the assignment shall be padded on the left with zeros. This is consistent with the padding that occurs during assignment of nonstring values.

## Example:

The following example declares a string variable large enough to hold 14 characters and assigns a value to it. The example then manipulates the string using the concatenation operator.

```
module string_test;
reg [ 8*14:1] stringvar;
initial begin
    stringvar = "Hello world";
    \$display("\%s is stored as \%h", stringvar, stringvar);
    stringvar = \{stringvar,"!!!"\};
    \$display("\%s is stored as \%h", stringvar, stringvar);
end
endmodule
```

The result of simulating the above description is

```
    Hello world is stored as 00000048656c6c6f20776f726c64
Hello world!!! is stored as 48656c6c6f20776f726c64212121
```


### 4.2.3.1 String operations

The common string operations copy, concatenate, and compare are supported by Verilog HDL operators. Copy is provided by simple assignment. Concatenation is provided by the concatenation operator. Comparison is provided by the equality operators.

When manipulating string values in vector regs, the regs should be at least $8^{*} \mathrm{n}$ bits (where n is the number of ASCII characters) in order to preserve the 8-bit ASCII code.

### 4.2.3.2 String value padding and potential problems

When strings are assigned to variables, the values stored shall be padded on the left with zeros. Padding can affect the results of comparison and concatenation operations. The comparison and concatenation operators shall not distinguish between zeros resulting from padding and the original string characters ( $\backslash 0$, ASCII NULL).

Examples:
The following example illustrates the potential problem.

```
reg [ 8* 10:1] s1, s2;
initial begin
    s1 = "Hello";
    s2 = " world!";
    if ({s1,s2} == "Hello world!")
        $display("strings are equal");
end
```

The comparison in this example fails because during the assignment the string variables are padded as illustrated in the next example:

```
s1 = 000000000048656c6c6f
s2 = 00000020776f726c6421
```

The concatenation of $s 1$ and $s 2$ includes the zero padding, resulting in the following value:
$000000000048656 c 6 c 6 f 00000020776 f 726 c 6421$

Since the string Hello world! contains no zero padding, the comparison fails, as shown in the following example:


This comparison yields a result of zero, which is equivalent to false.

### 4.2.3.3 Null string handling

The null string ("") shall be considered equivalent to the ASCII NULL (" $\backslash 0$ ") which has a value zero (0), which is different from a string " 0 ".

### 4.3 Minimum, typical, and maximum delay expressions

Verilog HDL delay expressions can be specified as three expressions separated by colons and enclosed by parentheses. This is intended to represent minimum, typical, and maximum values in that order. The syntax is given in Syntax 4-2.

```
constant_expression \(::=(\) From Annex A-A.8.3)
        constant_primary
        unary_operator \{ attribute_instance \} constant_primary
        constant_expression binary_operator \{ attribute_instance \} constant_expression
        constant_expression ? \{ attribute_instance \} constant_expression
            constant_expression
        \(\mid\) string
constant_mintypmax_expression ::=
        constant_expression
        | constant_expression : constant_expression : constant_expression
expression ::=
        primary
        | unary_operator \{ attribute_instance \} primary
        expression binary_operator \{ attribute_instance \} expression
        conditional_expression
        | string
mintypmax_expression ::=
        expression
    | expression : expression : expression
constant_primary \(::=(\) From Annex \(A-\) A.8.4 \()\)
        constant_concatenation
        constant_function_call
        ( constant_mintypmax_expression )
        constant_multiple_concatenation
        genvar_identifier
        number
        | parameter_identifier
        | specparam_identifier
primary ::=
        number
        | hierarchical_identifier
        hierarchical_identifier [ expression ] \{ [ expression ] \}
        hierarchical_identifier [ expression ] \{ [ expression ] \} [range_expression ]
        hierarchical_identifier [ range_expression ]
        concatenation
        multiple_concatenation
        function_call
        | system_function_call
        constant_function_call
    ( mintypmax_expression )
```

Syntax 4-2 Syntax for mintypmax expression

Verilog HDL models typically specify three values for delay expressions. The three values allow a design to be tested with minimum, typical, or maximum delay values.

Values expressed in min:typ:max format can be used in expressions. The min:typ:max format can be used wherever expressions can appear.

## Examples:

Example 1 This example shows an expression that defines a single triplet of delay values. The minimum value is the sum of $a+d$; the typical value is $b+e$; the maximum value is $c+f$, as follows:

$$
(a: b: c)+(d: e: f)
$$

Example 2 The next example shows a typical expression that is used to specify min:typ:max format values:

```
val - (32' d 50: 32' d 75: 32' d 100)
```


### 4.4 Expression bit lengths

Controlling the number of bits that are used in expression evaluations is important if consistent results are to be achieved. Some situations have a simple solution; for example, if a bit-wise and operation is specified on two 16 -bit regs, then the result is a 16-bit value. However, in some situations it is not obvious how many bits are used to evaluate an expression, or what size the result should be.

For example, should an arithmetic add of two 16-bit values perform the evaluation using 16 bits, or should the evaluation use 17 bits in order to allow for a possible carry overflow? The answer depends on the type of device being modeled, and whether that device handles carry overflow. The Verilog HDL uses the bit length of the operands to determine how many bits to use while evaluating an expression. The bit length rules are given in 4.4.1. In the case of the addition operator, the bit length of the largest operand, including the lefthand side of an assignment, shall be used.

## Examples:

```
reg [ 15:0] a, b; // 16-bit regs
reg [ 15:0] sumA; // 16-bit reg
reg [ 16:0] sumB; // 17-bit reg
sumA = a + b; // expression evaluates using 16 bits
sumB = a + b; // expression evaluates using 17 bits
```


### 4.4.1 ules for expression bit lengths

The rules governing the expression bit lengths have been formulated so that most practical situations have a natural solution.

The number of bits of an expression (known as the size of the expression) shall be determined by the operands involved in the expression and the context in which the expression is given.

A self-determined expression is one where the bit length of the expression is solely determined by the expression itself for example, an expression representing a delay value.

A context-determined expression is one where the bit length of the expression is determined by the bit length of the expression and by the fact that it is part of another expression. For example, the bit size of the righthand side expression of an assignment depends on itself and the size of the left-hand side.

Table 29 shows how the form of an expression shall determine the bit lengths of the results of the expression. In Table 29, i, j, and k represent expressions of an operand, and L (i) represents the bit length of the operand represented by i.

Table 29—Bit lengths resulting from self-determined expressions

| Expression | Bit length | Comments |
| :---: | :---: | :---: |
| Unsized constant number* | Same as integer |  |
| Sized constant number | As given |  |
| i op $j$, where op is: $+-* / \% \& \mid \wedge \wedge \sim \sim \wedge$ | $\max (\mathrm{L}(\mathrm{i}), \mathrm{L}(\mathrm{j})$ ) |  |
| op $i$, where op is: + - ~ | L(i) |  |
| i op j , where op is: <br> === !== == != \&\& \|| >>=<<= | 1 bit | Operands are sized to max(L(i),L(j)) |
| op i , where op is: <br> \& ~\& \| ~| ^ ~^ ^~! | 1 bit | All operands are self-determined |
| i op j , where op is: $\ggg \ll *$ | L(i) | j is self-determined |
| i ? $\mathrm{j}: \mathrm{k}$ | $\max (\mathrm{L}(\mathrm{j}), \mathrm{L}(\mathrm{k})$ ) | i is self-determined |
| \{i,..., j\} | $\mathrm{L}(\mathrm{i})+. .+\mathrm{L}(\mathrm{j})$ | All operands are self-determined |
| $\{\mathrm{i}\{\mathrm{j}, . . \mathrm{k}\}$ \} | i * (L(j)+.. $+\mathrm{L}(\mathrm{k})$ ) | All operands are self-determined |

*If an unsized constant is part of an expression that is longer than 32 bits, then if the most significant bit is unknown ( X or x ) or three-state ( Z or z ) the most significant bit is extended up to the size of the expression, otherwise signed constants are sign extended and unsigned constants are zero extended.

NOTE Multiplication without losing any overflow bits is still possible simply by assigning the result to something wide enough to hold it.

### 4.4.2 An example of an expression bit-length problem

During the evaluation of an expression, interim results shall take the size of the largest operand (in case of an assignment, this also includes the left-hand side). Care has to be taken to prevent loss of a significant bit during expression evaluation. The example below describes how the bit lengths of the operands could result in the loss of a significant bit.

Given the following declarations

```
reg [ 15:0] a, b, answer; // 16-bit regs
```

The intent is to evaluate the expression

```
answer = (a + b) >> 1; //will not work properly
```

where a and b are to be added, which may result in an overflow, and then shifted right by 1 bit to preserve the carry bit in the 16-bit answer.

A problem arises, however, because all operands in the expression are of a 16 -bit width. Therefore, the expression
$(\mathrm{a}+\mathrm{b})$ produces an interim result that is only 16 bits wide, thus losing the carry bit before the evaluation performs the 1 -bit right shift operation.

The solution is to force the expression $(\mathrm{a}+\mathrm{b})$ to evaluate using at least 17 bits. For example, adding an integer value of 0 to the expression will cause the evaluation to be performed using the bit size of integers. The following example will produce the intended result:

```
answer = (a + b + 0) >> 1; //will work correctly
```

In the following example:

```
module bitlength();
    reg [ 3:0] a,b,c;
    reg [ 4:0] d;
    initial begin
        a = 9;
        b = 8;
        c = 1;
        $display("answer = %b", c ? (a&b) : d);
        end
    endmodule
```

the $\$$ display statement will display:
answer $=01000$
By itself, the expression a\&b would have the bit length 4 , but since it is in the context of the conditional expression, which uses the maximum bit-length, the expression a\&b actually has length 5 , the length of d .

### 4.4.3 Example of self-determined expressions

```
reg [ 3:0] a;
reg [ 5:0] b;
reg [ 15:0] c;
initial begin
        \(a=4^{\prime} h F ;\)
        \(\mathrm{b}=6^{\prime} \mathrm{ha;}\)
        \$display ("a* \(\mathrm{b}=\frac{\circ \mathrm{x} \text { ", }}{}\)
                a*b); // expression size is self determined
        \(c=\left\{a^{* *} b\right\}\); // expression \(a^{* *} b\) is self determined
        \$display ("a**b=\%x", c); // due to \{\}
        c = a**b; // expression size is determined by c
        \$display (" \(\mathrm{C}=\% \mathrm{x}\) ", C );
    end
```

Simulator output for this example:

```
    a*b=16 // 96 was truncated since expression size
is 6
    a**b=1 // expression size is 4 bits (size of a)
    c=21 // example size is 16 bits (size of c)
```


### 4.5 Signed expressions

Controlling the sign of an expression is important if consistent results are to be achieved. In addition to the rules outlined in the following sections, two system functions shall be used to handle type casting on expressions: \$signed() and \$unsigned(). These functions shall evaluate the input expression and return a value with the same size and value of the input expression and the type defined by the function:
\$signed - returned value is signed
\$unsigned - returned value is unsigned
Example:

```
reg [ 7:0] regA;
reg signed [ 7:0] regS;
regA = $unsigned(-4); // regA = 4'b1100
regS = $signed(4'b1100); // regS = -4
```


### 4.5.1 ules for expression types

The following are the rules for determining the resulting type of an expression:
Expression type depends only on the operands. It does not depend on the LHS (if any).
Decimal numbers are signed.
Based_numbers are unsigned, except where the $s$ notation is used in the base specifier (as in "4'sd12").
Bit-select results are unsigned, regardless of the operands.
Part-select results are unsigned, regardless of the operands.
NOTE-This is true even if the part-select specifies the entire vector.

```
reg [ 15:0] a;
reg signed [ 7:0] b;
initial
\(\mathrm{a}=\mathrm{b}[7: 0] ; / / \mathrm{b}[7: 0]\) is unsigned and therefore zero-extended
```

Concatenate results are unsigned, regardless of the operands.
Comparison results $(1,0)$ are unsigned, regardless of the operands.
Reals converted to integers by type coercion are signed
The sign and size of any self-determined operand is determined by the operand itself and independent of the remainder of the expression.
For non-self-determined operands the following rules apply:
if any operand is real, the result is real;
if any operand is unsigned, the result is unsigned, regardless of the operator; if all operands are signed, the result will be signed, regardless of operator, except as noted.

### 4.5.2 Steps for evaluating an expression

Determine the expression size based upon the standard rules of expression size determination.
Determine the sign of the expression using the rules outlined in 4.5.1.
Coerce the type of each operand of the expression (excepting those which are self-determined) to the type of the expression.

Extend the size of each operand (excepting those which are self-determined) to the size of the expression. Perform sign extension if and only if the operand type (after type coercion) is signed.

### 4.5.3 Steps for evaluating an assignment

Determine the size of the RHS by the standard assignment size determination rules (see 4.4)
If needed, extend the size of the RHS, performing sign extension if and only if the type of the RHS is signed.

### 4.5.4 Handling $X$ and $Z$ in signed expressions

If a signed operand is to be resized to a larger signed width and the value of the sign bit is $X$, the resulting value shall be bit-filled with Xs. If the sign bit of the value is $Z$, then the resulting value shall be bit-filled with Zs . If any bit of a signed value is X or Z , then any non logical operation involving the value shall result in the entire resultant value being an $X$ and the type consistent with the expression's type.

## 5. Scheduling semantics

### 5.1 Execution of a model

The balance of the sections of this standard describe the behavior of each of the elements of the language. This section gives an overview of the interactions between these elements, especially with respect to the scheduling and execution of events.

The elements that make up the Verilog HDL can be used to describe the behavior, at varying levels of abstraction, of electronic hardware. An HDL has to be a parallel programming language. The execution of certain language constructs is defined by parallel execution of blocks or processes. It is important to understand what execution order is guaranteed to the user, and what execution order is indeterminate.

Although the Verilog HDL is used for more than simulation, the semantics of the language are defined for simulation, and everything else is abstracted from this base definition.

### 5.2 Event simulation

The Verilog HDL is defined in terms of a discrete event execution model. The discrete event simulation is described in more detail in this section to provide a context to describe the meaning and valid interpretation of Verilog HDL constructs. These resulting definitions provide the standard Verilog reference model for simulation, which all compliant simulators shall implement. Note, though, that there is a great deal of choice in the definitions that follow, and differences in some details of execution are to be expected between different simulators. In addition, Verilog HDL simulators are free to use different algorithms than those described in this section, provided the user-visible effect is consistent with the reference model.

A design consists of connected threads of execution or processes. Processes are objects that can be evaluated, that may have state, and that can respond to changes on their inputs to produce outputs. Processes include primitives, modules, initial and always procedural blocks, continuous assignments, asynchronous tasks, and procedural assignment statements.

Every change in value of a net or variable in the circuit being simulated, as well as the named event, is considered an update event.

Processes are sensitive to update events. When an update event is executed, all the processes that are sensitive to that event are evaluated in an arbitrary order. The evaluation of a process is also an event, known as an evaluation event.

In addition to events, another key aspect of a simulator is time. The term simulation time is used to refer to the time value maintained by the simulator to model the actual time it would take for the circuit being simulated. The term time is used interchangeably with simulation time in this section.

Events can occur at different times. In order to keep track of the events and to make sure they are processed in the correct order, the events are kept on an event queue, ordered by simulation time. Putting an event on the queue is called scheduling an event.

### 5.3 The stratified event queue

The Verilog event queue is logically segmented into five different regions. Events are added to any of the five regions but are only removed from the active region.

1) Events that occur at the current simulation time and can be processed in any order. These are the active events.
2) Events that occur at the current simulation time, but that shall be processed after all the active events are processed. These are the inactive events.
3) Events that have been evaluated during some previous simulation time, but that shall be assigned at this simulation time after all the active and inactive events are processed. These are the non blocking assign update events.
4) Events that shall be processed after all the active, inactive, and non blocking assign update events are processed. These are the monitor events.
5) Events that occur at some future simulation time. These are the future events. Future events are divided into future inactive events, and future non blocking assignment update events.

The processing of all the active events is called a simulation cycle.
The freedom to choose any active event for immediate processing is an essential source of nondeterminism in the Verilog HDL.

An explicit zero delay ( $\# 0$ ) requires that the process be suspended and added as an inactive event for the current time so that the process is resumed in the next simulation cycle in the current time.

A non blocking assignment (see 9.2.2) creates a non blocking assign update event, scheduled for current or a later simulation time.

The \$monitor and \$strobe system tasks (see 17.1) create monitor events for their arguments. These events are continuously re-enabled in every successive time step. The monitor events are unique in that they cannot create any other events.

The call back procedures scheduled with PLI routines such as $t f$ synchronize() (see 25.58) or vpi_register_cb (cb_readwrite) (see 27.33) shall be treated as inactive events.

### 5.4 The Verilog simulation reference model

In all the examples that follow, T refers to the current simulation time, and all events are held in the event queue, ordered by simulation time.

```
while (there are events) {
    if (no active events) {
        if (there are inactive events) {
                            activate all inactive events;
    } else if (there are non blocking assign update events) {
                activate all non blocking assign update events;
            } else if (there are monitor events) {
                activate all monitor events;
    } else {
                advance T to the next event time;
                activate all inactive events for time T;
    }
    }
    E = any active event;
    if (E is an update event) {
            update the modified object;
            add evaluation events for sensitive processes to event queue;
    } else {/* shall be an evaluation event */
            evaluate the process;
            add update events to the event queue;
    }
}
```


### 5.4.1 Determinism

This standard guarantees a certain scheduling order.

1) Statements within a begin-end block shall be executed in the order in which they appear in that begin-end block. Execution of statements in a particular begin-end block can be suspended in favor of other processes in the model; however, in no case shall the statements in a begin-end block be executed in any order other than that in which they appear in the source.
2) Non blocking assignments shall be performed in the order the statements were executed. Consider the following example:
```
initial begin
    a <= 0;
    a <= 1;
end
```

When this block is executed, there will be two events added to the non blocking assign update queue. The previous rule requires that they be entered on the queue in source order; this rule requires that they be taken from the queue and performed in source order as well. Hence, at the end of time step 1 , the variable a will be assigned 0 and then 1 .

### 5.4.2 Nondeterminism

One source of nondeterminism is the fact that active events can be taken off the queue and processed in any order. Another source of nondeterminism is that statements without time-control constructs in behavioral blocks do not have to be executed as one event. Time control statements are the \# expression and @ expression constructs (see 9.7). At any time while evaluating a behavioral statement, the simulator may suspend execution and place the partially completed event as a pending active event on the event queue. The effect of this is to allow the interleaving of process execution. Note that the order of interleaved execution is nondeterministic and not under control of the user.

## 5.5 ace conditions

Because the execution of expression evaluation and net update events may be intermingled, race conditions are possible:

```
\(\operatorname{assign} p=q\);
initial begin
    \(q\) = 1 ;
    \#1 q = 0;
    \$display (p) ;
end
```

The simulator is correct in displaying either a 1 or a 0 . The assignment of 0 to $q$ enables an update event for $p$. The simulator may either continue and execute the $\$$ display task or execute the update for $p$, followed by the \$display task.

### 5.6 Scheduling implication of assignments

Assignments are translated into processes and events as follows.

### 5.6.1 Continuous assignment

A continuous assignment statement (Section 6) corresponds to a process, sensitive to the source elements in the expression. When the value of the expression changes, it causes an active update event to be added to the event queue, using current values to determine the target.

### 5.6.2 rocedural continuous assignment

A procedural continuous assignment (which are the assign or force statement; see 9.3) corresponds to a process that is sensitive to the source elements in the expression. When the value of the expression changes, it causes an active update event to be added to the event queue, using current values to determine the target.

A deassign or a release statement deactivates any corresponding assign or force statement(s).

### 5.6.3 Blocking assignment

A blocking assignment statement (see 9.2.1) with a delay computes the right-hand side value using the current values, then causes the executing process to be suspended and scheduled as a future event. If the delay is 0 , the process is scheduled as an inactive event for the current time.

When the process is returned (or if it returns immediately if no delay is specified), the process performs the assignment to the left-hand side and enables any events based upon the update of the left-hand side. The values at the time the process resumes are used to determine the target(s). Execution may then continue with the

## 4 Non bq 6.3 Blocking assignment

qifnZqfi\}q A[\}2]Gqb\{28q\{\} A[\{z]\}qassi28ent A[\{z]\}qsta28ent A[\}2]Gqa28eee A2)Gqb\{28q\{lwayz\{7fod8lay A[\}2]Gqcomp\{28

### 5.6.6 ort connections

Ports connect processes through implicit continuous assignment statements or implicit bidirectional connections. Bidirectional connections are analogous to an always-enabled tran connection between the two nets, but without any strength reduction. Port connection rules require that a value receiver be a net or a structural net expression.

Ports can always be represented as declared objects connected as follows:
If an input port, then a continuous assignment from an outside expression to a local (input) net If an output port, then a continuous assignment from a local output expression to an outside net If an inout, then a nonstrength-reducing transistor connecting the local net to an outside net

### 5.6.7 Functions and tasks

Task and function parameter passing is by value, and it copies in on invocation and copies out on return. The copy out on the return function behaves in the same manner as does any blocking assignment.

## 6. Assignments

The assignment is the basic mechanism for placing values into nets and variables. There are two basic forms of assignments:

The continuous assignment, which assigns values to nets
The procedural assignment, which assigns values to variables
There are two additional forms of assignments, assign / deassign and force / release which are called procedural continuous assignments, described in 9.3.

An assignment consists of two parts, a left-hand side and a right-hand side, separated by the equals ( = ) character; or, in the case of non blocking procedural assignment, the less-than-equals ( $<=$ ) character pair. The right-hand side can be any expression that evaluates to a value. The left-hand side indicates the variable to which the right-hand side value is to be assigned. The left-hand side can take one of the forms given in Table 30, depending on whether the assignment is a continuous assignment or a procedural assignment.

Table 30-Legal left-hand side forms in assignment statements

| Statement type | Left-hand side (LHS) |
| :--- | :--- |
| Continuous assignment | Net (vector or scalar) <br> Constant bit select of a vector net <br> Constant part select of a vector net <br> Constant indexed part select of a vector net <br> Concatenation of any of the above four LHS |
| Procedural assignment | Variables (vector or scalar) <br> Bit-select of a vector reg, integer, or time variable <br> Constant part select of a vector reg, integer, or time <br> variable <br> Memory word <br> Indexed part select of a vector reg, integer, or time <br> variable <br> Concatenation of regs; bit or part selects of regs |

### 6.1 Continuous assignments

Continuous assignments shall drive values onto nets, both vector and scalar. This assignment shall occur whenever the value of the right-hand side changes. Continuous assignments provide a way to model combinational logic without specifying an interconnection of gates. Instead, the model specifies the logical expression that drives the net.

The syntax for continuous assignments is given in Syntax 6-1.

```
net_declaration ::= (From Annex A - A.2.1.3)
        net_type [ signed ]
                [ delay3 ] list_of_net_identifiers ;
    | net_type [ drive_strength ] [ signed ]
        [ delay3 ] list_of_net_decl_assignments ;
    | net_type [ vectored | scalared ] [ signed ]
        range [ delay3 ] list_of_net_identifiers ;
    | net_type [ drive_strength ] [ vectored | scalared ] [ signed ]
        range [ delay3 ] list_of_net_decl_assignments ;
    | trireg [ charge_strength ] [ signed ]
        [ delay3 ] list_of_net_identifiers ;
    | trireg [ drive_strength ] [ signed ]
        [ delay3 ] list_of_net_decl_assignments ;
    | trireg [ charge_strength ] [ vectored | scalared ] [ signed ]
        range [ delay3 ] list_of_net_identifiers ;
    | trireg [ drive_strength ] [ vectored | scalared ] [ signed ]
        range [ delay3 ] list_of_net_decl_assignments ;
list_of_net_decl_assignments ::= (From Annex A - A.2.3)
    net_decl_assignment {, net_decl_assignment }
net_decl_assignment ::= (From Annex A - A.2.4)
    net_identifier = expression
continuous_assign ::= (From Annex A - A.6.1)
    assign [ drive_strength ] [ delay3 ] list_of_net_assignments ;
list_of_net_assignments ::=
    net_assignment {, net_assignment }
net_assignment ::=
    net_lvalue = expression
```


## Syntax 6-1 Syntax for continuous assignment

### 6.1.1 The net declaration assignment

The first two alternatives in the net declaration are discussed in see 3.2. The third alternative, the net declaration assignment, allows a continuous assignment to be placed on a net in the same statement that declares the net.

## Example:

The following is an example of the net declaration form of a continuous assignment:

```
wire (strong1, pull0) mynet = enable ;
```

NOTE Because a net can be declared only once, only one net declaration assignment can be made for a particular net. This contrasts with the continuous assignment statement; one net can receive multiple assignments of the continuous assignment form.

### 6.1.2 The continuous assignment statement

The continuous assignment statement shall place a continuous assignment on a net data type. The net may be explicitly declared, or may inherit an implicit declaration in accordance with the implicit declarations rules defined in 3.5.

Assignments on nets shall be continuous and automatic. This means that whenever an operand in the righthand side expression changes value, the whole right-hand side shall be evaluated and if the new value is different from the previous value, then the new value shall be assigned to the left-hand side.

## Examples:

Example 1 The following is an example of a continuous assignment to a net that has been previously declared:

```
wire mynet ;
assign (strong1, pull0) mynet = enable ;
```

Example 2 The following is an example of the use of a continuous assignment to model a 4-bit adder with carry. The assignment could not be specified directly in the declaration of the nets because it requires a concatenation on the left-hand side.

```
module adder (sum_out, carry_out, carry_in, ina, inb);
output [ 3:0] sum out;
output carry_out;
input [ 3:0] ina, inb;
input carry_in;
wire carry_out, carry_in;
wire [ 3:0] sum out, ina, inb;
assign { carry_out, sum_out} = ina + inb + carry_in;
endmodule
```

Example 3 The following example describes a module with one 16-bit output bus. It selects between one of four input busses and connects the selected bus to the output bus.

```
module select bus(busout, bus0, bus1, bus2, bus3, enable, s);
parameter n = 16;
parameter Zee = 16' bz;
output [ 1:n] busout;
input [ 1:n] bus0, bus1, bus2, bus3;
input enable;
input [ 1:2] s;
tri [1:n] data; // net declaration
// net declaration with continuous assignment
tri [ 1:n] busout = enable ? data : Zee;
// assignment statement with four continuous assignments
assign
    data = (s == 0) ? bus0 : Zee,
    data = (s == 1) ? bus1 : Zee,
    data = (s == 2) ? bus2 : Zee,
    data = (s == 3) ? bus3 : Zee;
endmodule
```

The following sequence of events is experienced during simulation of this example:
a) The value of $s$, a bus selector input variable, is checked in the assign statement. Based on the value of $s$, the net data receives the data from one of the four input buses.
b) The setting of data net triggers the continuous assignment in the net declaration for busout. If enable is set, the contents of data are assigned to busout; if enable is 0 , the contents of Zee are assigned to busout.

### 6.1.3 Delays

A delay given to a continuous assignment shall specify the time duration between a right-hand side operand value change and the assignment made to the left-hand side. If the left-hand side references a scalar net, then the delay shall be treated in the same way as for gate delays that is, different delays can be given for the output rising, falling, and changing to high impedance (see 7.).

If the left-hand side references a vector net, then up to three delays can be applied. The following rules determine which delay controls the assignment:

If the right-hand side makes a transition from nonzero to zero, then the falling delay shall be used. If the right-hand side makes a transition to $z$, then the turn-off delay shall be used
For all other cases, the rising delay shall be used.
Specifying the delay in a continuous assignment that is part of the net declaration shall be treated differently from specifying a net delay and then making a continuous assignment to the net. A delay value can be applied to a net in a net declaration, as in the following example:

```
wire #10 wireA;
```

This syntax, called a net delay, means that any value change that is to be applied to wireA by some other statement shall be delayed for ten time units before it takes effect. When there is a continuous assignment in a declaration, the delay is part of the continuous assignment and is not a net delay. Thus, it shall not be added to the delay of other drivers on the net. Furthermore, if the assignment is to a vector net, then the rising and falling delays shall not be applied to the individual bits if the assignment is included in the declaration.

In situations where a right-hand side operand changes before a previous change has had time to propagate to the left-hand side, then the following steps are taken:
a) The value of the right-hand side expression is evaluated.
b) If this RHS value differs from the value currently scheduled to propagate to the left-hand side, then the currently scheduled propagation event is descheduled.
c) If the new RHS value equals the current left-hand side value, no event is scheduled.
d) If the new RHS value differs from the current LHS value, a delay is calculated in the standard way using the current value of the left-hand side, the newly calculated value of the right-hand side, and the delays indicated on the statement; a new propagation event is then scheduled to occur delay time units in the future.

### 6.1.4 Strength

The driving strength of a continuous assignment can be specified by the user. This applies only to assignments to scalar nets of the following types:

| wire | tri | trireg |
| :--- | :--- | :--- |
| wand | triand | tri0 |
| wor | trior | tri1 |

Continuous assignments driving strengths can be specified in either a net declaration or in a stand-alone assignment, using the assign keyword. The strength specification, if provided, shall immediately follow the keyword (either the keyword for the net type or assign) and precede any delay specified. Whenever the continuous assignment drives the net, the strength of the value shall be simulated as specified.

A drive strength specification shall contain one strength value that applies when the value being assigned to the net is 1 and a second strength value that applies when the assigned value is 0 . The following keywords shall specify the strength value for an assignment of 1 :
supply1
strong1
pull1
weak1
highz1

The following keywords shall specify the strength value for an assignment of 0 :
supply0 strong0 pull0 weak0 highz0

The order of the two strength specifications shall be arbitrary. The following two rules shall constrain the use of drive strength specifications:

The strength specifications ( highz1, highz0) and (highz0, highz1) shall be treated as illegal constructs.
If drive strength is not specified, it shall default to (strong1, strong0).

## 6.2 rocedural assignments

The primary discussion of procedural assignments is in 9.2. However, a description of the basic ideas in this clause highlights the differences between continuous assignments and procedural assignments.

As stated in 6.1, continuous assignments drive nets in a manner similar to the way gates drive nets. The expression on the right-hand side can be thought of as a combinatorial circuit that drives the net continuously. In contrast, procedural assignments put values in variables. The assignment does not have duration; instead, the variable holds the value of the assignment until the next procedural assignment to that variable.

Procedural assignments occur within procedures such as always, initial (see Clause 9), task, and function (see Clause 10) and can be thought of as triggered assignments. The trigger occurs when the flow of execution in the simulation reaches an assignment within a procedure. Reaching the assignment can be controlled by conditional statements. Event controls, delay controls, if statements, case statements, and looping statements can all be used to control whether assignments are evaluated. Clause 9 gives details and examples.

### 6.2.1 Variable declaration assignment

The variable declaration assignment is a special case of procedural assignment as it assigns a value to a variable. It allows an initial value to be placed in a variable in the same statement that declares the variable. The assignment shall be to a constant expression. The assignment does not have duration; instead, the variable holds the value until the next assignment to that variable. Variable declaration assignments to an array are not allowed. Variable declaration assignments are only allowed at the module level.

## Examples.

Example 1 Declare a 4 bit reg and assign it the value 4.

$$
\operatorname{reg}[3: 0] a=4 \text { 'h4; }
$$

This is equivalent to writing:

```
reg[ 3:0] a;
initial a = 4'h4;
```

Example 2 The following example is not legal.
reg [ 3:0] array [ 3:0] $=0$;

Example 3 Declare two integers, the first is assigned the value of 0 .
integer $i=0, j ;$
Example 4 Declare two real variables, assigned to the values 2.5 and 300,000.
real $r 1=2.5, \mathrm{n} 300 \mathrm{k}=3 \mathrm{E} 6$;

Example 5 Declare a time variable and realtime variable with initial values.
time t1 = 25;
realtime rt1 $=2.5$;

NOTE-If the same variable is assigned different values both in an initial block and in a variable declaration assignment, the order of the evaluation is undefined.

### 6.2.2 Variable declaration syntax

The syntax for variable declaration assignments is given in Syntax 6-2.

```
integer_declaration ::= (From Annex A - A.2.1.3)
    integer list_of_variable_identifiers ;
real_declaration ::=
    real list_of_real_identifiers ;
realtime_declaration ::=
    realtime list_of_real_identifiers ;
reg_declaration ::=
    reg [ signed ] [ range ] list_of_variable_identifiers ;
time_declaration ::=
    time list_of_variable_identifiers ;
real_type ::= (From Annex A - A.2.2.1)
            real_identifier [ = constant_expression ]
        | real_identifier dimension {dimension }
variable_type ::=
        variable_identifier [ = constant_expression ]
        | variable_identifier dimension { dimension }
list_of_real_identifiers ::= (From Annex A - A.2.3)
        real_type {, real_type }
list_of_variable_identifiers ::=
        variable_type {, variable_type }
```

Syntax 6-2 Syntax for reg declaration assignment

## 7. Gate and switch level modeling

This clause describes the syntax and semantics of these built-in primitives and how a hardware design can be described using these primitives.

There are 14 logic gates and 12 switches predefined in the Verilog HDL to provide the gate and switch level modeling facility. Modeling with logic gates and switches has the following advantages:

Gates provide a much closer one-to-one mapping between the actual circuit and the model.
There is no continuous assignment equivalent to the bidirectional transfer gate.

### 7.1 Gate and switch declaration syntax

Syntax 7-1 shows the gate and switch declaration syntax.
A gate or a switch instance declaration shall have the following specifications:
The keyword that names the type of gate or switch primitive
An optional drive strength
An optional propagation delay
An optional identifier that names each gate or switch instance
An optional range for array of instances
The terminal connection list

Multiple instances of the one type of gate or switch primitive can be declared as a comma-separated list. All such instances shall have the same drive strength and delay specification.

```
gate_instantiation ::= (From Annex A - A.3.1)
    cmos_switchtype [delay3] cmos_switch_instance {, cmos_switch_instance } ;
    enable_gatetype [drive_strength] [delay3] enable_gate_instance {, enable_gate_instance };
    mos_switchtype [delay3] mos_switch_instance {, mos_switch_instance };
    n_input_gatetype [drive_strength] [delay2] n_input_gate_instance {, n_input_gate_instance };
    |_output_gatetype [drive_strength] [delay2] n_output_gate_instance
        {, n_output_gate_instance };
    pass_en_switchtype [delay2] pass_enable_switch_instance {, pass_enable_switch_instance };
    pass_switchtype pass_switch_instance {, pass_switch_instance };
    pulldown [pulldown_strength] pull_gate_instance {, pull_gate_instance };
    pullup [pullup_strength] pull_gate_instance { , pull_gate_instance } ;
cmos_switch_instance ::= [ name_of_gate_instance ]
    ( output_terminal, input_terminal, ncontrol_terminal , pcontrol_terminal )
enable_gate_instance ::= [ name_of_gate_instance ]
    ( output_terminal , input_terminal, enable_terminal )
mos_switch_instance ::= [ name_of_gate_instance ]
    ( output_terminal , input_terminal , enable_terminal )
n_input_gate_instance ::= [ name_of_gate_instance ]
    ( output_terminal, input_terminal {, input_terminal })
n_output_gate_instance ::= [ name_of_gate_instance ]
    (output_terminal {, output_terminal }, input_terminal )
pass_switch_instance ::= [ name_of_gate_instance ] ( inout_terminal , inout_terminal )
pass_enable_switch_instance ::= [ name_of_gate_instance ]
    ( inout_terminal, inout_terminal, enable_terminal )
pull_gate_instance ::= [ name_of_gate_instance ] ( output_terminal )
name_of_gate_instance ::= gate_instance_identifier [range ]
pulldown_strength ::= (From Annex A-A.3.2)
    ( strength0, strength1 )
    ( strength1, strength0 )
    ( strength0)
pullup_strength ::= ( strength0, strength1 )
    |(strength1, strength0 )
    ( strength1)
enable_terminal ::= (From Annex A - A.3.3)
    expression
inout_terminal ::= net_lvalue
input_terminal ::= expression
ncontrol_terminal ::= expression
output_terminal ::= net_lvalue
pcontrol_terminal ::= expression
cmos_switchtype ::= (From Annex A - A.3.4)
    cmos remos
enable_gatetype ::= bufif0 | bufif1 | notif0 | notif1
mos_switchtype ::= nmos | pmos | rnmos | rpmos
n_input_gatetype ::= and | nand | or | nor | xor | xnor
n_output_gatetype ::= buf | not
pass_en_switchtype ::= tranif0 | tranif1 | rtranif1 | rtranif0
pass_switchtype ::= tran | rtran
```


### 7.1.1 The gate type specification

A gate or switch instance declaration shall begin with the keyword that specifies the gate or switch primitive being used by the instances that follow in the declaration. Table 31 lists the keywords that shall begin a gate or a switch instance declaration.

Table 31-Built-in gates and switches

| n_input gates | n_output gates | three-state <br> gates | pull gates | MOS switches | bidirectional <br> switches |
| :--- | :--- | :--- | :--- | :--- | :--- |
| and | buf | bufif0 | pulldown | cmos | rtran |
| nand |  | bufif1 | pullup | nmos | rtranif0 |
| nor |  | notif1 |  | pmos | rtranif1 |
| or |  |  | rnmos | tran |  |
| xnor |  |  | rpmos | tranif0 |  |
| xor |  |  | tranif1 |  |  |

Explanations of the built-in gates and switches shown in Table 31 begin in 7.2.

### 7.1.2 The drive strength specification

An optional drive strength specification shall specify the strength of the logic values on the output terminals of the gate instance. Only the instances of the gate primitives shown in Table 32 can have the drive strength specification.

Table 32-Valid gate types for strength specifications

| and | nand | buf | not | pulldown |
| :--- | :--- | :--- | :--- | :--- |
| or | nor | bufif0 | notif0 | pullup |
| xor | xnor | bufif1 | notif1 |  |

The drive strength specification for a gate instance, with the exception of pullup and pulldown, shall have a strength 1 specification and a strength 0 specification. The strength1 specification shall specify the strength of signals with a logic value 1 , and the strength 0 specification shall specify the strength of signals with a logic value 0 . The strength specification shall follow the gate type keyword and precede any delay specification. The strength0 specification can precede or follow the strength1 specification. The strength1 and strength0 specifications shall be separated by a comma and enclosed within a pair of parentheses.

The pullup gate can have only strengthl specification; strength 0 specification shall be optional. The pulldown gate can have only strength0 specification; strength 1 specification shall be optional.

The strength1 specification shall be one of the following keywords:

```
supply1 strong1 pull1 weak1
```

The strength0 specification shall be one of the following keywords:

$$
\text { supply0 } \quad \text { strong0 } \quad \text { pull0 } \quad \text { weak0 }
$$

Specifying highz1 as strength1 shall cause the gate or switch to output a logic value $z$ in place of a 1 . Specifying highz0 shall cause the gate to output a logic value $z$ in place of a 0 . The strength specifications
(highz0, highz1) and (highz1, highz0) shall be considered invalid.
In the absence of a strength specification, the instances shall have the default strengths strong1 and strong0.

## Example:

The following example shows a drive strength specification in a declaration of an open collector nor gate:

```
nor (highz1,strong0) n1(out1,in1,in2);
```

In this example, the nor gate outputs a $z$ in place of a 1.
Logic strength modeling is discussed in more detail in 7.9 through 7.13.

### 7.1.3 The delay specification

An optional delay specification shall specify the propagation delay through the gates and switches in a declaration. Gates and switches in declarations with no delay specification shall have no propagation delay. A delay specification can contain up to three delay values, depending on the gate type. The pullup and pulldown instance declarations shall not include delay specifications. Delays are discussed in more detail in 7.14.

### 7.1.4 The primitive instance identifier

An optional name can be given to a gate or switch instance. If multiple instances are declared as an array of instances, an identifier shall be used to name the instances.

### 7.1.5 The range specification

There are many situations when repetitive instances are required. These instances shall differ from each other only by the index of the vector to which they are connected.

In order to specify an array of instances, the instance name shall be followed by the range specification. The range shall be specified by two constant expressions, left-hand index (lhi) and right-hand index (rhi), separated by a colon and enclosed within a pair of square brackets. A [ lhi : rhi] range specification shall represent an array of abs (lhi-rhi) +1 instances. Neither of the two constant expressions are required to be zero, and lhi is not required to be larger than rhi. If both constant expressions are equal, only one instance shall be generated.

An array of instances shall have a continuous range. One instance identifier shall be associated with only one range to declare an array of instances.

The range specification shall be optional. If no range specification is given, a single instance shall be created.

## Example:

A declaration shown below is illegal:

```
nand #2 t_nand[ 0:3] ( ... ), t_nand[ 4:7] ( ... );
```

It could be declared correctly as one array of eight instances, or two arrays with unique names of four elements each:

```
nand #2 t nand[ 0:7] ( ... );
nand #2 x_nand[ 0:3] ( ... ), y_nand[ 4:7] ( ... );
```


### 7.1.6 rimitive instance connection list

The terminal list describes how the gate or switch connects to the rest of the model. The gate or switch type can limit these expressions. The connection list shall be enclosed in a pair of parentheses, and the terminals shall be separated by commas. The output or bidirectional terminals shall always come first in the terminal list, followed by the input terminals.

The terminal connections for an array of instances shall follow these rules:
The bit length of each port expression in the declared instance-array shall be compared with the bit length of each single-instance port or terminal in the instantiated module or primitive.
For each port or terminal where the bit length of the instance-array port expression is the same as the bit length of the single-instance port, the instance-array port expression shall be connected to each single-instance port.
If bit lengths are different, each instance shall get a part-select of the port expression as specified in the range, starting with the right-hand index.
Too many or too few bits to connect to all the instances shall be considered an error.
An individual instance from an array of instances shall be referenced in the same manner as referencing an element of an array of regs.

## Examples.

Example 1 The following declaration of nand_array declares four instances that can be referenced by nand_array[ 1], nand_array[ 2], nand_array[ 3], and nand_array[ 4] respectively.

```
nand #2 nand_array[ 1:4] ( ... ) ;
```

Example 2 The two module descriptions that follow are equivalent except for indexed instance names, and they demonstrate the range specification and connection rules for declaring an array of instances:

```
module driver (in, out, en);
input [ 3:0] in;
output [ 3:0] out;
input en;
bufif0 ar[ 3:0] (out, in, en); // array of three-state buffers
endmodule
module driver equiv (in, out, en);
input [ 3:0] in;
output [ 3:0] out;
input en;
bufif0 ar3 (out[ 3], in[ 3], en); // each buffer declared separately
bufif0 ar2 (out[ 2], in[ 2], en);
bufif0 ar1 (out[ 1], in[ 1], en);
bufif0 ar0 (out[ 0], in[ 0], en);
endmodule
```

Example 3 The two module descriptions that follow are equivalent except for indexed instance names, and they demonstrate how different instances within an array of instances are connected when the port sizes do not match.

```
module busdriver (busin, bushigh, buslow, enh, enl);
input [ 15:0] in;
output [ 7:0] bushigh, buslow;
input enh, enl;
driver busar3 (busin[ 15:12], bushigh[ 7:4], enh);
driver busar2 (busin[ 11:8], bushigh[ 3:0] , enh);
driver busar1 (busin[ 7:4], buslow[ 7:4], enl);
driver busar0 (busin[ 3:0], buslow[ 3:0], enl);
endmodule
module busdriver_equiv (busin, bushigh, buslow, enh, enl);
input [ 15:0] busin;
output [ 7:0] bushigh, buslow;
input enh, enl;
driver busar[ 3:0] (.out({ bushigh, buslow}), .in(busin),
    .en({ enh, enh, enl, enl}));
endmodule
```

Example 4 This example demonstrates how a series of modules can be chained together. Figure 4 shows an equivalent schematic interconnection of DFF instances.

```
module dffn (q, d, clk);
parameter bits = 1;
input [bits-1:0] d;
output [bits-1:0] q;
input clk ;
DFF dff[ bits-1:0] (q, d, clk); // create a row of D flip-flops
endmodule
module MxN_pipeline (in, out, clk);
parameter M = 3, N = 4; // M=width,N=depth
input [ M-1:0] in;
output [ M-1:0] out;
input clk;
wire [ M* (N-1):1] t;
// #(M) redefines the bits parameter for dffn
// create p[ 1:N] columns of dffn rows (pipeline)
dffn #(M) p[ 1:N] ({ out, t}, {t, in}, clk);
endmodule
```



Figure 4-Schematic diagram of interconnections in array of instances

## 7.2 and, nand, nor, or, xor, and xnor gates

The instance declaration of a multiple input logic gate shall begin with one of the following keywords:
and nand nor or xor xnor

The delay specification shall be zero, one, or two delays. If the specification contains two delays, the first delay shall determine the output rise delay, the second delay shall determine the output fall delay, and the smaller of the two delays shall apply to output transitions to $x$. If only one delay is specified, it shall specify both the rise delay and the fall delay. If there is no delay specification, there shall be no propagation delay through the gate.

These six logic gates shall have one output and one or more inputs. The first terminal in the terminal list shall connect to the output of the gate and all other terminals connect to its inputs.

The truth tables for these gates, showing the result of two input values, appear in Table 33.

Table 33-Truth tables for multiple input logic gates


| xnor | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{x}$ | $\mathbf{z}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\mathbf{0}$ | 1 | 0 | x | x |
| $\mathbf{1}$ | 0 | 1 | x | x |
| $\mathbf{x}$ | x | x | x | x |
| $\mathbf{z}$ | x | x | x | x |

Versions of these six logic gates having more than two inputs shall have a natural extension, but the number of inputs shall not alter propagation delays.

## Example:

The following example declares a two input and gate:

```
and a1 (out, in1, in2);
```

The inputs are in1 and in2. The output is out. The instance name is a1.

## 7.3 buf and not gates

The instance declaration of a multiple output logic gate shall begin with one of the following keywords:

```
buf not
```

The delay specification shall be zero, one, or two delays. If the specification contains two delays, the first delay shall determine the output rise delay, the second delay shall determine the output fall delay, and the smaller of the two delays shall apply to output transitions to x . If only one delay is specified, it shall specify both the rise delay and the fall delay. If there is no delay specification, there shall be no propagation delay through the gate.

These two logic gates shall have one input and one or more outputs. The last terminal in the terminal list shall connect to the input of the logic gate, and the other terminals shall connect to the outputs of the logic gate.

Truth tables for these logic gates with one input and one output are shown in Table 34.

Table 34-Truth tables for multiple output logic gates

| buf |  |
| :---: | :---: |
| input | output |
| $\mathbf{0}$ | 0 |
| $\mathbf{1}$ | 1 |
| $\mathbf{x}$ | $\mathbf{x}$ |
| $\mathbf{z}$ |  |


| not |  |
| :---: | :---: |
| input | output |
| 0 | 1 |
| 1 | 0 |
| $\mathbf{x}$ | $\mathbf{x}$ |

## Example:

The following example declares a two output buf:

```
buf b1 (out1, out2, in);
```

The input is in. The outputs are out1 and out2. The instance name is b1.

## 7.4 bufif1, bufif0, notif1, and notif0 gates

The instance declaration of these three-state logic gates shall begin with one of the following keywords:
bufif0 bufif1 notif1 notif0

These four logic gates model three-state drivers. In addition to logic values 1 and 0 , these gates can output z.

The delay specification shall be zero, one, two, or three delays. If the delay specification contains three delays, the first delay shall determine the rise delay, the second delay shall determine the fall delay, the third delay shall determine the delay of transitions to $z$, and the smallest of the three delays shall determine the delay of transitions to $x$. If the specification contains two delays, the first delay shall determine the output rise delay, the second delay shall determine the output fall delay, and the smaller of the two delays shall apply to output transitions to $x$ and $z$. If only one delay is specified, it shall specify the delay for all output transitions. If there is no delay specification, there shall be no propagation delay through the gate.

Some combinations of data input values and control input values can cause these gates to output either of two values, without a preference for either value (see 7.10.2). These logic tables for these gates include two symbols representing such unknown results. The symbol L shall represent a result that has a value 0 or z . The symbol H shall represent a result that has a value 1 or z . Delays on transitions to H or L shall be treated the same as delays on transitions to x .

These four logic gates shall have one output, one data input, and one control input. The first terminal in the terminal list shall connect to the output, the second terminal shall connect to the data input, and the third terminal shall connect to the control input.

Table 35 presents the logic tables for these gates.

Table 35-Truth tables for three-state logic gates


| bufif <br> $\mathbf{1}$ | CONTROL |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
|  |  | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{x}$ | $\mathbf{z}$ |
|  | 0 | z | 0 | L | L |
|  | T | 1 | z | 1 | H |
| T | H |  |  |  |  |
|  | x | z | x | x | x |
|  | z | z | x | x | x |



## Example:

The following example declares an instance of bufif1:

```
bufif1 bf1 (outw, inw, controlw);
```

The output is outw, the input is inw, and the control is controlw. The instance name is $b f 1$.

### 7.5 M S switches

The instance declaration of a MOS switch shall begin with one of the following keywords:
cmos nmos pmos remos rnmos rpmos

The cmos and remos switches are described in 7.7.

The pmos keyword stands for the P-type metal-oxide semiconductor (PMOS) transistor and the nmos keyword stands for the N-type metal-oxide semiconductor (NMOS) transistor. PMOS and NMOS transistors have relatively low impedance between their sources and drains when they conduct. The rpmos keyword stands for resistive PMOS transistor and the rnmos keyword stands for resistive NMOS transistor. Resistive

PMOS and resistive NMOS transistors have significantly higher impedance between their sources and drains when they conduct than PMOS and NMOS transistors have. The load devices in static MOS networks are examples of rpmos and rnmos transistors. These four switches are unidirectional channels for data similar to the bufif gates.

The delay specification shall be zero, one, two, or three delays. If the delay specification contains three delays, the first delay shall determine the rise delay, the second delay shall determine the fall delay, the third delay shall determine the delay of transitions to $z$, and the smallest of the three delays shall determine the delay of transitions to $x$. If the specification contains two delays, the first delay shall determine the output rise delay, the second delay shall determine the output fall delay, and the smaller of the two delays shall apply to output transitions to x and z . If only one delay is specified, it shall specify the delay for all output transitions. If there is no delay specification, there shall be no propagation delay through the switch.

Some combinations of data input values and control input values can cause these switches to output either of two values, without a preference for either value. The logic tables for these switches include two symbols representing such unknown results. The symbol L represents a result that has a value 0 or z . The symbol H represents a result that has a value 1 or $z$. Delays on transitions to $H$ and $L$ shall be the same as delays on transitions to x .

These four switches shall have one output, one data input, and one control input. The first terminal in the terminal list shall connect to the output, the second terminal shall connect to the data input, and the third terminal shall connect to the control input.

The nmos and pmos switches shall pass signals from their inputs and through their outputs with a change in the strength of the signal in only one case, as discussed in 7.11. The rnmos and rpmos switches shall reduce the strength of signals that propagate through them, as discussed in 7.12.

Table 36 presents the logic tables for these switches.

Table 36-Truth tables for MOS switches


## Example:

The following example declares a pmos switch:

```
pmos p1 (out, data, control);
```

The output is out, the data input is data, and the control input is control. The instance name is p 1 .

### 7.6 Bidirectional pass switches

The instance declaration of a bidirectional pass switch shall begin with one of the following keywords:

| tran | tranif1 | tranif0 |
| :--- | :--- | :--- |
| rtran | rtranif1 | rtranif0 |

The bidirectional pass switches shall not delay signals propagating through them. When tranif0, tranif1, rtranif0, or rtranif1 devices are turned off they shall block signals, and when they are turned on they shall pass signals. The tran and rtran devices cannot be turned off, and they shall always pass signals.

The delay specifications for tranif1, tranif0, rtranif1, and rtranif0 devices shall be zero, one, or two delays. If the specification contains two delays, the first delay shall determine the turn-on delay, and the second delay shall determine the turn-off delay, and the smaller of the two delays shall apply to output transitions to x and z . If only one delay is specified, it shall specify both the turn-on and the turn-off delays. If there is no delay specification, there shall be no turn-on or turn-off delay for the bidirectional pass switch.

The bidirectional pass switches tran and rtran shall not accept delay specification.
The tranif1, tranif0, rtranif1, and rtranif0 devices shall have three items in their terminal lists. The first two shall be bidirectional terminals that conduct signals to and from the devices, and the third terminal shall connect to a control input. The tran and rtran devices shall have terminal lists containing two bidirectional terminals. Both bidirectional terminals shall unconditionally conduct signals to and from the devices, allowing signals to pass in either direction through the devices. The bidirectional terminals of all six devices shall be connected only to scalar nets or bit-selects of vector nets.

The tran, tranif0, and tranif1 devices shall pass signals with an alteration in their strength in only one case, as discussed in 7.11. The rtran, rtranif0, and rtranif1 devices shall reduce the strength of the signals passing through them according to rules discussed in 7.12.

## Example:

The following example declares an instance of tranif1:

```
tranif1 t1 (inout1,inout2,control);
```

The bidirectional terminals are inout1 and inout2. The control input is control. The instance name is t1.

### 7.7 CM S switches

The instance declaration of a CMOS switch shall begin with one of the following keywords:

## cmos remos

The delay specification shall be zero, one, two, or three delays. If the delay specification contains three delays, the first delay shall determine the rise delay, the second delay shall determine the fall delay, the third delay shall determine the delay of transitions to $z$, and the smallest of the three delays shall determine the delay of transitions to x . Delays in transitions to H or L are the same as delays in transitions to x . If the specification contains two delays, the first delay shall determine the output rise delay, the second delay shall determine the output fall delay, and the smaller of the two delays shall apply to output transitions to x and z . If only one delay is specified, it shall specify the delay for all output transitions. If there is no delay specification, there shall be no propagation delay through the switch.

The cmos and remos switches shall have a data input, a data output, and two control inputs. In the terminal list, the first terminal shall connect to the data output, the second terminal shall connect to the data input, the
third terminal shall connect to the $n$-channel control input, and the last terminal shall connect to the p-channel control input.

The cmos gate shall pass signals with an alteration in their strength in only one case, as discussed in 7.11. The remos gate shall reduce the strength of signals passing through it according to rules described in 7.12.

The cmos switch shall be treated as the combination of a pmos switch and an nmos switch. The remos switch shall be treated as the combination of an rpmos switch and an rnmos switch. The combined switches in these configurations shall share data input and data output terminals, but they shall have separate control inputs.

## Example:

The equivalence of the cmos gate to the pairing of an nos gate and a pmos gate is shown in the following example:


## 7.8 pullup and pulldown sources

The instance declaration of a pullup or a pulldown source shall begin with one of the following keywords:

## pullup pulldown

A pullup source shall place a logic value 1 on the nets connected in its terminal list. A pulldown source shall place a logic value 0 on the nets connected in its terminal list. The signals that these sources place on nets shall have pull strength in the absence of a strength specification. If conflicting strength specification is declared, it shall be ignored. There shall be no delay specifications for these sources.

## Example:

The following example declares two pullup instances:

```
pullup (strong1) p1 (neta), p2 (netb);
```

In this example, the p 1 instance drives neta and the p 2 instance drives netb.

### 7.9 Logic strength modeling

The Verilog HDL provides for accurate modeling of signal contention, bidirectional pass gates, resistive MOS devices, dynamic MOS, charge sharing, and other technology-dependent network configurations by allowing scalar net signal values to have a full range of unknown values and different levels of strength or combinations of levels of strength. This multiple-level logic strength modeling resolves combinations of signals into known or unknown values to represent the behavior of hardware with improved accuracy.

A strength specification shall have two components
a) The strength of the 0 portion of the net value, called strength0, designated as one of the following:
supply0
strong0
pullo
weak0
highz0
b) The strength of the 1 portion of the net value, called strength1, designated as one of the following:

```
supply1 strong1 pull1 weak1 highz1
```

The combinations (highz0, highz1) and (highz1, highz0) shall be considered illegal.

Despite this division of the strength specification, it is helpful to consider strength as a property occupying regions of a continuum in order to predict the results of combinations of signals.

Table 37 demonstrates the continuum of strengths. The left column lists the keywords used in specifying strengths. The right column gives correlated strength levels.

Table 37-Strength levels for scalar net signal values

| Strength name | Strength level |
| :--- | :--- |
| supply0 | 7 |
| strong0 | 6 |
| pull0 | 5 |
| large0 | 4 |
| weak0 | 3 |
| medium0 | 2 |
| small0 | 1 |
| highz0 | 0 |
| highz1 | 0 |
| small1 | 1 |
| medium1 | 2 |
| weak1 | 3 |
| large1 | 4 |
| pull1 | 5 |
| strong1 | 6 |
| supply1 | 7 |
|  |  |

In Table 37, there are four driving strengths:
supply strong pull weak
Signals with driving strengths shall propagate from gate outputs and continuous assignment outputs.
In Table 37, there are three charge storage strengths:

```
large medium small
```

Signals with the charge storage strengths shall originate in the trireg net type.
It is possible to think of the strengths of signals in the preceding table as locations on the scale in Figure 5.

| strength0 |  |  |  |  |  |  |  | strength1 |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | 0 | 1 | 2 | 3 | 4 | 5 | 6 | 7 |
| Su0 | St0 | PuO | La0 | We0 | MeO | Sm0 | Hizo | HiZ1 | Sm1 | Me1 | We1 | La1 | Pu1 | St1 | Su1 |

Figure 5-Scale of strengths

Discussions of signal combinations later in this section employs graphics similar to those used in Figure 5.

If the signal value of a net is known, all of its strength levels shall be in either the strength 0 part of the scale represented by Figure 5, or all strength levels shall be in its strength1 part. If the signal value of a net is unknown, it shall have strength levels in both the strength0 and the strength1 parts. A net with a signal value $z$ shall have a strength level only in one of the 0 subdivisions of the parts of the scale.

### 7.10 Strengths and values of combined signals

In addition to a signal value, a net shall have either a single unambiguous strength level or an ambiguous strength consisting of more than one level. When signals combine, their strengths and values shall determine the strength and value of the resulting signal in accordance with the principles in 7.10.1 through 7.10.4.

### 7.10.1 Combined signals of unambiguous strength

This subclause deals with combinations of signals in which each signal has a known value and a single strength level.

If two or more signals of unequal strength combine in a wired net configuration, the stronger signal shall dominate all the weaker drivers and determine the result. The combination of two or more signals of like value shall result in the same value with the greater of all the strengths. The combination of signals identical in strength and value shall result in the same signal.

The combination of signals with unlike values and the same strength can have three possible results. Two of the results occur in the presence of wired logic and the third occurs in its absence. Wired logic is discussed in 7.10.4. The result in the absence of wired logic is the subject of Figure 7.

Example:


Figure 6-Combining unequal strengths
In Figure 6, the numbers in parentheses indicate the relative strengths of the signals. The combination of a pull 1 and a strong 0 results in a strong 0 , which is the stronger of the two signals.

### 7.10.2 Ambiguous strengths: sources and combinations

There are several classifications of signals possessing ambiguous strengths
Signals with known values and multiple strength levels
Signals with a value $x$, which have strength levels consisting of subdivisions of both the strength1 and the strength 0 parts of the scale of strengths in Figure 5
Signals with a value $L$, which have strength levels that consist of high impedance joined with strength levels in the strength0 part of the scale of strengths in Figure 5
Signals with a value $H$, which have strength levels that consist of high impedance joined with strength levels in the strength1 part of the scale of strengths in Figure 5

Many configurations can produce signals of ambiguous strength. When two signals of equal strength and opposite value combine, the result shall be a value x , along with the strength levels of both signals and all the smaller strength levels.

## Examples:

Figure 7 shows the combination of a weak signal with a value 1 and a weak signal with a value 0 yielding a signal with weak strength and a value x .


Figure 7-Combination of signals of equal strength and opposite values
This output signal is described in Figure 8.


Figure 8-Weak x signal strength
An ambiguous signal strength can be a range of possible values. An example is the strength of the output from the three-state drivers with unknown control inputs as shown in Figure 9.


Figure 9-Bufifs with control inputs of $x$
The output of the bufif1 in Figure 9 is a strong H, composed of the range of values described in Figure 10.


Figure 10 -Strong H range of values
The output of the bufif0 in Figure 9 is a strong L, composed of the range of values described in Figure 11.


Figure 11-Strong L range of values
The combination of two signals of ambiguous strength shall result in a signal of ambiguous strength. The resulting signal shall have a range of strength levels that includes the strength levels in its component signals. The combination of outputs from two three-state drivers with unknown control inputs, shown in Figure 12, is an example.


Figure 12-Combined signals of ambiguous strength
In Figure 12, the combination of signals of ambiguous strengths produces a range that includes the extremes of the signals and all the strengths between them, as described in Figure 13.

| strength0 |  |  |  |  |  |  |  | strength1 |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | 0 | 1 | 2 | 3 | 4 | 5 | 6 | 7 |
| Su0 | St0 | Pu0 | La0 | We0 | Me0 | Sm0 | Hizo | Hiz1 | Sm1 | Me1 | We1 | La1 | Pu1 | St1 | Su1 |

Figure 13-Range of strengths for an unknown signal
The result is a value x because its range includes the values 1 and 0 . The number 35 , which precedes the x , is a concatenation of two digits. The first is the digit 3 , which corresponds to the highest strength 0 level for the result. The second digit, 5 , corresponds to the highest strength1 level for the result.

Switch networks can produce a ranges of strengths of the same value, such as the signals from the upper and lower configurations in Figure 14.


Figure 14-Ambiguous strengths from switch networks
In Figure 14, the upper combination of a reg, a gate controlled by a reg of unspecified value, and a pullup produces a signal with a value of 1 and a range of strengths (651) described in Figure 15.


Figure 15-Range of two strengths of a defined value
In Figure 14, the lower combination of a pulldown, a gate controlled by a reg of unspecified value, and an and gate produces a signal with a value 0 and a range of strengths (530) described in Figure 16.


Figure 16-Range of three strengths of a defined value

When the signals from the upper and lower configurations in Figure 14 combine, the result is an unknown with a range (56x) determined by the extremes of the two signals shown in Figure 17.


Figure 17-Unknown value with a range of strengths
In Figure 14, replacing the pulldown in the lower configuration with a supply0 would change the range of the result to the range (StX) described in Figure 18.

The range in Figure 18 is strong $x$, because it is unknown and the extremes of both its components are strong. The extreme of the output of the lower configuration is strong because the lower pmos reduces the strength of the supply0 signal. This modeling feature is discussed in 7.11.


Figure 18 - Strong $X$ range
Logic gates produce results with ambiguous strengths as well as three-state drivers. Such a case appears in Figure 19. The and gate N1 is declared with highz0 strength, and N2 is declared with weak0 strength.


Figure 19-Ambiguous strength from gates
In Figure 19, reg b has an unspecified value, so input to the upper and gate is strong $x$. The upper and gate has a strength specification including highz0. The signal from the upper and gate is a strong $H$ composed of the values as described in Figure 20.

| strength0 |  |  |  |  |  |  |  | strength1 |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | 0 | 1 | 2 | 3 | 4 | 5 | 6 | 7 |
| SuO | Sto | PuO | La0 | We0 | Me0 | Sm0 | Hizo | Hiz1 | Sm1 | Me1 | We1 | La1 | Pu1 | St1 | Su1 |

Figure 20-Ambiguous strength signal from a gate
HiZO is part of the result, because the strength specification for the gate in question specified that strength for an output with a value 0 . A strength specification other than high impedance for the 0 value output results in a gate output value x . The output of the lower and gate is a weak 0 as described in Figure 21.


Figure 21-Weak 0
When the signals combine, the result is the range (36x) as described in Figure 22.


Figure 22-Ambiguous strength in combined gate signals
Figure 22 presents the combination of an ambiguous signal and an unambiguous signal. Such combinations are the topic of 7.10.3.

### 7.10.3 Ambiguous strength signals and unambiguous signals

The combination of a signal with unambiguous strength and known value with another signal of ambiguous strength presents several possible cases. To understand a set of rules governing this type of combination, it is necessary to consider the strength levels of the ambiguous strength signal separately from each other and relative to the unambiguous strength signal. When a signal of known value and unambiguous strength combines with a component of a signal of ambiguous strength, these shall be the effects:
a) The strength levels of the ambiguous strength signal that are greater than the strength level of the unambiguous signal shall remain in the result.
b) The strength levels of the ambiguous strength signal that are smaller than or equal to the strength level of the unambiguous signal shall disappear from the result, subject to rule c .
c) If the operation of rule $a$ and rule $b$ results in a gap in strength levels because the signals are of opposite value, the signals in the gap shall be part of the result.

The following figures show some applications of the rules.


Figure 23-Elimination of strength levels
In Figure 23, the strength levels in the ambiguous strength signal that are smaller than or equal to the strength level of the unambiguous strength signal disappear from the result, demonstrating rule $b$.


Figure 24-Result demonstrating a range and the elimination of strength levels of two values

In Figure 24, rules $a, b$, and $c$ apply. The strength levels of the ambiguous strength signal that are of opposite value and lesser strength than the unambiguous strength signal disappear from the result. The strength levels in the ambiguous strength signal that are less than the strength level of the unambiguous strength signal, and of the same value, disappear from the result. The strength level of the unambiguous strength signal and the greater extreme of the ambiguous strength signal define a range in the result.


Figure 25-Result demonstrating a range and the elimination of strength levels of one value

In Figure 25, rules $a$ and $b$ apply. The strength levels in the ambiguous strength signal that are less than the strength level of the unambiguous strength signal disappear from the result. The strength level of the unambiguous strength signal and the strength level at the greater extreme of the ambiguous strength signal define a range in the result.

Combining the two signals above results in the following signal:
Combining the two signals above results in the following signal:


Figure 26-A range of both values
In Figure 26, rules $a, b$, and $c$ apply. The greater extreme of the range of strengths for the ambiguous strength signal is larger than the strength level of the unambiguous strength signal. The result is a range defined by the greatest strength in the range of the ambiguous strength signal and by the strength level of the unambiguous strength signal.

### 7.10.4 Wired logic net types

The net types triand, wand, trior, and wor shall resolve conflicts when multiple drivers have the same strength. These net types shall resolve signal values by treating signals as inputs of logic functions.

## Examples:

Consider the combination of two signals of unambiguous strength in Figure 27.


Figure 27-Wired logic with unambiguous strength signals

The combination of the signals in Figure 27, using wired and logic, produces a result with the same value as the result produced by an and gate with the value of the two signals as its inputs. The combination of signals using wired or logic produces a result with the same value as the result produced by an or gate with the values of the two signals as its inputs. The strength of the result is the same as the strength of the combined signals in both cases. If the value of the upper signal changes so that both signals in Figure 27 possess a value 1 , then the results of both types of logic have a value 1 .

When ambiguous strength signals combine in wired logic, it is necessary to consider the results of all combinations of each of the strength levels in the first signal with each of the strength levels in the second signal, as shown in Figure 28.

Signal 1

Signal 2

| signal1 |  | signal2 |  | result |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| strength | value | strength | value | strength | value |
| 5 | 0 | 5 | 1 | 5 | 0 |
| 6 | 0 | 5 | 1 | 6 | 0 |

The result is the following signal:

| signal1 |  | signal2 |  | result |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| strength | value | strength | value | strength | value |
| 5 | 0 | 5 | 1 | 5 | 1 |
| 6 | 0 | 5 | 1 | 6 | 0 |

Figure 28-Wired logic and ambiguous strengths

### 7.11 Strength reduction by nonresistive devices

The nmos, pmos, and cmos switches shall pass the strength from the data input to the output, except that a supply strength shall be reduced to a strong strength.

The tran, tranif0, and tranif1 switches shall not affect signal strength across the bidirectional terminals, except that a supply strength shall be reduced to a strong strength.

### 7.12 Strength reduction by resistive devices

The rnmos, rpmos, remos, rtran, rtranif1, and rtranif0 devices shall reduce the strength of signals that pass through them according to Table 38.

Table 38-Strength reduction rules

| Input strength | Reduced strength |
| :--- | :--- |
| Supply drive | Pull drive |
| Strong drive | Pull drive |
| Pull drive | Weak drive |
| Large capacitor | Medium capacitor |
| Weak drive | Medium capacitor |
| Medium capacitor | Small capacitor |
| Small capacitor | Small capacitor |
| High impedance | High impedance |

### 7.13 Strengths of net types

The tri0, tri1, supply0, and supply1 net types shall generate signals with specific strength levels. The trireg declaration can specify either of two signal strength levels other than a default strength level.

### 7.13.1 tri0 and tri1 net strengths

The tri0 net type models a net connected to a resistive pulldown device. In the absence of an overriding source, such a signal shall have a value 0 and a pull strength. The tri1 net type models a net connected to a resistive pullup device. In the absence of an overriding source, such a signal shall have a value 1 and a pull strength.

### 7.13.2 trireg strength

The trireg net type models charge storage nodes. The strength of the drive resulting from a trireg net that is in the charge storage state (that is, a driver charged the net and then went to high impedance) shall be one of these three strengths: large, medium, or small. The specific strength associated with a particular trireg net shall be specified by the user in the net declaration. The default shall be medium. The syntax of this specification is described in 3.4.1.

### 7.13.3 supply0 and supply1 net strengths

The supply0 net type models ground connections. The supply1 net type models connections to power supplies. The supply0 and supply1 net types shall have supply driving strengths.

### 7.14 Gate and net delays

Gate and net delays provide a means of more accurately describing delays through a circuit. The gate delays specify the signal propagation delay from any gate input to the gate output. Up to three values per output representing rise, fall, and turn-off delays can be specified (see 7.2 through 7.8).

Net delays refer to the time it takes from any driver on the net changing value to the time when the net value is updated and propagated further. Up to three delay values per net can be specified.

For both gates and nets, the default delay shall be zero when no delay specification is given. When one delay value is given, then this value shall be used for all propagation delays associated with the gate or the net. When two delays are given, the first delay shall specify the rise delay and the second delay shall specify the fall delay. The delay when the signal changes to high impedance or to unknown shall be the lesser of the two delay values.

For a three-delay specification
The first delay refers to the transition to the 1 value (rise delay).
The second delay refers to the transition to the 0 value (fall delay).
The third delay refers to the transition to the high-impedance value.
When a value changes to the unknown (x) value, the delay is the smallest of the three delays. The strength of the input signal shall not affect the propagation delay from an input to an output.

Table 39 summarizes the from-to propagation delay choice for the two- and three-delay specifications.
Table 39-Rules for propagation delays

|  |  | Delay used if there are |  |
| :--- | :--- | :--- | :--- |
| From value: | To value: | 2 delays | 3 delays |
| 0 | 1 | d1 | d 1 |
| 0 | x | $\min (\mathrm{d} 1, \mathrm{~d} 2)$ | $\min (\mathrm{d} 1, \mathrm{~d} 2, \mathrm{~d} 3)$ |
| 0 | z | $\min (\mathrm{d} 1, \mathrm{~d} 2)$ | d 3 |
| 1 | 0 | d 2 | d 2 |
| 1 | x | $\min (\mathrm{d} 1, \mathrm{~d} 2)$ | $\min (\mathrm{d} 1, \mathrm{~d} 2, \mathrm{~d} 3)$ |
| 1 | 0 | $\min (\mathrm{~d} 1, \mathrm{~d} 2)$ | d 3 |
| x | z | d 2 | d 2 |
| x | z | d 1 | d 1 |
| x | z | $\min (\mathrm{d} 1, \mathrm{~d} 2)$ | d 3 |
| z | 0 | d 2 | d 2 |
| z | 1 | d 1 | d 1 |
| z | x | $\min (\mathrm{d} 1, \mathrm{~d} 2)$ | $\min (\mathrm{d} 1, \mathrm{~d} 2, \mathrm{~d} 3)$ |

## Examples:

Example 1 The following is an example of a delay specification with one, two, and three delays:

```
and #(10) a1 (out, in1, in2); // only one delay
and #(10,12) a2 (out, in1, in2); // rise and fall delays
bufif0 #(10,12,11) b3 (out, in, ctrl);// rise, fall, and turn-off
delays
```

Example 2 The following example specifies a simple latch module with three-state outputs, where individual delays are given to the gates. The propagation delay from the primary inputs to the outputs of the module will be cumulative, and it depends on the signal path through the network.

```
module tri_latch (qout, nqout, clock, data, enable);
output qout, nqout;
input clock, data, enable;
tri qout, nqout;
not #5 n1 (ndata, data);
nand #(3,5) n2 (wa, data, clock),
    n3 (wb, ndata, clock);
nand # (12,15) n4 (q, nq, wa),
    n5 (nq, q, wb);
bufif1 #(3,7,13) q_drive (qout, q, enable),
    nq_drive (nqout, nq, enable);
```

endmodule

### 7.14.1 min:typ:max delays

The syntax for delays on gate primitives (including user-defined primitives; see Clause 8), nets, and continuous assignments shall allow three values each for the rising, falling, and turn-off delays. The minimum, typical, and maximum values for each delay shall be specified as constant expressions separated by colons. There shall be no required relation (e.g., min $\leq t y p \leq \max$ ) between the expressions for minimum, typical, and maximum delays. These can be any three constant expressions.

## Examples:

The following example shows min: typ:max values for rising, falling, and turn-off delays:

```
module iobuf (io1, io2, dir);
bufif0 #(5:7:9, 8:10:12, 15:18:21) b1 (io1, io2, dir);
bufif1 #(6:8:10, 5:7:9, 13:17:19) b2 (io2, io1, dir);
    ...
endmodule
```

The syntax for delay controls in procedural statements (see 9.7) also allows minimum, typical, and maximum values. These are specified by expressions separated by colons. The following example illustrates this concept.

```
parameter min_hi = 97, typ_hi = 100, max_hi = 107;
reg clk;
always begin
    \#(95:100:105) clk = 1;
    \#(min_hi:typ_hi:max_hi) clk = 0;
end
```


### 7.14.2 trireg net charge decay

Like all nets, the delay specification in a trireg net declaration can contain up to three delays. The first two delays shall specify the delay for transition to the 1 and 0 logic states when the trireg net is driven to these states by a driver. The third delay shall specify the charge decay time instead of the delay in a transition to the $z$ logic state. The charge decay time specifies the delay between when the drivers of a trireg net turn off and when its stored charge can no longer be determined.

A trireg net does not need a turn-off delay specification because a trireg net never makes a transition to the $z$ logic state. When the drivers of a trireg net make transitions from the 1,0 , or $x$ logic states to off, the trireg net shall retain the previous 1,0 , or $x$ logic state that was on its drivers. The $z$ value shall not propagate from the drivers of a trireg net to a trireg net. A trireg net can only hold a $z$ logic state when $z$ is the initial logic state of the trireg net or when the trireg net is forced to the $z$ state with a force statement (see 9.3.2).

A delay specification for charge decay models a charge storage node that is not ideal a charge storage node whose charge leaks out through its surrounding devices and connections.

The following subclauses describe the charge decay process and the delay specification for charge decay.

### 7.14.2.1 The charge decay process

Charge decay is the cause of transition of a 1 or 0 that is stored in a trireg net to an unknown value $(x)$ after a specified delay. The charge decay process shall begin when the drivers of the trireg net turn off and the trireg net starts to hold charge. The charge decay process shall end under the following two conditions:
a) The delay specified by charge decay time elapses and the trireg net makes a transition from 1 or 0 to x .
b) The drivers of trireg net turn on and propagate a 1,0 , or x into the trireg net.

### 7.14.2.2 The delay specification for charge decay time

The third delay in a trireg net declaration shall specify the charge decay time. A three-valued delay specification in a trireg net declaration shall have the following form:
\#(d1, d2, d3) // (rise_delay, fall_delay, charge_decay_time)

The charge decay time specification in a trireg net declaration shall be preceded by a rise and a fall delay specification.

## Examples:

Example 1 The following example shows a specification of the charge decay time in a trireg net declaration:
trireg (large) \# ( $0,0,50$ ) cap1;

This example declares a trireg net named cap1. This trireg net stores a large charge. The delay specifications for the rise delay is 0 , the fall delay is 0 , and the charge decay time specification is 50 time units.

Example 2 The next example presents a source description file that contains a trireg net declaration with a charge decay time specification. Figure 29 shows an equivalent schematic for the source description.


Figure 29-Trireg net with capacitance

```
module capacitor;
reg data, gate;
// trireg declaration with a charge decay time of 50 time units
trireg (large) # (0,0,50) cap1;
nmos nmos1 (cap1, data, gate); // nmos that drives the trireg
initial begin
    $monitor("%0d data=%v gate=%v cap1=%v", $time, data, gate, cap1);
    data = 1;
    // Toggle the driver of the control input to the nmos switch
        gate = 1;
    #10 gate = 0;
    #30 gate = 1;
    #10 gate = 0;
    #100 $finish;
end
endmodule
```


## 8. User-defined primitives (UD s)

This clause describes a modeling technique to augment the set of predefined gate primitives by designing and specifying new primitive elements called user-defined primitives (UDPs). Instances of these new UDPs can be used in exactly the same manner as the gate primitives to represent the circuit being modeled.

The following two types of behavior can be represented in a user-defined primitive:
a) Combinational modeled by a combinational UDP
b) Sequential modeled by a sequential UDP

A combinational UDP uses the value of its inputs to determine the next value of its output. A sequential UDP uses the value of its inputs and the current value of its output to determine the value of its output. Sequential UDPs provide a way to model sequential circuits such as flip-flops and latches. A sequential UDP can model both level-sensitive and edge-sensitive behavior.

Each UDP has exactly one output, which can be in one of three states: 0,1 , or $x$. The three-state value $z$ is not supported. In sequential UDPs, the output always has the same value as the internal state.

The $z$ values passed to UDP inputs shall be treated the same as x values.

### 8.1 UD definition

UDP definitions are independent of modules; they are at the same level as module definitions in the syntax hierarchy. They can appear anywhere in the source text, either before or after they are instantiated inside a module. They shall not appear between the keywords module and endmodule.

NOTE Implementations may limit the maximum number of UDP definitions in a model, but they shall allow at least 256.

The formal syntax of the UDP definition is given in Syntax 8-1.

```
udp_declaration ::= (From Annex A - A.5.1)
        { attribute_instance } primitive udp_identifier ( udp_port_list ) ;
        udp_port_declaration { udp_port_declaration }
        udp_body
        endprimitive
    | { attribute_instance } primitive udp_identifier (udp_declaration_port_list ) ;
        udp_body
        endprimitive
udp_port_list ::= (From Annex A - A.5.2)
    output_port_identifier, input_port_identifier { , input_port_identifier }
udp_declaration_port_list ::=
    udp_output_declaration, udp_input_declaration { , udp_input_declaration
```


### 8.1.1 UD header

A UDP definition shall have one of two alternate forms. The first form shall begin with the keyword primitive, followed by an identifier, which shall be the name of the UDP. This in turn shall be followed by a comma-separated list of port names enclosed in parentheses, which shall be followed by a semicolon. The UDP definition header shall be followed by port declarations and a state table. The UDP definition shall be terminated by the keyword endprimitive.

The second form shall begin with the keyword primitive, followed by an identifier, which shall be the name of the UDP. This in turn shall be followed by a comma separated list of port declarations enclosed in parenthesis, followed by a semicolon. The UDP definition header shall be followed by a state table. The UDP definition shall be terminated by the keyword endprimitive.

UDPs have multiple input ports and exactly one output port; bidirectional inout ports are not permitted on UDPs. All ports of a UDP shall be scalar; vector ports are not permitted.

The output port shall be the first port in the port list.

### 8.1.2 UD port declarations

UDPs shall contain input and output port declarations. The output port declaration begins with the keyword output, followed by one output port name. The input port declaration begins with the keyword input, followed by one or more input port names.

Sequential UDPs shall contain a reg declaration for the output port, either in addition to the output declaration, when the UDP is declared using the first form of a UDP Header, or as part of the output_declaration, in either case. Combinational UDPs cannot contain a reg declaration. The initial value of the output port can be specified in an initial statement in a sequential UDP (see 8.1.3).

NOTE-Implementations may limit the maximum number of inputs to a UDP, but they shall allow at least 9 inputs for sequential UDPs and 10 inputs for combinational UDPs.

### 8.1.3 Sequential UD initial statement

The sequential UDP initial statement specifies the value of the output port when simulation begins. This statement begins with the keyword initial. The statement that follows shall be an assignment statement that assigns a single-bit literal value to the output port.

### 8.1.4 UD state table

The state table defines the behavior of a UDP. It begins with the keyword table and is terminated with the keyword endtable. Each row of the table is terminated by a semicolon.

Each row of the table is created using a variety of characters (see Table 40), which indicate input values and output state. Three states 0,1 , and x are supported. The z state is explicitly excluded from consideration in user-defined primitives. A number of special characters are defined to represent certain combinations of state possibilities. These are described in Table 40.

The order of the input state fields of each row of the state table is taken directly from the port list in the UDP definition header. It is not related to the order of the input port declarations.

Combinational UDPs have one field per input and one field for the output. The input fields are separated from the output field by a colon (:). Each row defines the output for a particular combination of the input values (see 8.2).

Sequential UDPs have an additional field inserted between the input fields and the output field. This additional field represents the current state of the UDP and is considered equivalent to the current output value. It is delimited by colons. Each row defines the output based on the current state, particular combinations of input values, and at most one input transition (see 8.4). A row such as the one shown below is illegal:

$$
\text { (01) (10) } 0: 0: 1 \text {; }
$$

If all input values are specified as $x$, then the output state shall be specified as $x$.
It is not necessary to explicitly specify every possible input combination. All combinations of input values that are not explicitly specified result in a default output state of x .

It is illegal to have the same combination of inputs, including edges, specified for different outputs.

### 8.1.5 Z values in UD

The $z$ value in a table entry is not supported and it is considered illegal. The $z$ values passed to UDP inputs shall be treated the same as $x$ values.

### 8.1.6 Summary of symbols

To improve the readability and to ease writing of the state table, several special symbols are provided. Table 40 summarizes the meaning of all the value symbols that are valid in the table part of a UDP definition.

Table 40-UDP table symbols

| Symbol | Interpretation | Comments |
| :--- | :--- | :--- |
| 0 | Logic 0 | Logic 1 |
| 1 | Unknown | Permitted in the input fields of all <br> UDPs and in the current state field <br> of sequential UDPs. |
| x | Iteration of 0, 1, and x | Not permitted in output field. |
| $?$ | Iteration of 0 and 1 | Permitted in the input fields of all <br> UDPs and in the current state field <br> of sequential UDPs. Not permitted <br> in the output field. |
| b | No change | Permitted only in the output field of <br> a sequential UDP. |
| - | Value change from v to w | v and w can be any one of 0, <br> $1, \mathrm{x}, ?$, or b, and are only permitted <br> in the input field. |
| $(\mathrm{vw})$ | Same as $(? ?)$ | Any value change on input. |
| $*$ | Same as $(01)$ | Rising edge on input. |
| r | Same as (10) | Falling edge on input. |
| f | Iteration of $(01),(0 \mathrm{x})$ and (x1) | Potential positive edge on the input. |
| p | Iteration of (10), (1x)and (x0 | Potential negative edge on the input. |
| n |  |  |

### 8.2 Combinational UD s

In combinational UDPs, the output state is determined solely as a function of the current input states. Whenever an input state changes, the UDP is evaluated and the output state is set to the value indicated by the row in the state table that matches all the input states. All combinations of the inputs that are not explicitly specified will drive the output state to the unknown value x .

## Examples:

The following example defines a multiplexer with two data inputs and a control input.

```
primitive multiplexer (mux, control, dataA, dataB);
output mux;
input control, dataA, dataB;
table
// control dataA dataB mux
            0 1 0 : 1;
            0 1 1 : 1 ;
            0 1 x : 1 ;
            0 0 0 : 0 ;
            0 0 1 : 0 ;
            0 0 x : 0 ;
            1 0 1 : 1 ;
            1 1 1 : 1 ;
            1 x 1 : 1 ;
            1 0 0 : 0 ;
            1 1 0 : 0 ;
            1 x 0 : 0 ;
            x 0 0 : 0 ;
            x 1 1 : 1 ;
```

endtable
endprimitive

The first entry in this example can be explained as follows: when control equals 0 , and dataA equals 1 , and dataB equals 0 , then output mux equals 1 .

The input combination $0 x x$ (control=0, dataA $=x, \operatorname{dataB}=x$ ) is not specified. If this combination occurs during simulation, the value of output port mux will become x .

Using ? , the description of a multiplexer can be abbreviated as

```
primitive multiplexer (mux, control, dataA, dataB);
output mux;
input control, dataA, dataB;
table
// control dataA dataB mux
    0 1 ? : 1 ; // ? = 0 1 x
```


### 8.3 Level-sensitive sequential UD s

Level-sensitive sequential behavior is represented the same way as combinational behavior, except that the output is declared to be of type reg, and there is an additional field in each table entry. This new field represents the current state of the UDP. The output field in a sequential UDP represents the next state.

## Example:

Consider the example of a latch:

```
primitive latch (q, clock, data);
output \(q\); reg \(q\);
input clock, data;
table
// clock data q q+
    \(0 \quad 1\) : ? : 1 ;
    00 : ? : 0 ;
    1 ? : ? : - ; // - = no change
endtable
endprimitive
```

This description differs from a combinational UDP model in two ways. First, the output identifier $q$ has an additional reg declaration to indicate that there is an internal state $q$. The output value of the UDP is always the same as the internal state. Second, a field for the current state, which is separated by colons from the inputs and the output, has been added.

### 8.4 Edge-sensitive sequential UD s

In level-sensitive behavior, the values of the inputs and the current state are sufficient to determine the output value. Edge-sensitive behavior differs in that changes in the output are triggered by specific transitions of the inputs. This makes the state table a transition table.

Each table entry can have a transition specification on at most one input. A transition is specified by a pair of values in parenthesis such as ( 01 ) or a transition symbol such as r. Entries such as the following are illegal:

$$
(01)(01) 0: 0: 1 \text {; }
$$

All transitions that do not affect the output shall be explicitly specified. Otherwise, such transitions cause the value of the output to change to x . All unspecified transitions default to the output value x .

If the behavior of the UDP is sensitive to edges of any input, the desired output state shall be specified for all edges of all inputs.

## Example:

The following example describes a rising edge D flip-flop:

```
primitive d_edge_ff (q, clock, data);
output q; reg q;
input clock, data;
table
// clock data q q+
    // obtain output on rising edge of clock
    (01) 0 : ? : 0 ;
    (01) 1 : ? : 1 ;
    (0?) 1 : 1 : 1 ;
    (0?) 0 : 0 : 0 ;
    // ignore negative edge of clock
    (? 0) ? : ? : - ;
    // ignore data changes on steady clock
    ? (??) : ? : - ;
endtable
endprimitive
```

The terms such as (01) represent transitions of the input values. Specifically, (01) represents a transition from 0 to 1 . The first line in the table of the preceding UDP definition is interpreted as follows: when clock changes value from 0 to 1 , and data equals 0 , the output goes to 0 no matter what the current state.

The transition of clock from 0 to x with data equal to 0 and current state equal to 1 will result in the output $q$ going to x .

### 8.5 Sequential UD initialization

The initial value on the output port of a sequential UDP can be specified with an initial statement that provides a procedural assignment. The initial statement is optional.

Like initial statements in modules, the initial statement in UDPs begin with the keyword initial. The valid contents of initial statements in UDPs and the valid left-hand and right-hand sides of their procedural assignment statements differ from initial statements in modules. A partial list of differences between these two types of initial statements is described in Table 41.

Table 41-Initial statements in UDPs and modules

| Initial statements in UDPs | Initial statements in modules |
| :--- | :--- |
| Contents limited to one procedural assignment <br> statement | Contents can be one procedural statement of <br> any type or a block statement that contains <br> more than one procedural statement |
| The procedural assignment statement shall <br> assign a value to a reg whose identifier <br> matches the identifier of an output terminal | Procedural assignment statements in initial <br> statements can assign values to a reg whose <br> identifier does not match the identifier of an <br> output terminal |
| The procedural assignment statement shall <br> assign one of the following values: $1 \mathrm{~b} 1,1 \mathrm{~b} 0$, <br> 1 bx, 1,0 | Procedural assignment statements can assign <br> values of any size, radix, and value |

## Examples:

Example 1 The following example shows a sequential UDP that contains an initial statement.

```
primitive srff (q, s, r);
output q; reg q;
input s, r;
initial q = 1' b1;
table
// s r q q+
    1 0 : ? : 1 ;
    f 0 : 1 : - ;
    0 r : ? : 0 ;
    0 f : 0 : - ;
    1 1 : ? : 0 ;
endtable
endprimitive
```

The output $q$ has an initial value of 1 at the start of the simulation; a delay specification on an instantiated UDP does not delay the simulation time of the assignment of this initial value to the output. When simulation starts, this value is the current state in the state table. Delays are not permitted in a UDP initial statement.

Example 2 The following example and figure show how values are applied in a module that instantiates a sequential UDP with an initial statement.

```
primitive dff1 (q, clk, d);
input clk, d;
output q; reg q;
initial q = 1' b1;
table
// clk d q q+
    r 0 : ? : 0 ;
    r 1 : ? : 1 ;
    f ? : ? : - ;
    ? * : ? : - ;
endtable
endprimitive
module dff (q, qb, clk, d);
input clk, d;
output q, qb;
    dff1 g1 (qi, clk, d);
    buf #3 g2 (q, qi);
    not #5 g3 (qb, qi);
endmodule
```

The UDP $d f f 1$ contains an initial statement that sets the initial value of its output to 1 . The module $d f f$ contains an instance of UDP $d f f 1$.

Figure 30 shows the schematic of the preceding module and the simulation propagation times of the initial value of the UDP output.


Figure 30-Module schematic and simulation times of initial value propagation
In Figure 30, the fanout from the UDP output qi includes nets $q$ and $q b$. At simulation time 0 , qi changes value to 1 . That initial value of qi does not propagate to net $q$ until simulation time 3 , and it does not propagate to net qb until simulation time 5 .

### 8.6 UD instances

The syntax for creating a UDP instance is shown in Syntax 8-2.

```
udp_instantiation ::= (From Annex A- A.5.4)
    udp_identifier [ drive_strength ] [ delay2 ]
        [attribute_instance] udp_instance \{ , udp_instance \};
udp_instance ::=
    [ name_of_udp_instance ] ( output_terminal , input_terminal
                \{ , input_terminal \})
name_of_udp_instance ::=
    udp_instance_identifier [ range ]
```

Instances of user-defined primitives are specified inside modules in the same manner as gates (see 7.1). The instance name is optional, just as for gates. The port connection order is as specified in the UDP definition. Only two delays can be specified because z is not supported for UDPs. An optional range may be specified for an array of UDP instances. The port connection rules remain the same as outlined in 7.1.

Example:
The following example creates an instance of the D-type flip-flop d_edge_ff (defined in 8.4).

```
module flip;
reg clock, data;
parameter p1 = 10;
parameter p2 = 33;
parameter p3 = 12;
d_edge_ff #p3 d_inst (q, clock, data);
```

initial begin
data $=1 ;$
clock = 1;
\#(20 * p1) \$finish;
end
always \#p1 clock = ~clock;
always \#p2 data = ~data;
endmodule

### 8.7 Mixing level-sensitive and edge-sensitive descriptions

UDP definitions allow a mixing of the level-sensitive and the edge-sensitive constructs in the same table. When the input changes, the edge-sensitive cases are processed first, followed by level-sensitive cases. Thus, when level-sensitive and edge-sensitive cases specify different output values, the result is specified by the level-sensitive case.

## Example:

primitive jk_edge_ff (q, clock, j, k, preset, clear);
output $q$; reg q;
input clock, j, k, preset, clear;
table
// clock jk pc state output/next state
? ?? 01 : ? : 1 ; // preset logic
? ?? * 1 : 1 : 1 ;
? ?? 10 : ? : 0 ; // clear logic
? ?? 1* : 0 : 0 ;
r $0000: 0$ : 1 ; // normal clocking cases
r 0011 : ? : - ;
r 0111 : ? : 0 ;
r 1011 : ? : 1 ;
r 1111 : 0 : 1 ;
r 1111 : 1 : 0 ;
f ?? ?? : ? : - ;
b *? ?? : ? : - ; // j and k transition cases
b ?* ? ? : ? : - ;
endtable
endprimitive

In this example, the preset and clear logic is level-sensitive. Whenever the preset and clear combination is 01 , the output has value 1 . Similarly, whenever the preset and clear combination has value 10 , the output has value 0 .

The remaining logic is sensitive to edges of the clock. In the normal clocking cases, the flip-flop is sensitive to the rising clock edge, as indicated by an $r$ in the clock field in those entries. The insensitivity to the falling edge of clock is indicated by a hyphen ( - ) in the output field (see Table 40) for the entry with an $f$ as the value of clock. Remember that the desired output for this input transition shall be specified to avoid unwanted $x$ values at the output. The last two entries show that the transitions in $j$ and $k$ inputs do not change the output on a steady low or high clock.

### 8.8 Level-sensitive dominance

Table 42 shows level-sensitive and edge-sensitive entries in the example from 8.7, their level-sensitive or edge-sensitive behavior, and a case of input values that each includes.

Table 42-Mixing of level-sensitive and edge-sensitive entries

| Entry | Included case | Behavior |
| :--- | :--- | :--- |
| ? ?? 01: ?: 1; | $00001: 0: 1 ;$ | Level-sensitive |
| f ?? ??: ?: -; | f $0001: 0: 0 ;$ | Edge-sensitive |

The included cases specify opposite next state values for the same input and current state combination. The level-sensitive included case specifies that when the inputs clock, $j k$, and pc values are 0,00 , and 01 and the current state is 0 , the output changes to 1 . The edge-sensitive included case specifies that when clock falls from 1 to 0 , the other inputs $j \mathrm{k}$ and pc are 00 and 01 , and the current state is 0 , then the output changes to 0 .

When the edge-sensitive case is processed first, followed by the level-sensitive case, the output changes to 1.

## 9. Behavioral modeling

The language constructs introduced so far allow hardware to be described at a relatively detailed level. Modeling a circuit with logic gates and continuous assignments reflects quite closely the logic structure of the circuit being modeled; however, these constructs do not provide the power of abstraction necessary for describing complex high-level aspects of a system. The procedural constructs described in this section are well suited to tackling problems such as describing a microprocessor or implementing complex timing checks.

This section starts with a brief overview of a behavioral model to provide a context for many types of behavioral statements in the Verilog HDL.

### 9.1 Behavioral model overview

Verilog behavioral models contain procedural statements that control the simulation and manipulate variables of the data types previously described. These statements are contained within procedures. Each procedure has an activity flow associated with it.

The activity starts at the control constructs initial and always. Each initial construct and each always construct starts a separate activity flow. All of the activity flows are concurrent to model the inherent concurrence of hardware. These constructs are formally described in 9.9.

The following example shows a complete Verilog behavioral model.

```
module behave;
reg [1:0] a, b;
initial begin
    a = 'b1;
    b = 'b0;
end
always begin
    #50 a = ~a;
end
always begin
    #100 b = ~b;
end
endmodule
```

During simulation of this model, all of the flows defined by the initial and always constructs start together at simulation time zero. The initial constructs execute once, and the always constructs execute repetitively.

In this model, the reg variables $a$ and $b$ initialize to 1 and 0 respectively at simulation time zero. The initial construct is then complete and does not execute again during this simulation run. This initial construct contains a begin-end block (also called a sequential block) of statements. In this begin-end block a is initialized first, followed by b.

The always constructs also start at time zero, but the values of the variables do not change until the times specified by the delay controls (introduced by \#) have elapsed. Thus, reg a inverts after 50 time units and reg b inverts after 100 time units. Since the always constructs repeat, this model will produce two square waves. The reg a toggles with a period of 100 time units, and reg b toggles with a period of 200 time units. The two always constructs proceed concurrently throughout the entire simulation run.

## 9.2 rocedural assignments

As described in Clause 6, procedural assignments are used for updating reg, integer, time, real, realtime, and memory data types. There is a significant difference between procedural assignments and continuous assignments:

Continuous assignments drive nets and are evaluated and updated whenever an input operand changes value.
Procedural assignments update the value of variables under the control of the procedural flow constructs that surround them.

The right-hand side of a procedural assignment can be any expression that evaluates to a value. The lefthand side shall be a variable that receives the assignment from the right-hand side. The left-hand side of a procedural assignment can take one of the following forms:
reg, integer, real, realtime, or time data type: an assignment to the name reference of one of these data types.
Bit-select of a reg, integer, or time data type: an assignment to a single bit that leaves the other bits untouched.
Part-select of a reg, integer, or time data type: a part-select of one or more contiguous bits that leaves the rest of the bits untouched.
Memory word: a single word of a memory.
Concatenation of any of the above: a concatenation of any of the previous four forms can be speci-
fied, which effectively partitions the result of the right-hand side expression and assigns the partition parts, in order, to the various parts of the concatenation.

NOTE When the right-hand side evaluates to fewer bits than the left-hand side, then if the right-hand side is signed (see 4.5), it shall be sign-extended to the size of the left-hand side.

The Verilog HDL contains two types of procedural assignment statements:
Blocking procedural assignment statements
Non blocking procedural assignment statements

Blocking and non blocking procedural assignment statements specify different procedural flows in sequential blocks.

### 9.2.1 Blocking procedural assignments

A blocking procedural assignment statement shall be executed before the execution of the statements that follow it in a sequential block (see 9.8.1). A blocking procedural assignment statement shall not prevent the execution of statements that follow it in a parallel block (see 9.8.2).

The syntax for a blocking procedural assignment is given in Syntax 9-1.

```
blocking_assignment \(::=(\) From Annex A-A.6.2)
        variable_lvalue \(=[\) delay_or_event_control ] expression
delay_control \(::=(\) From Annex \(A-\) A.6.5 \()\)
        \# delay_value
        | \# ( mintypmax_expression )
delay_or_event_control ::=
        delay_control
        | event_control
        repeat ( expression ) event_control
event_control ::=
        © event_identifier
        @ ( event_expression)
        @*
        @ (*)
event_expression ::=
        expression
        | hierarchical_identifier
        posedge expression
        negedge expression
        event_expression or event_expression
        event_expression, event_expression
variable_lvalue \(::=(\) From Annex \(A-A .8 .5)\)
        hierarchical_variable_identifier
        hierarchical_variable_identifier [ expression ] \{ [ expression ] \}
        | hierarchical_variable_identifier [ expression ] \{ [ expression ] \}
        [ range_expression ]
        | hierarchical_variable_identifier [ range_expression ]
        | variable_concatenation
```


## Syntax 9-1 Syntax for blocking assignments

In this syntax, reg_lvalue is a data type that is valid for a procedural assignment statement, $=$ is the assignment operator, and delay_or_event_control is the optional intra-assignment timing control. The control can be either a delay control (e.g., \#6) or an event_control (e.g., @ (posedge clk)). The expression is the right-hand side value that shall be assigned to the left-hand side. If reg_lvalue requires an evaluation, it shall be evaluated at the time specified by the intra-assignment timing control.

The $=$ assignment operator used by blocking procedural assignments is also used by procedural continuous assignments and continuous assignments.

Example:

The following examples show blocking procedural assignments.

```
rega = 0;
rega[3] = 1; // a bit-select
rega[3:5] = 7; // a part-select
mema[address] = 8'hff; // assignment to a mem element
{carry, acc} = rega + regb; // a concatenation
```


### 9.2.2 The non blocking procedural assignment

The non blocking procedural assignment allows assignment scheduling without blocking the procedural flow. The non blocking procedural assignment statement can be used whenever several variable assignments within the same time step can be made without regard to order or dependence upon each other.

The syntax for a non blocking procedural assignment is given in Syntax 9-2.

```
nonblocking_assignment \(::=\) (From Annex A-A.6.2)
        variable_lvalue \(<=\) [ delay_or_event_control ] expression
    delay_control \(::=\) (From Annex A-A.6.5)
        \# delay_value
        | \# ( mintypmax_expression )
    delay_or_event_control ::=
        delay_control
        | event_control
        repeat ( expression ) event_control
    event_control ::=
        (a) event_identifier
        @ ( event_expression)
    © *
    - (*)
event_expression ::=
        expression
        | hierarchical_identifier
        posedge expression
        negedge expression
        event_expression or event_expression
        event_expression, event_expression
    variable_lvalue \(::=(\) From Annex \(A-A .8 .5)\)
        hierarchical_variable_identifier
        | hierarchical_variable_identifier [ expression ] \{[ expression ] \}
        | hierarchical_variable_identifier [ expression ] \{[expression] \}
        [ range_expression]
        | hierarchical_variable_identifier [ range_expression ]
        | variable_concatenation
```


## Syntax 9-2 Syntax for non blocking assignments

In this syntax, reg_lvalue is a data type that is valid for a procedural assignment statement, $<=$ is the non blocking assignment operator, and delay_or_event_control is the optional intra-assignment timing control. If reg_lvalue requires an evaluation, it shall be evaluated at the same time as the expression on the right-hand side. The order of evaluation of the reg_lvalue and the expression on the right-hand side is undefined if timing control is not specified.

The non blocking assignment operator is the same operator as the less-than-or-equal-to relational operator. The interpretation shall be decided from the context in which <= appears. When $<=$ is used in an expression, it shall be interpreted as a relational operator, and when it is used in a non blocking procedural assignment, it shall be interpreted as an assignment operator.

The non blocking procedural assignments shall be evaluated in two steps as discussed in Clause 5. These two steps are shown in the following example.

Example 1:
output out;
reg $a, b, c ;$
initial begin
$\mathrm{a}=0$;
b = 1;
c $=0$;
end
always $c=\# 5 \sim c ;$
a <= b; // evaluates, schedules,
b <= a; // and executes in two steps
end
endmodule
module evaluates2 (out);
always @ (posedge c) begin

At posedge c, the simulator evaluates the right-hand sides of the non blocking assignments and schedules the assignments of the new values at the end of the non blocking assign update events (see 5.4).

When the simulator activates the non blocking assign update events, the simulator updates the left-hand side of each non blocking assignment statement.

| Non block- |
| :--- |
| ing assign- |
| ment |
| schedules <br> changes at |

$a=0$
$b=1$
Assignment values are:

$$
\begin{aligned}
& a=1 \\
& b=0
\end{aligned}
$$

At the end of the time step means that the non blocking assignments are the last assignments executed in a time step with one exception. Non blocking assignment events can create blocking assignment events. These blocking assignment events shall be processed after the scheduled non blocking events.

Unlike an event or delay control for blocking assignments, the non blocking assignment does not block the procedural flow. The non blocking assignment evaluates and schedules the assignment, but it does not block the execution of subsequent statements in a begin-end block.

Example 2:


As shown in the previous example, the simulator evaluates and schedules assignments for the end of the current time step and can perform swapping operations with the non blocking procedural assignments.

Example 3:

```
//non_block1.v Step 1: The simulator evaluates the right-
module non_block1; hand side of the non blocking
reg a, b;
initial begin
    a = 0;
    b = 1;
    a <= b; // evaluates, schedules, and
    b <= a; // executes in two steps
end
initial begin
    $monitor ($time, ,"a = %b b = %b", a, b);
    #100 $finish;
end
endmodule
assignment values are:
```


## Step 2:

```
At the end of the current time step, the simulator updates the left-hand side of each non blocking assignment statement.
```


## end

```
endmodule
\[
\begin{aligned}
& a=1 \\
& b=0
\end{aligned}
\]
```

The order of the execution of distinct non blocking assignments to a given variable shall be preserved. This means that if there is clear ordering of the execution of a set of non blocking assigments, then the order of the resulting updates of the destination of the non blocking assignments shall be the same as the ordering of the execution.

Example 4:

```
module multiple2;
reg a;
initial a = 1;
// The assigned value of the reg is determinate
initial begin
    a <= #4 0; // schedules a = 0 at time 4
    a <= #4 1; // schedules a = 1 at time 4
end // At time 4, a = 1
endmodule
```

If the simulator executes two procedural blocks concurrently, and if these procedural blocks contain non blocking assignment operators to the same variable, the final value of that variable is indeterminate. For example, the value of reg a is indeterminate in the following example.

Example 5:

```
module multiple3 ;
reg a;
initial a = 1;
initial a <= #4 0; // schedules 0 at time 4
initial a <= #4 1; // schedules 1 at time 4
// At time 4, a = ??
// The assigned value of the reg is indeterminate
endmodule
```

NOTE-The fact that two non blocking assignments targeting the same variable are in different blocks is not by itself sufficient to make the order of assignments to a variable indeterminate. For example, the value of reg a at the end of time cycle 16 is determinate in the following example:

```
moduīe multiple2 ;
reg a;
initial #8 a <= #8 1;// executed at time 8; schedules
    // an update of 1 at time 16
initial #12 a <= #4 0;// executed at time 12; schedules
    // an update of 0 at time 16
    // Because it is determinate that the update of a to
    // the value 1 is scheduled before the update of a to
    // the value 0, then it is determinate that a will have the
    // value 0 at the end of time slot 16.endmodule
endmodule
```

The following example shows how the value of $i[0$ ] is assigned to $r 1$ and how the assignments are scheduled to occur after each time delay.

## Example 6:

```
module multiple;
reg r1;
reg [2:0] \(i\);
initial begin
// starts at time 0, does not hold the block
    r1 = 0;
// makes assignments to r1 without cancelling previous assignments
    for ( \(i=0\); \(i<=5\); \(i=i+1\) )
        r1 <= \# (i*10) i[0];
end
endmodule
```



## 9.3 rocedural continuous assignments

The procedural continuous assignments (using keywords assign and force) are procedural statements that allow expressions to be driven continuously onto variables or nets. The syntax for these statements is given in Syntax 9-3.

```
net_assignment ::= (From Annex A-A.6.1)
    net_lvalue \(=\) expression
procedural_continuous_assignments ::= (From Annex A - A.6.2)
        assign variable_assignment
        deassign variable_lvalue
        force variable_assignment
        force net_assignment
        release variable_lvalue
        release net_lvalue
variable_assignment ::= (From Annex A - A.6.3)
        variable_lvalue \(=\) expression
net_lvalue ::= (From Annex A - A.8.5)
        hierarchical_net_identifier
        hierarchical_net_identifier [ constant_expression ] \{ [ constant_expression ] \}
        | hierarchical_net_identifier [ constant_expression ] \{ [ constant_expression ] \}
            [ constant_range_expression ]
        | hierarchical_net_identifier [ constant_range_expression ]
        net_concatenation
variable_lvalue ::=
        hierarchical_variable_identifier
        hierarchical_variable_identifier [ expression ] \{ [ expression ] \}
        | hierarchical_variable_identifier [ expression ] \{ [ expression ] \}
        [ range_expression ]
    | hierarchical_variable_identifier [ range_expression ]
    variable_concatenation
```


## Syntax 9-3 Syntax for procedural continuous assignments

The left-hand side of the assignment in the assign statement shall be a variable reference or a concatenation of variables. It shall not be a memory word (array reference) or a bit-select or a part-select of a variable.

In contrast, the left-hand side of the assignment in the force statement can be a variable reference or a net reference. It can be a concatenation of any of the above. Bit-selects and part-selects of vector variables are not allowed.

### 9.3.1 The assign and deassign procedural statements

The assign procedural continuous assignment statement shall override all procedural assignments to a variable. The deassign procedural statement shall end a procedural continuous assignment to a variable. The value of the variable shall remain the same until the reg is assigned a new value through a procedural assignment or a procedural continuous assignment. The assign and deassign procedural statements allow, for example, modeling of asynchronous clear/preset on a D-type edge-triggered flip-flop, where the clock is inhibited when the clear or preset is active.

If the keyword assign is applied to a variable for which there is already a procedural continuous assignment, then this new procedural continuous assignment shall deassign the variable before making the new procedural continuous assignment.

## Example:

The following example shows a use of the assign and deassign procedural statements in a behavioral description of a D-type flip-flop with preset and clear inputs.

```
module dff (q, d, clear, preset, clock);
output q;
input d, clear, preset, clock;
reg q;
always @(clear or preset)
    if (!clear)
        assign q = 0;
    else if (!preset)
        assign q = 1;
    else
        deassign q;
always @(posedge clock)
    q = d;
endmodule
```

If either clear or preset is low, then the output $q$ will be held continuously to the appropriate constant value and a positive edge on the clock will not affect $q$. When both the clear and preset are high, then q is deassigned.

### 9.3.2 The force and release procedural statements

Another form of procedural continuous assignment is provided by the force and release procedural statements. These statements have a similar effect to the assign-deassign pair, but a force can be applied to nets as well as to variables. The left-hand side of the assignment can be a variable, a net, a constant bit-select of a vector net, a part-select of a vector net, or a concatenation. It cannot be a memory word (array reference) or a bit-select or a part-select of a vector variable.

A force statement to a variable shall override a procedural assignment or procedural continuous assignment that takes place on the variable until a release procedural statement is executed on the variable. After the release procedural statement is executed, the variable shall not immediately change value (as would a net that is assigned with a procedural continuous assignment). The value specified in the force statement shall be maintained in the variable until the next procedural assignment takes place, except in the case where a procedural continuous assignment is active on the variable.

A force procedural statement on a net overrides all drivers of the net gate outputs, module outputs, and continuous assignments until a release procedural statement is executed on the net.

Releasing a variable that currently has an active procedural continuous assignment shall re-establish that assignment.

## Example:

```
module test;
reg a, b, c, d;
wire e;
and and1 (e, a, b, c);
initial begin
    $monitor("%d d=%b,e=%b", $stime, d, e);
    assign d = a & b & c;
    a = 1;
    b = 0;
    c = 1;
    #10;
    force d = (a | b | c);
    force e = (a | b | c);
    #10 $stop;
    release d;
    release e;
    #10 $finish;
end
endmodule
Results:
    0 d=0,e=0
    10 d=1,e=1
    20 d=0,e=0
```

In this example, an and gate instance and 1 is patched as an or gate by a force procedural statement that forces its output to the value of its logical or inputs, and an assign procedural statement of logical and values is patched as an assign procedural statement of logical or values.

The right-hand side of a procedural continuous assignment or a force statement can be an expression. This shall be treated just as a continuous assignment; that is, if any variable on the right-hand side of the assignment changes, the assignment shall be re-evaluated while the assign or force is in effect. For example:
force $a=b+f(c) ;$
Here, if $b$ changes or $c$ changes, $a$ will be forced to the new value of the expression $b+f(c)$.

### 9.4 Conditional statement

The conditional statement (or if-else statement) is used to make a decision as to whether a statement is executed or not. Formally, the syntax is given in Syntax 9-4.

```
conditional_statement \(::=\) (From Annex \(A-\) A.6.6)
        if ( expression )
        statement_or_null [ else statement_or_null ]
        | if_else_if_statement
function_conditional_statement \(::=\) (From Annex A - A.6.6)
    if ( expression ) function_statement_or_null
        [ else function_statement_or_null]
        | function_if_else_if_statement
```


## Syntax 9-4 Syntax of if statement

If the expression evaluates to true (that is, has a nonzero known value), the first statement shall be executed. If it evaluates to false (has a zero value or the value is $x$ or $z$ ), the first statement shall not execute. If there is an else statement and expression is false, the else statement shall be executed.

Since the numeric value of the if expression is tested for being zero, certain shortcuts are possible. For example, the following two statements express the same logic:

```
if (expression)
if (expression != 0)
```

Because the else part of an if-else is optional, there can be confusion when an else is omitted from a nested if sequence. This is resolved by always associating the else with the closest previous if that lacks an else. In the example below, the else goes with the inner if, as shown by indentation.

```
if (index > 0)
    if (rega > regb)
        result = rega;
    else // else applies to preceding if
        result = regb;
```

If that association is not desired, a begin-end block statement shall be used to force the proper association, as shown below.

```
if (index > 0) begin
    if (rega > regb)
        result = rega;
end
else result = regb;
```


### 9.4.1 If-else-if construct

The following construction occurs so often that it is worth a brief separate discussion:

```
if_else_if_statement \(::=(\) From Annex A - A.6.6)
    if ( expression ) statement_or_null
        \{ else if ( expression ) statement_or_null \}
        [ else statement_or_null ]
function_if_else_if_statement \(::=\) (From Annex A - A.6.6)
    if ( expression ) function_statement_or_null
        \{ else if ( expression ) function_statement_or_null \}
        [ else function_statement_or_null ]
```


## Syntax 9-5 Syntax of if-else-if construct

This sequence of if statements (known as an if-else-if construct) is the most general way of writing a multiway decision. The expressions shall be evaluated in order; if any expression is true, the statement associated with it shall be executed, and this shall terminate the whole chain. Each statement is either a single statement or a block of statements.

The last else part of the if-else-if construct handles the none-of-the-above or default case where none of the other conditions were satisfied. Sometimes there is no explicit action for the default; in that case, the trailing else statement can be omitted or it can be used for error checking to catch an impossible condition.

## Example:

The following module fragment uses the if-else statement to test the variable index to decide whether one of three modify_segn regs has to be added to the memory address, and which increment is to be added to the index reg. The first ten lines declare the regs and parameters.

```
// declare regs and parameters
reg [31:0] instruction, segment_area[255:0];
reg [7:0] index;
reg [5:0] modify_seg1,
    modify_seg2,
    modify_seg3;
parameter
    segment1 = 0, inc_seg1 = 1,
    segment2 = 20, inc_seg2 = 2,
    segment3 = 64, inc_seg3 = 4,
    data = 128;
// test the index variable
if (index < segment2) begin
    instruction = segment_area [index + modify_seg1];
    index = index + inc_seg1;
end
else if (index < segment3) begin
    instruction = segment_area [index + modify_seg2];
    index = index + inc_seg2;
end
else if (index < data) begin
    instruction = segment_area [index + modify_seg3];
    index = index + inc_seg3;
end
else
    instruction = segment_area [index];
```


### 9.5 Case statement

The case statement is a multiway decision statement that tests whether an expression matches one of a number of other expressions and branches accordingly. The case statement has the syntax shown in Syntax 9-6.

```
case_statement ::= (From Annex A - A.6.7)
        case ( expression)
        case_item { case_item } endcase
    | casez ( expression )
        case_item { case_item } endcase
    | casex ( expression )
        case_item { case_item } endcase
case_item ::=
            expression {, expression } : statement_or_null
            default [:] statement_or_null
function_case_statement ::=
        case ( expression)
        function_case_item { function_case_item } endcase
    | casez ( expression )
        function_case_item { function_case_item } endcase
    | casex ( expression )
        function_case_item { function_case_item } endcase
function_case_item ::=
    expression {, expression } : function_statement_or_null
    | default [:] function_statement_or_null
```


## Syntax 9-6 Syntax for case statement

The default statement shall be optional. Use of multiple default statements in one case statement shall be illegal.

The case expression and the case item expression can be computed at runtime; neither expression is required to be a constant expression.

## Examples:

A simple example of the use of the case statement is the decoding of reg rega to produce a value for result as follows:

```
reg [15:0] rega;
reg [9:0] result;
case (rega)
    16'd0: result = 10'b0111111111;
    16'd1: result = 10'b1011111111;
    16'd2: result = 10'b1101111111;
    16'd3: result = 10'b1110111111;
    16'd4: result = 10'b1111011111;
    16'd5: result = 10'b1111101111;
    16'd6: result = 10'b1111110111;
    16'd7: result = 10'b1111111011;
    16'd8: result = 10'b11111111101;
    16'd9: result = 10'b11111111110;
    default result = 'bx;
endcase
```

The case item expressions shall be evaluated and compared in the exact order in which they are given. During the linear search, if one of the case item expressions matches the case expression given in parentheses, then the statement associated with that case item shall be executed. If all comparisons fail and the default item is given, then the default item statement shall be executed. If the default statement is not given and all of the comparisons fail, then none of the case item statements shall be executed.

Apart from syntax, the case statement differs from the multiway if-else-if construct in two important ways:
a) The conditional expressions in the if-else-if construct are more general than comparing one expression with several others, as in the case statement.
b) The case statement provides a definitive result when there are x and z values in an expression.

In a case expression comparison, the comparison only succeeds when each bit matches exactly with respect to the values $0,1, x$, and $z$. As a consequence, care is needed in specifying the expressions in the case statement. The bit length of all the expressions shall be equal so that exact bit-wise matching can be performed. The length of all the case item expressions, as well as the case expression in the parentheses, shall be made equal to the length of the longest case expression and case item expression.

NOTE The default length of x and z is same as the default length of an integer.

The reason for providing a case expression comparison that handles the x and z values is that it provides a mechanism for detecting such values and reducing the pessimism that can be generated by their presence.

## Examples:

Example 1 The following example illustrates the use of a case statement to handle x and z values properly.

```
case (select[1:2])
    2'b00: result = 0;
    2'b01: result = flaga;
    2'b0x,
    2'b0z: result = flaga ? 'bx : 0;
    2'b10: result = flagb;
    2'bx0,
    2'bz0: result = flagb ? 'bx : 0;
    default result = 'bx;
endcase
```

In this example, if select[1] is 0 and flaga is 0 , then whether the value of select[2] is x or z , result should be 0 which is resolved by the third case.

Example 2 The following example shows another way to use a case statement to detect x and z values.

```
case (sig)
    1'bz: $display("signal is floating");
    1'bx: $display("signal is unknown");
    default: $display("signal is %b", sig);
endcase
```


### 9.5.1 Case statement with don t-cares

Two other types of case statements are provided to allow handling of don t-care conditions in the case comparisons. One of these treats high-impedance values ( $z$ ) as don t-cares, and the other treats both high-impedance and unknown (x) values as don t-cares.

These case statements can be used in the same way as the traditional case statement, but they begin with keywords casez and casex respectively.

Don t -care values ( z values for casez, z and x values for casex) in any bit of either the case expression or the case items shall be treated as don t-care conditions during the comparison, and that bit position shall not be considered. The don t-care conditions in case expression can be used to control dynamically which bits should be compared at any time.

The syntax of literal numbers allows the use of the question mark (?) in place of $z$ in these case statements. This provides a convenient format for specification of don $t$-care bits in case statements.

## Examples:

Example 1 The following is an example of the casez statement. It demonstrates an instruction decode, where values of the most significant bits select which task should be called. If the most significant bit of ir is a 1 , then the task instruction 1 is called, regardless of the values of the other bits of ir.

```
reg [7:0] ir;
casez (ir)
    8'b1???????: instruction1(ir);
    8'b01??????: instruction2(ir);
    8'b00010???: instruction3(ir);
    8'b000001??: instruction4(ir);
endcase
```

Example 2 The following is an example of the casex statement. It demonstrates an extreme case of how don t-care conditions can be dynamically controlled during simulation. In this case, if $r$ $8^{\prime} \mathrm{b} 01100110$, then the task stat2 is called.

```
reg [7:0] r, mask;
mask = 8'bx0x0x0x0;
casex (r ^ mask)
    8'b001100xx: stat1;
    8'b1100xx00: stat2;
    8'b00xx0011: stat3;
    8'bxx010100: stat4;
endcase
```


### 9.5.2 Constant expression in case statement

A constant expression can be used for case expression. The value of the constant expression shall be compared against case item expressions.

## Example:

The following example demonstrates the usage by modeling a 3-bit priority encoder.

```
reg [2:0] encode ;
case (1)
    encode[2] : $display("Select Line 2") ;
    encode[1] : $display("Select Line 1") ;
    encode[0] : $display("Select Line 0") ;
    default $display("Error: One of the bits expected ON");
endcase
```

Note that the case expression is a constant expression (1). The case items are expressions (bit-selects) and are compared against the constant expression for a match.

### 9.6 Looping statements

There are four types of looping statements. These statements provide a means of controlling the execution of a statement zero, one, or more times.
forever Continuously executes a statement.
repeat Executes a statement a fixed number of times. If the expression evaluates to unknown or high impedance, it shall be treated as zero, and no statement shall be executed.
while Executes a statement until an expression becomes false. If the expression starts out false, the statement shall not be executed at all.
for Controls execution of its associated statement(s) by a three-step process, as follows:
a) Executes an assignment normally used to initialize a variable that controls the number of loops executed.
b) Evaluates an expression if the result is zero, the for-loop shall exit, and if it is not zero, the for-loop shall execute its associated statement(s) and then perform step c. If the expression evaluates to an unknown or high-impedance value, it shall be treated as zero.
c) Executes an assignment normally used to modify the value of the loop-control variable, then repeats step b.

Syntax 9-7 shows the syntax for various looping statements.

```
function_loop_statement ::= (From Annex A - A.6.8)
    forever function_statement
    repeat ( expression ) function_statement
    while ( expression ) function_statement
    for ( variable_assignment ; expression ; variable_assignment )
        function_statement
loop_statement ::=
        forever statement
        repeat ( expression ) statement
        while ( expression ) statement
        for ( variable_assignment ; expression ; variable_assignment )
        statement
```


## Syntax 9-7 Syntax for the looping statements

The rest of this clause presents examples for three of the looping statements. The forever loop should only be used in conjunction with the timing controls or the disable statement, therefore, this example is presented in 9.7.2.

Examples:
Example 1 Repeat statement: In the following example of a repeat loop, add and shift operators implement a multiplier.

```
parameter size = 8, longsize = 16;
reg [size:1] opa, opb;
reg [longsize:1] result;
begin : mult
    reg [longsize:1] shift_opa, shift_opb;
    shift_opa = opa;
    shift_opb = opb;
    result = 0;
    repeat (size) begin
            if (shift_opb[1])
                result = result + shift_opa;
            shift_opa = shift_opa << 1;
            shift_opb = shift_opb >> 1;
    end
end
```

Example 2 While statement: The following example counts the number of logic 1 values in rega.

```
begin : count1s
    reg [7:0] tempreg;
    count = 0;
    tempreg = rega;
    while (tempreg) begin
        if (tempreg[0])
            count = count + 1;
        tempreg = tempreg >> 1;
    end
end
```

Example 3 For statement: The for statement accomplishes the same results as the following pseudo-code that is based on the while loop:

```
begin
    initial_assignment;
    while (condition) begin
        statement
        step_assignment;
        end
end
```

The for loop implements this logic while using only two lines, as shown in the pseudo-code below.

```
for (initial_assignment; condition; step_assignment)
    statement
```


## 9.7 rocedural timing controls

The Verilog HDL has two types of explicit timing control over when procedural statements can occur. The first type is a delay control, in which an expression specifies the time duration between initially encountering the statement and when the statement actually executes. The delay expression can be a dynamic function of the state of the circuit, but it can be a simple number that separates statement executions in time. The delay control is an important feature when specifying stimulus waveform descriptions. It is described in 9.7.1 and 9.7.7.

The second type of timing control is the event expression, which allows statement execution to be delayed until the occurrence of some simulation event occurring in a procedure executing concurrently with this procedure. A simulation event can be a change of value on a net or variable (an implicit event) or the occurrence of an explicitly named event that is triggered from other procedures (an explicit event). Most often, an event control is a positive or negative edge on a clock signal. Event control is discussed in 9.7.2 through 9.7.7.

The procedural statements encountered so far all execute without advancing simulation time. Simulation time can advance by one of the following three methods:

A delay control, which is introduced by the symbol \#
An event control, which is introduced by the symbol @
The wait statement, which operates like a combination of the event control and the while loop
Syntax 9-8 describes timing control in procedural statements.

```
delay_control ::=(From Annex A-A.6.5)
        \# delay_value
    | \# ( mintypmax_expression )
delay_or_event_control ::=
        delay_control
        | event_control
        repeat ( expression ) event_control
event_control ::=
        (a) event_identifier
        @ ( event_expression)
        @*
        @ (*)
event_expression ::=
        expression
        hierarchical_identifier
        posedge expression
        negedge expression
        event_expression or event_expression
        | event_expression, event_expression
```


## Syntax 9-8 Syntax for procedural timing control

The gate and net delays also advance simulation time, as discussed in Clause 6. The next subclauses discuss the three procedural timing control methods.

### 9.7.1 Delay control

A procedural statement following the delay control shall be delayed in its execution with respect to the procedural statement preceding the delay control by the specified delay. If the delay expression evaluates to an unknown or high-impedance value, it shall be interpreted as zero delay. If the delay expression evaluates to a negative value, it shall be interpreted as a 2 s complement unsigned integer of the same size as a time variable. Specify parameters are permitted in the delay expression. They may be overridden by SDF annotation, in which case the expression is reevaluated.

## Examples:

Example 1 The following example delays the execution of the assignment by 10 time units:

```
#10 rega = regb;
```

Example 2 The next three examples provide an expression following the number sign (\#). Execution of the assignment is delayed by the amount of simulation time specified by the value of the expression.

```
#d rega = regb; // d is defined as a parameter
#((d+e)/2) rega = regb;// delay is average of d and e
#regr regr = regr + 1; // delay is the value in regr
```


### 9.7.2 Event control

The execution of a procedural statement can be synchronized with a value change on a net or variable or the occurrence of a declared event. The value changes on nets and variable can be used as events to trigger the execution of a statement. This is known as detecting an implicit event. The event can also be based on the direction of the change that is, towards the value 1 ( posedge) or towards the value 0 (negedge). The behavior of posedge and negedge event is shown in Table 43 and can be described as follows:

A negedge shall be detected on the transition from 1 to $\mathrm{x}, \mathrm{z}$, or 0 , and from x or z to 0
A posedge shall be detected on the transition from 0 to $\mathbf{x}, \mathbf{z}$, or 1 , and from x or $\mathbf{z}$ to 1
Table 43-Detecting posedge and negedge

|  | To | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{x}$ | $\mathbf{z}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| From |  |  |  |  |  |
| $\mathbf{0}$ |  | No edge | posedge | posedge | posedge |
| $\mathbf{1}$ |  | negedge | No edge | negedge | negedge |
| $\mathbf{x}$ |  | negedge | posedge | No edge | No edge |
| $\mathbf{z}$ |  | negedge | posedge | No edge | No edge |

If the expression evaluates to more than a 1-bit result, the edge transition shall be detected on the least significant bit of the result. The change of value in any of the operands without a change in the value of the least significant bit of the expression result shall not be detected as an edge.

## Example:

The following example shows illustrations of edge-controlled statements.

```
@r rega = regb; // controlled by any value change in the reg r
@(posedge clock) rega = regb; // controlled by posedge on clock
forever @(negedge clock) rega = regb; // controlled by negative edge
```


### 9.7.3 Named events

A new data type, in addition to nets and variables, called event can be declared. An identifier declared as an event data type is called a named event. A named event can be triggered explicitly. It can be used in an event expression to control the execution of procedural statements in the same manner as event control described in 9.7.1. Named events can be made to occur from a procedure. This allows control over the enabling of multiple actions in other procedures.

An event name shall be declared explicitly before it is used. Syntax $9-9$ gives the syntax for declaring events.

```
event_declaration ::= (From Annex A-A.2.1.3)
    event list_of_event_identifiers ;
list_of_event_identifiers ::= (From Annex A-A.2.3)
        event_identifier [ dimension \{ dimension \}]
            \{, event_identifier [dimension \(\{\) dimension \(\}]\}\)
dimension ::= (From Annex A-A.2.5)
    [ dimension_constant_expression : dimension_constant_expression ]
```

Syntax 9-9 Syntax for event declaration

An event shall not hold any data. The following are the characteristics of a named event:

It can be made to occur at any particular time
It has no time duration
Its occurrence can be recognized by using the event control syntax described in
A declared event is made to occur by the activation of an event triggering statement with the syntax given in Syntax 9-10. An event is not made to occur by changing the index of an event array in an event control expression.

```
event_trigger ::= (From Annex A - A.6.5)
    -> hierarchical_event_identifier ;
```


## Syntax 9-10 Syntax for event trigger

An event-controlled statement (for example, @trig rega = regb; shall cause simulation of its containing procedure to wait until some other procedure executes the appropriate event-triggering statement (for example,

```
-> trig).
```

Named events and event control give a powerful and efficient means of describing the communication between, and synchronization of, two or more concurrently active processes. A basic example of this is a small waveform clock generator that synchronizes control of a synchronous circuit by signalling the occurrence of an explicit event periodically while the circuit waits for the event to occur.

### 9.7.4 Event or operator

The logical or of any number of events can be expressed such that the occurrence of any one of the events triggers the execution of the procedural statement that follows it. The keyword or or a comma character (,) is used as an event logical or operator. A combination of these can be used in the same event expression. Comma-separated sensitivity lists shall be synonymous to or-separated sensitivity lists.

## Examples:

The next two examples show the logical or of two and three events respectively.

```
@(trig or enable) rega = regb; // controlled by trig or
enable
@(posedge clk_a or posedge clk_b or trig) rega = regb;
```

The following examples show the use of the comma (, ) as an event logical or operator.

```
always @(a, b, c, d, e)
always @(posedge clk, negedge rstn)
always @(a or b, c, d or e)
```


### 9.7.5 Implicit event_expression list

The event_expression list of an event control is a common source of bugs in RTL simulations. Users tend to forget to add some of the nets or variables read in the timing control statement. This is often found when comparing RTL and gate level versions of a design. The implicit event_expression, @*, is a convenient shorthand that eliminates these problems by adding all nets and variables which are read by the statement (which can be a statement group) of a procedural_timing_control_statement to the event_expression.

All net and variable identifiers which appear in the statement will be automatically added to the event expression with these exceptions:

Identifiers which only appear in wait or event expressions.
Identifiers which only appear as a hierarchical_reg_identifier in the reg_lvalue of the left hand side of assignments.

Nets and variables which appear on the right hand side of assignments, in function and task calls, or case and conditional expressions shall all be included by these rules.

Examples:
Example 1

```
always @(*) // equivalent to @(a or b or c or d or f)
    y = (a & b) | (c & d) | myfunction(f);
```

Example 2

```
always @* begin // equivalent to @(a or b or c or d or tmp1 or tmp2)
    tmp1 = a & b;
    tmp2 = c & d;
    y = tmp1 | tmp2;
end
```

Example 3

```
always @* begin // equivalent to @(b)
    @(i) kid = b; // i is not added to @*
end
```


## Example 4

```
always @* begin // equivalent to @(a or b or c or d)
    x = a ^ b;
    @* // equivalent to @(c or d)
        x = c ^ d;
end
```


### 9.7.6 Level-sensitive event control

The execution of a procedural statement can also be delayed until a condition becomes true. This is accomplished using the wait statement, which is a special form of event control. The nature of the wait statement is level-sensitive, as opposed to basic event control (specified by the @ character), which is edge-sensitive.

The wait statement shall evaluate a condition, and, if it is false, the procedural statements following the wait statement shall remain blocked until that condition becomes true before continuing. The wait statement has the form given in Syntax 9-11.

```
wait_statement ::= (From Annex A - A.6.5)
    wait ( expression ) statement_or_null
```


## Syntax 9-11 Syntax for wait statement

## Example:

The following example shows the use of the wait statement to accomplish level-sensitive event control.

```
begin
    wait (!enable) #10 a = b;
    #10 c = d;
end
```

If the value of enable is 1 when the block is entered, the wait statement will delay the evaluation of the next statement (\#10 a = b; ) until the value of enable changes to 0 . If enable is already 0 when the begin-end block is entered, then the assignment $a=b$; is evaluated after a delay of 10 and no additional delay occurs.

### 9.7.7 Intra-assignment timing controls

The delay and event control constructs previously described precede a statement and delay its execution. In contrast, the intra-assignment delay and event controls are contained within an assignment statement and modify the flow of activity in a different way. This subclause describes the purpose of intra-assignment timing controls and the repeat timing control that can be used in intra-assignment delays.

An intra-assignment delay or event control shall delay the assignment of the new value to the left-hand side, but the right-hand side expression shall be evaluated before the delay, instead of after the delay. The syntax for intra-assignment delay and event control is given in Syntax 9-12.

```
blocking_assignment \(::=(\) From Annex A-A.6.2)
    variable_lvalue = [ delay_or_event_control ] expression
nonblocking_assignment ::=
    variable_lvalue <= [ delay_or_event_control ] expression
delay_control \(::=(\) From Annex \(A-\) A.6.5 \()\)
        \# delay_value
        | \# ( mintypmax_expression )
delay_or_event_control ::=
        delay_control
        | event_control
        repeat ( expression ) event_control
event_control ::=
        (a) event_identifier
        @ ( event_expression)
        ©*
        - (*)
event_expression ::=
        expression
        hierarchical_identifier
        posedge expression
        negedge expression
        event_expression or event_expression
        event_expression, event_expression
```

Syntax 9-12 Syntax for intra-assignment delay and event control

The intra-assignment delay and event control can be applied to both blocking assignments and non blocking assignments. The event expression shall be resolved to a 1-bit value. The repeat event control shall specify an intra-assignment delay of a specified number of occurrences of an event. If the repeat count literal, or signed reg holding the repeat count, is less than or equal to 0 at the time of evaluation, the assignment occurs as if there is no repeat construct.

## Examples:

```
repeat (-3) @ (event_expression)
    // will not execute event_expression.
repeat (a) @ (event_expression)
    // if a is assigned -3 it will execute the event_expression
    // if a is declared as an unsigned reg but not if it is
signed.
```

This construct is convenient when events have to be synchronized with counts of clock signals.

## Examples:

Table 44 illustrates the philosophy of intra-assignment timing controls by showing the code that could accomplish the same timing effect without using intra-assignment.

Table 44-Intra-assignment timing control equivalence

| Intra-assignment timing control |  |
| :---: | :---: |
| With intra-assignment construct | Without intra-assignment construct |
| $\mathrm{a}=$ \#5 b; | $\begin{aligned} & \text { begin } \\ & \text { temp }=\mathrm{b} ; \\ & \# 5 \mathrm{a}=\text { temp; } \\ & \text { end } \end{aligned}$ |
| $\mathrm{a}=@($ posedge clk) b ; | begin <br> temp = b; <br> @(posedge clk) a = temp; <br> end |
| $a=\underset{@(\text { posedge clk) }}{\operatorname{repeat}} \text {; }$ | ```begin temp = b; @(posedge clk); @(posedge clk); @(posedge clk) a = temp; end``` |

The next three examples use the fork-join behavioral construct. All statements between the keywords fork and join execute concurrently. This construct is described in more detail in 9.8.2.

The following example shows a race condition that could be prevented by using intra-assignment timing control:

```
fork
        #5 a = b;
        #5 b = a;
join
```

The code in this example samples and sets the values of both a and b at the same simulation time, thereby creating a race condition. The intra-assignment form of timing control used in the next example prevents this race condition.

|  | $\mathrm{a}=$ \#5 b; |
| :---: | :---: |
|  | $\mathrm{b}=$ \#5 a ; |
| join |  |

Intra-assignment timing control works because the intra-assignment delay causes the values of $a$ and $b$ to be evaluated before the delay, and the assignments to be made after the delay. Some existing tools that implement intra-assignment timing control use temporary storage in evaluating each expression on the right-hand side.

Intra-assignment waiting for events is also effective. In the following example, the right-hand side expressions are evaluated when the assignment statements are encountered, but the assignments are delayed until the rising edge of the clock signal.

```
fork // data shift
    \(\mathrm{a}=\mathrm{@}(\) posedge clk) b ;
    b = @(posedge clk) c;
join
```

The following is an example of a repeat event control as the intra-assignment delay of a non blocking assignment:

```
a <= repeat(5) @(posedge clk) data;
```

Figure 31 illustrates the activities that result from this repeat event control.


Figure 31-Repeat event control utilizing a clock edge
In this example, the value of data is evaluated when the assignment is encountered. After five occurrences of posedge clk , a is assigned the value of data.

The following is an example of a repeat event control as the intra-assignment delay of a procedural assignment:

```
a = repeat(num) @(clk) data;
```

In this example, the value of data is evaluated when the assignment is encountered. After the number of transitions of clk equals the value of num, a is assigned the value of data.

The following is an example of a repeat event control with expressions containing operations to specify both the number of event occurrences and the event that is counted:
$a<=\operatorname{repeat}(a+b) @(p o s e d g e$ phi1 or negedge phi2) data;
In this example, the value of data is evaluated when the assignment is encountered. After the sum of the positive edges of phil and the negative edges of phi2 equals the sum of $a$ and $b, a$ is assigned the value of data. Even if posedge phi1 and negedge phi2 occurred at the same simulation time, each will be detected separately.

### 9.8 Block statements

The block statements are a means of grouping two or more statements together so that they act syntactically like a single statement. There are two types of blocks in the Verilog HDL:

## Sequential block, also called begin-end block

Parallel block, also called fork-join block
The sequential block shall be delimited by the keywords begin and end. The procedural statements in sequential block shall be executed sequentially in the given order.

The parallel block shall be delimited by the keywords fork and join. The procedural statements in parallel block shall be executed concurrently.

### 9.8.1 Sequential blocks

A sequential block shall have the following characteristics:
Statements shall be executed in sequence, one after another
Delay values for each statement shall be treated relative to the simulation time of the execution of the previous statement
Control shall pass out of the block after the last statement executes
Syntax 9-13 gives the formal syntax for a sequential block.

```
function_seq_block ::= (From Annex \(A\) - A.6.3)
            begin [ : block_identifier
                \{ block_item_declaration \} ] \{ function_statement \} end
seq_block ::=
        begin [ : block_identifier
            \{ block_item_declaration \}] \{ statement \} end
block_item_declaration ::= (From Annex A-A.2.8)
            \{ attribute_instance \} block_reg_declaration
            | \{ attribute_instance \} event_declaration
            | \{ attribute_instance \} integer_declaration
            | \{ attribute_instance \} local_parameter_declaration
            | \{ attribute_instance \} parameter_declaration
            | \{ attribute_instance \} real_declaration
            | \{ attribute_instance \} realtime_declaration
                        | \{ attribute_instance \} time_declaration
```

Syntax 9-13 Syntax for the sequential block

## Examples:

Example 1 A sequential block enables the following two assignments to have a deterministic result:

```
begin
    areg = breg;
    creg = areg; // creg stores the value of breg
end
```

The first assignment is performed and areg is updated before control passes to the second assignment.
Example 2 Delay control can be used in a sequential block to separate the two assignments in time.

```
begin
    areg = breg;
    @(posedge clock) creg = areg; // assignment delayed until
end // posedge on clock
```

Example 3 The following example shows how the combination of the sequential block and delay control can be used to specify a time-sequenced waveform.

```
parameter d = 50; // d declared as a parameter and
reg [7:0] r; // r declared as an 8-bit reg
begin // a waveform controlled by sequential delay
    #d r = 'h35;
    #d r = 'hE2;
    #d r = 'h00;
    #d r = 'hF7;
    #d -> end_wave;//trigger an event called end_wave
end
```


### 9.8.2 arallel blocks

A parallel block shall have the following characteristics:
Statements shall execute concurrently
Delay values for each statement shall be considered relative to the simulation time of entering the block
Delay control can be used to provide time-ordering for assignments
Control shall pass out of the block when the last time-ordered statement executes
Syntax 9-14 gives the formal syntax for a parallel block.

```
par_block ::= (From Annex A - A.6.3)
    fork [ : block_identifier
            {block_item_declaration } ] { statement } join
block_item_declaration ::= (From Annex A - A.2.8)
        { attribute_instance } block_reg_declaration
        | { attribute_instance } event_declaration
        | attribute_instance } integer_declaration
        | { attribute_instance } local_parameter_declaration
        { attribute_instance } parameter_declaration
        | { attribute_instance } real_declaration
        | { attribute_instance } realtime_declaration
        | { attribute_instance } time_declaration
```


## Syntax 9-14 Syntax for the parallel block

The timing controls in a fork-join block do not have to be ordered sequentially in time.

## Example:

The following example codes the waveform description shown in example 3 of 9.8.1 by using a parallel block instead of a sequential block. The waveform produced on the reg is exactly the same for both implementations.

```
fork
    #50 r = 'h35;
    #100 r = 'hE2;
    #150 r = 'h00;
    #200 r = 'hF7;
    #250 -> end_wave;
join
```


### 9.8.3 Block names

Both sequential and parallel blocks can be named by adding : name_of_block after the keywords begin or fork. The naming of blocks serves several purposes:

It allows local variables, parameters, and named events to be declared for the block.
It allows the block to be referenced in statements such as the disable statement (Section 11).
All variables shall be static that is, a unique location exists for all variables and leaving or entering blocks shall not affect the values stored in them.

The block names give a means of uniquely identifying all variables at any simulation time.

### 9.8.4 Start and finish times

Both sequential and procedural blocks have the notion of a start and finish time. For sequential blocks, the start time is when the first statement is executed, and the finish time is when the last statement has been executed. For parallel blocks, the start time is the same for all the statements, and the finish time is when the last time-ordered statement has been executed.

Sequential and parallel blocks can be embedded within each other, allowing complex control structures to be expressed easily and with a high degree of structure. When blocks are embedded within each other, the timing of when a block starts and finishes is important. Execution shall not continue to the statement following a block until the finish time for the block has been reached that is, until the block has completely finished executing.

## Examples.

Example 1 The following example shows the statements from the example in 9.8 .2 written in the reverse order and still producing the same waveform.

```
fork
    #250 -> end_wave;
    #200 r = 'hF7;
    #150 r = 'h00;
    #100 r = 'hE2;
    #50 r = 'h35;
join
```

Example 2 When an assignment is to be made after two separate events have occurred, known as the joining of events, a fork-join block can be useful.

```
begin
    fork
            @Aevent;
            @Bevent;
        join
        areg = breg;
end
```

The two events can occur in any order (or even at the same simulation time) and the fork-join block will complete and the assignment will be made. In contrast to this, if the fork-join block was a begin-end block and the Bevent occurred before the Aevent, then the block would be waiting for the next Bevent.

Example 3 This example shows two sequential blocks, each of which will execute when its controlling event occurs. Because the event controls are within a fork-join block, they execute in parallel and the sequential blocks can therefore also execute in parallel.

```
fork
    @enable_a
        begin
            #ta wa = 0;
            #ta wa = 1;
            #ta wa = 0;
        end
    @enable_b
        begin
            #tb wb = 1;
            #tb wb = 0;
            #tb wb = 1;
        end
join
```


### 9.9 Structured procedures

All procedures in the Verilog HDL are specified within one of the following four statements:
initial construct
always construct
Task
Function

The initial and always constructs are enabled at the beginning of a simulation. The initial construct shall execute only once and its activity shall cease when the statement has finished. In contrast, the always construct shall execute repeatedly. Its activity shall cease only when the simulation is terminated. There shall be no implied order of execution between initial and always constructs. The initial constructs need not be scheduled and executed before the always constructs. There shall be no limit to the number of initial and always constructs that can be defined in a module.

Tasks and functions are procedures that are enabled from one or more places in other procedures. Tasks and functions are described in Section 10.

### 9.9.1 Initial construct

The syntax for the initial construct is given in Syntax 9-15.

```
initial_construct ::= (From Annex A - A.6.2)
    initial statement
```


## Syntax 9-15 Syntax for initial construct

## Examples:

The following example illustrates use of the initial construct for initialization of variables at the start of simulation.

```
initial begin
    areg = 0; // initialize a reg
    for (index = 0; index < size; index = index + 1)
        memory[index] = 0; //initialize memory word
end
```

Another typical usage of the initial construct is specification of waveform descriptions that execute once to provide stimulus to the main part of the circuit being simulated.

```
initial begin
    inputs = 'b000000; //initialize at time zero
    #10 inputs = 'b011001; // first pattern
    #10 inputs = 'b011011; // second pattern
    #10 inputs = 'b011000; // third pattern
    #10 inputs = 'b001000; // last pattern
end
```


### 9.9.2 Always construct

The always construct repeats continuously throughout the duration of the simulation. Syntax 9-16 shows the syntax for the always construct.

$$
\begin{gathered}
\text { always_construct }::=\text { (From Annex } A-A .6 .2) \\
\text { always statement }
\end{gathered}
$$

## Syntax 9-16 Syntax for always construct

The always construct, because of its looping nature, is only useful when used in conjunction with some form of timing control. If an always construct has no control for simulation time to advance, it will create a simulation deadlock condition.

The following code, for example, creates a zero-delay infinite loop.
always areg $=$ ~areg;
Providing a timing control to the above code creates a potentially useful description as shown in the following:
always \#half_period areg = ~areg;

## 10. Tasks and functions

Tasks and functions provide the ability to execute common procedures from several different places in a description. They also provide a means of breaking up large procedures into smaller ones to make it easier to read and debug the source descriptions. This clause discusses the differences between tasks and functions, describes how to define and invoke tasks and functions, and presents examples of each.

### 10.1 Distinctions between tasks and functions

The following rules distinguish tasks from functions:
A function shall execute in one simulation time unit; a task can contain time-controlling statements.
A function cannot enable a task; a task can enable other tasks and functions.
A function shall have at least one input type argument and shall not have an output or inout type argument; a task can have zero or more arguments of any type.
A function shall return a single value; a task shall not return a value.
The purpose of a function is to respond to an input value by returning a single value. A task can support multiple goals and can calculate multiple result values. However, only the output or inout type arguments pass result values back from the invocation of a task. A function is used as an operand in an expression; the value of that operand is the value returned by the function.

## Example:

Either a task or a function can be defined to switch bytes in a 16 -bit word. The task would return the switched word in an output argument, so the source code to enable a task called switch_bytes could look like the following example:

```
switch_bytes (old_word, new_word);
```

The task switch_bytes would take the bytes in old_word, reverse their order, and place the reversed bytes in new_word.

A word-switching function would return the switched word as the return value of the function. Thus, the function call for the function switch_bytes could look like the following example:

```
new_word = switch_bytes (old_word);
```


### 10.2 Tasks and task enabling

A task shall be enabled from a statement that defines the argument values to be passed to the task and the variables that receive the results. Control shall be passed back to the enabling process after the task has completed. Thus, if a task has timing controls inside it, then the time of enabling a task can be different from the time at which the control is returned. A task can enable other tasks, which in turn can enable still other tasks with no limit on the number of tasks enabled. Regardless of how many tasks have been enabled, control shall not return until all enabled tasks have completed.

### 10.2.1 Task declarations

The syntax for defining tasks is given in Syntax 10-1.
task_declaration $::=$ (From Annex A-A.2.7)
task [ automatic ] task_identifier ; \{ task_item_declaration \} statement
endtask
task [ automatic ] task_identifier ( task_port_list ) ; \{block_item_declaration \} statement endtask
task_item_declaration ::=
block_item_declaration
| \{ attribute_instance \} tf_input_declaration ;
| \{ attribute_instance \} tf_output_declaration ;
| \{ attribute_instance \} tf_inout_declaration ;
task_port_list ::=
task_port_item \{ , task_port_item \}
task_port_item ::=
\{ attribute_instance \} tf_input_declaration
| \{ attribute_instance \} tf_output_declaration
| \{ attribute_instance \} tf_inout_declaration
tf_input_declaration ::=
input [ reg ] [ signed ] [ range ] list_of_port_identifiers
| input [ task_port_type ] list_of_port_identifiers
tf_output_declaration ::=
output [ reg ] [ signed ] [ range ] list_of_port_identifiers
output [ task_port_type ] list_of_port_identifiers
tf_inout_declaration ::=
inout [ reg ] [ signed ] [ range ] list_of_port_identifiers
| inout [ task_port_type ] list_of_port_identifiers
task_port_type ::=
time $\mid$ real | realtime $\mid$ integer
block_item_declaration $::=$ (From Annex A-A.2.8)
\{ attribute_instance \} block_reg_declaration
| \{ attribute_instance \} event_declaration
| \{ attribute_instance \} integer_declaration
| \{ attribute_instance \} local_parameter_declaration
| \{ attribute_instance \} parameter_declaration
| \{ attribute_instance \} real_declaration
| \{ attribute_instance \} realtime_declaration
| \{ attribute_instance \} time_declaration
block_reg_declaration ::=
reg [ signed ] [ range ]
list_of_block_variable_identifiers ;
list_of_block_variable_identifiers ::=
block_variable_type \{ , block_variable_type \}
block_variable_type ::=
variable_identifier
| variable_identifier dimension \{dimension \}

There are two alternate task declaration syntaxes.
The first syntax shall begin with the keyword task, followed by the optional keyword automatic, followed by a name for the task and a semicolon, and ending with the keyword endtask. The keyword automatic declares an automatic task that is reentrant with all the task declarations allocated dynamically for each concurrent task entry. Task item declarations can specify the following:

```
Input arguments
Output arguments
Inout arguments
All data types that can be declared in a procedural block
```

The second syntax shall begin with the keyword task, followed by a name for the task and a parenthesis enclosed task_port_list. The task_port_list shall consist of zero or more comma separated task_port_items. There shall be a semicolon after the close parenthesis. The task body shall follow and then the keyword endtask.

In both syntaxes, the port declarations shall have the same syntax as defined by the tf_input_declaration, $t f$ _output_declaration and tf_inout_declaration, as detailed in Syntax 10-1 above.

Tasks without the optional keyword automatic are static tasks, with all declared items being statically allocated. These items shall be shared across all uses of the task executing concurrently. Task with the optional keyword automatic are automatic tasks. All items declared inside automatic tasks are allocated dynamically for each invocation. Automatic task items can not be accessed by hierarchical references. Automatic tasks can be invoked through use of their hierarchical name.

### 10.2.2 Task enabling and argument passing

The task enabling statement shall pass arguments as a comma-separated list of expressions enclosed in parentheses. The formal syntax of the task enabling statement is given in Syntax 10-2.

```
task_enable ::= (From Annex A - A.6.9)
    hierarchical_task_identifier [ ( expression { , expression } )];
```


## Syntax 10-2 Syntax of the task enabling statement

The list of arguments for a task enabling statement shall be optional. If the list of arguments is provided, the list shall be an ordered list of expressions that has to match the order of the list of arguments in the task definition.

If an argument in the task is declared as an input, then the corresponding expression can be any expression. The order of evaluation of the expressions in the argument list is undefined. If the argument is declared as an output or an inout, then the expression shall be restricted to an expression that is valid on the left-hand side of a procedural assignment (see 9.2). The following items satisfy this requirement:
reg, integer, real, realtime, and time variables
Memory references
Concatenations of reg, integer, real, realtime and time variables
Concatenations of memory references
Bit-selects and part-selects of reg, integer, and time variables

The execution of the task enabling statement shall pass input values from the expressions listed in the enabling statement to the arguments specified within the task. Execution of the return from the task shall pass values from the task output and inout type arguments to the corresponding variables in the task enabling statement. All arguments to the task shall be passed by value rather than by reference (that is, a pointer to the value).

## Examples:

Example 1 The following example illustrates the basic structure of a task definition with five arguments.

```
task my_task;
input \(a, b ;\)
inout c ;
output d, e;
begin
    . . . // statements that perform the work of the task
    - • •
    c = fool; // the assignments that initialize result regs
    d = foo2;
    e = foo3;
end
endtask
```

Or using the second form of a task declaration, the task could be defined as:

```
task my_task (input a, b, inout c, output d, e);
begin
    . . . // statements that perform the work of the task
    • • -
    c = fool; // the assignments that initialize result regs
    d = foo2;
    e = foo3;
end
endtask
```

The following statement enables the task:

```
my_task (v, w, x, y, z);
```

The task enabling arguments ( $\mathrm{v}, \mathrm{w}, \mathrm{x}, \mathrm{y}$, and z ) correspond to the $\operatorname{arguments}(\mathrm{a}, \mathrm{b}, \mathrm{c}, \mathrm{d}$, and e ) defined by the task. At task enabling time, the input and inout type arguments ( $\mathrm{a}, \mathrm{b}, \mathrm{and} \mathrm{c}$ ) receive the values passed in $v, w$, and $x$. Thus, execution of the task enabling call effectively causes the following assignments:

$$
\begin{aligned}
& \mathrm{a}=\mathrm{v} ; \\
& \mathrm{b}=\mathrm{w} ; \\
& \mathrm{c}=\mathrm{x} ;
\end{aligned}
$$

As part of the processing of the task, the task definition for my_task shall place the computed result values into $c$, $d$, and $e$. When the task completes, the following assignments to return the computed values to the calling process are performed:

$$
\begin{aligned}
& x=c ; \\
& y=d ; \\
& z=e ;
\end{aligned}
$$

Example 2 The following example illustrates the use of tasks by describing a traffic light sequencer:

```
module traffic_lights;
reg clock, red, amber, green;
parameter on = 1, off = 0, red_tics = 350,
            amber_tics = 30, green_tics = 200;
// initialize colors.
initial red = off;
initial amber = off;
initial green = off;
always begin // sequence to control the lights.
    red = on; // turn red light on
    light(red, red_tics); // and wait.
    green = on; // turn green light on
    light(green, green_tics); // and wait.
    amber = on; // turn amber light on
    light(amber, amber_tics); // and wait.
end
// task to wait for 'tics' positive edge clocks
// before turning 'color' light off.
task light;
output color;
input [31:0] tics;
begin
    repeat (tics)@ (posedge clock);
    color = off; // turn light off.
end
endtask
always begin // waveform for the clock.
    #100 clock = 0;
    #100 clock = 1;
end
endmodule // traffic_lights.
```


### 10.2.3 Task memory usage and concurrent activation

A task may be enabled more than once concurrently. All variables of an automatic task shall be replicated on each concurrent task invocation to store state specific to that invocation. All variables of a static task shall be static in that there shall be a single variable corresponding to each declared local variable in a module instance, regardless of the number of concurrent activations of the task. However, static tasks in different instances of a module shall have separate storage from each other.

Variables declared in static tasks shall retain their values between invocations. They shall be initialized to the default initialization value as described in 3.2.2. Variables declared in automatic tasks shall be initialized to the default initialization value whenever execution enters their scope.

Because variables declared in automatic tasks are deallocated at the end of the task invocation, they shall not be used in certain constructs that might refer to them after that point.

They shall not be assigned values using non blocking assignments or procedural continuous assignments.
They shall not be referenced by procedural continuous assignments or procedural force statements.

They shall not be referenced in intra-assignment event controls of non blocking assignments.
They shall not be traced with system tasks such as \$monitor and \$dumpvars.

### 10.3 Functions and function calling

The purpose of a function is to return a value that is to be used in an expression. The rest of this clause explains how to define and use functions.

### 10.3.1 Function declarations

The syntax for defining a function is given in Syntax 10-3.

```
function_declaration \(::=(\) From Annex A - A.2.6)
            function [ automatic ] [ signed ] [ range_or_type ]
                function_identifier ;
                function_item_declaration \{ function_item_declaration \}
                function_statement
            endfunction
            | function [ automatic ] [ signed ] [ range_or_type ]
                function_identifier ( function_port_list ) ;
                block_item_declaration \{ block_item_declaration \}
                function_statement
        endfunction
function_item_declaration ::=
        block_item_declaration
            | tf_input_declaration ;
function_port_list ::=
        \{ attribute_instance \} tf_input_declaration
                \{, \{ attribute_instance \}tf_input_declaration \}
tf_input_declaration ::=
        input [ reg ] [ signed ] [ range ] list_of_port_identifiers
        | input [ task_port_type ] list_of_port_identifiers
range_or_type ::=
    range \(\mid\) integer \(\mid\) real \(\mid\) realtime \(\mid\) time
block_item_declaration ::= (From Annex A-A.2.8)
        \{ attribute_instance \} block_reg_declaration
        | \{ attribute_instance \} event_declaration
        | \{ attribute_instance \} integer_declaration
        | \{ attribute_instance \} local_parameter_declaration
        | \{ attribute_instance \} parameter_declaration
        | \{ attribute_instance \} real_declaration
        | \{ attribute_instance \} realtime_declaration
        | \{ attribute_instance \} time_declaration
block_reg_declaration ::=
            reg [ signed ] [ range ]
                list_of_block_variable_identifiers ;
list_of_block_variable_identifiers ::=
        block_variable_type \{ , block_variable_type \}
block_variable_type ::=
        variable_identifier
        variable_identifier dimension \{ dimension \}
```


## Syntax 10-3 Syntax for function declaration

A function definition shall begin with the keyword function, followed by the optional keyword automatic, followed by the optional signed designator, followed by the range or type of the return value from the function, followed by the name of the function, and then either a semicolon, or a function port list enclosed in parenthesis, and then a semicolon, and then shall end with the keyword endfunction. The use of a range_or_type shall be optional. A function specified without a range or type defaults to a one bit reg for the return value. If used, range_or_type shall specify the return value of the function is a real, an integer, a time, a realtime or a value with a range of [ $n: m$ ] bits. A function shall have at least one input declared.

The keyword automatic declares a recursive function with all the function declarations allocated dynamically for each recursive call. Automatic function items can not be accessed by hierarchical references. Automatic functions can be invoked through the use of their hierarchical name.

Function inputs shall be declared one of two ways. The first method shall have the name of the function followed by a semicolon. After the semicolon one or more input declarations optionally mixed with block item declarations shall follow. After the function item declarations there shall be a behavioral statement and then the endfunction keyword.

The second method shall have the name of the function, followed by an open parenthesis, and one or more input declarations, separated by commas. After all the input declarations, there shall be a close parenthesis, and a semicolon. After the semicolon, there shall be zero or more block item declarations, followed by a behavioral statement, and then the endfunction keyword.

## Example:

The following example defines a function called getbyte, using a range specification.

```
function [7:0] getbyte;
input [15:0] address;
begin
    // code to extract low-order byte from addressed word
    • • •
    getbyte = result_expression;
end
endfunction
```

Or using the second form of a function declaration, the function could be defined as:

```
function [7:0] getbyte (input [15:0] address);
begin
    // code to extract low-order byte from addressed word
    • • •
    getbyte = result_expression;
end
endfunction
```


### 10.3.2 eturning a value from a function

The function definition shall implicitly declare a variable, internal to the function, with the same name as the function. This variable either defaults to a 1-bit reg or is the same type as the type specified in the function declaration. The function definition initializes the return value from the function by assigning the function result to the internal variable with the same name as the function.

It is illegal to declare another object with the same name as the function in the scope where the function is declared. Inside a function, there is an implied variable with the name of the function, which may be used in expressions within the function. It is, therefore, also illegal to declare another object with the same name as the function inside the function scope.

The following line from the example in 10.3.1 illustrates this concept:

```
getbyte = result_expression;
```


### 10.3.3 Calling a function

A function call is an operand within an expression. The function call has the syntax given in Syntax 10-4.

```
function_call ::= (From Annex A - A.8.2)
    hierarchical_function_identifier{ attribute_instance } (expression {, expression })
    Syntax 10-4 Syntax for function call
```

The order of evaluation of the arguments to a function call is undefined.

## Example:

The following example creates a word by concatenating the results of two calls to the function getbyte (defined in 10.3.1):

```
word = control ? {getbyte(msbyte), getbyte(lsbyte)}:0;
```


### 10.3.4 Function rules

Functions are more limited than tasks. The following six rules govern their usage:
a) A function definition shall not contain any time-controlled statements that is, any statements introduced with \#, @, or wait.
b) Functions shall not enable tasks.
c) A function definition shall contain at least one input argument.
d) A function definition shall not have any argument declared as output or inout.
e) A function definition shall include an assignment of the function result value to the internal variable that has the same name as the function name.
f) A function shall not have any non blocking assignments.

## Example:

This example defines a function called factorial that returns an integer value. The factorial function is called iteratively and the results are printed.

```
module tryfact;
// define the function
function automatic integer factorial;
input [ 31:0] operand;
integer i;
if (operand >= 2)
    factorial = factorial (operand - 1) * operand;
else
    factorial = 1;
endfunction
// test the function
integer result;
integer n;
initial begin
    for ( }\textrm{n}=0;\textrm{n}<=7;\textrm{n}=\textrm{n}+1)\mathrm{ begin
        result = factorial(n);
        $display("%0d factorial=%0d", n, result);
    end
end
endmodule // tryfact
```

The simulation results are as follows:

```
0 factorial=1
1 factorial=1
2 factorial=2
3 factorial=6
4 factorial=24
5 factorial=120
factorial=720
7 factorial=5040
```


### 10.3.5 Use of constant functions

Constant function calls are used to support the building of complex calculations of values at elaboration time (see 12.1.3). A constant function call shall be a function invocation of a constant function local to the calling module where the arguments to the function are constant expressions. Constant functions are a subset of normal Verilog functions that shall meet the following constraints:

They shall contain no hierarchical references.
Any function invoked within a constant function shall be a constant function local to the current module. System functions shall not be invoked.
All system tasks within a constant function shall be ignored.
All system functions within a constant function shall be illegal.
The only system task that may be invoked is \$display, and it shall be ignored when invoked at elaboration time.

All parameter values used within the function shall be defined before the use of the invoking constant function call (i.e. any parameter use in the evaluation of a constant function call constitutes a use of that parameter at the site of the original constant function call).
All identifiers which are not parameters or functions shall be declared locally to the current function.
If they use any parameter value that is affected directly or indirectly by a defparam statement (see 12.2.1), the result is undefined. This can produce an error or the constant function can return an indeterminate value.
They shall not be declared inside a generate scope.
They shall not themselves use constant functions in any context requiring a constant expression.
Constant function calls are evaluated at elaboration time. Their execution has no effect on the initial values of the variables used either at simulation time or among multiple invocations of a function at elaboration time. In each of these cases, the variables are initialized as they would be for normal simulation.

## Example:

This example defines a function called clogb2 that returns an integer which has the value of the ceiling of the $\log$ base 2 .

```
module ram_model (address, write, chip_select, data);
    parameter data_width = 8;
    parameter ram_depth = 256;
    localparam adder_width = clogb2(ram_depth);
    input [adder_width - 1:0] address;
    input write, chip_select;
    inout [data_width - 1:0] data;
    //define the clogb2 function
    function integer clogb2;
        input depth;
        integer i,result;
        begin
            for (i = 0; 2 ** \(i<d e p t h ; i=i+1)\)
                    result = i + 1;
            clogb2 = result;
        end
    endfunction
```

    reg [data_width - 1:0] data_store[0:ram_depth - 1];
    //the rest to the ram model
    An instance of this ram_model with parameters assigned:

```
ram_model #(32,421) ram_a0(a_addr,a_wr,a_cs,a_data);
```


## 11. Disabling of named blocks and tasks

The disable statement provides the ability to terminate the activity associated with concurrently active procedures, while maintaining the structured nature of Verilog HDL procedural descriptions. The disable statement gives a mechanism for terminating a task before it executes all its statements, breaking from a looping statement, or skipping statements in order to continue with another iteration of a looping statement. It is useful for handling exception conditions such as hardware interrupts and global resets.

The disable statement has the syntax form shown in Syntax 11-1.

```
disable_statement ::= (From Annex A - A.6.5)
    disable hierarchical_task_identifier ;
    | disable hierarchical_block_identifier ;
```


## Syntax 11-1 Syntax of disable statement

Either form of disable statement shall terminate the activity of a task or a named block. Execution shall resume at the statement following the block or following the task enabling statement. All activities enabled within the named block or task shall be terminated as well. If task enable statements are nested that is, one task enables another, and that one enables yet another then disabling a task within the chain shall disable all tasks downward on the chain. If a task is enabled more than once, then disabling such a task shall disable all activations of the task.

The results of the following activities that may be initiated by a task are not specified if the task is disabled:

> Results of output and inout arguments
> Scheduled, but not executed, non blocking assignments
> Procedural continuous assignments ( assign and force statements)

The disable statement can be used within blocks and tasks to disable the particular block or task containing the disable statement. The disable statement can be used to disable named blocks within a function, but cannot be used to disable functions. In cases where a disable statement within a function disables a block or a task that called the function, the behavior is undefined. Disabling an automatic task or a block inside an automatic task proceeds as for regular tasks for all concurrent executions of the task.

## Examples:

Example 1 This example illustrates how a block disables itself.

```
begin : block_name
    rega = regb;
    disable block_name;
    regc = rega; // this assignment will never execute
end
```

Example 2 This example shows the disable statement being used within a named block in a manner similar to a forward goto. The next statement executed after the disable statement is the one following the named block.

```
begin : block_name
    if (a == 0)
        disable block_name;
    . . .
end // end of named block
// continue with code following named block
```

Example 3 This example shows the disable statement being used as an early return from a task. However, a task disabling itself using a disable statement is not a short-hand for the return statement found in programming languages.

```
task proc_a;
begin
    ...
    ...
    if (a == 0)
        disable proc_a; // return if true
    ..
end
endtask
```

Example 4 This example shows the disable statement being used in an equivalent way to the two statements continue and break in the C programming language. The example illustrates control code that would allow a named block to execute until a loop counter reaches $n$ iterations or until the variable a is set to the value of b . The named block break contains the code that executes until $\mathrm{a}==\mathrm{b}$, at which point the disable break; statement terminates execution of that block. The named block continue contains the code that executes for each iteration of the for loop. Each time this code executes the disable continue; statement, the continue block terminates and execution passes to the next iteration of the for loop. For each iteration of the cont inue block, a set of statements executes if (a $!=0$ ). Another set of statements executes if $(\mathrm{a}!=\mathrm{b})$.

```
begin : break
    for \((i=0\); \(i<n\); \(i=i+1)\) begin : continue
        @clk
            if ( \(\mathrm{a}==0\) ) // "continue" loop
                        disable continue;
            statements
            statements
        @clk
            if ( \(\mathrm{a}==\mathrm{b}\) ) // "break" from loop
                    disable break;
            statements
            statements
        end
end
```

Example 5 This example shows the disable statement being used to disable concurrently a sequence of timing controls and the task action, when the reset event occurs. The example shows a fork/join block within which is a named sequential block (event_expr) and a disable statement that waits for occurrence of the event reset. The sequential block and the wait for reset execute in parallel. The event_expr block waits for one occurrence of event ev1 and three occurrences of event trig. When these four events have happened, plus a delay of $d$ time units, the task action executes. When the event reset occurs, regardless of events within the sequential block, the fork/join block terminates including the task action.

```
fork
    begin : event_expr
        @ev1;
        repeat(3)@trig;
        #d action (areg, breg);
    end
    @reset disable event_expr;
join
```

Example 6 The next example is a behavioral description of a retriggerable monostable. The named event retrig restarts the monostable time period. If retrig continues to occur within 250 time units, then $q$ will remain at 1 .

```
always begin : monostable
    #250 q = 0;
end
always @retrig begin
    disable monostable;
    q = 1;
end
```


## 12. Hierarchical structures

The Verilog HDL supports a hierarchical hardware description structure by allowing modules to be embedded within other modules. Higher-level modules create instances of lower-level modules and communicate with them through input, output, and bidirectional ports. These module input/output ports can be scalar or vector.

As an example of a module hierarchy, consider a system consisting of printed circuit boards (PCBs). The system would be represented as the top-level module and would create instances of modules that represent the boards. The board modules would, in turn, create instances of modules that represent ICs, and the ICs could, in turn, create instances of modules such as flip-flops, mux s, and alu s.

To describe a hierarchy of modules, the user provides textual definitions of the various modules. Each module definition stands alone; the definitions are not nested. Statements within the module definitions create instances of other modules, thus describing the hierarchy.

### 12.1 Modules

This clause gives the formal syntax for a module definition and then gives the syntax for module instantiation, along with an example of a module definition and a module instantiation.

A module definition shall be enclosed between the keywords module and endmodule. The identifier following the keyword module shall be the name of the module being defined. The optional list of parameter definitions shall specify an ordered list of the parameters for the module. The optional list of ports or port declarations shall specify an ordered list of the ports for the module. The order used used in defining the list of parameters in the module_parameter_port_list and in the list of ports can be significant when instantiating the module (see 12.2.2.1 and 12.3.5). The identifiers in this list shall be declared in input, output, and inout statements within the module definition. Ports declared in the list of port declarations shall not be redeclared within the body of the module. The module items define what constitutes a module and they include many different types of declarations and definitions; many of which have already been introduced.

The keyword macromodule can be used interchangeably with the keyword module to define a module. An implementation can choose to treat module definitions beginning with macromodule keyword differently.

```
module_declaration \(::=(\) From Annex \(A-\) A.1.3)
    \{ attribute_instance \} module_keyword module_identifier [ module_parameter_port_list ]
        [ list_of_ports ] ; \{ module_item \}
        endmodule
    | \{ attribute_instance \} module_keyword module_identifier [ module_parameter_port_list ]
        [ list_of_port_declarations ] ; \{non_port_module_item \}
        endmodule
module_keyword \(::=\) module \(\mid\) macromodule
module_parameter_port_list ::= (From Annex A -A.1.4
    \# ( parameter_declaration \{ , parameter_declaration \} )
list_of_ports ::=( port \{, port \})
list_of_port_declarations ::=( port_declaration \{, port_declaration \})|()
port \(::=\) [ port_expression ] | . port_identifier ( [ port_expression ] )
port_expression ::= port_reference | \{ port_reference \{, port_reference \} \}
port_reference \(::=\) port_identifier \(\mid\) port_identifier [ constant_expression ]
    | port_identifier [ range_expression ]
port_declaration \(::=\{\) attribute_instance \(\}\) inout_declaration
    \{attribute_instance\} input_declaration
    \{attribute_instance\} output_declaration
module_item \(::=\) module_or_generate_item (From Annex \(A\) - A.1.5)
    port_declaration ;
    | \{ attribute_instance \} generated_instantiation
    \{ attribute_instance \} local_parameter_declaration
    \{ attribute_instance \} parameter_declaration
    \{ attribute_instance \} specify_block
    \{ attribute_instance \} specparam_declaration
module_or_generate_item \(::=\{\) attribute_instance \(\}\) module_or_generate_item_declaration
    | \{ attribute_instance \} parameter_override
    | \{ attribute_instance \} continuous_assign
    \{ attribute_instance \} gate_instantiation
    \{ attribute_instance \} udp_instantiation
    \{ attribute_instance \} module_instantiation
    | \{ attribute_instance \} initial_construct
    | \{ attribute_instance \} always_construct
module_or_generate_item_declaration ::= net_declaration
    reg_declaration
    integer_declaration
    real_declaration
    time_declaration
    realtime_declaration
    event_declaration
    genvar_declaration
    task_declaration
    function_declaration
non_port_module_item \(::=\{\) attribute_instance \(\}\) generated_instantiation
    | \{ attribute_instance \} local_parameter_declaration
    | \{ attribute_instance \} module_or_generate_item
    | \{ attribute_instance \} parameter_declaration
    | \{ attribute_instance \} specify_block
    | \{ attribute_instance \} specparam_declaration
parameter_override ::= defparam list_of_param_assignments ;
```

See 12.3 for the definitions of ports.

### 12.1.1 Top-level modules

Top-level modules are modules that are included in the source text but are not instantiated, as described in 12.1.2.

### 12.1.2 Module instantiation

Instantiation allows one module to incorporate a copy of another module into itself. Module definitions do not nest. That is, one module definition shall not contain the text of another module definition within its module-endmodule keyword pair. A module definition nests another module by instantiating it. The module instantiation statement creates one or more named instances of a defined module.

For example, a counter module might instantiate a D flip-flop module to create multiple instances of the flipflop.

Syntax 12-2 gives the syntax for specifying instantiations of modules.

```
module_instantiation ::= (From Annex A - A.4.1)
    module_identifier [ parameter_value_assignment ]
        module_instance {, module_instance };
parameter_value_assignment ::=
    # ( list_of_parameter_assignments )
list_of_parameter_assignments ::=
            ordered_parameter_assignment { , ordered_parameter_assignment }
        | named_parameter_assignment {, named_parameter_assignment }
ordered_parameter_assignment ::=
    expression
named_parameter_assignment ::=
    . parameter_identifier ( [ expression ] )
module_instance ::=
    name_of_instance ([ list_of_port_connections ])
name_of_instance ::=
        module_instance_identifier [ range ]
list_of_port_connections ::=
            ordered_port_connection { , ordered_port_connection }
        | named_port_connection {, named_port_connection }
ordered_port_connection ::=
    { attribute_instance } [ expression ]
named_port_connection ::=
    { attribute_instance } .port_identifier ([ expression ] )
```


## Syntax 12-2 Syntax for module instantiation

The instantiations of modules can contain a range specification. This allows an array of instances to be created. The array of instances are described in 7.1. The syntax and semantics of arrays of instances defined for gates and primitives apply for modules as well.

One or more module instances (identical copies of a module) can be specified in a single module instantiation statement.

The list of port connections shall be provided only for modules defined with ports. The parentheses, however, are always required. When a list of port connections is given using the ordered port connection method, the first element in the list shall connect to the first port declared in the module, the second to the second port, and so on. See 12.3 for a more detailed discussion of ports and port connection rules.

A connection can be a simple reference to a variable or a net identifier, an expression, or a blank. An expression can be used for supplying a value to a module input port. A blank port connection shall represent the situation where the port is not to be connected.

When connecting ports by name, an unconnected port can be indicated either by omitting it in the port list, or by providing no expression in the parentheses [i.e., . port_name ()].

## Examples:

Example 1 The following example illustrates a circuit (the lower-level module) being driven by a simple waveform description (the higher-level module) where the circuit module is instantiated inside the waveform module.

```
// Lower level module:
// module description of a nand flip-flop circuit
module ffnand (q, qbar, preset, clear);
output q, qbar; //declares 2 circuit output nets
input preset, clear; //declares 2 circuit input nets
// declaration of two nand gates and their interconnections
nand g1 (q, qbar, preset),
    g2 (qbar, q, clear);
endmodule
// Higher-level module:
// a waveform description for the nand flip-flop
module ffnand wave;
wire out1, out2; //outputs from the circuit
reg in1, in2; //variables to drive the circuit
parameter d = 10;
// instantiate the circuit ffnand, name it "ff",
// and specify the IO port interconnections
ffnand ff(out1, out2, in1, in2);
// define the waveform to stimulate the circuit
initial begin
    #d in1 = 0; in2 = 1;
    #d in1 = 1;
    #d in2 = 0;
    #d in2 = 1;
end
endmodule
```

Example 2 The following example creates two instances of the flip-flop module ffnand defined in example 1 . It connects only to the $q$ output in one instance and only to the qbar output in the other instance.

```
// a waveform description for testing
// the nand flip-flop, without the output ports
module ffnand_wave;
reg in1, in2; //variables to drive the circuit
parameter d = 10;
// make two copies of the circuit ffnand
// ff1 has qbar unconnected, ff2 has q unconnected
ffnand ff1(out1, , in1, in2),
    ff2(.qbar(out2), .clear(in2), .preset(in1), .q());
    // ff3(.q(out3),.clear(in1),,,); is illegal
// define the waveform to stimulate the circuit
initial begin
    #d in1 = 0; in2 = 1;
    #d in1 = 1;
    #d in2 = 0;
    #d in2 = 1;
end
endmodule
```


### 12.1.3 Generated instantiation

After a Verilog design has been parsed, but before simulation begins, the design must have the modules being instantiated linked to the modules being defined, the parameters propagated among the various modules, and hierarchical references resolved. This phase in understanding a Verilog description is termed elaboration.

Generate instantiations are resolved during elaboration because that is when the parameters associated with a module become defined, hence, allowing the definition of the generated statements and declarations. Genvars are only defined during the evaluation of the generate instantiations and do not exist during simulation of a design.

Generate statements facilitate the creation of parameterized models. When used with constant functions (see 10.3.5), parameters can be used to constrain other parameter(s) or localparam(s) in a generated design.

All generate instantiations are coded within a module scope and require the keywords generate - endgenerate.

Generate statements allow control over the declaration of variables, functions and tasks, as well as control over instantiations. Generated instantiations are one or more: modules, user defined primitives, Verilog gate primitives, continuous assignments, initial blocks and always blocks. Generated declarations and instantiations can be conditionally instantiated into a design. Generated variable declarations and instantiations can be multiply instantiated into a design. Generated instances have unique identifier names and can be referenced hierarchically as described in 12.4.

To support the interconnection between structural elements and/or procedural blocks, generate statements permit the following Verilog data types to be declared within the generate scope: net, reg, integer, real, time, realtime, and event. Generated data types have unique identifier names and can be referenced hierarchically as described in 12.4 .

Parameter redefinition using by the ordered or named parameter = value assignment or defparam statements can also be declared within the generate scope. However, a defparam statement within the generate scope or within a hierarchy instantiated within the generate scope shall only modify the value of a parameter declared within the generate scope or within a hierarchy instantiated within the generate scope.

Tasks and functions declarations shall also be permitted within the generate scope, however not in a generate loop. Generated tasks and functions shall have unique identifier names and may be referenced hierarchically as described in 12.4.

Module declarations and module items that shall not be permitted in a generate statement include: parameters, local parameters, input declarations, output declarations, inout declarations and specify blocks.

Connections to generated module instances are handled the same way as they are handled with normal module instances as described in 12.1.2.

Generated statements are created using one of the following three methods: generate-loop, generate-conditional, or generate-case.

The syntax for generate instantiations is given in Syntax 12-3.

```
module_item ::= (From Annex A-A.1.5)
    module_or_generate_item
    port_declaration ;
    { attribute_instance } generated_instantiation
    | { attribute_instance } local_parameter_declaration
    { attribute_instance } parameter_declaration
    { { attribute_instance } specify_block
    | { atrribute_instance } specparam_declaration
module_or_generate_item ::=
        { attribute_instance } module_or_generate_item_declaration
    | {attribute_instance } parameter_override
    { attribute_instance } continuous_assign
    { attribute_instance } gate_instantiation
    | { atribute_instance } udp_instantiation
    { attribute_instance } module_instantiation
    { {attribute_instance } initial_construct
    { { attribute_instance } always_construct
module_or_generate_item_declaration ::=
        net_declaration
        | reg_declaration
        integer_declaration
        real_declaration
        time_declaration
        realtime_declaration
        event_declaration
        genvar_declaration
        task_declaration
        | function_declaration
generated_instantiation ::= (From Annex A -A.4.2)
    generate { generate_item } endgenerate
generate_item_or_null ::=
    generat__item |;
generate_item ::=
        generate_conditional_statement
        generate_case_statement
        generate_loop_statement
        generate_block
        | module_or_generate_item
generate_conditional_statement ::=
    if ( constant_expression ) generate_item_or_null [ else generate_item_or_null ]
generate_case_statement ::= case ( constant_expression)
        genvar_case_item { genvar_case_item } endcase
genvar_case_item ::= constant_expression {, constant_expression }:
        generate_item_or_null | default [:] generate_item_or_null
generate_loop_statement ::=
        for (genvar_assignment ; constant_expression ; genvar_assignment )
            begin : generate_block_identifier { generate_item } end
genvar_assignment ::=
        genvar_identifier = constant_expression
generate_block ::=
        begin [ : generate_block_identifier ] { generate_item } end
```


### 12.1.3.1 genvar - generate statement index variable

An index variable that shall only be declared for use in generate statements shall be declared as a genvar and is referred to as a genvar in the rest of this section.

The syntax for generate statement index variable declarations is given in Syntax 12-4.

```
genvar_declaration ::= (From Annex A - A.2.1.3)
    genvar list_of_genvar_identifiers ;
list_of_genvar_identifiers ::= (From Annex A - A.2.3)
    genvar_identifier {, genvar_identifier }
```

Syntax 12-4 Syntax for generate statement index variable declaration

A genvar shall be declared within the module where the genvar is used. A genvar can be declared either inside or outside of a generate scope. A genvar is an integer that is local to and shall only be used within a generate loop that uses it as an index variable. If any bit of the genvar ever is set to an X or Z or if the genvar is set to a negative value, this shall be an error.

Genvars are only defined during the evaluation of the generate blocks (see 12.1.3), and do not exist during simulation of a Verilog design.

The value of a genvar shall only be defined by a generate loop. Two generate loops using the same genvar as an index variable shall not be nested. The value of a genvar can be referenced in any context where the value of a parameter could be referenced.

### 12.1.3.2 generate-loop

A generate-loop permits one or more variable declarations, modules, user defined primitives, gate primitives, continuous assignments, initial blocks and always blocks to be instantiated multiple times using a forloop. The index loop variable used in a generate for-loop shall be declared as a genvar. Both genvar assignments in the for-loop shall assign to the same genvar, which is the loop index variable. The first genvar assignment in the for-loop shall not reference the loop index variable on the right hand side.

## Examples.

Example 1 A parameterized gray-code to binary-code converter module using a loop to generate continuous assignments

```
module gray2bin1 (bin, gray);
    parameter SIZE = 8; // this module is parameterizable
    output [SIZE-1:0] bin;
    input [SIZE-1:0] gray;
    genvar i;
    generate for (i=0; i<SIZE; i=i+1) begin:bit
        assign bin[ i] = ^gray[SIZE-1:i] ;
    end endgenerate
endmodule
```

Example 2 The same gray-code to binary-code converter module in example 1 is built using a loop to generate always blocks

```
module gray2bin2 (bin, gray);
    parameter SIZE = 8; // this module is parameterizable
    output [SIZE-1:0] bin;
    input [SIZE-1:0] gray;
    reg [SIZE-1:0] bin;
    genvar i;
    generate for (i=0; i<SIZE; i=i+1) begin:bit
            always @(gray[SIZE-1:i]) // fixed part select
                bin[ i] = ^gray[ SIZE-1:i] ;
        end endgenerate
    endmodule
```

The models in examples 3 and 4 are parameterized modules of ripple adders using a loop to generate Verilog gate primitives. Example 3 uses a two dimensional net declaration outside of the generate loop to make the connections between the gate primitives while example 4 makes the net declaration inside of the generate loop to generate the wires needed to connect the gate primitives for each iteration of the loop.

Example 3 Generated ripple adder with two-dimensional net declaration outside of the generate loop

```
module addergen1 (co, sum, a, b, ci);
    parameter SIZE = 4;
    output [ SIZE-1:0] sum;
    output co;
    input [SIZE-1:0] a, b;
    input Ci;
    wire [SIZE :0] c;
    wire [SIZE-1:0] t [ 1:3];
    genvar i;
    assign C[ 0] = Ci;
    // Generated instance names are:
    // xor gates: bit[ 0] .g1 bit[ 1] .g1 bit[ 2] .g1 bit[ 3] .g1
    // bit[ 0] .g2 bit[ 1] .g2 bit[ 2] .g2 bit[ 3] .g2
    // and gates: bit[ 0] .g3 bit[ 1] .g3 bit[ 2] .g3 bit[ 3] .g3
    // bit[ 0] .g4 bit[ 1] .g4 bit[ 2] .g4 bit[ 3] .g4
    // or gates: bit[ 0] .g5 bit[ 1] .g5 bit[ 2] .g5 bit[ 3] .g5
    // Generated instances are connected with
    // multi-dimensional nets t[ 1][ 3:0] t[ 2][ 3:0] t[ 3][ 3:0]
    // (12 multi-dimensional nets total)
    generate
        for(i=0; i<SIZE; i=i+1) begin:bit
            xor g1 ( t[ 1][ i], a[ i], b[ i] );
            xor g2 ( sum[ i], t[ 1][ i], c[ i] );
            and g3 ( t[ 2][ i], a[ i], b[ i] );
            and g4 ( t[ 3][ i], t[ 1][ i], c[ i] );
            or g5 ( c[ i+1], t[ 2][ i], t[ 3][ i] );
        end
    endgenerate
```

    assign \(\mathrm{CO}=\mathrm{C}[\) SIZE];
    endmodule

Example 4 Generated ripple adder with net declaration inside of the generate loop

```
module addergen1 (co, sum, a, b, ci);
    parameter SIZE = 4;
    output [SIZE-1:0] sum;
    output CO;
    input [SIZE-1:0] a, b;
    input Ci;
    wire [SIZE :0] c;
    genvar i;
    assign C[ 0] = ci;
    // Generated instance names are:
    // xor gates: bit[ 0] .g1 bit[ 1] .g1 bit[ 2] .g1 bit[ 3] .g1
    // bit[ 0] .g2 bit[ 1] .g2 bit[ 2] .g2 bit[ 3] .g2
    // and gates: bit[ 0] .g3 bit[ 1] .g3 bit[ 2] .g3 bit[ 3] .g3
    // bit[ 0] .g4 bit[ 1] .g4 bit[ 2] .g4 bit[ 3] .g4
    // or gates: bit[ 0] .g5 bit[ 1] .g5 bit[ 2] .g5 bit[ 3] .g5
    // Generated instances are connected with
    // generated nets: bit[ 0].t1 bit[ 1].t1 bit[ 2].t1 bit[ 3] .t1
    // bit[ 0].t2 bit[ 1].t2 bit[ 2].t2 bit[ 3].t2
    // bit[ 0] .t3 bit[ 1] .t3 bit[ 2] .t3 bit[ 3] .t3
    generate
        for(i=0; i<SIZE; i=i+1) begin:bit
            wire t1, t2, t3; // generated net declaration
            xor g1 ( t1, a[ i], b[ i] );
            xor g2 ( sum[ i], t1, c[ i] );
            and g3 ( t2, a[ i] , b[ i] );
            and g4 ( t3, t1, c[ i] );
            or g5 ( c[ i+1], t2, t3);
        end
    endgenerate
```

    assign \(\mathrm{CO}=\mathrm{C}[\) SIZE];
    endmodule

The generated instance names in a multi-level generate loop are shown in example 5. The generated name for the scope at each generate loop is created by adding the "[genvar s value]" string to the end of the generate block identifier for the loop. The generated names are now generated identifiers (see 2.7.2) which can be used in hierarchical path names (see 12.4).

Example 5 A multi-level generate loop

```
parameter SIZE = 2;
genvar i, j, k, m;
generate
    for (i=0; i<SIZE+1; i=i+1) begin:B1 // scope B1[ i]
            M1 N1(); // instantiates B1[ i].N1[ i]
            for (j=0; j<SIZE; j=j+1) begin:B2 // scope B1[ i].B2[ j]
                M2 N2(); // instantiates B1[ i] .B2[ j] .N2
                    for (k=0; k<SIZE; k=k+1) begin:B3 // scope B1[ i].B2[ j].B3[ k]
                    M3 N3(); // instantiates B1[ i] .B2[ j] .B3[ k] .N3
                    end
            end
            if (i>0)
                for (m=0; m<SIZE; m=m+1) begin:B4 // scope B1[ i] .B4[m]
                    M4 N4(); // instantiates B1[ i] .B4[ m] .N4
            end
    end
endgenerate
    // some of the generated instance names are:
    // B1[ 0].N1 B1[ 1].N1
    // B1[ 0].B2[ 0].N2 B1[ 0].B2[ 1] .N2
    // B1[ 0].B2[ 0].B3[ 0] .N3 B1[ 0] .B2[ 0] .B3[ 1] .N3
    // B1[ 0] .B2[ 1] . B3[ 0] .N3
    // B1[ 1].B4[ 0] .N4 B1[ 1].B4[ 1] .N4
```


### 12.1.3.3 generate-conditional

A generate-conditional is an if-else-if generate construct that permits modules, user defined primitives, Verilog gate primitives, continuous assignments, initial blocks and always blocks to be conditionally instantiated into another module based on an expression that is deterministic at the time the design is elaborated.

Example 6 shows the implementation of a parameterized module. If either of the multiplier s a_width or b_width parameters are less than 8 (bits), a CLA multiplier is instantiated. If both of the multiplier s a_width or b_width parameters are greater than or equal to 8 (bits), a Wallace tree multiplier is instantiated.

Example 6 An implementation of a parameterized multiplier module

```
module multiplier(a,b,product);
parameter a_width = 8, b_width = 8;
localparam product_width = a_width+b_width; // can not be modified
// directly with the defparam statement
// or the module instance statement #
input [ a_width-1:0] a;
input [ b_width-1:0] b;
output [ product_width-1:0] product;
generate
    if((a_width < 8) || (b_width < 8))
        CL\overline{A} multiplier #(a width,b width) ul(a, b, product);
        // instance a CLA multiplier
        else
            WALLACE_multiplier #(a_width,b_width) ul(a, b, product);
            // instance a Wallace-tree multiplier
endgenerate
// The generated instance name is ul
endmodule
```


### 12.1.3.4 generate-case

A generate case construct permits modules, user defined primitives, Verilog gate primitives, continuous assignments, initial blocks and always blocks to be conditionally instantiated into another module based on a select one-of-many case construct. The selecting case expression must be deterministic at the time the design is elaborated.

Example 7 Generate with a case to handle widths less that 3

```
generate
    case (WIDTH)
            1: adder_1bit x1(co, sum, a, b, ci);
// 1-bit adde\overline{r implementation}
            2: adder 2bit x1(co, sum, a, b, ci);
// 2-bit adde\overline{r implementation}
            default: adder_cla #(WIDTH) x1(co, sum, a, b, ci);
// others - carry look-ahead adder
    endcase
// The generated instance name is x1
endgenerate
```

Example 8 A module of memory dimm

```
module dimm;
    parameter [ 31:0] MEM_SIZE = 8, // in mbytes
                            MEM WIDTH = 16;
        input [ 11:0] adr;
        input [ 1:0] ba;
        input rasx, casx, csx, wex;
        input [ 7:0] dqm;
        input cke;
        input [ 7:0] ds;
        inout [ 63:0] data;
        input [ 3:0] clk;
        wire rasb, casb, csb, web;
        wire [ 7:0] bex;
        genvar i;
        generate
                case ({ MEM_SIZE, MEM_WIDTH})
                {32' d8, - 32' d16} : /// 8Meg 16 bits wide.
                    begin
                    for (i=0;i<4;i = i + 1)
                            begin:word
                        sms_16b216t0 p
                        (.clk(clk), .csb(csx), .cke(cke), .ba(ba[ 0] ),
                        .addr(adr[ 10:0] ),...rasb(rasx), .casb(casx),
                        .web(wex),.udqm(dqm[ 2*i+1]), .ldqm(dqm[ 2* i] ),
                        ...dqi(data[ 15+16*i:16*i]), .dev_id(dev_id3[ 4:0] )
                );
                end
                task read_mem;
                        input [ 31:0] address;
                        output [ 63:0] data;
                        begin
                        word[ 3] .p.read_mem(address, data[ 63:48]);
                        word[ 2] .p.read mem(address, data[ 47:32]);
                        word[ 1].p.read_mem(address, data[ 31:16]);
                        word[ 0].p.read_mem(address, data[ 15:0]);
                        end
                endtask
                end
```

```
// The generated instance names are word[ 3] .p, word[ 2] .p,
// word[ 1] .p, word[ 0] .p, and the task read_mem
            {32' d16, 32' d8} : // 16Meg 8 bits wide.
                begin
                for (i=0;i<4;i = i + 1)
                    begin:byte
                    sms_16b208t0 p
                    (.clk(clk), .csb(csx), .cke(cke), .ba(ba[ 0]),
                        .addr(adr[ 10:0]),
                    ...rasb(rasx), .casb(casx), .web(wex), .dqm(dqm[ i] ),
                    .dqi(data[ 8+8*i:8*i]),...dev_id(dev_id7[ 4:0] )
                    );
            end
                task read_mem;
            input [ \}11:0] address
            output [ 63:0] data;
            begin
            byte[ 7] .p.read mem(address, data[ 63:56]);
            byte[ 6].p.read_mem(address, data[ 55:48]);
            byte[ 5].p.read_mem(address, data[ 47:40]);
            byte[ 4].p.read_mem(address, data[ 39:32]);
            byte[ 3].p.read_mem(address, data[ 31:24]);
            byte[ 2].p.read_mem(address, data[ 23:16]);
            byte[ 1].p.read_mem(address, data[ 15:8]);
            byte[ 0].p.read_mem(address, data[ 7:0]);
                    end
                endtask
            .....
            endcase
        endgenerate
    // The generated instance names are byte[ 7] .p, byte[ 6] .p,
    // byte[ 5] .p, byte[ 4] .p, byte[ 3] .p, byte[ 2] .p, byte[ 1] .p,
    // byte[ 0].p and the task read_mem
```

endmodule

## 12.2 verriding module parameter values

There are two different ways that parameters can be defined. The first is the module_parameter_port_list (see 12.1), and the second is as a module_item (see 3.11). A module declaration can contain parameter definitions of either or both types, or no parameter definitions.

A module parameter can have a type specification and a range specification. The effect of parameter overrides on a parameter s type and range shall be in accordance with the following rules:

A parameter declaration with no type or range specification shall default to the type and range of the final override value assigned to the parameter.
A parameter with a range specification, but with no type specification, shall be the range of the parameter declaration and shall be unsigned. An override value shall be converted to the type and range of the parameter.
A parameter with a type specification, but with no range specification, shall be of the type specified. An override value shall be converted to the type of the parameter. A signed parameter shall default to the range of the final override value assigned to the parameter.

A parameter with a signed type specification and with a range specification shall be a signed, and shall be the range of its declaration. An override value shall be converted to the type and range of the parameter.

## Examples:

```
module generic fifo
    #(parameter MSB=3, LSB=0, DEPTH=4) // These parameters can be
overridden
    (
            input [ MSB:LSB] in,
            input clk, read, write, reset,
            output [ MSB:LSB] out,
            output full, empty
            );
    localparam FIFO_MSB = DEPTH*MSB; // These parameters are local,
and
    localparam FIFO_LSB = LSB; // cannot be overridden. They can
be
                    // affected by altering the public
                    // parameters above, and the module
                        // will work correctly.
        reg [ FIFO_MSB:FIFO_LSB] fifo;
        reg [ LOG2(DEPTH):0] depth;
```

        always@(posedge clk or reset) begin
            casex (\{read,write, reset\})
            // implementation of fifo
        endcase
        end
    endmodule
    There are two ways to alter non-local parameter values: the defparam statement, which allows assignment to parameters using their hierarchical names, and the module instance parameter value assignment, which allows values to be assigned inline during module instantiation. If a defparam assignment conflicts with a module instance parameter, the parameter in the module will take the value specified by the defparam. The module instance parameter value assignment comes in two forms, by ordered list or by name. The next two subclauses describe these two methods.

There are two kinds of parameter declarations. The first kind of parameter declaration has a type and or range qualification, and second does not. When an untyped and unranged parameter $s$ value is overridden, the parameter takes on the size and type of the override.

When a typed and/or ranged parameter is overriden, the new value is converted to the type and size of the destination, and assigned to that parameter.

## Example:

```
module foo (a,b);
    real \(r 1, r 2\);
    parameter [ 2:0] \(A=3^{\prime} h 2\);
    parameter \(B=3^{\prime} h 2\);
    initial begin
        \(r 1=A\);
        r2 \(=B\);
```

```
        $display("r1 is %f r2 is %f",r1,r2);
    end
endmodule // foo
module bar;
    wire a,b;
    defparam f1.A = 3.1415;
    defparam f1.B = 3.1415;
    foo f1(a,b);
endmodule // bar
```

Parameter A is a typed and/or ranged parameter, so when its value is redefined, the parameter retains its original type and sign. Therefore, the defparam of f1. A with the value 3.1415 is performed by converting the floating point number 3.1415 into a fixed point number 3 and then the low 3 bits of 3 are assigned to A.

Parameter B is not a typed and/or ranged parameter, so when its value is redefined, the parameter type and range take on the type and range of the new value. Therefore, the defparam of $f 1$. $B$ with the value 3.1415 replaces B s current value of 3 h 2 with the floating point number 3.1415.

### 12.2.1 defparam statement

Using the defparam statement, parameter values can be changed in any module instance throughout the design using the hierarchical name of the parameter. However, a defparam statement in a hierarchy under a generate scope or array of instances shall not change a parameter value outside that hierarchy. See 12.4 for hierarchical names.

The expression on the right-hand side of the defparam assignments shall be a constant expression involving only numbers and references to parameters. The referenced parameters (on the right-hand side of the defparam) shall be declared in the same module as the defparam statement.

The defparam statement is particularly useful for grouping all of the parameter value override assignments together in one module.

In the case of multiple defparams for a single parameter, the parameter takes the value of the last defparam statement encountered in the source text. When defparams are encountered in multiple source files, e.g., found by library searching, the defparam from which the parameter takes it s value is undefined.

## Example:

```
module top;
reg clk;
reg [ 0:4] in1;
reg [ 0:9] in2;
wire [ 0:4] o1;
wire [ 0:9] 02;
vdff m1 (o1, in1, clk);
vdff m2 (o2, in2, clk);
endmodule
module vdff (out, in, clk);
parameter size = 1, delay = 1;
input [ 0:size-1] in;
input clk;
output [ 0:size-1] out;
reg [ 0:size-1] out;
always @(posedge clk)
    # delay out = in;
endmodule
module annotate;
defparam
    top.m1.size = 5,
    top.m1.delay = 10,
    top.m2.size = 10,
    top.m2.delay = 20;
endmodule
```

The module annotate has the defparam statement which overrides size and delay parameter values for instances $m 1$ and $m 2$ in the top-level module top. The modules top and annotate would both be considered top-level modules.

### 12.2.2 Module instance parameter value assignment

An alternative method for assigning values to parameters within module instances is to use one of the two forms of module instance parameter value assignment. They are assignment by ordered list and assignment by name. The two types of module instance parameter value assignment shall not be mixed; parameter assignments to a particular module instance shall be entirely by order or entirely by name.

Module instance parameter value assignment by ordered list is similar in appearance to the assignment of delay values to gate instances and assignment by name is similar to connecting module ports by name. It supplies values for particular instances of a module to any parameters that have been specified in the definition of that module.

### 12.2.2.1 arameter value assignment by ordered list

The order of the assignments in the module instance parameter value assignment by ordered list shall follow the order of declaration of the parameters within the module. It is not necessary to assign values to all of the parameters within a module when using this method. However, it is not possible to skip over a parameter.

Therefore, to assign values to a subset of the parameters declared within a module, the declarations of the parameters that make up this subset shall precede the declarations of the remaining parameters. An alternative is to assign values to all of the parameters, but to use the default value (the same value assigned in the declaration of the parameter within the module definition) for those parameters that do not need new values.

## Example:

Consider the following example, where the parameters within module instance mod_a are changed during instantiation.

```
module m;
reg clk;
wire[ 0:4] out_c, in_c;
wire[ 1:10] out_a, in_a;
wire[ 1:5] out_b, in_b;
// create an instance and set parameters
vdff #(10,15) mod_a(out_a, in_a, clk);
// create an instance leaving default values
vdff mod_b(out_b, in_b, clk);
// create an instance and set one parameter
vdff #(.delay(12)) mod_c(out_c, in_c, clk);
endmodule
module vdff (out, in, clk);
parameter size = 5, delay = 1;
input [ 0:size-1] in;
input clk;
output [ 0:size-1] out;
reg [ 0:size-1] out;
always @(posedge clk)
    # delay out = in;
endmodule
```

In this example, the name of the module being instantiated is vdff. The construct \# $(10,15)$ assigns values to parameters used in the mod_a instance of vdff. The parameter size is assigned the value 10 and the parameter delay is assigned the value 15 for the instance of module vdff called mod_a. The construct \# (. delay (12)) assigns the parameter delay the value 12 in the instance of module $\overline{\text { v }}$ dff called mod_c.

### 12.2.2.2 arameter value assignment by name

Parameter assignment by name consists of explicitly linking the parameter name and it's new value. The name of the parameter shall be the name specified in the instantiated module.

It is not necessary to assign values to all of the parameters within a module when using this method. Only those parameters that are assigned new values need to be specified.

The parameter expression is optional so that the instantiating module can document the existence of a parameter without assigning anything to it. The parentheses are required and in this case the parameter retains its default value. Once a parameter is assigned a value, there shall not be another assignment to this parameter name.

### 12.2.3 arameter dependence

A parameter (for example, memory_size) can be defined with an expression containing another parameter (for example, word_size). Since memory_size depends on the value of word_size, a modification of word_size changes the value of memory_size. For example, in the following parameter declaration, an update of word_size, whether by defparam statement or in an instantiation statement for the module that defined these parameters, automatically updates memory_size.

```
parameter
```

```
word_size = 32,
```

word_size = 32,
memory_size = word_size * 4096;

```
memory_size = word_size * 4096;
```


## 12.3 orts

Ports provide a means of interconnecting a hardware description consisting of modules, primitives, and macromodules. For example, module A can instantiate module B, using port connections appropriate to module A. These port names can differ from the names of the internal nets and variables specified in the definition of module B.

### 12.3.1 ort definition

The syntax for ports and a list of ports is given in Syntax 12-5.

```
list_of_ports ::= (From Annex A - A.1.4)
    ( port \{, port \})
list_of_port_declarations ::=
        ( port_declaration \(\{\), port_declaration \} )
        ()
port ::=
        [ port_expression ]
    |. port_identifier ( [ port_expression ] )
port_expression ::=
        port_reference
    | \{ port_reference \{ , port_reference \} \}
port_reference ::=
        port_identifier
        port_identifier [ constant_expression ]
        | port_identifier [ range_expression ]
port_declaration \(::=\)
        \{attribute_instance\} inout_declaration
        | \{attribute_instance\} input_declaration
        | \{attribute_instance\} output_declaration
```


## Syntax 12-5 Syntax for port

### 12.3.2 List of ports

The port reference for each port in the list of ports at the top of each module declaration can be one of the following:

A simple identifier or escaped identifier
A bit-select of a vector declared within the module
A part-select of a vector declared within the module
A concatenation of any of the above

The port expression is optional because ports can be defined that do not connect to anything internal to the module. Once a port has been defined, there shall not be another port definition with this same name.

The first type of module port with only a port expression is an implicit port. The second type is the explicit port. This explicitly specifies the port_identifier used for connecting module instance ports by name (see 12.3.6) and the port_expression which contains identifiers declared inside the module as described in 12.3.3. Use of named port connections shall not be used for implicit ports unless the port_expression is a simple port_identifier.

### 12.3.3 ort declarations

Each port_expression in the list of ports for the module declaration shall also be declared in the body of the module as one of the following port declarations: input, output, or inout (bidirectional). This is in addition to any other data type declaration for a particular port for example, a reg or wire. The syntax for port declarations is given in Syntax 12-6.

```
inout_declaration ::= (From Annex A - A.2.1.2)
    inout [ net_type ] [ signed ] [ range ] list_of_port_identifiers
input_declaration ::=
    input [ net_type ] [ signed ] [ range ] list_of_port_identifiers
output_declaration ::=
        output [ net_type ] [ signed ] [ range ]
            list_of_port_identifiers
    | output [ reg ] [ signed ] [ range ]
        list_of_port_identifiers
    | output reg [ signed ] [ range ]
        list_of_variable_port_identifiers
    | output [ output_variable_type ]
        list_of_port_identifiers
    | output output_variable_type
        list_of_variable_port_identifiers
list_of_port_identifiers ::= (From Annex A - A.2.3)
    port_identifier {, port_identifier }
```


## Syntax 12-6 Syntax for port declarations

If a port declaration includes a net or variable type, then the port is considered completely declared and it is an error for the port to be declared again as a variable or net data type declaration. Because of this, all other aspects of the port shall be declared in such a port declaration, including the signed and range definitions if needed.

If a port declaration does not include a net or variable type, then the port can be again declared in a net or variable declaration. If the net or variable is declared as a vector, the range specification between the two declarations of a port shall be identical. Once a name is used in a port declaration it shall not be declared again in another port declaration or in a data type declaration.

NOTE-Implementations may limit maximum number of ports in a module definition, but they will at least be 256 .

## Example:

```
input aport; // First declaration - okay.
input aport; // Error - multiple declaration, port declaration
output aport; // Error - multiple declaration, port declaration
```

The signed attribute can be attached either to a port declaration or to the corresponding net or reg declaration, or to both. If either the port or the net/reg is declared as signed, then the other shall also be considered signed.

Implicit nets shall be considered unsigned. Nets connected to ports without an explicit net declaration shall be considered unsigned, unless the port is declared as signed.

## Example:

```
module test(a,b,c,d,e,f,g,h);
input [ 7:0] a; // no explicit declaration - net is unsigned
input [ 7:0] b;
input signed [ 7:0] c;
input signed [ 7:0] d; // no explicit net declaration - net is
signed
output [ 7:0] e; // no explicit declaration - net is unsigned
output [ 7:0] f;
output signed [ 7:0] g;
output signed [ 7:0] h; // no explicit net declaration - net is
signed
wire signed [ 7:0] b; // port b inherits signed attribute from net
decl.
wire [ 7:0] c; // net c inherits signed attribute from port
reg signed [ 7:0] f; // port f inherits signed attribute from reg
decl.
reg [ 7:0] g; // reg g inherits signed attribute from port
endmodule
module complex_ports ({ c,d}, .e(f)); // Nets {c,d} receive the
first
                            // port bits. Name ' f' is declared inside the
module.
    // Name ' e' is defined outside the module.
    // Can't use named port connections of first port.
module split_ports (a[ 7:4], a[ 3:0]); // First port is upper 4
bits of
    // ' a' .
    // Second port is lower 4 bits of ' a'.
    // Can't use named port connections because
    // of part-select port ' a' .
module same port (.a(i), .b(i)); // Name ' i' is declared
inside the
                                    // module as a inout port. Names ' a' and 'b' are
                                    // defined for port connections.
module renamed_concat (.a({ b, c}), f, .g(h[ 1] ));
    // Names ' b', ' C' , ' f' ,' 'h' are defined inside the
module.
    // Names ' a', 'f', ' g' are defined for port
connections.
    // Can use named port connections.
module same_input (a,a);
input a; - // This is legal. The inputs are ored together.
```


### 12.3.4 List of ports declarations

An alternate syntax which minimizes the duplication of data can be used to specify the ports of a module. Each module shall either be declared entirely with the list of ports syntax as described in 12.3.2 or entirely using the list_of_port_declarations as described in this section.

Each declared port provides the complete information about the port. The port s direction, width, net, or variable type, and whether the port is signed or unsigned is completely described. The same syntax for input, inout, and output declarations is used in the module header as would be used for the list of port style declaration, except the list_of_port_declarations is included in the module header rather than separately (after the ; which terminates the module header).

As an example, the module named test given in the previous example could alternatively be declared as:

## Example:

```
module test (
    input [ 7:0] a,
    input signed [ 7:0] b, c, d, // multiple ports that share all
                                    // attributes can be declared together
    output [ 7:0] e, // every attribute of the declaration
                            // must be in the one declaration
    output signed reg [ 7:0] f, g,
    output signed [ 7:0] h) ;
            // It is illegal to redeclare any ports of the module in the
body
    // of the module.
endmodule
```

The port_reference type of module port declaration shall not be done using list_of_port_declarations style of module declarations. Also ports declared using the list_of_port_declarations shall only be simple identifiers. They shall not be bit-selects, part-selects, or concatenations (as in the example complex_ports); nor can a port be split (as in the example split_ports); nor can they be named ports (as in the example same_port).

Designs may freely mix modules declared using each syntax; hence implementations desiring the above special cases of port declaration can be done using the first list_of ports syntax.

### 12.3.5 Connecting module instance ports by ordered list

One method of making the connection between the port expressions listed in a module instantiation and the ports declared within the instantiated module is the ordered list that is, the ports expressions listed for the module instance shall be in the same order as the ports listed in the module declaration.

## Example:

The following example illustrates a top-level module (topmod) that instantiates a second module (modB). Module modB has ports that are connected by an ordered list. The connections made are as follows:

```
Port wa in the modB definition connects to the bit-select v[ 0] in the topmod module.
Port wb connects to v[ 3].
Port c connects to w.
Port d connects to v[ 4] .
```

In the modB definition, ports wa and wb are declared as inouts while ports c and d are declared as input.

```
module topmod;
    wire [ 4:0] v;
    wire a,b,c,w;
    modB b1 (v[ 0], v[ 3] , w, v[ 4]);
    endmodule
    module modB (wa, wb, c, d);
    inout wa, wb;
    input c, d;
    tranif1 g1 (wa, wb, cinvert);
    not #(2, 6) n1 (cinvert, int);
    and #(6, 5) g2 (int, c, d);
endmodule
```

During simulation of the b1 instance of modb, the and gate g 2 activates first to produce a value on int. This value triggers the not gate n 1 to produce output on cinvert, which then activates the tranif1 gate g1.

### 12.3.6 Connecting module instance ports by name

The second way to connect module ports consists of explicitly linking the two names for each side of the connection, the port declaration name from the module declaration to the expression the name used in the module declaration, followed by the name used in the instantiating module. This compound name is then placed in the list of module connections. The port name shall be the name specified in the module declaration. The port name cannot be a bit-select, a part-select, or a concatenation of ports. If the module port declaration was implicit, the port_expression shall be a simple port_identifer which is used as the port name. If the module port declaration was explicit, the explicit name is used as the name of port.

The port expression can be any valid expression.
The port expression is optional so that the instantiating module can document the existence of the port without connecting it to anything. The parentheses are required.

The two types of module port connections shall not be mixed; connections to the ports of a particular module instance shall be all by order or all by name.

## Examples:

Example 1 In the following example, the instantiating module connects its signals topA and topB to the ports In1 and Out defined by the module ALPHA. At least one port provided by ALPHA is unused; it is named In2. There could be other unused ports not mentioned in the instantiation.

ALPHA instance1 (.Out(topB),.In1(topA),.In2());

Example 2 This example defines the modules modB and topmod, and then topmod instantiates modB using ports connected by name.

```
module topmod;
    wire [ 4:0] v;
    wire a,b,c,w;
    modB b1 (.wb(v[ 3] ),.wa(v[ 0] ),.d(v[ 4] ),.c(w));
endmodule
module modB(wa, w.b, c, d);
    inout wa, wb;
    input c, d;
    tranif1 gl(wa, wb, cinvert);
    not #(6, 2) n1(cinvert, int);
    and # (5, 6) g2(int, c, d);
endmodule
```

Since these connections are made by name, the order in which they appear is irrelevant.
Multiple module instance port connections are not allowed, e.g., the following example is illegal:
Example 3 This example shows illegal port connections.

```
module test;
    a ia (.i (a),.i (b), // illegal connection of input port
twice.
    .O (c), .O (d), // illegal connection of output port
twice.
            .e (e), .e (f)); // illegal connection of inout port
twice.
endmodule
```


### 12.3.7 eal numbers in port connections

The real data type shall not be directly connected to a port. It shall be connected indirectly, as shown in the following example. The system functions \$realtobits and \$bitstoreal shall be used for passing the bit patterns across module ports. (See 17.8 for a description of these system tasks.)

## Example:

```
module driver (net_r);
    output net_r;
    real \(r\);
    wire [ 64:1] net_r = \$realtobits (r);
endmodule
module receiver (net_r);
    input net_r;
    wire [ \(64: \overline{1}]\) net_r;
    real \(r\);
    initial assign \(r=\) \$bitstoreal (net_r);
endmodule
```


### 12.3.8 Connecting dissimilar ports

A port of a module can be viewed as providing a link or connection between two items (nets, regs, expressions, etc.) one internal to the module instance and one external to the module instance.

Examination of the port connection rules described in 12.3 .9 will show that the item receiving the value through the port (the internal item for inputs, the external item for outputs) shall be a structural net expression. The item that provides the value can be any expression.

NOTE A port that is declared as input (output) but used as an output (input) or inout may be coerced to inout. If not coerced to inout, a warning has to be issued.

### 12.3.9 ort connection rules

The following rules shall govern the way module ports are declared and the way they are interconnected.

### 12.3.9.1 ule 1

An input or inout port shall be of type net.

### 12.3.9.2 ule 2

Each port connection shall be a continuous assignment of source to sink, where one connected item shall be a signal source and the other shall be a signal sink. The assignment shall be a continuous assignment from source to sink for input or output ports. The assignment is a nonstrength reducing transistor connection for inout ports. Only nets or structural net expressions shall be the sinks in an assignment.

A structural net expression is a port expression whose operands can be the following:
A scalar net
A vector net
A constant bit-select of a vector net
A part-select of a vector net
A concatenation of structural net expressions

The following external items shall not be connected to the output or inout ports of modules:
Variables
Expressions other than
i) A scalar net
ii) A vector net
iii) A constant bit-select of a vector net
iv) A part-select of a vector net
v) A concatenation of the expressions listed above

### 12.3.10 Net types resulting from dissimilar port connections

When different net types are connected through a module port, the nets on both sides of the port can take on the same type. The resulting net type can be determined as shown in Table 45. In the table, external net means the net specified in the module instantiation, and internal net means the net specified in the module definition. The net whose type is used is said to be the dominating net. The net whose type is changed is said to be the dominated net. It is permissible to merge the dominating and dominated nets into a single net, whose type shall be that of the dominating net. The resulting net is called the simulated net, and the dominated net is called a collapsed net.

The simulated net shall take the delay specified for the dominating net. If the dominating net is of the type trireg, any strength value specified for the trireg net shall apply to the simulated net.

### 12.3.10.1 Net type resolution rule

When the two nets connected by a port are of different net type, the resulting single net can be assigned one of the following:

The dominating net type if one of the two nets is dominating, or The net type external to the module

When a dominating net type does not exist, the external net type shall be used.

### 12.3.10.2 Net type table

Table 45 shows the net type dictated by net type resolution rule.
The simulated net shall take the net type specified in the table and the delay specified for that net. If the simulated net selected is a trireg, any strength value specified for the trireg net applies to the simulated net.

## Table 45-Net types resulting from dissimilar port connections

| Internal net | External net |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | wire, tri | wand, triand | wor, trior | trireg | tri0 | tri1 | supply0 | supply1 |
| wire, tri | ext | ext | ext | ext | ext | ext | ext | ext |
| wand, triand | int | ext | warn | warn | warn | warn | ext | ext |

Table 45-Net types resulting from dissimilar port connections (continued)

| Internal net | External net |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | wire, tri | wand, triand | wor, trior | trireg | tri0 | tri1 | supply0 | supply1 |
| wor, trior | int | warn | ext | warn | warn | warn | ext | ext |
| trireg | int | warn | warn | ext | ext | ext | ext | ext |
| tri0 | int | warn | warn | int | ext | warn | ext | ext |
| tri1 | int | warn | warn | int | warn | ext | ext | ext |
| supply0 | int | int | int | int | int | int | ext | warn |
| supply 1 | int | int | int | int | int | int | warn | ext |
| KEY: <br> ext = The external net type is used <br> int = The internal net type is used <br> warn = A warning is issued and the external net type is used |  |  |  |  |  |  |  |  |

### 12.3.11 Connecting signed values via ports

The sign attribute shall not cross hierarchy. In order to have the signed type cross hierarchy, the signed keyword must be used in the object's declaration at the different levels of hierarchy. Any expressions on a port shall be treated as any other expression in an assignment. It shall be typed, sized, evaluated and the resulting value assigned to the object on the other side of the port using the same rules as an assignment.

### 12.4 Hierarchical names

Every identifier in a Verilog HDL description shall have a unique hierarchical path name. The hierarchy of modules and the definition of items such as tasks and named blocks within the modules shall define these names. The hierarchy of names can be viewed as a tree structure, where each module instance, generated instance, task, function, or named begin-end or fork-join block defines a new hierarchical level, or scope, in a particular branch of the tree.

At the top of the name hierarchy are the names of one or more root modules of which no instances have been created. This root or these parallel root modules make up one or more hierarchies in a design description or description. Inside any module, each module instance (including an arrayed or generated instance), task definition, function definition, and named begin-end or fork-join block shall define a new branch of the hierarchy. Named blocks within named blocks and within tasks and functions shall create new branches. Only non-recursively referenced automatic tasks and/or functions create visible branches that can be referenced. Recursively called tasks and functions, declared using the automatic keyword and recursively called from within the same task or function, do not create visible branches that can be referenced. See 10.2.1 and 10.3.1 for a discussion of automatic tasks and functions.

Each node in the hierarchical name tree shall be a separate scope with respect to identifiers. A particular identifier can be declared at most once in any scope. See 12.6 for a discussion of scope rules and 3.12 for a discussion of name spaces.

Any named Verilog object or hierarchical name reference can be referenced uniquely in its full form by concatenating the names of the modules, module instance names, tasks, functions, or named blocks that contain it. The period character shall be used to separate each of the names in the hierarchy, except for escaped identifiers embedded in the hierarchical name reference, which are followed by separators composed of white space and a period-character. The complete path name to any object shall start at a top-level (root) module. This path name can be used from any level in the hierarchy or from a parallel hierarchy. The first node name in a path name can also be the top of a hierarchy that starts at the level where the path is being used (which allows and enables downward referencing of items) with the exceptions of items of automatic tasks and automatic task item declarations. These declarations can not be accessed by their hierarchical names.

The syntax for hierarchical path names is given in Syntax 12-7.

```
escaped_hierarchical_identifier \({ }^{*}::=\) (From Annex A - A.9.3)
    escaped_hierarchical_branch
        [ \{ .simple_hierarchical_branch | .escaped_hierarchical_branch \}]
escaped_identifier ::=
    \ \{Any_ASCII_character_except_white_space\} white_space
hierarchical_identifier ::=
        simple_hierarchical_identifier
        | escaped_hierarchical_identifier
simple_hierarchical_identifier ::=
        simple_hierarchical_branch [.escaped_identifier ]
simple_identifier \(::=[\mathbf{a}-\mathbf{z A}-\mathbf{Z}-]\left\{\left[\mathbf{a - z A} \mathbf{- Z 0} \mathbf{- 9} \_\$\right]\right\}\)
simple_hierarchical_branch ::=(From Annex A-A.9.4)
        simple_identifier [ [ unsigned_number ] ]
        [ \{ .simple_identifier [ [ unsigned_number ] ] \}]
escaped_hierarchical_branch \({ }^{*}::=\)
        escaped_identifier [ [ unsigned_number ] ]
            [ \{ .escaped_identifier [ [ unsigned_number ] ] \}]
white_space \(::=\) (From Annex \(A\) - A.9.5)
        space | tab | newline | eof
```

*The period in escaped_hierarchical_identifier and escaped_hierarchical_branch shall be preceded by white_space, but shall not be followed by white_space.
The period (.) in simple_hierarchical_identifier and simple_hierarchical_branch shall not be preceded or followed by white_space.
A simple_identifier and arrayed_reference shall start with an alpha or underscore (_) character, shall have at least one character, and shall not have any spaces.
End of file.
Syntax 12-7 Syntax for hierarchical path names

## Examples:

Example 1 The code in this example defines a hierarchy of module instances and named blocks.

```
module mod (in); module cct (stim1, stim2);
input in;
always @(posedge in) begin : keep
reg hold;
    hold = in;
end
endmodule
module wave;
reg stim1, stim2;
cct a(stim1, stim2); // instantiate cct
initial begin :wave1
    #100 fork :innerwave
                reg hold;
            join
        #150 begin
            stim1 = 0;
    end
end
endmodule
```

Figure 32 illustrates the hierarchy implicit in this Verilog code.


Figure 32-Hierarchy in a model

Figure 33 is a list of the hierarchical forms of the names of all the objects defined in the code.

```
wave wave.a.bmod
wave.stim1 wave.a.bmod.in
wave.stim2 wave.a.bmod.keep
wave.a wave.a.bmod.keep.hold
wave.a.stim1 wave.wave1
wave.a.stim2 wave.wave1.innerwave
wave.a.amod wave.wavel.innerwave.hold
wave.a.amod.in
wave.a.amod.keep
wave.a.amod.keep.hold
```

Figure 33-Hierarchical path names in a model
Hierarchical name referencing allows free data access to any object from any level in the hierarchy. If the unique hierarchical path name of an item is known, its value can be sampled or changed from anywhere within the description.

Example 2 The next example shows how a pair of named blocks can refer to items declared within each other.

```
begin
    fork : \(\bmod 1\)
        reg x ;
        mod_2.x = 1;
    join
    fork :mod_2
        reg x ;
        mod_1.x = 0;
    join
end
```


### 12.5 Upwards name referencing

The name of a module or module instance is sufficient to identify the module and its location in the hierarchy. A lower-level module can reference items in a module above it in the hierarchy.Variables can be referenced if the name of the higher-level module or its instance name is known. For tasks, functions, and named blocks, Verilog shall look in the enclosing module for the name until it is found or until the root of the hierarchy is reached. It shall only search in higher enclosing modules for the name, not instances.

The syntax for an upward reference is given in Syntax 12-8.

```
upward_name_reference ::=
    module_identifier.item_name
item_name ::=
        function_identifier
        block_identifier
        net_identifier
        parameter_identifier
        port_identifier
        task_identifier
        variable_identifier
```

Upwards name references can also be done with names of the form

```
module_instance_name.item_name
```

A name of this form shall be resolved as follows:
a) Look in the current module for a module instance named module_instance_name. If found, this name reference shall be treated as a downward reference, and the item name shall be resolved in the corresponding module.
b) Look in the parent module for a module instance named module_instance_name. If found, the item name shall be resolved from that instance, which is the sibling of the module containing the reference.
c) Repeat step b), going up the hierarchy.

There shall be no spaces within the hierarchical name reference, except for escaped identifiers embedded in the hierarchical name reference, which are followed by separators composed of white space and a periodcharacter.

## Example:

In this example, there are four modules, a, b, c, and d. Each module contains an integer i. The highestlevel modules in this segment of a model hierarchy are a and $d$. There are two copies of module b because module $a$ and $d$ instantiate $b$. There are four copies of $c$. i because each of the two copies of $b$ instantiates c twice.

```
module a;
integer i;
b a b1();
endmodule
module b;
integer i;
c b_c1(), b_c2();
initial // downward path references two copies of i:
    #10 b_c1.i = 2;// a.a_b1.b_c1.i, d.d_b1.b_c1.i
endmodule
module c;
integer i;
initial begin // local name references four copies of i:
    i = 1; // a.a_b1.b_c1.i, a.a_b1.b_c2.i,
    // d.d_b1.b_c1.i, d.d_b1.b_c2.i
    b.i = 1; // upwārd päth refereñces twwo copies of i:
        // a.a_b1.i, d.d_b1.i
    end
endmodule
module d;
integer i;
b d_b1();
initial begin // full path name references each copy of i
    a.i = 1; d.i = 5;
    a.a_b1.i = 2; d.d_b1.i = 6;
    a.a_b1.b_c1.i = 3; d.d_b1.b_c1.i = 7;
    a.a_b1.b_c2.i = 4; d.d_b1.b_c2.i = 8;
end
endmodule
```


### 12.6 Scope rules

The following four elements define a new scope in Verilog:

Modules<br>Tasks<br>Functions<br>Named blocks

An identifier shall be used to declare only one item within a scope. This rule means it is illegal to declare two or more variables that have the same name, or to name a task the same as a variable within the same module, or to give a gate instance the same name as the name of the net connected to its output.

If an identifier is referenced directly (without a hierarchical path) within a task, function, or named block, it shall be declared either locally within the task, function, or named block, or within a module, task or named block that is higher in the same branch of the name tree that contains the task, function, or named block. If it is declared locally, then the local item shall be used; if not, the search shall continue upward until an item by that name is found or until a module boundary is encountered. If the item is a variable, it shall stop at a module boundary; if the item is a task, function, or named block it continues to search higher-level modules until
found. The search shall cross named block, task, and function boundaries but not module boundaries. This fact means that tasks and functions can use and modify the variables within the containing module by name, without going through their ports.

If an identifier is referenced with a hierarchical name, the path can start with an module name, instance name, task, function, or named block. The names shall be searched first at the current level, then in higherlevel modules until found. Since both module names and instance names can be used, precedence is given to instance names if there is a module named the same as an instance name.

Because of the upward searching, path names which are not strictly on a downward path can be used.

## Example:

Example 1 In Figure 34, each rectangle represents a local scope. The scope available to upward searching extends outward to all containing rectangles with the boundary of the module A as the outer limit. Thus block $G$ can directly reference identifiers in $\mathrm{F}, \mathrm{E}$, and A ; it cannot directly reference identifiers in $\mathrm{H}, \mathrm{B}, \mathrm{C}$, and D.


Figure 34-Scopes available to upward name referencing

Example 2 The following example shows an incompletely defined downward reference that can be accessed.

```
task t;
reg r, s;
begin : b
    // redundant assignments to reg r
    t.b.r = 0; // poorly defined but found by upward search
    t.s = 0; // fully defined downward reference
end
endtask
```


## 13. Configuring the contents of a design

### 13.1 Introduction

To facilitate both the sharing of Verilog designs between designers and/or design groups, and the repeatability of the exact contents of a given simulation (or other tool) session, the concept of configurations is used in the Verilog language. A configuration is simply an explicit set of rules to specify the exact source description to be used to represent each instance in a design. The operation of selecting a source representation for an instance is referred to as binding the instance.

The example below shows a simple configuration problem.
Example:

```
    file top.v
    module top();
    adder a1(...);
description
    adder a2(...);
    endmodule
```

```
file adder.v
```

file adder.v
module adder(...);
module adder(...);
// rtl adder description
// rtl adder description
...
...
endmodule
endmodule
endmodule

```

Consider using the rtl adder description in adder.v for instance al in module top and the gate-level adder description in adder. vg for instance a2. In order to specify this particular set of instance bindings and to avoid having to change the source description to specify a new set, a configuration can be used.
```

    config cfgl; // specify rtl adder for top.al, gate-level adder for
    top.a2
design rtlLib.top;
default liblist rtlLib;
instance top.a2 liblist gateLib;
endconfig

```

The elements of a config are explained in subsequent sections, but this simple example illustrates some important points about configs. As evidenced by the config-endconfig syntax, the config is a design element, similar to a module, which exists in the Verilog namespace. The config contains a set of rules which are applied when searching for a source description to bind to a particular instance of the design.

A Verilog design description starts with a top-level module (or modules), which is not instantiated elsewhere in the design. From this module s source description, the instantiated modules (or children) are found, and then the source descriptions for the module definitions of these subinstances shall be located, and so on until every instance in the design is mapped to a source description.

\subsection*{13.1.1 Library notation}

In order to map a Verilog instance to a source description, the concept of a symbolic library, which is simply a logical collection of design elements (such as modules, macromodules, primitives, or configs) can be used. These design elements can be referred to as cells. The cell name shall be the same as the name of the module/macromodule/primitive/config being processed. Syntax 13-1 specifies a cell from a given library.

> library_cell ::=
> \(\quad\) [library_identifier.]cell_identifier[:config]

\section*{Syntax 13-1 Syntax for cell}

This notation gives a symbolic method of referring to source descriptions; the method of mapping source descriptions into libraries is shown in greater detail in 13.2.1. The optional : config extension shall be used explicitly to refer to a config in the case where a config has the same name as a module/macromodule/ primitive.

For the purposes of this example, suppose the files top.v and adder.v, the rtl descriptions, have been mapped into the library rtlLib, and the file adder.vg, the gate-level description of the adder, has been mapped into the library gateLib. The actual mechanism for mapping source descriptions to libraries is detailed in 13.2.

\subsection*{13.1.2 Basic configuration elements}

The design statement in config cfg1 of the first example of 13.1 specifies the top-level module in the design and what source description is to be used. In this example, the rtlLib.top notation indicates the top-level module description shall be taken from rtlLib. Since top.v and adder.v were mapped to this library, the actual description for the module is known to come from top.v.

The default statement coupled with the liblist clause specifies, by default, all subinstances of top (i.e., top.a1 and top.a2) shall be taken from rtlLib, which means the descriptions in top.v and adder.v, which were mapped to this library, shall be used. For a basic design, which can be completely rtl, this can be sufficient to specify completely the binding for the entire design. However, here the top. a 2 instance of adder to the gate-level description shall be bound.

The instance statement specifies, for the particular instance top.a2, the source description shall be taken from gateLib. The instance statement overrides the default rule for this particular instance. Since adder.vg was mapped to gateLib, this statement dictates the gate-level description in adder.vg be used for instance top.a2.

\subsection*{13.2 Libraries}

As mentioned in the previous section, a library is a logical collection of cells which are mapped to particular source description files. The symbolic lib.cell[:config] notation supports the separate compilation of source files by providing a file system-independent name to refer to source descriptions when instances in a design are bound. It also allows multiple tools, which can have different invocation use-models, to share the same configuration.

\subsection*{13.2.1 Specifying libraries - the library map file}

When parsing a source description file (or files), the parser shall first read the library mapping information from a pre-defined file prior to reading any source files. The name of this file and the mechanism for reading it shall be tool-specific, but all compliant tools shall provide a mechanism to specify one or more library mapping files to be used for a particular invocation of the tool. If multiple mapping files are specified, then they shall be read in the order in which they are specified.

For the purposes of this discussion, assume the existence of a file named lib.map in the current working directory, which is automatically read by the parser prior to parsing any source files specified on the command line. The syntax for declaring a library in the library map file is shown in Syntax 13-2.
```

escaped_hierarchical_identifier ${ }^{*}::=$ (From Annex $A$ - A.1.1)
library_text ::=
\{ library_descriptions \}
library_descriptions $::=$
library_declaration
include_statement
config_declaration
library_declaration ::=
library library_identifier file_path_spec [ \{ , file_path_spec \}]
[ -incdir file_path_spec [ \{, file_path_spec \}];
file_path_spec ::=
file_path
include_statement ::=
include <file_path_spec> ;

```
*The period in escaped_hierarchical_identifier and escaped_hierarchical_branch shall be preceded by white_space, but shall not be followed by white_space.

\section*{Syntax 13-2 Syntax for declaring library in the library map file}

\section*{NOTES}

1 The file_path uses file system-specific notation to specify an absolute or relative path to a particular file or set of files. The following shortcuts/wildcards can be used:
? single character wildcard (matches any single character)
* multiple character wildcard (matches any number of characters in a directory/file name)
. . . hierarchical wildcard (matches any number of hierarchical directories)
. . specifies the parent directory
. \(\quad\) specifies the directory containing the lib.map
Paths which end in / shall include all files in the specified directory. Identical to /*.
Paths which do not begin with / are relative to the directory in which the current lib.map file is located.
2 The paths . /*.v and *.v are identical and both specify all files with a . v suffix in the current directory.
Any file encountered by the compiler which does not match any library s file_path specification shall by default be compiled into a library named work.

To perform the library mapping discussed in the example in 13.1, use the following library definitions in the lib.map file:
library \(r\) tlLib *.v; // matches all files in the current directory with a \(\cdot \mathrm{v}\) suffix
library gateLib . /*.vg; // matches all files in the current directory with a . vg suffix

\subsection*{13.2.1.1 File path resolution}

If a file name potentially matches multiple file path specifications, the path specifications shall be resolved in the following order:
a) File path specifications which end with an explicit filename
b) File path specifications which end with a wildcarded filename
c) File path specifications which end with a directory

If a file name matches path specifications in multiple library definitions (after the above resolution rules have been applied), it shall be an error.

Using these rules with the library definitions in the lib.map file, all source files encountered by the parser/ compiler can be mapped to a unique library. Once the source descriptions have been mapped to libraries, the cells defined therein are available for binding.

NOTE Tool implementers may find it convenient to provide a command-line argument to explicitly specify the library into which the file being parsed is to be mapped, which shall override any library definitions in the lib.map file. If these libraries do not exist in the lib. map file, they can only be accessed via an explicit config.

If multiple cells with the same name map to the same library, then the LAST cell encountered shall be written to the library. This is to support a separate-compile use-model (see 13.4.3), where it is assumed encountering a cell after it has previously been compiled is intended to be a recompiling of the cell. In the case where multiple modules with the same name are mapped to the same library in a single invocation of the compiler, then a warning message shall be issued.

\subsection*{13.2.2 Using multiple library mapping files}

In addition to specifying library mapping information, a lib. map file can also include references to other lib. map files. The include command is used to insert the entire contents of a library mapping file in another file during parsing. The result is as though the contents of the included mapping file appear in place of the include command.

The syntax of a lib. map file is limited to library specifications, include statements, and standard Verilog comment syntax. Syntax 13-3 shows the syntax for the include command.
```

include_statement ::= (From Annex A - A.1.1)
include <file_path_spec> ;

```

\section*{Syntax 13-3 Syntax for include command}

If the file path specification, whether in an include or library statement, describes a relative path, it shall be relative to the location of the file which contains the file path. Library providers shall include a local library mapping file in addition to the source contents of the library. Individual users can then simply include the provider s library mapping file in their own map file to gain access to the contents of the provided library.

\subsection*{13.2.3 Mapping source files to libraries}

For each cell definition encountered during parsing/compiling, the name of the source file being parsed is compared to the file path specifications of the library declarations in all of the library map files being used. The cell is mapped into the library whose file path specification matches the source file name.

\subsection*{13.3 Configurations}

As mentioned in the introduction of this chapter, a configuration is simply a set of rules to apply when searching for library cells to which to bind instances. The syntax for configurations is shown in 13.3.1

\subsection*{13.3.1 Basic configuration syntax}

The configuration syntax is shown in Syntax 13-4.
```

config_declaration $::=$ (From Annex A -A.1.2)
config config_identifier ;
design_statement
\{config_rule_statement \}
endconfig
design_statement ::=
design \{ [library_identifier.]cell_identifier \} ;
config_rule_statement $::=$
default_clause liblist_clause
inst_clause liblist_clause
inst_clause use_clause
cell_clause liblist_clause
cell_clause use_clause

```

\section*{Syntax 13-4 Syntax for configuration}

\subsection*{13.3.1.1 Design statement}

The design statement names the library and cell of the top-level module or modules in the design hierarchy configured by the config. There shall be one and only one design statement, but multiple top-level modules can be listed in the design statement. The cell or cells identified can not be configurations themselves. It is possible the design identified can have the same name as configs, however.

The design statement shall appear before any config rule statements in the config.
If the library identifier is omitted, then the library which contains the config shall be used to search for the cell.

\subsection*{13.3.1.2 The default clause}

The syntax for the default clause is specified in Syntax 13-5.
```

default_clause ::= (From Annex A - A.1.2)
default

```

\section*{Syntax 13-5 Syntax for default clause}

The default clause selects all instances which do not match a more specific selection clause. The use expansion clause (see 13.3.1.6) can not be used with a default selection clause. For other expansion clauses, there can not be more than one default clause which specifies the expansion clause.

For simple design configurations, it might be sufficient to specify a default liblist (see 13.3.1.5).

\subsection*{13.3.1.3 The instance clause}

The instance clause is used to specify the specific instance to which the expansion clause shall apply.The syntax for the instance clause is specified in Syntax 13-6.
```

inst_clause ::= (From Annex A - A.1.2)
instance inst_name
inst_name ::=
topmodule_identifier \{.instance_identifier\}

```

\section*{Syntax 13-6 Syntax for instance clause}

The instance name associated with the instance clause is a Verilog hierarchical name, starting at the toplevel module of the config (i.e., the name of the cell in the design statement).

\subsection*{13.3.1.4 The cell clause}

The cell selection clause names the cell to which it applies. The syntax for the cell clause is specified in Syntax 13-7.
```

cell_clause ::= (From Annex A - A.1.2)
cell [ library_identifier.]cell_identifier

```

\section*{Syntax 13-7 Syntax for cell clause}

If the optional library name is specified then the selection rule applies to any instance which is bound or is under consideration for being bound to the selected library and cell. It is an error if a library name is included in a cell selection clause and the corresponding expansion clause is a library list expansion clause.

\subsection*{13.3.1.5 The liblist clause}

The liblist clause defines an ordered set of libraries to be searched to find the current instance. The syntax for the liblist clause is specified in Syntax 13-8.
```

liblist_clause ::= (From Annex A - A.1.2)
liblist [{library_identifier}]

```

\section*{Syntax 13-8 Syntax for liblist clause}
liblists are inherited hierarchically downward as instances are bound. When searching for a cell to bind to the current unbound instance, and in the absence of an applicable binding expansion clause, the specified library list is searched in the specified order.

The current library list is selected by the selection clauses. If no library list clause is selected, or the selected library list is empty, then the library list contains the single name which is the library in which the cell containing the unbound instance is found (i.e., the parent cell s library).

\subsection*{13.3.1.6 The use clause}

The use clause specifies a specific binding for the selected cell. The syntax for the use clause is specified in Syntax 13-9.
```

use_clause ::= (From Annex A-A.1.2)
use [library_identifier.]cell_identifier[:config]

```

\section*{Syntax 13-9 Syntax for use clause}

A use clause can only be used in conjunction with an instance or cell selection clause. It specifies the exact library and cell to which a selected cell or instance is bound.

The use clause has no effect on the current value of the library list. It can be common in practice to specify multiple config rule statements, one of which specifies a binding and the other of which specifies a library list.

If the lib.cell being referred to by the use clause is a config which has the same name as a module/macromodule/primitive in the same library, then the optional : config suffix can be added to the lib. cell to specify the config explicitly.

If the library name is omitted, the library shall be inherited from the parent cell.

NOTE-The binding statement can create situations where the unbound instance's module name and the cell name to which it is bound are different.

\subsection*{13.3.2 Hierarchical configurations}

For situations where it is desirable to specify a special set of configuration rules for a subsection of a design, it is possible to bind a particular instance directly to a configuration using the binding clause:
```

instance top.a1.foo use lib1.foo:config;
// bind to the config foo in library lib1

```
specifies the instance top.a1.foo is to be replaced with the design hierarchy specified by the configuration libl.foo:config. The design statement in lib1.foo:config shall specify the actual binding for the instance top.a1.foo, and the rules specified in the config shall determine the configuration of all other subinstances under top.a1.foo.

It shall be an error for an instance clause to specify a hierarchical path to an instance which occurs within a hierarchy specified by another config.
```

config bot;
design lib1.bot;
default liblist lib1 lib2;
instance bot.a1 liblist lib3;
endconfig
config top;
design lib1.top;
default liblist lib2 lib1;
instance top.bot use lib1.bot:config;
instance top.bot.a1 liblist lib4;
// ERROR - can't set liblist for top.bot.al from this config
endconfig

```

\subsection*{13.4 Using libraries and configs}

The following subclause describes potential use-models for referencing configs on the command line. It is included for clarification purposes.

The traditional Verilog simulation use-model takes a file-based approach, where the source descriptions for all cells in the design are specified on the command line for each invocation of the tool. With the advent of compiled-code simulators, the configuration mechanism shall also support a use-model which allows for the source files to be pre-compiled and then for the pre-compiled design objects to be referenced on the command line. This subclause shall explain how configurations can be used in both of these scenarios.

\subsection*{13.4.1 recompiling in a single-pass use-model}

The single-pass use-model is the traditional use-model with which most users are familiar. In this use-model, all of the source description files shall be provided to the simulator via the command line, and only these source descriptions can be used to bind cell instances in the current design. A precompiling strategy in this scenario actually parses every cell description provided on the command line and maps it into the library without regard to whether the cell actually is used in the design. The tool can optionally check to see if the cell already exists in the library, and if it is up-to-date (i.e. the source description has not changed since the last time the cell was compiled) the tool can skip recompiling the cell. After all cells on the command line have been compiled, then the tool can locate the top-level cell (discussed in Section 12), and proceed down the hierarchy, binding each instance as it is encountered in the hierarchy.

NOTE With this use-model it is not necessary for library objects to persist from one tool invocation to another (although for performance considerations it is recommended they do).

\subsection*{13.4.2 Elaboration-time compiling in a single-pass use-model}

An alternate strategy which can be used with a single-pass tool is to parse the source files only to find the top-level module(s), without actually compiling anything into the library during this scanning process. Once the top-level module(s) has been found, then it can be compiled into the library, and the tool can proceed down the hierarchy, only compiling the source descriptions necessary to bind the design successfully. Based on the binding rules in place, only the source files which match the current library specification need to be parsed to find the current cell s source description to compile. As with the precompiled single-pass usemodel, it is not necessary for library cells to persist from one invocation to another using this strategy.

\subsection*{13.4.3 recompiling using a separate compilation tool}

When using a separate compilation tool, it is essential library cells persist, and the compiled forms shall therefore exist somewhere in the file system. The exact format and location for holding these compiled forms shall be vendor/tool-specific. Using this separate compiler strategy, the source descriptions shall be parsed and compiled into the library using one or more invocations of the compiler tool. The only restriction is all cells in a design shall be precompiled prior to binding the design (typically via an invocation of a separate tool). Using this strategy, the tool which actually does the binding only needs to be told the top-level module(s) of the design to be bound, and then it shall use the precompiled form of the cell description(s) from the library to determine the subinstances and descend hierarchically down the design binding each cell as it is located.

\subsection*{13.4.4 Command line considerations}

In each of the three preceding strategies, the binding rules can either be specified via a config, or the default rules (from the library map file) can be used. In the single-pass use-models, the config can be specified by including its source description file on the command line. In the case where the config includes a design statement, then the specified cell shall be the top-level module, regardless of the presence of any uninstantiated cells in the rest of the source files. When using a separate compilation tool, the tool which actually does the binding only needs to be given the lib.cell specification for the top-level cell(s) and/or the config to be used. In this strategy, the config itself shall also be precompiled.

\subsection*{13.5 Configuration examples}

Consider the following set of source descriptions:
Example:
\begin{tabular}{|c|c|c|c|}
\hline \begin{tabular}{l}
file top.v \\
module top (...);
\end{tabular} & \begin{tabular}{l}
file adder.v \\
module adder(...);
\end{tabular} & file adder.vg module adder(...); & \begin{tabular}{l}
file lib.map \\
library rtlLib top.v;
\end{tabular} \\
\hline & . // rtl & . // gate-level & library aLib adder.*; \\
\hline adder a1(...); & foo f1(...); & foo f1(...); & library gateLib \\
\hline adder a2(...); & foo f2 (...); & foo f2 (...); & adder.vg; \\
\hline endmodule & endmodule & endmodule & \\
\hline module foo(...); & module foo(...); & module foo(...); & \\
\hline // rtl & // rtl & . // gate-level & \\
\hline endmodule & endmodule & endmodule & \\
\hline
\end{tabular}

All of the examples in this section shall assume the top.v, adder.v and adder.vg files get compiled with the given lib. map file. This yields the following library structure:
```

rtlLib.top // from top.v
rtlLib.foo // from top.v
aLib.adder // from adder.v
aLib.foo // rtl from adder.v
gateLib.adder // from adder.vg
gateLib.foo // from adder.vg

```

\subsection*{13.5.1 Default configuration from library map file}

With no configuration, the libraries are searched according to the library declaration order in the library map file. This means all instances of module adder shall use aLib. adder (since aLib is the first library specified which contains a cell named adder), and all instances of module foo shall use rtlLib. foo (since rtlLib is the first library which contains foo).

\subsection*{13.5.2 Using the default clause}

To always use the foo definition from file adder. v , use the following simple configuration:
```

config cfg1;
design rtlLib.top
default liblist aLib rtlLib;
endconfig

```

The default liblist statement overrides the library search order in the lib. map file, so aLib is always searched before rtlLib. Since the gateLib library is not included in the liblist, the gate-level descriptions of adder and foo shall not be used.

To use the gate-level representations of adder and foo, add to the config as follows:
```

config cfg2;
design rtlLib.top
default liblist gateLib aLib rtlLib;
endconfig

```

This shall cause the gate representation always to be taken before the \(r t l\) representation, using the module definitions for adder and foo from adder.vg. The rtl view of top shall be taken since there is no gate representation available.

\subsection*{13.5.3 Using the cell clause}

To modify the config to use the rtl view of adder and the gate-level representation of foo from gateLib:
```

config cfg3;
design rtlLib.top
default liblist aLib rtlLib;
cell foo use gateLib.foo;
endconfig

```

The cell clause selects all cells named foo and explicitly binds them to the gate representation in gateLib.

\subsection*{13.5.4 Using the instance clause}

To modify the config so the top. al adder (and its descendants) use the gate representation and the top. a 2 adder (and its descendants) use the rtl representation from aLib:
```

config cfg4
design rtlLib.top
default liblist gateLib rtlLib;
instance top.a2 liblist aLib;
endconfig

```

Since the liblist is inherited, all of the descendants of top. a 2 inherit its liblist from the instance selection clause.

\subsection*{13.5.5 Using a hierarchical config}

Now suppose all this work has only been on the adder module by itself and a config which uses the \(r t l\) Lib. foo cell for \(f 1\), and the gateLib. foo cell for \(f 2\) has already been developed. Then use:
```

config cfg5;
design aLib.adder;
default liblist gateLib aLib;
instance adder.f1 liblist rtlLib;
endconfig

```

To use this configuration cfg 5 for the top. a 2 instance of adder and take the full default aLib adder for the top.a1 instance, use the following config:
```

config cfg6;
design rtlLib.top;
default liblist aLib rtlLib;
instance top.a2 use work.cfg5:config
endconfig

```

The binding clause specifies the work.cfg5:config configuration is to be used to resolve the bindings of instance top. a2 and its descendants. It is the design statement in config cfg 5 which defines the exact binding for the top. 22 instance itself. The rest of cfg 5 defines the rules to bind the descendants of top.a2. Notice the instance clause in cfg 5 is relative to its own top-level module, adder.

\subsection*{13.6 Displaying library binding information}

It shall be possible to display the actual library binding information for module instances during simulation. The format specifier \(\% 1\) or \(\% \mathrm{~L}\) shall print out the library.cell binding information for the module instance containing the display (or other textual output) command. This is similar to the \(\% \mathrm{~m}\) format specifier which prints out the hierarchical path name of the module containing it.

It shall also be able to use the VPI interface to display the binding information. The following new vpiProperties shall exist for objects of type vpiModule:
vpiUseBinding - the library. cell binding information for a module instance vpiLibrary - the library name into which the module was compiled vpiCell - the name of the cell bound to the module instance vpiConfig - the library.cell name of the config controlling the binding of the module instance

These properties shall be of string type, similar to the vpiName and vpiFullName properties.

\subsection*{13.7 Library mapping examples}

In the absence of a configuration, it is possible to perform basic control of the library searching order when binding a design.

When a config is used, the config overrides the rules specified here.

\subsection*{13.7.1 Using the command line to control library searching}

In the absence of a configuration, it shall be necessary for all compliant tools to provide a mechanism of specifying a library search order on the command line which overrides the default order from the library mapping file. This mechanism shall include specification of library names only, with the definitions of these libraries to be taken from the library mapping file.

NOTE It is recommended all compliant tools use "-L <library_name>" to specify this search order.

\subsection*{13.7.2 File path specification examples}

\section*{Example:}

Given the following set of files:
```

/proj/lib1/rtl/a.v
/proj/lib2/gates/a.v
/proj/lib1/rtl/b.v
/proj/lib2/gates/b.v

```

From the /proj library, the following absolute file_path_specs are resolved as shown:
```

/proj/lib*/*/a.v =/proj/lib1/rtl/a.v, /proj/lib2/gates/a.v
.../a.v =/proj/lib1/rtl/a.v, /proj/lib2/gates/a.v
/proj/.../b.v =/proj/lib1/rtl/b.v, /proj/lib2/gates/b.v
.../rtl/*.v =/proj/lib1/rtl/a.v, /proj/lib1/rtl/b.v

```

From the /proj/lib1 directory, the following relative file_path_specs are resolved as shown:
```

../lib2/gates/*.v = /proj/lib2/gates/a.v, /proj/lib2/gates/b.v
./rtl/?.v = /proj/lib1/rtl/a.v, /proj/lib1/rtl/b.v
./rtl/ = /proj/lib1/rtl/a.v, /proj/lib1/rtl/b.v

```

\subsection*{13.7.3 esolving multiple path specifications}

Example:
```

library lib1 "/proj/lib1/foo*.v";
library lib2 "/proj/lib1/foo.v";
library lib3 "../lib1/";
library lib4 "/proj/lib1/*ver.v";

```

When evaluated from the directory / proj / tb directory, the following source files shall map into the specified library:
../lib1/foobar.v - lib1 // potentially matches lib1 and lib3. Since lib1 includes a filename and lib3 only specifies a directory; lib1 takes precedence
/proj/lib1/foo.v - lib2 // takes precedence over lib1 and lib3 path specifications
/proj/lib1/bar.v - lib3
/proj/lib1/barver.v - lib4 // takes precedence over lib3 path specification
/proj/lib1/foover.v - ERROR // matches lib1 and lib4
/test/tb/tb.v - work// does not match any library specifications.

\section*{14. Specify blocks}

Two types of HDL constructs are often used to describe delays for structural models such as ASIC cells. They are as follows:

Distributed delays, which specify the time it takes events to propagate through gates and nets inside the module (see 7.14)
Module path delays, which describe the time it takes an event at a source (input port or inout port) to propagate to a destination (output port or inout port)

This clause describes how paths are specified in a module and how delays are assigned to these paths.

\subsection*{14.1 Specify block declaration}

A block statement called the specify block is the vehicle for describing paths between a source and a destination and for assigning delays to these paths. The syntax for specify block is shown in Syntax 14-1.
```

specify_item ::= (From Annex A - A.7.1)
specparam_declaration
pulsestyle_declaration
|sowcancelled_declaration
path_declaration
system_timing_check

```

\section*{Syntax 14-1 Syntax of specify block}

The specify block shall be bounded by the keywords specify and endspecify, and it shall appear inside a module declaration. The specify block can be used to perform the following tasks:

Describe various paths across the module.
Assign delays to those paths.
Perform timing checks to ensure that events occurring at the module inputs satisfy the timing constraints of the device described by the module (see Section 15).

The paths described in the specify block, called module paths, pair a signal source with a signal destination. The source may be unidirectional (an input port) or bidirectional (an inout port) and is referred to as the module path source. Similarly, the destination may be unidirectional (an output port) or bidirectional (an inout port) and is referred to as the module path destination.

\section*{Example:}
```

specify
specparam tRise_clk_q = 150, tFall_clk_q = 200;
specparam tSetup = 70;
(clk => q) = (tRise_clk_q, tFall_clk_q);
\$setup(d, posedge clk, tSetup);
endspecify

```

The first two lines following the keyword specify declare specify parameters, which are discussed in 3.11.3. The line following the declarations of specify parameters describes a module path and assigns delays to that module path. The specify parameters determine the delay assigned to the module path. Specifying module paths is presented in 14.2. Assigning delays to module paths is discussed in 14.3. The line preceding the keyword endspecify instantiates one of the system timing checks, which are discussed further in Section 15.

\subsection*{14.2 Module path declarations}

There are two steps required to set up module path delays in a specify block:
a) Describe the module paths
b) Assign delays to those paths (see 14.3)

The syntax of the module path declaration is described in Syntax 14-2.
```

path_declaration ::= (From Annex A - A.7.2)
simple_path_declaration ;
edge_sensitive_path_declaration;
state_dependent_path_declaration ;

```

\section*{Syntax 14-2 Syntax of the module path declaration}

A module path may be described as a simple path, an edge sensitive path, or a state dependent path. A module path shall be defined inside a specify block as a connection between a source signal and a destination signal. Module paths can connect any combination of vectors and scalars.

\section*{Example:}

Figure 35 illustrates a circuit with module path delays. More than one source (A, B, C, and D) may have a module path to the same destination (Q), and different delays may be specified for each input to output path.


Figure 35-Module path delays

\subsection*{14.2.1 Module path restrictions}

Module paths have the following restrictions:
The module path source shall be a net that is connected to a module input port or inout port.
The module path destination shall be a net or variable that is connected to a module output port or inout port.
The module path destination shall have only one driver inside the module.

\subsection*{14.2.2 Simple module paths}

The syntax for specifying a simple module path is given in Syntax 14-3.
```

simple_path_declaration ::= (From Annex A - A.7.2)
parallel_path_description = path_delay_value
| full_path_description = path_delay_value
parallel_path_description ::=
( specify_input_terminal_descriptor [ polarity_operator ] =>
specify_output_terminal_descriptor )
full_path_description ::=
( list_of_path_inputs [ polarity_operator ] *> list_of_path_outputs )
list_of_path_inputs ::=
specify_input_terminal_descriptor $\{$, specify_input_terminal_descriptor \}
list_of_path_outputs ::=
specify_output_terminal_descriptor \{ , specify_output_terminal_descriptor \}
specify_input_terminal_descriptor ::= (From Annex A - A.7.3)
input_identifier
input_identifier [ constant_expression ]
input_identifier [ range_expression]
specify_output_terminal_descriptor ::=
output_identifier
output_identifier [ constant_expression ]
output_identifier [ range_expression ]
input_identifier ::=
input_port_identifier | inout_port_identifier
output_identifier ::=
output_port_identifier | inout_port_identifier
polarity_operator ::= (From Annex A - A.7.4)
$+\mid-$

```

\section*{Syntax 14-3 Syntax for simple module path}

Simple path can be declared in one of the two forms:
```

source *> destination
source => destination

```

The symbols *> and => each represent a different kind of connection between the module path source and the module path destination. The operator *> establishes a full connection between source and destination. The operator \(\Rightarrow\) establishes a parallel connection between source and destination. Refer to 14.2.5 for a description of full connection and parallel connection paths.

\section*{Example:}

The following three examples illustrate valid simple module path declarations.
```

(A => Q) = 10;
(B => Q) = (12);
(C, D *> Q) = 18;

```

\subsection*{14.2.3 Edge-sensitive paths}

When a module path is described using an edge transition at the source, it is called an edge-sensitive path. The edge-sensitive path construct is used to model the timing of input to output delays, which only occur when a specified edge occurs at the source signal.

The syntax of the edge-sensitive path declaration is shown in Syntax 14-4.
```

edge_sensitive_path_declaration ::= (From Annex A - A.7.4)
parallel_edge_sensitive_path_description = path_delay_value
| full_edge_sensitive_path_description = path_delay_value
parallel_edge_sensitive_path_description ::=
([ edge_identifier ] specify_input_terminal_descriptor =>
specify_output_terminal_descriptor [ polarity_operator ] : data_source_expression )
full_edge_sensitive_path_description ::=
([ edge_identifier ] list_of_path_inputs *>
list_of_path_outputs [ polarity_operator ] : data_source_expression )
data_source_expression ::=
expression
edge_identifier ::=
posedge negedge

```

\section*{Syntax 14-4 Syntax of the edge-sensitive path declaration}

The edge identifier may be one of the keywords posedge or negedge, associated with an input terminal descriptor, which may be any input port or inout port. If a vector port is specified as the input terminal descriptor, the edge transition shall be detected on the least significant bit. If the edge transition is not specified, the path shall be considered active on any transition at the input terminal.

An edge-sensitive path may be specified with full connections ( \(*>\) ) or parallel connections ( \(=>\) ). For parallel connections (=>), the destination shall be any scalar output or inout port or the bit-select of a vector output or inout port. For full connections \((*>)\), the destination shall be a list of one or more of the vector or scalar output and inout ports, and bit-selects or part-selects of vector output and inout ports. Refer to 14.2.5 for a description of parallel paths and full connection paths.

The data source expression is an arbitrary expression, which serves as a description of the flow of data to the path destination. This arbitrary data path description does not affect the actual propagation of data or events through the model; how an event at the data path source propagates to the destination depends on the internal logic of the module. The polarity operator describes whether the data path is inverting or noninverting.

\section*{Examples:}

Example 1 The following example demonstrates an edge-sensitive path declaration with a positive polarity operator:
( posedge clock \(\Rightarrow\) ( out \(+:\) in ) ) \(=(10,8)\);
In this example, at the positive edge of clock, a module path extends from clock to out using a rise delay of 10 and a fall delay of 8 . The data path is from in to out, and in is not inverted as it propagates to out.

Example 2 The following example demonstrates an edge-sensitive path declaration with a negative polarity operator:
```

( negedge clock[ 0] => ( out -: in ) ) = (10, 8);

```

In this example, at the negative edge of clock[ 0], a module path extends from clock[ 0] to out using a rise delay of 10 and a fall delay of 8 . The data path is from in to out, and in is inverted as it propagates to out.

Example 3 The following example demonstrates an edge-sensitive path declaration with no edge identifier:
( clock \(=>(\) out \(:\) in \()=(10,8) ;\)
In this example, at any change in clock, a module path extends from clock to out.

\subsection*{14.2.4 State-dependent paths}

A state-dependent path makes it possible to assign a delay to a module path that affects signal propagation delay through the path only if specified conditions are true.

A state-dependent path description includes the following items:
A conditional expression that, when evaluated true, enables the module path
A module path description
A delay expression that applies to the module path

The syntax for the state-dependent path declaration is shown in Syntax 14-5.
state_dependent_path_declaration ::= (From Annex A-A.7.4)
if ( module_path_expression ) simple_path_declaration
if ( module_path_expression ) edge_sensitive_path_declaration
ifnone simple_path_declaration

\section*{Syntax 14-5 Syntax of state-dependent paths}

\subsection*{14.2.4.1 Conditional expression}

The operands in the conditional expression shall be constructed from the following:
Scalar or vector module input ports or inout ports or their bit-selects or part-selects
Locally defined variables or nets or their bit-selects or part-selects
Compile time constants (constant numbers and specify parameters)

Table 46 contains a list of valid operators that may be used in conditional expressions:
Table 46-List of valid operators in state dependent path delay expression
\begin{tabular}{|l|l|l|l|}
\hline Operator & \multicolumn{1}{|c|}{ Description } & Operator & \multicolumn{1}{c|}{ Description } \\
\hline \hline\(\sim\) & bit-wise negation & \(\&\) & reduction and \\
\hline\(\&\) & bit-wise and & \(\mid\) & reduction or \\
\hline \(\mid\) & bit-wise or & \(\wedge\) & reduction xor \\
\hline\(\wedge\) & bit-wise xor & \(\sim \&\) & reduction nand \\
\hline\(\wedge \sim \sim \wedge\) & bit-wise xnor & \(\sim \mid\) & reduction nor \\
\hline\(==\) & logical equality & \(\wedge \sim \sim \wedge\) & reduction xnor \\
\hline\(!=\) & logical inequality & \(\}\) & concatenation \\
\hline\(\& \&\) & logical and & \(\{\}\}\) & replication \\
\hline\(\|\) & logical or & \(?:\) & conditional \\
\hline\(!\) & logical not & & \\
\hline
\end{tabular}

A conditional expression shall evaluate to true (1) for the state-dependent path to be assigned a delay value. If the conditional expression evaluates to \(x\) or \(z\), it shall be treated as true. If the conditional expression evaluates to multiple bits, the least significant bit shall represent the result. The conditional expression can have any number of operands and operators.

\subsection*{14.2.4.2 Simple state-dependent paths}

If the path description of a state-dependent path is a simple path, then it is called a simple state-dependent path. The simple path description is discussed in 14.2.2.

\section*{Examples:}

Example 1 The following example uses state-dependent paths to describe the timing of an XOR gate.
```

module XORgate (a, b, out);
input a, b:
output out;
xor x1 (out, a, b);
specify
specparam noninvrise = 1, noninvfall = 2
specparam invertrise = 3, invertfall = 4;
if (a) (b => out) = (invertrise, invertfall);
if (b) (a => out) = (invertrise, invertfall);
if (~a)(b => out) = (noninvrise, noninvfall);
if (~b)(a => out) = (noninvrise, noninvfall);
endspecify
endmodule

```

In this example, first two state-dependent paths describe a pair of output rise and fall delay times when the XOR gate ( x 1 ) inverts a changing input. The last two state-dependent paths describe another pair of output rise and fall delay times when the XOR gate buffers a changing input.

Example 2 The following example models a partial ALU. The state-dependent paths specify different delays for different ALU operations.
```

module ALU (o1, i1, i2, opcode);
input [ 7:0] i1, i2;
input [ 2:1] opcode;
output [ 7:0] o1;
//functional description omitted
specify
// add operation
if (opcode == 2'b00) (i1,i2 *> o1) = (25.0, 25.0);
// pass-through il operation
if (opcode == 2' b01) (i1 => o1) = (5.6, 8.0);
// pass-through i2 operation
if (opcode == 2' b10) (i2 => o1) = (5.6, 8.0);
// delays on opcode changes
(opcode => o1) = (6.1, 6.5);
endspecify
endmodule

```

In the preceding example, the first three path declarations declare paths extending from operand inputs i1 and i2 to the ol output. The delays on these paths are assigned to operations on the basis of the operation specified by the inputs on opcode. The last path declaration declares a path from the opcode input to the o1 output.

\subsection*{14.2.4.3 Edge-sensitive state-dependent paths}

If the path description of a state-dependent path describes an edge-dependent path, then the state-dependent path is called an edge-sensitive state-dependent path. The edge-sensitive paths are discussed in 14.2.3.

Different delays can be assigned to the same edge-sensitive path as long as the following criteria are met:
The edge, condition, or both make each declaration unique.
The port is referenced in the same way in all path declarations (entire port, bit-select, or part-select).

\section*{Examples:}

\section*{Example 1}
```

if ( !reset \&\& !clear )
( posedge clock => ( out +: in ) ) = (10, 8) ;

```

In this example, if the positive edge of clock occurs when reset and clear are low, and a module path extends from clock to out using a rise delay of 10 and a fall delay of 8 .

Example 2 The following example shows four edge-sensitive path declarations. Note that each path has a unique edge or condition.
```

specify
( posedge clk => ( q[ 0] : data ) ) = (10, 5);
( negedge clk => ( q[ 0] : data ) ) = (20, 12);
if (reset)
( posedge clk => ( q[ 0] : data ) ) = (15, 8);
if (!reset \&\& cntrl)
( posedge clk => ( q[ 0] : data ) ) = (6, 2);
endspecify

```

Example 3 The two state-dependent path declarations shown below are not legal because even though they have different conditions, the destinations are not specified in the same way: the first destination is a partselect, the second is a bit-select.
```

specify
if (reset)
(posedge clk => (q[ 3:0] :data)) = (10,5);
if (!reset)
(posedge clk => (q[ 0] :data)) = (15,8);
endspecify

```

\subsection*{14.2.4.4 The ifnone condition}

The ifnone keyword is used to specify a default state-dependent path delay when all other conditions for the path are false. The ifnone condition shall specify the same module path source and destination as the statedependent module paths. The following rules apply to module paths specified with the ifnone condition:

Only simple module paths may be described with an ifnone condition.
The state-dependent paths that correspond to the ifnone path may be either simple module paths or edge-sensitive paths.
If there are no corresponding state-dependent module paths to the ifnone module path, then the ifnone module path shall be treated the same as an unconditional simple module path.
It is illegal to specify both an ifnone condition for a module path and an unconditional simple module path for the same module path.

\section*{Examples:}

Example 1 The following are valid state-dependent path combinations.
```

if (C1) (IN => OUT) = (1,1);
ifnone (IN => OUT) = (2,2);
// add operation
if (opcode == 2'b00) (i1,i2 *> o1) = (25.0, 25.0);
// pass-through il operation
if (opcode == 2'b01) (i1 => o1) = (5.6, 8.0);
// pass-through i2 operation
if (opcode == 2'b10) (i2 => o1) = (5.6, 8.0);
// all other operations
ifnone (i2 => o1) = (15.0, 15.0);
(posedge CLK => (Q +: D)) = (1,1);
ifnone (CLK => Q) = (2,2);

```

Example 2 The following module path description combination is illegal because it combines a statedependent path using an ifnone condition and an unconditional path for the same module path.
```

if (a) (b => out) = (2,2);
if (b) (a => out) = (2,2);
ifnone (a => out) = (1,1);
(a => out) = (1,1);

```

\subsection*{14.2.5 Full connection and parallel connection paths}

The operator * \(>\) shall be used to establish a full connection between source and destination. In a full connection, every bit in the source shall connect to every bit in the destination. The module path source need not have the same number of bits as the module path destination.

The full connection can handle most types of module paths, since it does not restrict the size or number of source signals and destination signals. The following situations require the use of full connections:

To describe a module path between a vector and a scalar
To describe a module path between vectors of different sizes
To describe a module path with multiple sources or multiple destinations in a single statement (see 14.2.6)

The operator \(=>\) shall be used to establish a parallel connection between source and destination. In a parallel connection, each bit in the source shall connect to one corresponding bit in the destination. Parallel module paths can be created only between sources and destinations that contain the same number of bits.

Parallel connections are more restrictive than full connections. They only connect one source to one destination, where each signal contains the same number of bits. Therefore, a parallel connection may only be used to describe a module path between two vectors of the same size. Since scalars are one bit wide, either \(*>\) or \(=>\) may be used to set up bit-to-bit connections between two scalars.

\section*{Examples:}

Example 1 Figure 36 illustrates how a parallel connection differs from a full connection between two 4-bit vectors.
\begin{tabular}{|c|c|}
\hline arallel module path & Full module path \\
\hline  &  \\
\hline \multicolumn{2}{|c|}{\(\mathrm{N}=\) number of bits \(=4\)} \\
\hline Number of paths \(=\mathrm{N}=\) 4
\(\square\) Use \(=>\) to define path bit-to-bit connections & Number of paths = N * \(\mathrm{N}=\) Use \(\square\) to define path bit-to-vector connections \\
\hline
\end{tabular}

Figure 36-The difference between parallel and full connection paths
Example 2 The following example shows module paths for a \(2: 1\) multiplexor with two 8-bit inputs and one 8 -bit output.
```

module mux8 (in1, in2, s, q) ;
output [ 7:0] q;
input [ 7:0] in1, in2;
input s;
// Functional description omitted ...
specify
(in1 => q) = (3, 4) ;
(in2 => q) = (2, 3) ;
(s *> q) = 1;
endspecify
endmodule

```

The module path from \(s\) to \(q\) uses a full connection ( \(*>\) ) because it connects a scalar source the 1-bit select line to a vector destination the 8-bit output bus. The module paths from both input lines in1 and in2 to \(q\) use a parallel connection ( \(=>\) ) because they set up parallel connections between two 8-bit buses.

\subsection*{14.2.6 Declaring multiple module paths in a single statement}

Multiple module paths may be described in a single statement by using the symbol \({ }^{*}>\) to connect a commaseparated list of sources to a comma-separated list of destinations. When describing multiple module paths in one statement, the lists of sources and destinations may contain a mix of scalars and vectors of any size.

The connection in a multiple module path declaration is always a full connection.

\section*{Example:}
\[
\left(\mathrm{a}, \mathrm{~b}, \mathrm{c}^{*}>\mathrm{q} 1, \mathrm{q} 2\right)=10 ;
\]
is equivalent to the following six individual module path assignments:
\((\mathrm{a} *>\mathrm{q} 1)=10 ;\)
\((\mathrm{b} *>\mathrm{q} 1)=10 ;\)
\((\mathrm{c} *>\mathrm{q} 1)=10 ;\)
\((\mathrm{a} *>\mathrm{q} 2)=10 ;\)
\((\mathrm{b} *>\mathrm{q} 2)=10 ;\)
\((\mathrm{c} *>\mathrm{q} 2)=10 ;\)

\subsection*{14.2.7 Module path polarity}

The polarity of a module path is an arbitrary specification indicating whether or not the direction of a signal transition is inverted as it propagates from the input to the output. This arbitrary polarity description does not affect the actual propagation of data or events through the model; how a rise or a fall at the source propagates to the destination depends on the internal logic of the module.

Module paths may specify any of three polarities:
Unknown polarity
Positive polarity
Negative polarity

\subsection*{14.2.7.1 Unknown polarity}

By default, module paths shall have unknown polarity that is, a transition at the path source may propagate to the destination in an unpredictable way, as follows:

A rise at the source may cause either a rise transition, a fall transition, or no transition at the destination.
A fall at the source may cause either a rise transition, a fall transition, or no transition at the destination.

A module path specified either as a full connection or a parallel connection, but without a polarity operator + or -, shall be treated as a module path with unknown polarity.

\subsection*{14.2.7.2 ositive polarity}

For module paths with positive polarity, any transition at the source may cause the same transition at the destination, as follows:

A rise at the source may cause either a rise transition or no transition at the destination.
A fall at the source may cause either a fall transition or no transition at the destination.
A module path with positive polarity shall be specified by prefixing the + polarity operator to \(=>\) or \(*>\).

\subsection*{14.2.7.3 Negative polarity}

For module paths with negative polarity, any transition at the source may cause the opposite transition at the destination, as follows:

A rise at the source may cause either a fall transition or no transition at the destination. A fall at the source may cause either a rise transition or no transition at the destination.

A module path with negative polarity shall be specified by prefixing the - polarity operator to \(=>\) or \(*>\).

\section*{Examples:}

The following examples show each type of path polarity:
```

// Positive polarity
(In1 +=> q) = In to q ;
(s +*> q) = s_\overline{to_\overline{q}};
// Negative polarity
(In1 -=> q) = In_to_q ;
(s -*> q) = s_to_q ;
// Unknown polarity
(In1 => q) = In_to_q ;
(s *> q) = s_to_q ;

```

\subsection*{14.3 Assigning delays to module paths}

The delays that occur at the module outputs where paths terminate shall be specified by assigning delay values to the module path descriptions. The syntax for specifying delay values is shown in Syntax 14-6.
```

path_delay_value ::= (From Annex A - A.7.4)
list_of_path_delay_expressions
| ( list_of_path_delay_expressions )
list_of_path_delay_expressions ::=
t_path_delay_expression
| trise_path_delay_expression,tfall_path_delay_expression
| trise_path_delay_expression, tfall_path_delay_expression, tz_path_delay_expression
| t01_path_delay_expression,t10_path_delay_expression, t0z_path_delay_expression,
tz1_path_delay_expression, t1z_path_delay_expression, tz0_path_delay_expression
| t01_path_delay_expression,t10_path_delay_expression,t0z_path_delay_expression,
tz1_path_delay_expression, t1z_path_delay_expression, tz0_path_delay_expression
t0x_path_delay_expression,tx1_path_delay_expression, t1x_path_delay_expression,
tx0_path_delay_expression,txz_path_delay_expression, tzx_path_delay_expression
t_path_delay_expression ::=
path_delay_expression

```

In module path delay assignments, a module path description (see 14.2) is specified on the left-hand side, and one or more delay values are specified on the right-hand side. The delay values may be optionally enclosed in a pair of parentheses. There may be one, two, three, six, or twelve delay values assigned to a module path, as described in 14.3.1. The delay values shall be constant expressions containing literals or specparams, and there may be a delay expression of the form min:typ:max.

\section*{Example:}
```

specify
// Specify Parameters
specparam tRise_clk_q = 45:150:270, tFall_clk_q=60:200:350;
specparam tRise_Control = 35:40:45, tFall_control=40:50:65;
// Module Path Assignments
(clk => q) = (tRise_clk_q, tFall_clk_q);
(clr, pre *> q) = (tRise_control, tFall_control);
endspecify

```

In the example above, the specify parameters declared following the specparam keyword specify values for the module path delays. The module path assignments assign those module path delays to the module paths.

\subsection*{14.3.1 Specifying transition delays on module paths}

Each path delay expression may be a single value representing the typical delay or a colon-separated list of three values representing a minimum, typical, and maximum delay, in that order. If the path delay expression results in a negative value, it shall be treated as zero. Table 47 describes how different path delay values shall be associated with various transitions. The path delay expression names refer to the names used in Syntax 14-6.

Table 47-Associating path delay expressions with transitions
\begin{tabular}{|c|c|c|c|c|c|}
\hline \multirow[b]{2}{*}{Transitions} & \multicolumn{5}{|r|}{Number of path delay expressions specified} \\
\hline & 1 & 2 & 3 & 6 & 12 \\
\hline \(0->1\) & t & trise & trise & t01 & t01 \\
\hline \(1->0\) & t & tfall & tfall & t10 & t10 \\
\hline \(0->\mathrm{z}\) & t & trise & tz & t0z & t0z \\
\hline \(\mathrm{z} \mathrm{->} 1\) & t & trise & trise & tz1 & tz1 \\
\hline \(1->\mathrm{z}\) & t & tfall & tz & tlz & tlz \\
\hline \(\mathrm{z} \mathrm{->} 0\) & t & tfall & tfall & tz0 & tz0 \\
\hline \(0->\mathrm{x}\) & * & * & * & * & t0x \\
\hline \(\mathrm{x}->1\) & * & * & * & * & tx1 \\
\hline \(1->\mathrm{x}\) & * & * & * & * & t1x \\
\hline \(\mathrm{x}->0\) & * & * & * & * & tx0 \\
\hline X -> z & * & * & * & * & txz \\
\hline \(\mathrm{z} \mathrm{->} \mathrm{x}\) & * & * & * & * & tzx \\
\hline
\end{tabular}
* See 14.3.2.

\section*{Example:}
```

// one expression specifies all transitions
(C => Q) = 20;
(C => Q) = 10:14:20;
// two expressions specify rise and fall delays
specparam tPLH1 = 12, tPHL1 = 25;
specparam tPLH2 = 12:16:22, tPHL2 = 16:22:25;
(C => Q) = ( tPLH1, tPHL1 ) ;
(C => Q) = ( tPLH2, tPHL2 ) ;
// three expressions specify rise, fall, and z transition delays
specparam tPLH1 = 12, tPHL1 = 22, tPz1 = 34;
specparam tPLH2 = 12:14:30, tPHL2 = 16:22:40, tPz2 = 22:30:34;
(C => Q) = (tPLH1, tPHL1, tPz1);
(C => Q) = (tPLH2, tPHL2, tPz2);
// six expressions specify transitions to/from 0, 1, and z
specparam t01 = 12, t10=16, t0z = 13,
tz1 = 10, t1z = 14, tz0 = 34 ;
(C => Q) = ( t01, t10, t0z, tz1, t1z, tz0) ;
specparam T01 = 12:14:24, T10 = 16:18:20, T0z = 13:16:30 ;
specparam Tz1 = 10:12:16, T1z = 14:23:36, Tz0 = 15:19:34 ;
(C => Q) = ( T01, T10, T0z, Tz1, T1z, Tz0) ;
// twelve expressions specify all transition delays explicitly
specparam t01=10, t10=12, t0z=14, tz1=15, t1z=29, tz0=36,
t0x=14, tx1=15, t1x=15, tx0=14, txz=20, tzx=30 ;
(c => Q) = (t01, t10, t0z, tz1, t1z, tz0,
t0x, tx1, t1x, tx0, txz, tzx) ;

```

\subsection*{14.3.2 Specifying \(x\) transition delays}

If the x transition delays are not explicitly specified, the calculation of delay values for x transitions is based on the following two pessimistic rules:

Transitions from a known state to \(x\) shall occur as quickly as possible that is, the shortest possible delay shall be used for any transition to \(x\).
Transitions from x to a known state shall take as long as possible that is, the longest possible delay shall be used for any transition from \(x\).

Table 48 presents the general algorithm for calculating delay values for x transitions, along with specific examples. The following two groups of x transitions are represented in the table:
a) Transition from a known state \(s\) to \(x\) : \(s\) x
b) Transition from \(x\) to a known state \(s\) : \(x ~ s\)

Table 48-Calculating delays for x transitions
\begin{tabular}{|c|c|}
\hline X transition & Delay value \\
\hline \multicolumn{2}{|r|}{General algorithm} \\
\hline s -> x & minimum ( \(\mathrm{s}->\) other known signals) \\
\hline x -> s & maximum (other known signals -> s) \\
\hline \multicolumn{2}{|r|}{Specific transitions} \\
\hline \(0->\mathrm{x}\) & minimum ( \(0->\mathrm{z}\) delay, \(0->1\) delay \()\) \\
\hline \(1->\mathrm{x}\) & minimum ( \(1->\mathrm{z}\) delay, \(1->0\) delay \\
\hline \(\mathrm{z} \mathrm{->} \mathrm{x}\) & minimum (z -> 1 delay, z -> 0 delay) \\
\hline x -> 0 & maximum (z -> 0 delay, 1 -> 0 delay) \\
\hline x -> 1 & maximum (z -> 1 delay, 0 -> 1 delay) \\
\hline x -> z & maximum ( \(1->\mathrm{z}\) delay, \(0->\mathrm{z}\) delay ) \\
\hline \multicolumn{2}{|r|}{Usage: \(\quad(C=>Q)=(5,12,17,10,6,22)\);} \\
\hline \(0->\mathrm{x}\) & minimum \((17,5)=5\) \\
\hline \(1->\mathrm{x}\) & minimum \((6,12)=6\) \\
\hline \(\mathrm{z} \mathrm{->} \mathrm{x}\) & minimum \((10,22)=10\) \\
\hline x -> 0 & maximum \((22,12)=22\) \\
\hline x -> 1 & \(\operatorname{maximum}(10,5)=10\) \\
\hline x -> z & maximum \((6,17)=17\) \\
\hline
\end{tabular}

\subsection*{14.3.3 Delay selection}

The simulator shall determine the proper delay to use when a specify path output must be scheduled to transition. There may be specify paths to the output from more than one input, and the simulator must decide which specify path to use.

The simulator shall do this by first determining which specify paths to the output are active. Active specify paths are those whose input has transitioned most recently in time, and which have either no condition or whose conditions are true. In the presence of simultaneous input transitions, it is possible for many specify paths to an output to be simultaneously active.

Once the active specify paths are identified, a delay must be selected from among them. This is done by comparing the correct delay for the specific transition being scheduled from each specify path, and choosing the smallest.

\section*{Examples:}

Example 1:
\[
\begin{aligned}
& (A=Y)=(6,9) ; \\
& (B=Y)=(5,11) ;
\end{aligned}
\]

For a Y transition from 0 to 1 , if A transitioned more recently than B a delay of 6 will be chosen. But if B transitioned more recently than A, a delay of 5 will be chosen. And if the last time they transitioned A and B did so simultaneously, then the smallest of the two rise delays would be chosen, which is the rise delay from \(B\) of 5 . The fall delay from \(A\) of 9 would be chosen if \(Y\) was instead to transition from 1 to 0 .

Example 2:
```

if (MODE < 5) (A => Y) = (5, 9);
if (MODE < 4) (A => Y) = (4, 8);
if (MODE < 3) (A => Y) = (6, 5);
if (MODE < 2) (A => Y) = (3, 2);
if (MODE < 1) (A => Y) = (7, 7);

```

Anywhere from zero to five of these specify paths might be active depending upon the value of MODE. For instance, when MODE is 2 the first three specify paths are active. A rise transition would select a delay of 4, because that is the smallest rise delay among the first three. A fall transition would select a delay of 5, because that is the smallest fall delay among the first three.

\subsection*{14.4 Mixing module path delays and distributed delays}

If a module contains module path delays and distributed delays (delays on primitive instances within the module), the larger of the two delays for each path shall be used.

\section*{Examples:}

Example 1 Figure 37 illustrates a simple circuit modeled with a combination of distributed delays and path delays (only the D input to Q output path is illustrated). Here, the delay on the module path from input D to output \(\mathrm{Q}=22\), while the sum of the distributed delays \(=0+1=1\). Therefore, a transition on Q caused by a transition on \(D\) will occur 22 time units after the transition on \(D\).


Figure 37-Module path delays longer than distributed delays
Example 2 In Figure 38, the delay on the module path from \(D\) to \(Q=22\), but the distributed delays along that module path now add up to \(10+20=30\). Therefore, an event on \(Q\) caused by an event on \(D\) will occur 30 time units after the event on \(D\).


Figure 38-Module path delays shorter than distributed delays

\subsection*{14.5 Driving wired logic}

Module path output nets shall not have more than one driver within the module. Therefore, wired logic is not allowed at module path outputs.

Figure 40 illustrates a violation of this wired-output rule and a method of avoiding the rule violation.


Figure 39-Legal and illegal module paths
In Figure 39 (a), any module path to \(S\) is illegal because the path destination has two drivers.
Assuming signal S in Figure 39 (a) is a wired-and, this limitation can be circumvented by replacing wired logic with gated logic to create a single driver to the output. Figure 39 (b) shows how adding a third and gate the shaded gate solves the problem for the module in Figure 39 (a).

The example in Figure 40 is also illegal. In this example, when the outputs \(Q\) and \(R\) are wired together, it creates a condition where both paths have multiple drivers from within the same module.


Figure 40-IIlegal module paths
Although multiple output drivers to a path destination are prohibited inside the same module, they are allowed outside the module. The example in Figure 41 is legal since \(Q\) and \(R\) each have only one driver within the module in which the module paths are specified.


Figure 41—Legal module paths

\subsection*{14.6 Detailed control of pulse filtering behavior}

Two consecutive scheduled transitions closer together in time than the module path delay is deemed a pulse. By default, pulses on a module path output are rejected. Consecutive transitions cannot be closer together than the module path delay, and this is known as the inertial delay model of pulse propagation.

Pulse width ranges control how to handle a pulse presented at a module path output. They are:
A pulse width range for which a pulse shall be rejected
A pulse width range for which a pulse shall be allowed to propagate to the path destination
A pulse width range for which a pulse shall generate a logic \(x\) on the path destination

Two pulse limit values define the pulse width ranges associated with each module path transition delay. The pulse limit values are called the error limit (e-limit) and the rejection limit (r-limit). The e-limit must always be at least as large as the r-limit. Pulses greater than or equal to the e-limit pass unfiltered. Pulses less than the e-limit but greater than or equal to the r-limit are filtered to X . Pulses less than the r -limit are rejected and no pulse emerges. By default, both the e-limit and the r-limit are set equal to the delay. These default values yield full inertial pulse behavior, rejecting all pulses smaller than the delay.

\section*{Example:}


The rise delay from input A to output \(Y\) is 7, and the fall delay is 9 . By default, the e-limit and the \(r\)-limit for the rise delay are both 7 . The e-limit and the r -limit for the fall delay are both 9 . The pulse limits associated with the delay forming the trailing edge of the pulse determine if and how the pulse should be filtered. Waveform \(Y^{\prime}\) shows the waveform resulting from no pulse filtering. The width of the pulse is 2 , which is less than the reject limit for the rise delay of 7, and so the pulse is filtered as shown in waveform Y.

There are three ways to modify the pulse limits from their default values. First, the Verilog language provides the PATHPULSE\$ specparam to modify the pulse limits from their default values. Second, invocation options can specify percentages applying to all module path delays to form the corresponding e-limits and r-limits. Third, SDF annotation can individually annotate the e-limit and r-limit of each module path transition delay.

\subsection*{14.6.1 Specify block control of pulse limit values}

Pulse limit values may be set from within the specify block with the PATHPULSE \(\$\) specparam. The syntax for using PATHPULSE\$ to specify the reject limit and error limit values is given in Syntax 14-7.
```

pulse_control_specparam ::= (From Annex A - A.2.4)
PATHPULSES = (reject_limit_value [ , error_limit_value ] );
|PATHPULSE$specify_input_terminal_descriptor$specify_output_terminal_descriptor
=( reject_limit_value [, error_limit_value ]);
error_limit_value ::=
limit_value
reject_limit_value ::=
limit_value
limit_value ::=
constant_mintypmax_expression

```

If only the reject limit value is specified, it shall apply to both the reject limit and the error limit.
The reject limit and error limit may be specified for a specific module path. When no module path is specified, the reject limit and error limit shall apply to all module paths defined in a module. If both path-specific PATHPULSE \(\$\) specparams and a non-path-specific PATHPULSE \(\$\) specparam appear in the same module, then the path-specific specparams shall take precedence for the specified paths.

The module path input terminals and output terminals shall conform to the rules for module path inputs and outputs, with the following restriction: the terminals may not be a bit-select or part-select of a vector.

When a module path declaration declares multiple paths, the PATHPULSES specparam shall only be specified for the first path input terminal and the first path output terminal. The reject limit and error limit specified shall apply to all other paths in the multiple path declaration. A PATHPULSE\$ specparam which specifies anything other than the first path input and path output terminals shall be ignored.

\section*{Example:}

In the following example, the path ( \(c l k=>q\) ) acquires a reject limit of 2 and an error limit of 9 , as defined by the first PATHPULSE\$ declaration. The paths ( \(c l r^{*}>q\) ) and ( \(p r e^{*}>q\) ) receive a reject limit of 0 and an error limit of 4, as specified by the second PATHPULSE\$ declaration. The path (data=>q) is not explicitly defined in any of the PATHPULSE\$ declarations, and so it acquires reject and error limit of 3, as defined by the last PATHPULSE \(\$\) declaration.
```

specify
(clk => q) = 12;
(data => q) = 10;
(clr, pre *> q) = 4;
specparam
PATHPULSE$clk$q = (2,9),
PATHPULSE$clr$q = (0,4),
PATHPULSE\$ = 3;
endspecify

```

\subsection*{14.6.2 Global control of pulse limit values}

Two invocation options can specify percentages applying globally to all module path transition delays. The error limit invocation option specifies the percentage of each module path transition delay used for its error limit value. The reject limit invocation option specifies the percentage of each module path transition delay used for its reject limit value. The percentage values shall be an integer between 0 and 100 .

The default values for both the reject and error limit invocation options are \(100 \%\). When neither option is present then \(100 \%\) of each module transition delay is used as the reject and error limits.

It is an error if the error limit percentage is smaller than the reject limit percentage. In such cases the error limit percentage is set equal to the reject limit percentage.

When both PATHPULSE\$ and global pulse limit invocation options are present, the PATHPULSES values shall take precedence.

\subsection*{14.6.3 SDF annotation of pulse limit values}

SDF annotation can be used to specify the pulse limit values of module path transition delays. Section 16 describes this in greater detail.

When both PATHPULSES, global pulse limit invocation options, and SDF annotation of pulse limit values are present, SDF annotation values shall take precedence.

\subsection*{14.6.4 Detailed pulse control capabilities}

The default style of pulse filtering behavior has two drawbacks. First, pulse filtering to the \(X\) state may be insufficiently pessimistic with an \(X\) state duration too short to be useful. Second, unequal delays can result in pulse rejection whenever the trailing edge precedes the leading edge, leaving no indication that a pulse was rejected. This section introduces more detailed pulse control capabilities.

\subsection*{14.6.4.1 n-event versus on-detect pulse filtering}

When an output pulse must be filtered to X , greater pessimism can be expressed if the module path output transitions immediately to \(X\) (on-detect) instead of at the already scheduled transition time of the leading edge of the pulse (on-event).

The on-event method of pulse filtering to \(X\) is the default. When an output pulse must be filtered to \(X\), the leading edge of the pulse becomes a transition to X and the trailing edge a transition from X . The times of transition of the edges do not change.

Just like on-event, the on-detect method of pulse filtering changes the leading edge of the pulse into a transition to \(X\) and the trailing edge to a transition from \(X\), but the time of the leading edge is changed to occur immediately upon detection of the pulse.

Figure 42 illustrates this behavior using a simple buffer with asymmetric rise/fall times and both the r-limits and e-limits equal to 0 . An output waveform is shown for both on-detect and on-event approaches.


Figure 42-On-detect -vs.- on-event
On-detect versus on-event behavior can be selected in two different ways. First, one may be selected globally for all module path outputs through use of the on-detect or on-event invocation option. Second, one may be selected locally through use of specify block pulse style declarations.

The syntax for pulse style declarations is shown in Syntax 14-8.

> pulsestyle_declaration :: = (From Annex A-A.7.1) pulsestyle_onevent list_of_path_outputs; \(\mid\) pulsestyle_ondetect list_of_path_outputs;

\section*{Syntax 14-8 Syntax for pulse style declarations}

It is an error if a module path output appears in a pulse style declaration after it has already appeared in a module path declaration.

The pulse style invocation options take precedence over pulse style specify block declarations.

\subsection*{14.6.4.2 Negative pulse detection}

When the delays to a module path output are unequal, it is possible for the trailing edge of a pulse to be scheduled for a time earlier than the schedule time of the leading edge, yielding a pulse with a negative width. Under normal operation, if the schedule for a trailing pulse edge is earlier than the schedule for a leading pulse edge, then the leading edge is cancelled. No transition takes place when the initial and final states of the pulse are the same, leaving no indication a schedule was ever present.

Negative pulses can be indicated with the \(X\) state by use of the showcancelled style of behavior. When the trailing edge of a pulse would be scheduled before the leading edge, this style causes the leading edge to be scheduled to X , and the trailing edge to be scheduled from X . With on-event pulse style, the schedule to X replaces the leading edge schedule. With on-detect pulse style, the schedule to \(X\) is made immediately upon detection of the negative pulse.
showcancelled behavior can be enabled in two different ways. First, it may be enabled globally for all module path outputs through use of the showcancelled and noshowcancelled invocation options. Second, it may be enabled locally through use of specify block showcancelled declarations.

The syntax for showcancelled declarations is shown in Syntax 14-9.
```

showcancelled_declaration ::= (From Annex A- A.7.1)
showcancelled list_of_path_outputs ;
| noshowcancelled list_of_path_outputs ;

```

\section*{Syntax 14-9 Syntax for showcancelled declarations}

It is an error if a module path output appears in a showcancelled declaration after it has already appeared in a module path declaration. The showcancelled invocation options take precedence over the showcancelled specify block declarations.

The showcancelled behavior is illustrated in Figure 43, which shows a narrow pulse presented at the input to a buffer with unequal rise/fall delays. This causes the trailing edge of the pulse to be scheduled earlier than leading edge. The leading edge of the input pulse schedules an output event 6 units later at the point marked by A. The pulse trailing edge occurs one time unit later, which schedules an output event 4 units later marked by point \(B\). This second schedule on the output is for a time prior to the already existing schedule for the leading output pulse edge.

The output waveform is shown for three different operating modes. The first waveform shows the default behavior with showcancelled behavior not enabled and with the default on-event style. The second waveform shows showcancelled behavior in conjunction with on-event. The last waveform shows showcancelled behavior in conjunction with on-detect.


Figure 43-Current event cancellation problem and correction
This same situation can also arise with nearly simultaneous input transitions, which is defined as two inputs transitioning closer together in time than the difference in their respective delays to the output. Figure 44 shows waveforms for a 2 -input NAND gate where initially \(A\) is high and \(B\) is low. \(B\) transitions \(0->1\) at time 10 , causing a \(1->0\) output schedule at time 24 . A transitions \(1->0\) at time 12 , causing a \(0->1\) schedule at time 22. Arrows mark the output transitions caused by the transitions on inputs A and B.

The output waveform is shown for three different operating modes. The first waveform shows the default behavior with showcancelled behavior not enabled and with the default on-event style. The second shows showcancelled behavior in conjunction with on-event. The third shows showcancelled behavior in conjunction with on-detect.


Figure 44-NAND gate with nearly simultaneous input switching where one event is scheduled prior to another that has not matured

One drawback of the on-event style with showcancelled behavior is that as the output pulse edges draw closer together, the duration of the resulting X state becomes smaller. Figure 45 illustrates how the on-detect style solves this problem.


Figure 45-Input NAND gate with nearly simultaneous input switching with output event scheduled at same time.

Examples:
Example 1:
```

specify
(a=>out)=(2,3);
(b =>out)=(3,4);
endspecify

```

Since no pulse style or showcancelled declarations appear within the specify block, the compiler applies the default modes of on-event and noshowcancelled.

Example 2:
```

specify
( $\mathrm{a}=>$ out) $=(2,3)$;
showcancelled out;
(b $=>$ out) $=(3,4)$;
endspecify

```

This showcancelled declaration is in error because it follows use of out in a module path declaration. It would be contradictory for out to have noshowcancelled behavior from input a, but showcancelled behavior from input b.

Example 2 Both these specify blocks produce the same result. Outputs out and out_b are both declared showcancelled and on_detect.
```

specify
showcancelled out;
pulsestyle_ondetect out;
(a =>out)= (2,3);
(a=>out)=(4,5);
showcancelled out_b;
pulsestyle_ondetect out_b;
(b=>out_b)=(5,6);
(b=>out_b)=(3,4);
endspecify
specify
showcancelled out,out_b;
pulsestyle_ondetect out,out_b;
(a =>out)=(2,3);
(b=>out)=(3,4);
(a =>out b)=(3,4);
(b=>out_\overline{b})=(5,6);
endspecify

```

\section*{15. Timing checks}

This section describes how timing checks are used in specify blocks to determine if signals obey the timing constraints.

\section*{15.1 verview}

Timing checks can be placed in specify blocks to verify the timing performance of a design by making sure critical events occur within given time limits. The syntax for system timing checks is given in Syntax 15-1.
system_timing_check \(::=(\) From Annex \(A-A .7 .5 .1)\)
\$setup_timing_check
\$hold _timing_check
| \$setuphold_timing_check
\$recovery_timing_check
\$removal_timing_check
\$recrem_timing_check
\$skew_timing_check
| \$timeskew_timing_check
\$ fullskew_timing_check
\$period_timing_check
\$width_timing_check
\$nochange_timing_check
\$setup_timing_check ::=
\$setup (data_event , reference_event, timing_check_limit [, [ notify_reg ] ]);
\$hold _timing_check ::=
\$hold (reference_event, data_event, timing_check_limit [, [ notify_reg ] ]);
\$setuphold_timing_check ::=
\$setuphold (reference_event, data_event , timing_check_limit, timing_check_limit [, [ notify_reg ] [, [ stamptime_condition ] [, [ checktime_condition ] [,[ delayed_reference ][,[delayed_data]]]]]]);

\section*{\$recovery_timing_check ::=}
\$recovery (reference_event, data_event, timing_check_limit [, [ notify_reg ] ]);
\$removal_timing_check ::=
\$removal (reference_event, data_event, timing_check_limit [, [ notify_reg ] ]) ;
\$recrem_timing_check ::=
\$recrem (reference_event, data_event, timing_check_limit, timing_check_limit
[, [ notify_reg ] [, [ stamptime_condition ] [, [ checktime_condition ]
[, [delayed_reference ][,[delayed_data ]]]]]]);
\$skew_timing_check ::=
\$skew (reference_event, data_event, timing_check_limit [, [ notify_reg ] ]);
\$timeskew_timing_check ::=
\$timeskew ( reference_event, data_event, timing_check_limit [, [ notify_reg ][,[ event_based_flag ][,[remain_active_flag ]]]]);
\$fullskew_timing_check ::=
\$fullskew (reference_event, data_event, timing_check_limit, timing_check_limit \([,[\) notify_reg \(][,[\) event_based_flag \(][,[\) remain_active_flag \(]]]]\);
\$period_timing_check \(::=\)
\$period ( controlled_reference_event, timing_check_limit [, [ notify_reg] ]) ;
\$width_timing_check ::=
\$width ( controlled_reference_event, timing_check_limit, threshold [, [ notify_reg]]);
\$nochange_timing_check ::=
\$nochange (reference_event, data_event, start_edge_offset, end_edge_offset [, [ notify_reg] ]);

Syntax 15-1 Syntax for system timing checks

The syntax for check time conditions and timing check events are given in Syntax 15-2.
```

checktime_condition ::= (From Annex A - A.7.5.2)
mintypmax_expression
controlled_reference_event ::=
controlled_timing_check_event
data_event ::=
timing_check_event
delayed_data ::=
terminal_identifier
| terminal_identifier [ constant_mintypmax_expression ]
delayed_reference ::=
terminal_identifier
| terminal_identifier [ constant_mintypmax_expression ]
end_edge_offset ::= mintypmax_expression
event_based_flag ::= constant_expression
notify_reg ::= variable_identifier
reference_event ::= timing_check_event
remain_active_flag ::= constant_mintypmax_expression
stamptime_condition ::= mintypmax_expression
start_edge_offset ::= mintypmax_expression
threshold ::=constant_expression
timing_check_limit ::= expression
timing_check_event ::= (From Annex A - A.7.5.3)
[timing_check_event_control] specify_terminal_descriptor [ \&\&\&\& timing_check_condition ]
controlled_timing_check_event ::=
timing_check_event_control specify_terminal_descriptor [\boldsymbol{\&\&\&}
timing_check_event_control ::= posedge | negedge | edge_control_specifier
specify_terminal_descriptor ::=
specify_input_terminal_descriptor
| specify_output_terminal_descriptor
edge_control_specifier ::= edge [ edge_descriptor [, edge_descriptor]]
edge_descriptor*::= 01 | 10 | z_or_x zero_or_one | zero_or_one z_or_x
zero_or_one ::= 0 | 1
z_or_x ::= = \ | X | z |
timing_check_condition ::=
scalar_timing_check_condition
| ( scalar_timing_check_condition )
scalar_timing_check_condition ::=
expression
|~ expression
| expression == scalar_constant
| expression === scalar_constant
| expression != scalar_constant
| expression !== scalar_constant
scalar_constant ::= 1'b0 | 1'b1 | 1'B0 | 1'B1 |'b0 | 'b1 |'B0 |'B1 | 1 | 0

```
*Embedded spaces are illegal.

\section*{Syntax 15-2 Syntax for check time conditions and timing check events}

For ease of presentation, the timing checks are divided into two groups. The first group of timing checks are described in terms of stability time windows:
\begin{tabular}{lll} 
\$setup & \$hold & \$setuphold \\
\$recovery & \$removal & \$recrem
\end{tabular}

The timing checks in the second group check clock and control signals, and are described in terms of the difference in time between two events (the \$nochange check involves three events):
\begin{tabular}{lll} 
\$skew & Stimeskew & \$fullskew \\
\$width & \$period & \$nochange
\end{tabular}

Although they begin with a \(\$\), timing checks are not system tasks. The leading \(\$\) is present because of historical reasons, and timing checks shall not be confused with system tasks. In particular, no system task can appear in a specify block, and no timing check can appear in procedural code.

Some timing checks can accept negative limit values. This topic is discussed in detail in 15.8.
All timing checks have both a reference event and a data event, and boolean conditions can be associated with each. Some of the checks have two signal arguments, one of which is the reference event and the other the data event. Other checks have only one signal argument, and the reference and data events are derived from it. Reference events and data events shall only be detected by timing checks when their associated conditions are true. See 15.6 for more information about conditions in timing checks.

Timing check evaluation is based upon the times of two events, called the timestamp event and the timecheck event. A transition on the timestamp event signal causes the simulator to record (stamp) the time of transition for future use in evaluating the timing check. A transition on the timecheck event signal causes the simulator to actually evaluate the timing check to determine whether a violation has taken place.

For some checks the reference event is always the timestamp event, while the data event is always the timecheck event, while for other checks the reverse is true. And for yet other checks the decision as to which is the timestamp and which the timecheck event is based upon factors to be discussed later in greater detail.

Every timing check can include an optional notifier which toggles whenever the timing check detects a violation. The model can use the notifier to make behavior a function of timing check violations. Notifiers are discussed in greater detail in 15.5.

Like expressions for module path delays, timing check limit values are constant expressions which can include specparams.

\subsection*{15.2 Timing checks using a stability window}

These timing checks are discussed in this section:
\begin{tabular}{lll} 
\$setup & \$hold & \$setuphold \\
\$recovery & \$removal & \$recrem
\end{tabular}

These checks accept two signals, the reference event and the data event, and define a time window with respect to one signal while checking the time of transition of the other signal with respect to the window. In general they all perform the following steps:
a) Define a time window with respect to the reference signal using the specified limit or limits;
b) Check the time of transition of the data signal with respect to the time window;
c) Report a timing violation if the data signal transitions within the time window.

\subsection*{15.2.1 \$setup}

The \$setup timing check syntax is shown in Syntax 15-3.
```

\$setup_timing_check $::=$ (From Annex A-A.7.5.1)
\$setup (data_event, reference_event, timing_check_limit [, [ notify_reg]]);
data_event $::=$ (From Annex $A$ - A.7.5.2)
timing_check_event
notify_reg ::=
variable_identifier
reference_event ::=
timing_check_event
timing_check_limit $::=$
expression

```

Syntax 15-3 Syntax for \$setup

Table 49 defines the \$setup timing check.
Table 49-\$setup arguments
\begin{tabular}{|l|l|}
\hline Argument & \multicolumn{1}{c|}{ Description } \\
\hline data_event & Timestamp event \\
\hline reference_event & Timecheck event \\
\hline limit & Non-negative constant expression \\
\hline notifier (optional) & Reg \\
\hline
\end{tabular}

The data event is usually a data signal, while the reference event is usually a clock signal.
The endpoints of the time window are determined as follows:
```

(beginning of time window) = (timecheck time) - limit
(end of time window) = (timecheck time)

```

The \$setup timing check reports a timing violation in the following case:
```

(beginning of time window) < (timestamp time) < (end of time
window)

```

The endpoints of the time window are not part of the violation region. When the limit is zero, the \$setup check shall never issue a violation.

\subsection*{15.2.2 \$hold}

The \$hold timing check syntax is shown in Syntax 15-4.
```

\$hold _timing_check $::=($ From Annex $A-$ A.7.5.1)
\$hold (reference_event, data_event, timing_check_limit [, [ notify_reg ] ]) ;
data_event $::=($ From Annex $A-$ A.7.5.2 $)$
timing_check_event
notify_reg ::=
variable_identifier
reference_event ::=
timing_check_event
timing_check_limit ::=
expression

```
Syntax 15-4 Syntax for \$hold

Table 50 defines the \$hold timing check.
Table 50-\$hold arguments
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ Argument } & \multicolumn{1}{c|}{ Description } \\
\hline reference_event & Timestamp event \\
\hline data_event & Timecheck event \\
\hline limit & Non-negative constant expression \\
\hline notifier (optional) & Reg \\
\hline
\end{tabular}

The data event is usually a data signal, while the reference event is usually a clock signal.
The endpoints of the time window are determined as follows:
```

(beginning of time window) = (timestamp time)
(end of time window) = (timestamp time) + limit

```

The \$hold timing check reports a timing violation in the following case:
```

(beginning of time window) <= (timecheck time) < (end of time
window)

```

Only the end of the time window is not part of the violation region. When the limit is zero, the \$hold check shall never issue a violation.

\subsection*{15.2.3 \$setuphold}

The \$setuphold timing check syntax is shown in Syntax 15-5.
```

\$setuphold_timing_check ::= (From Annex A - A.7.5.1)
\$setuphold ( reference_event, data_event, timing_check_limit, timing_check_limit
[, [ notify_reg ] [, [ stamptime_condition ] [, [ checktime_condition ]
[,[ delayed_reference ][,[delayed_data ]]]]]]);
checktime_condition ::= (From Annex $A$ - A.7.5.2)
mintypmax_expression
data_event ::=
timing_check_event
delayed_data ::=
terminal_identifier
| terminal_identifier [ constant_mintypmax_expression ]
delayed_reference ::=
terminal_identifier
| terminal_identifier [ constant_mintypmax_expression ]
notify_reg ::=
variable_identifier
reference_event ::=
timing_check_event
stamptime_condition ::=
mintypmax_expression
timing_check_limit ::=
expression

```

\section*{Syntax 15-5 Syntax for \$setuphold}

Table 51 defines the \$setuphold timing check.
Table 51-\$setuphold arguments
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ Argument } & \multicolumn{1}{c|}{ Description } \\
\hline reference_event & \begin{tabular}{l} 
Timecheck or timestamp event when setup limit is positive \\
Timestamp event when setup limit is negative
\end{tabular} \\
\hline data_event & \begin{tabular}{l} 
Timecheck or timestamp event when hold limit is positive \\
Timestamp event when hold limit is negative
\end{tabular} \\
\hline setup_limit & Constant expression \\
\hline hold_limit & Constant expression \\
\hline notifier (optional) & Reg \\
\hline timestamp_cond (optional) & Timestamp condition for negative timing checks \\
\hline timecheck_cond (optional) & Timecheck condition for negative timing checks \\
\hline delayed_reference (optional) & Delayed reference signal for negative timing checks \\
\hline delayed_data (optional) & Delayed data signal for negative timing checks \\
\hline
\end{tabular}

The \$setuphold timing check can accept negative limit values. This is discussed in greater detail in 15.8.
The data event is usually a data signal, while the reference event is usually a clock signal.

When both the setup limit and the hold limit are positive, either the reference event or the data event can be the timecheck event. It shall depend upon which occurs first in the simulation.

When either the setup limit or the hold limit is negative the restriction becomes:
```

setup_limit + hold_limit > (simulation unit of precision)

```

The \$setuphold timing check combines the functionality of the \$setup and \$hold timing checks into a single timing check. Therefore, the following invocation:
```

\$setuphold( posedge clk, data, tSU, tHLD );

```
is equivalent in functionality to the following, if \(t S U\) and \(t H L D\) are not negative:
```

\$setup( data, posedge clk, tSU );
\$hold( posedge clk, data, tHLD );

```

When both setup and hold limits are positive and the data event occurs first, the endpoints of the time window are determined as follows:
```

(beginning of time window) = (timecheck time) - limit
(end of time window) = (timecheck time)

```

And the \$setuphold timing check reports a timing violation in the following case:
```

(beginning of time window) < (timecheck time) <= (end of time
window)

```

Only the beginning of the time window is not part of the violation region. The \$setuphold check shall report a timing violation when the reference and data events occur simultaneously.

When both setup and hold limits are positive and the data event occurs second, the endpoints of the time window are determined as follows:
```

beginning of time window) = (timestamp time)
(end of time window) = (timestamp time) + limit

```

And the \$setuphold timing check reports a timing violation in the following case:
```

(beginning of time window) <= (timecheck time) < (end of time
window)

```

Only the end of the time window is not part of the violation region. The \$setuphold check shall report a timing violation when the reference and data events occur simultaneously.

When both limits are zero, the \$setuphold check shall never issue a violation.

\subsection*{15.2.4 \$removal}

The \$removal timing check syntax is shown in Syntax 15-6.
```

\$removal_timing_check ::= (From Annex A - A.7.5.1)
\$removal (reference_event, data_event, timing_check_limit [, [ notify_reg ] ] ) ;
data_event ::= (From Annex A - A.7.5.2)
timing_check_event
notify_reg ::=
variable_identifier
reference_event ::=
timing_check_event
timing_check_limit ::=
expression

```

\section*{Syntax 15-6 Syntax for \$removal}

Table 52 defines the \$removal timing check.
Table 52-\$removal arguments
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ Argument } & \multicolumn{1}{c|}{ Description } \\
\hline reference_event & Timestamp event \\
\hline data_event & Timecheck event \\
\hline limit & Non-negative constant expression \\
\hline notifier (optional) & Reg \\
\hline
\end{tabular}

The reference event is usually a control signal like clear, reset or set, while the data event is usually a clock signal.

The endpoints of the time window are determined as follows:
```

(beginning of time window) = (timecheck time) - limit
(end of time window) = (timecheck time)

```

The \$removal timing check reports a timing violation in the following case:
```

(beginning of time window) < (timestamp time) < (end of time
window)

```

The endpoints of the time window are not part of the violation region. When the limit is zero, the \$removal check shall never issue a violation.

\subsection*{15.2.5 \$recovery}

The \$recovery timing check syntax is shown in Syntax 15-7.
```

\$recovery_timing_check ::= (From Annex A - A.7.5.1)
\$recovery (reference_event, data_event, timing_check_limit [, [ notify_reg ] ]);
data_event ::= (From Annex A - A.7.5.2)
timing_check_event
notify_reg ::=
variable_identifier
reference_event ::=
timing_check_event
timing_check_limit ::=
expression

```

Syntax 15-7 Syntax for \$recovery

Table 53 defines the \$recovery timing check.
Table 53-\$recovery arguments
\begin{tabular}{|l|l|}
\hline Argument & \multicolumn{1}{c|}{ Description } \\
\hline reference_event & Timestamp event \\
\hline data_event & Timecheck event \\
\hline limit & Non-negative constant expression \\
\hline notifier (optional) & Reg \\
\hline
\end{tabular}

The reference event is usually a control signal like clear, reset or set, while the data event is usually a clock signal.

The endpoints of the time window are determined as follows:
```

(beginning of time window) = (timestamp time)
(end of time window) = (timestamp time) + limit

```

The \$recovery timing check reports a timing violation in the following case:
```

(beginning of time window) <= (timecheck time) < (end of time
window)

```

Only the end of the time window is not part of the violation region. When the limit is zero, the \$recovery check shall never issue a violation.

\subsection*{15.2.6 \$recrem}

The \$recrem timing check syntax is shown in Syntax 15-8.
```

\$recrem_timing_check ::= (From Annex A - A.7.5.1)
\$recrem (reference_event, data_event, timing_check_limit, timing_check_limit
[, [ notify_reg ] [, [ stamptime_condition ] [, [ checktime_condition ]
[,[delayed_reference ][,[delayed_data ]]]]]]);
checktime_condition ::= (From Annex A - A.7.5.2)
mintypmax_expression
data_event ::=
timing_check_event
delayed_data ::=
terminal_identifier
| terminal_identifier [ constant_mintypmax_expression ]
delayed_reference ::=
terminal_identifier
| terminal_identifier [ constant_mintypmax_expression ]
notify_reg ::=
variable_identifier
reference_event ::=
timing_check_event
stamptime_condition ::=
mintypmax_expression
timing_check_limit ::=
expression

```

\section*{Syntax 15-8 Syntax for \$recrem}

Table 54 defines the \$recrem timing check.
Table 54-\$recrem arguments
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ Argument } & \multicolumn{1}{c|}{ Description } \\
\hline reference_event & \begin{tabular}{l} 
Timecheck or timestamp event when removal limit is positive \\
Timestamp event when removal limit is negative
\end{tabular} \\
\hline data_event & \begin{tabular}{l} 
Timecheck or timestamp event when recovery limit is positive \\
Timestamp event when recovery limit is negative
\end{tabular} \\
\hline recovery_limit & Constant expression \\
\hline removal_limit & Constant expression \\
\hline notifier (optional) & Reg \\
\hline timestamp_cond (optional) & Timestamp condition for negative timing checks \\
\hline timecheck_cond (optional) & Timecheck condition for negative timing checks \\
\hline delayed_reference (optional) & Delayed reference signal for negative timing checks \\
\hline delayed_data (optional) & Delayed data signal for negative timing checks \\
\hline
\end{tabular}

The \$recrem timing check can accept negative limit values. This is discussed in greater detail in 15.8 .
When both the removal limit and the recovery limit are positive, either the reference event or the data event can be the timecheck event. It shall depend upon which occurs first in the simulation.

When either the removal limit or the recovery limit is negative the restriction becomes:
```

removal_limit + recovery_limit > (simulation unit of precision)

```

The \$recrem timing check combines the functionality of the \$removal and \$recovery timing checks into a single timing check. Therefore, the following invocation:
```

\$recrem( posedge clear, posedge clk, tREC, tREM );

```
is equivalent in functionality to the following, if \(\operatorname{tREC}\) and \(t R E M\) are not negative:
```

\$removal( posedge clear, posedge clk, tREM );
\$recovery( posedge clear, posedge clk, tREC );

```

When both removal and recovery limits are positive and the data event occurs first, the endpoints of the time window are determined as follows:
```

(beginning of time window) = (timecheck time) - limit
(end of time window) = (timecheck time)

```

And the \$recrem timing check reports a timing violation in the following case:
```

(beginning of time window) < (timecheck time) <= (end of time
window)

```

Only the beginning of the time window is not part of the violation region. The \$recrem check shall report a timing violation when the reference and data events occur simultaneously.

When both removal and recovery limits are positive and the data event occurs second, the endpoints of the time window are determined as follows:
```

(beginning of time window) = (timestamp time)
(end of time window) = (timestamp time) + limit

```

And the \$recrem timing check reports a timing violation in the following case:
```

(beginning of time window) <= (timecheck time) < (end of time
window)

```

Only the end of the time window is not part of the violation region. The \$recrem check shall report a timing violation when the reference and data events occur simultaneously.

When both limits are zero, the \$recrem check shall never issue a violation.

\subsection*{15.3 Timing checks for clock and control signals}

The following timing checks are discussed in this section:
\$skew \$timeskew \$fullskew \$period \$width \$nochange

These checks accept one or two signals and verify transitions on them are never separated by more than the limit. For those checks specifying only one signal, the reference event and data event are derived from that one signal. In general these checks all perform the following steps:
a) Determine the elapsed time between two events;
b) Compare the elapsed time to the specified limit;
c) Report a timing violation if the elapsed time violates the limit.

The skew checks have two different violation detection mechanisms, event-based and timer-based. Eventbased skew checking is performed only when a signal transitions, while timer-based skew checking takes place as soon as the simulation time equal to the skew limit has elapsed.

The \$nochange check involves three events rather than two.

\subsection*{15.3.1 \$skew}

The \$skew timing check syntax is shown in Syntax 15-9.
```

\$skew_timing_check ::= (From Annex A - A.7.5.1)
\$skew (reference_event, data_event,timing_check_limit [, [ notify_reg ]]);
data_event ::= (From Annex A - A.7.5.2)
timing_check_event
notify_reg ::=
variable_identifier
reference_event ::=
timing_check_event
timing_check_limit ::=
expression

```

\section*{Syntax 15-9 Syntax for \$skew}

Table 55 defines the \$skew timing check.
Table 55-\$skew arguments
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ Argument } & \multicolumn{1}{c|}{ Description } \\
\hline reference_event & Timestamp event \\
\hline data_event & Timecheck event \\
\hline limit & Non-negative constant expression \\
\hline notifier (optional) & Reg \\
\hline
\end{tabular}

The \$skew timing check reports a violation in the following case:
```

(timecheck time) - (timestamp time) > limit

```

Simultaneous transitions on the reference and data signals can never cause \$skew to report a timing violation, even when the skew limit value is zero.

The \$skew timing check is event-based; it is evaluated only after a data event. If there is never a data event (i.e., the data event is infinitely late), the \$skew timing check shall never be evaluated, and no timing violation shall ever be reported. In contrast, the \$timeskew and \$fullskew checks are timer-based by default, and they shall be used if violation reports are absolutely required and the data event can be very late or even absent altogether. These checks are discussed in 15.3.2 and 15.3.3.
\$skew shall wait indefinitely for the data event once it has detected a reference event and it shall not report a timing violation until the data event takes place. A second consecutive reference event shall cancel the old wait for the data event and begin a new one.

After a reference event, the \$skew timing check shall never stop checking data events for a timing violation. \$skew shall report timing violations for all data events occurring beyond the limit after a reference event.

\subsection*{15.3.2 \$timeskew}

The syntax for \$timeskew is shown in Syntax 15-10.
```

\$timeskew_timing_check ::= (From Annex A - A.7.5.1)
\$timeskew (reference_event, data_event, timing_check_limit
[,[ notify_reg ][,[ event_based_flag ][,[remain_active_flag ]]]]);
data_event ::= (From Annex A - A.7.5.2)
timing_check_event
event_based_flag ::=
constant_expression
notify_reg ::=
variable_identifier
reference_event ::=
timing_check_event
remain_active_flag ::=
constant_mintypmax_expression
timing_check_limit ::=
expression

```

\section*{Syntax 15-10 Syntax for \$timeskew}

Table 56 defines the \$timeskew timing check arguments.
Table 56-\$timeskew arguments
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ Argument } & \multicolumn{1}{c|}{ Description } \\
\hline reference_event & Timestamp event \\
\hline data_event & Timecheck event \\
\hline limit & Non-negative constant expression \\
\hline notifier (optional) & Reg \\
\hline event_based_flag (optional) & Constant expression \\
\hline remain_active_flag (optional) & Constant expression \\
\hline
\end{tabular}

The \$timeskew timing check reports a violation only in the following cases:
```

(timecheck time) - (timestamp time) > limit

```

Simultaneous transitions on the reference and data signals can never cause \$timeskew to report a timing violation, even when the skew limit value is zero.

The default behavior for \$timeskew is timer-based. Violations are reported immediately upon an elapse of time after the reference event equal to the limit, and the check shall become dormant and report no more violations (even in response to data events) until after the next reference event. This check shall also become dormant if it detects a reference event when its condition is false.

The \$timeskew check's default timer-based behavior can be altered to event-based using the event based flag. It behaves like the \$skew check when both the event based flag and the remain active flag are set. The \$timeskew check behaves like the \$skew check when only the event based flag is set, except it becomes dormant after reporting the first violation.

Example:
Stimeskew (posedge \(C P \& \& \& M O D E\), negedge \(C P N, 50)\);


Figure 46-Sample \$timeskew
Case 1: Event based flag and remain active flag not set.
After the first reference event on \(C P\) at \(A\), a violation is reported at \(B\) as soon as 50 time units have passed. No further violations are reported.

Case 2: Event based flag set, remain active flag not set.
The negative transition on CPN at point \(C\) shall cause a timing violation. Subsequent negative transitions at points \(D\) and \(E\) do not cause violations. The second reference event at \(F\) occurs while MODE is false, turning the \$timeskew check dormant, and no further violations are reported.

Case 3: Event based flag set, remain active flag set.
The first three negative transitions on CPN at points C, D and E shall cause timing violations. The second reference event at F occurs while MODE is false, turning the \$timeskew check dormant, and no further violations are reported.

Case 4: Event based flag and remain active flag both set.
Every negative edge on CPN is reported as a violation, which is identical to \$skew behavior.

\subsection*{15.3.3 \$fullskew}

The syntax for \$fullskew is shown in Syntax 15-11.
```

\$fullskew_timing_check ::= (From Annex A - A.7.5.1)
\$fullskew (reference_event , data_event, timing_check_limit , timing_check_limit
[,[ notify_reg ][,[ event_based_flag ][,[remain_active_flag ]]]]);
data_event ::= (From Annex A - A.7.5.2)
timing_check_event
event_based_flag ::=
constant_expression
notify_reg ::=
variable_identifier
reference_event ::=
timing_check_event
remain_active_flag ::=
constant_mintypmax_expression
timing_check_limit ::=
expression

```

Syntax 15-11 Syntax for \$fullskew

Table 57 defines the \$fullskew timing check arguments.
Table 57-\$fullskew arguments
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ Argument } & \multicolumn{1}{c|}{ Description } \\
\hline reference_event & Timestamp or timecheck event \\
\hline data_event & Timestamp or timecheck event \\
\hline limit 1 & Non-negative constant expression \\
\hline limit 2 & Non-negative constant expression \\
\hline notifier (optional) & Reg \\
\hline event_based_flag (optional) & Constant expression \\
\hline remain_active_flag (optional) & Constant expression \\
\hline
\end{tabular}

Sfullskew is identical to \$timeskew except the reference and data events can transition in either order. The first limit is the maximum time by which the data event can follow the reference event. The second limit is the maximum time by which the reference event can follow the data event.

The reference event is the timestamp event and the data event is the timecheck event when the reference event precedes the data event. The data event is the timestamp event and the reference event is the timecheck event when the data event precedes the reference event.

The \$fullskew timing check reports a violation only in the following case, where limit is set to limitl when the reference event transitions first, and to limit2 when the data event transitions first:
```

(timecheck time) - (timestamp time) > limit

```

Simultaneous transitions on the reference and data signals shall never cause \$fullskew to report a timing violation, even when the skew limit value is zero.

The default behavior for \(\$\) fullskew is timer-based. Violations shall be reported immediately upon an elapse of time after the timestamp event equal to the limit. It then becomes dormant and reports no more violations, even in response to timecheck events, until after the next timestamp event. This check shall also become dormant if it detects a timestamp event when the associated condition is false.

The \$fullskew check's default timer-based behavior can be altered to event-based using the event based flag. It behaves like the \$skew check when both the event based flag and the remain active flag are set. The \$timeskew check behaves like the \$skew check when only the event based flag is set, except it becomes dormant after it reports the first violation.

\section*{Example:}
\$fullskew (posedge \(C P \& \& \& M O D E\), negedge \(C P N, 50,70\) );


Figure 47-Sample \$fullskew
Case 1: Event based flag and remain active flag not set.
The transition at A of CP while MODE is true begins a wait for a negative transition on CPN, and a violation is reported at B as soon as a period of time equal to 50 time units has passed. This resets the check and readies it for the next active transition.

A negative transition on CPN occurs next at \(C\), beginning a wait for a positive transition on \(C P\) while MODE is true. At D a time equal to 70 time units has passed without a positive edge on CP while MODE is true, so a violation is reported and the check is again reset to await the next active transition.

A transition on CPN at E also results in a timing violation, as does the transition at F , because even though CP transitions, MODE is no longer true. Transitions at \(G\) and \(H\) also result in timing violations, but not the transition at I, because it is followed by a positive transition on CP while MODE is true.

Case 2: Event based flag set, remain active flag not set.
The transition at \(A\) of \(C P\) while MODE is true begins a wait for a negative transition on CPN, and a violation is reported at C on CPN because it occurs beyond the 50 time unit limit. This transition at C also begins a wait of 70 time units for a positive transition on CP while MODE is true. But for transitions on CPN at B
through \(H\) there is no positive transition on \(C P\) while \(M O D E\) is true, and so no timing violations are reported. The transition at I on CPN begins a wait of 70 time units, and this is satisfied by the positive transition on \(C P\) at \(J\) while \(M O D E\) is true.

Case 3: Event based flag and remain active flag both set.

The transition at \(A\) of \(C P\) while MODE is true begins a wait for a negative transition on CPN, and a violation is reported at C on \(C P N\), and it shall also begin a wait for a positive transition on \(C P\) while MODE is true. No such transition on CP ever takes place after CPN transitions \(C\) through \(H\), but no violations are reported because CP never experiences a positive transition while MODE is true. Transition I also reports no violation because a positive transition at I on CP while MODE is true occurs within the 70 time unit skew limit.

\subsection*{15.3.4 \$width}

The \$width timing check syntax is shown in Syntax 15-12.
```

\$width_timing_check ::= (From Annex A - A.7.5.1)
\$width (controlled_reference_event,timing_check_limit,
threshold [,[ notify_reg ]]);
controlled_reference_event ::= (From Annex A - A.7.5.2)
controlled_timing_check_event
notify_reg ::=
variable_identifier
threshold ::=
constant_expression
timing_check_limit ::=
expression

```

\section*{Syntax 15-12 Syntax for \$width}

If the comma before the threshold is present, the comma before the notifier shall also be present, even though both arguments are optional.

Table 58 defines the \(\$\) width timing check.
Table 58—\$width arguments
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ Argument } & \multicolumn{1}{c|}{ Description } \\
\hline reference_event & Timestamp edge triggered event \\
\hline (data_event - implicit) & Timecheck edge triggered event \\
\hline limit & Non-negative constant expression \\
\hline threshold (optional) & Non-negative constant expression \\
\hline notifier (optional) & Reg \\
\hline
\end{tabular}

The \$width timing check monitors the width of signal pulses by measuring the time from the timestamp event to the timecheck event. Since a data event is not passed to \$width, it is derived from the reference event, as follows:
```

data event = reference event signal with opposite edge

```

Because of the way the data event is derived for \$width, an edge triggered event has to be passed as the reference event. A compilation error shall occur if the reference event is not an edge specification.

While the \$width timing check can be defined in terms of a time window, it is simpler to express it as the difference between the timecheck and timestamp times. The \$width timing check reports a violation in the following case:
```

threshold < (timecheck time) - (timestamp time) < limit

```

The pulse width has to be greater than or equal to limit in order to avoid a timing violation, but no violation is reported for glitches smaller than the threshold.

The threshold argument shall be included if the notifier argument is required. It is permissible to not specify both the threshold and notifier arguments, making the default value for the threshold zero (0). If the notifier is present, a non-null value for the threshold shall also be present. Here is a legal \$width check when the notifier is required and the threshold is not:
\$width (posedge clk, 6, 0, ntfr_reg);
The data event and the reference event shall never occur at the same simulation time because these events are triggered by opposite transitions.

\section*{Example:}

The following example demonstrates some examples of legal and illegal calls:
```

// Legal Calls
\$width ( negedge clr, lim );
\$width ( negedge clr, lim, thresh, notif );
\$width ( negedge clr, lim, 0, notif );
// Illegal Calls
\$width ( negedge clr, lim, , notif );
\$width ( negedge clr, lim, notif );

```

\subsection*{15.3.5 \$period}

The \$period timing check syntax is shown in Syntax 15-13.
```

\$period_timing_check ::= (From Annex A-A.7.5.1)
\$period ( controlled_reference_event, timing_check_limit [, [ notify_reg]]);
controlled_reference_event $::=$ (From Annex A-A.7.5.2)
controlled_timing_check_event
notify_reg ::=
variable_identifier
timing_check_limit $::=$
expression

```

Syntax 15-13 Syntax for \$period

Table 59 defines the \$period timing check.
Table 59-\$period arguments
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ Argument } & \multicolumn{1}{c|}{ Description } \\
\hline reference_event & Timestamp edge triggered event \\
\hline (data_event - implicit) & Timestamp edge triggered event \\
\hline limit & Non-negative constant expression \\
\hline notifier (optional) & Reg \\
\hline
\end{tabular}

Since the data event is not passed as an argument to \$period, it is derived from the reference event, as follows:
```

data event = reference event signal with the same edge

```

Because of the way the data event is derived for \$period, an edge triggered event shall be passed as the reference event. A compilation error shall occur if the reference event is not an edge specification.

While the \$period timing check can be defined in terms of a time window, it is simpler to express it as the difference between the timecheck and timestamp times. The \$period timing check reports a violation in the following case:
(timecheck time) - (timestamp time) < limit

\subsection*{15.3.6 \$nochange}

The \$nochange syntax is shown in Syntax 15-14.
```

\$nochange_timing_check ::= (From Annex A - A.7.5.1)
\$nochange (reference_event, data_event, start_edge_offset,
end_edge_offset[,[notify_reg]]);
data_event ::= (From Annex A - A.7.5.2)
timing_check_event
end_edge_offset ::=
mintypmax_expression
notify_reg ::=
variable_identifier
reference_event ::=
timing_check_event
start_edge_offset ::=
mintypmax_expression

```

Table 60 defines the \$nochange timing check arguments.
Table 60-\$nochange arguments
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ Argument } & \multicolumn{1}{c|}{ Description } \\
\hline reference_event & Edge triggered timestamp and/or timecheck event \\
\hline data_event & Timestamp or timecheck event \\
\hline start_edge_offset & Constant expression \\
\hline end_edge_offset & Constant expression \\
\hline notifier (optional) & Reg \\
\hline
\end{tabular}

The \$nochange timing check reports a timing violation if the data event occurs during the specified level of the control signal (the reference event). The reference event can be specified with the posedge or the negedge keyword, but the edge control specifiers (see 15.4) can not be used.

The start edge and end edge offsets can expand or shrink the timing violation region, which is defined by the duration of the reference event signal after the edge. For example, if the reference event is a posedge, then the duration is the period during which the reference signal is high. A positive offset for start edge extends the region by starting the timing violation region earlier; a negative offset for start edge shrinks the region by starting the region later. Similarly, a positive offset for the end edge extends the timing violation region by ending it later, while a negative offset for the end edge shrinks the region by ending it earlier. If both the offsets are zero, the size of the region shall not change.

Unlike other timing checks, \$nochange involves three, rather than two, transitions. The leading edge of the reference event defines the beginning of the time window, while the trailing edge of the reference event defines the end of the time window. A violation results if the data event occurs anytime within the time window.

The endpoints of the time window are determined as follows:
```

(beginning of time window) =
(leading reference edge time) - start_edge_offset
(end of time window) = (trailing reference edge time) +
end_edge_offset

```

The \$nochange timing check reports a timing violation in the following case:
```

beginning of time window) < (data event time) < (end of time
window)

```

The endpoints of the time window are not included. The values of start edge offset and end_edge_offset play a significant role in determining which signal, the reference event or the data event, is the timestamp or timecheck event.

\section*{Example:}
\$nochange ( posedge clk, data, 0,0 ) ;
In this example, \$nochange system task shall report a violation if the data signal changes while \(\mathrm{cl} k\) is high. It shall not be a violation if posedge clk and a transition on data occur simultaneously.

\subsection*{15.4 Edge-control specifiers}

The edge-control specifiers can be used to control events in timing checks based on specific edge transitions between 0,1 , and x . Syntax 15-15 shows the syntax for edge-control specifiers.
```

edge_control_specifier ::= (From Annex A - A.7.5.3)
edge [ edge_descriptor [ , edge_descriptor ] ]
edge_descriptor }\mp@subsup{}{}{*}::
01
10
z_or_x zero_or_one
|ero_or_one z_or_x
zero_or_one ::= 0 | 1
z_or_x ::= =\mathbf{x | X | z |}

```
*Embedded spaces are illegal.

\section*{Syntax 15-15 Syntax for edge control specifier}

Edge-control specifiers contain the keyword edge followed by a square bracketed list of from one to six pairs of edge transitions between 0,1 and x , as follows:
```

0 1 ~ T r a n s i t i o n ~ f r o m ~ 0 ~ t o ~ 1 ~
0x Transition from 0 to x
10 Transition from 1 to 0
1x Transition from 1 to x
x0 Transition from x to 0
x1 Transition from x to 1

```

Edge transitions involving z are treated the same way as edge transitions involving x .

The posedge and negedge keywords can be used as a shorthand for certain edge-control specifiers. For example, the construct:
posedge clr
is equivalent to the following:
```

edge[ 01, 0x, x1] clr

```

Similarly, the construct
```

negedge clr

```
is the same as the following:
```

edge[ 10, x0, 1x] clr

```

However, edge-control specifiers offer the flexibility to declare edge transitions other than posedge and negedge.

\subsection*{15.5 Notifiers: user-defined responses to timing violations}

Timing check notifiers detect timing check violations behaviorally, and, therefore, take an action as soon as a violation occurs. Such notifiers can be used to print an informative error message describing the violation or to propagate an x value at the output of the device which reported the violation.

The notifier is a reg declared in the module where timing check tasks are invoked which is passed as the last argument to a system timing check. Whenever a timing violation occurs, the system task updates the value of the notifier.

The notifier is an optional argument to all system timing checks and can be omitted from the system task call without adversely affecting its operation.

Table 61 shows how the notifier values are toggled when timing violations occur.

\section*{Table 61-User-defined responses to timing violations}
\begin{tabular}{|l|l|}
\hline BEFORE violation & AFTER violation \\
\hline x & 0 \\
\hline 0 & 1 \\
\hline 1 & 0 \\
\hline z & z \\
\hline
\end{tabular}

Examples:
Example 1
\$setup ( data, posedge clk, 10, notify_reg ) ;
\$width ( posedge clk, 16, notify_reg ) ;

Example 2 Consider a more complex example of how to use notifiers in a behavioral model. The following example uses a notifier to set the D flip-flop output to x when a timing violation occurs in an edge-sensitive UDP.
```

primitive posdff_udp(q, clock, data, preset, clear, notifier);
output q; reg q;
input clock, data, preset, clear, notifier;
table
//clock data p c notifier state q
//--------------------------------------

| r | 0 | 1 | 1 | $?$ | $:$ | $?$ | $:$ | 0 | ; |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| r | 1 | 1 | 1 | $?$ | $:$ | $?$ | $:$ | 1 | ; |


| p | 1 | $?$ | 1 | $?$ | $:$ | 1 | $:$ | 1 | $;$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| p | 0 | 1 | $?$ | $?$ | $:$ | 0 | $:$ | 0 | $;$ |

    n ? ? ? ? : ? : - ;
    ? * ? ? ? : ? : - ;
    ? ? 0 1 ? : ? : 1 ;
    ? ? * 1 ? : 1 : 1 ;
    ? ? 1 0 ? : ? : 0 ;
    ? llllllllll
    endtable
endprimitive

```
```

module dff(q, qbar, clock, data, preset, clear);
output q, qbar;
input clock, data, preset, clear;
reg notifier;
and (enable, preset, clear);
not (qbar, ffout);
buf (q, ffout);
posdff_udp (ffout, clock, data, preset, clear, notifier);
specify
// Define timing check specparam values
specparam $\mathrm{t} S \mathrm{U}=10$, $\mathrm{tHD}=1$, $\mathrm{tPW}=25$, $\mathrm{tWPC}=10$, $\mathrm{tREC}=5$;
// Define module path delay rise and fall min:typ:max values
specparam $\mathrm{tPLHc}=4: 6: 9$, $\mathrm{tPHLc}=5: 8: 11$;
specparam $\mathrm{tPLHpc}=3: 5: 6, \mathrm{tPHLpc}=4: 7: 9$;
// Specify module path delays
(clock *> q,qbar) = (tPLHc, tPHLc);
(preset,clear *> q,qbar) = (tPLHpc, tPHLpc);
// Setup time : data to clock, only when preset and clear are 1
\$setup (data, posedge clock \&\&\& enable, tSU, notifier);
// Hold time: clock to data, only when preset and clear are 1
\$hold (posedge clock, data \&\&\& enable, tHD, notifier);
// Clock period check
\$period(posedge clock, tPW, notifier);
// Pulse width : preset, clear
\$width (negedge preset, tWPC, 0 , notifier);
\$width (negedge clear, tWPC, 0, notifier);
// Recovery time: clear or preset to clock
\$recovery (posedge preset, posedge clock, tREC, notifier);
\$recovery (posedge clear, posedge clock, tREC, notifier);
endspecify
endmodule

```

NOTE-This model applies to edge-sensitive UDPs only; for level-sensitive models, an additional UDP for \(\times\) propagation has to be generated.

\subsection*{15.5.1 equirements for accurate simulation}

In order to accurately model negative value timing checks:
a) A timing violation shall be triggered if the signal changes in the violation window, exclusive of the endpoints. Violation windows smaller than two units of simulation precision can not yield timing violations.
b) The value of the latched data shall be the one which is stable during the violation window, again, exclusive of the endpoints.

To facilitate these modeling requirements, delayed copies of the data and reference signals are generated in the timing checks, and these are used internally for timing check evaluation at run-time. The setup and hold times used internally are adjusted so as to shift the violation window and make it overlap the reference signal.

Delayed data and reference signals can be declared within the timing check so they can be used in the model's functional implementation to insure accurate simulation. If no delayed signals are declared in the timing check, and if a negative setup or hold value is present, then implicit delayed signals are created. Since implicit delayed signals can not be used in defining model behavior, such a model can possibly behave incorrectly.

\section*{Examples:}

Example 1:
```

\$setuphold (posedge CLK, DATA, -10, 20);

```

Implicit delayed signals shall be created for CLK and DATA, but it shall not be possible to access them. The \$setuphold check shall be properly evaluated, but functional behavior shall not always be accurate. The old DATA value shall be incorrectly clocked in if DATA transitions between posedge CLK and 10 time units later.

Example 2:
```

\$setuphold(posedge CLK, DATA1, -10, 20);
\$setuphold(posedge CLK, DATA2, -15, 18);

```

Implicit delayed signals shall be created for CLK, DATA1 and DATA2, one for each. Even though CLK is referenced in two different timing checks, only one implicit delayed signal is created, and it is used for both timing checks.

Example 3:
If a given signal has a delayed signal in some timing checks but not in others, the delayed signal shall be used in both cases:
```

\$setuphold (posedge CLK, DATA1, -10, 20,,,, del_CLK, del_DATA1);
\$setuphold(posedge CLK, DATA2, -15, 18);

```

Explicit delayed signals of del CLK and del DATA1 are created for CLK and DATA1, while an implicit delayed signal is created for DATA2. In other words, CLK has only one delayed signal created for it, del_CLK, rather than one explicit delayed signal for the first check, and another implicit delayed signal for the second check.

The delayed versions of the signals, whether implicit or explicit, shall be used in the \$setup, \$hold, \$setuphold, \$recovery, \$removal, \$recrem, \$width, \$period and \$nochange timing checks, and these checks shall have their limits adjusted accordingly. This ensures the notifier shall be toggled at the proper moment. If the adjusted limit becomes less than or equal to 0 , the limit shall be set to 0 and the simulator shall issue a warning.

The delayed versions of the signals shall not be used for the \$skew, \$fullskew and \$timeskew timing checks because it can possibly result in the reversal of the order of signal transitions. This causes the notifiers for these timing checks to toggle at the wrong time relative to the rest of the model, perhaps resulting in transitions to X due to a timing check violation being canceled. This issue shall be addressed in the model, possibly by using separate notifiers for these checks.

It is possible for a set of negative timing check values to be mutually inconsistent and produce no solution for the delay values of delayed signals. In these situations the simulator shall issue a warning message. The inconsistency shall be resolved by changing the smallest negative limit value to 0 and recalculating the delays for the delayed signals, and this shall be repeated until a solution is reached. This procedure shall always produce a solution because in the worst case all negative limit values become 0 , and no delayed signals are needed.

The delayed timing check signals are only actually delayed when negative limit values are present. If a timing check signal becomes delayed by more than the propagation delay from that signal to an output, that output shall take longer than its propagation delay to change. It shall instead transition at the same time which the delayed timing check signal changes. Thus, the output shall behave as if its specify path delay were equal to the delay applied to the timing check signal. This situation can only arise when unique setup/hold or removal/recovery times are given for each edge of the data signal.

\section*{Example:}
```

(CLK = Q) = 6;
\$setuphold (posedge CLK, posedge D, -3, 8, , , , dCLK, dD);
\$setuphold (posedge CLK, negedge D, -7, 13, , , , dCLK, dD);

```

The setup time of -7 (the larger in absolute value of -3 and -7 ) creates a delay of 7 for dCLK , and so output \(Q\) shall not change until 7 time units after a positive edge on CLK, rather than the 6 time units given in the specify path.

\subsection*{15.5.2 Conditions in negative timing checks}

Conditions can be associated with both the reference and data signals by using the \(\& \& \&\) operator, but when either the setup or hold time is negative the conditions need to be paired with reference and data signals in a more flexible way. This example illustrates why.

This pair of \$setup and \$hold checks work together to provide the same check as a single \$setuphold:
```

\$setup (data, clk\&\&\&cond1, tsetup, ntfr);
\$hold (clk, data\&\&\&cond1, thold, ntfr);

```
clk is the timecheck event for the \$setup check, while data is the timecheck event for the \$hold check. This can not be represented in a single \$setuphold check, and so additional arguments are provided to make this possible. These arguments are timestamp_cond and timecheck_cond, and they immediately follow the notifier (see 15.2.3). This \$setuphold check is equivalent to the separate \$setup and \$hold checks shown above:
```

\$setuphold( clk, data, tsetup, thold, ntfr, , cond1);

```

The timestamp_cond argument is null, while the timecheck_cond argument is cond1.
The timestamp_cond and timecheck_cond arguments are associated with either the reference or data signals based on which delayed version of these signals occurs first. timestamp_cond is associated with the delayed signal which transitions first, while timecheck_cond is associated with the delayed signal which transitions second.

Delayed signals are only created for the reference and data signals, and not for any condition signals associated with them. Therefore, timestamp_cond and timecheck_cond are not implicitly delayed by the simulator. Delayed condition signals for the timestamp_cond and timecheck_cond fields can be created by making them a function of the delayed signals.

\section*{Example:}
```

assign TE cond_D = (dTE !== 1' b1);
assign TE_cond_TI = (dTE !== 1'b0);
assign DXTI_con}d=(dTI !== dD)
specify
\$setuphold (posedge CP, D, -10, 20, notifier, ,TE_cond_D, dCP, dD);
\$setuphold (posedge CP, TI, 20, -10, notifier, ,TE_cond_TI, dCP,
dTI);
\$setuphold (posedge CP, TE, -4, 8, notifier, ,DXTI_cond, dCP,
dTE); endspecify

```

The assign statements create condition signals which are functions of the delayed signals. Creating delayed signal conditions synchronizes the conditions with the delayed versions of the reference and data signals used to perform the checks.

The first \$setuphold has a negative setup time, and so the timecheck condition TE_cond_D is associated with data signal D. The second \$setuphold has a negative hold time, and so the timecheck condition TE_cond_TI is associated with reference signals CP. The third \$setuphold has a negative setup time, and so the timecheck condition DXTI_cond is associated with data signal TE.

The violation windows for the example are shown in Figure 48.


Figure 48-Timing check violation windows
These are the delay values calculated for the delayed signals:
\begin{tabular}{lr} 
dCP & 10.01 \\
dD & 0.00 \\
dTI & 20.02 \\
dTE & 2.02
\end{tabular}

Use of delayed signals in creating the signals for the timestamp_cond and timecheck_cond arguments is not required, but it is usually closer to actual device behavior.

\subsection*{15.5.3 Notifiers in negative timing checks}

Because the reference and data signals are delayed internally, the detection of the timing violation is also delayed. Notifier regs in negative timing checks shall be toggled when the timing check detects a timing violation, which occurs when the delayed signals as measured by the adjusted timing check values are in violation, not when the undelayed signals at the model inputs as measured by the original timing check values are in violation.

\subsection*{15.5.4 ption behavior}

As already mentioned, the ability of Verilog simulators to handle negative values in \$setuphold and \$recrem timing checks shall be enabled with an invocation option. It is possible models written to accept negative timing check values with delayed reference and/or delayed data signals can be run without this invocation option enabled. In this circumstance the delayed reference and data signals become copies of the original reference and data signals. The same occurs if an invocation option turning off all timing checks is used.

\subsection*{15.6 Enabling timing checks with conditioned events}

A construct called a conditioned event ties the occurrence of timing checks to the value of a conditioning signal. Syntax 15-16 shows the syntax for controlled timing check event.
```

timing_check_event ::= (From Annex A - A.7.5.3)
[timing_check_event_control] specify_terminal_descriptor [ \&\&\&\& timing_check_condition ]
controlled_timing_check_event ::=
timing_check_event_control specify_terminal_descriptor [ \&\&\&\& timing_check_condition ]
timing_check_event_control ::=
posedge
|negedge
| edge_control_specifier
specify_terminal_descriptor ::=
specify_input_terminal_descriptor
| specify_output_terminal_descriptor
timing_check_condition ::=
scalar_timing_check_condition
| ( scalar_timing_check_condition )
scalar_timing_check_condition ::=
expression
| expression
expression == scalar_constant
expression === scalar_constant
expression != scalar_constant
| expression !== scalar_constant
scalar_constant ::=
1'b0 1'b1 1 1'B0 | 1'B1 |'b0 |'b1 |'B0 |'B1 | 1 0

```

\section*{Syntax 15-16 Syntax for controlled timing check event}

The comparisons used in the condition can be deterministic, as in \(===,!==, \sim\), or no operation, or nondeterministic, as in \(==\) or \(!=\). When comparisons are deterministic, an \(\times\) value on the conditioning signal shall not enable the timing check. For nondeterministic comparisons, an \(x\) on the conditioning signal shall enable the timing check.

The conditioning signal shall be a scalar net; if a vector net or an expression resulting in a multi-bit value is used, then the least significant bit of the vector net or the expression value is used.

If more than one conditioning signal is required for conditioning timing checks, appropriate logic shall be combined in a separate signal outside the specify block, which can be used as the conditioning signal.

\section*{Examples:}

Example 1 To illustrate the difference between conditioned and unconditioned timing check events, consider the following example with unconditioned timing check:
```

\$setup( data, posedge clk, 10 );

```

Here, a setup timing check shall occur every time there is a positive edge on the signal clk .
To trigger the setup check on the positive edge on the signal \(\mathrm{cl} k\) only when the signal clr is high, rewrite the command as
```

\$setup( data, posedge clk \&\&\&\& clr, 10 ) ;

```

Example 2 This example shows two ways to trigger the same timing check as in example 1 (on the positive clk edge) only when clr is low. The second method uses the \(===\) operator, which makes the comparison deterministic.
```

\$setup( data, posedge clk \&\&\& (~clr), 10 ) ;
\$setup( data, posedge clk \&\&\& (clr===0), 10 );

```

Example 3 To perform the previous sample setup check on the positive clk edge only when clr and set are high, add the following statement outside the specify block:
```

and new_gate( clr_and_set, clr, set );

```

Then add the condition to the timing check using the signal clr_and_set as follows:
```

\$setup( data, posedge clk $\boldsymbol{\&} \boldsymbol{\&} \boldsymbol{\&}$ clr_and_set, 10 );

```

\subsection*{15.7 Vector signals in timing checks}

Either or both signals in a timing check can be a vector. This shall be interpreted as a single timing check where the transition of one or more bits of a vector is considered a single transition of that vector.

\section*{Example:}
```

module DFF (Q, CLK, DAT);
input CLK;
input [ 7:0] DAT;
output [ 7:0] Q;
always @(posedge clk)
Q = DAT;
specify
\$setup (DAT, posedge CLK, 10);
endspecify
endmodule

```

If DAT transitions from 'b00101110 to 'b01010011 at time 100, and CLK transitions from 0 to 1 at time 105 , then the \$setup timing check shall still only report a single timing violation.

Simulators can provide an option causing vectors in timing checks to result in the creation of multiple sin-gle-bit timing checks. For timing checks with only a single signal, such as \$period or \$width, a vector of width \(N\) results in \(N\) unique timing checks. For timing checks with two signals, such as \$setup, \$hold, \$setuphold, \$skew, \$timeskew, \$fullskew, \$recovery, \$removal, \$recrem and \$nochange, where M and N are the widths of the signals, the result is \(M^{\star} N\) unique timing checks. If there is a notifier, all the timing checks trigger that notifier.

With such an option enabled, the above example yields six timing violation because six bits of DAT transitioned.

\subsection*{15.8 Negative timing checks}

Both the \$setuphold and \$recrem timing checks can accept negative values when the negative timing check option is enabled. The behavior of these two timing checks is identical with respect to negative values. The descriptions in this section are for the \$setuphold timing check, but apply equally to the \$recrem timing check.

The setup and hold timing check values define a timing violation window with respect to the reference signal edge during which the data shall remain constant. Any change of the data during the specified window causes a timing violation. The timing violation is reported and, through the notifier reg, other actions can take place in the model, such as forcing the output of a flip-flop to \(X\) when it detects a timing violation.

A positive value for both setup and hold times implies this violation window straddles the reference signal shown in Figure 49


Figure 49—Data constraint interval, positive setup/hold
A negative hold or setup time means the violation window is shifted to either before or after the reference edge. This can happen in a real device because of disparate internal device delays between the internal clock and data signal paths. These internal device delays are illustrated in Figure 50 showing how significant differences in these delays can cause negative setup or hold values.


Figure 50—Data constraint interval, negative setup/hold

\section*{16. Backannotation using the Standard Delay Format (SDF)}

SDF files contain timing values for specify path delays, specparam values, timing check constraints, and interconnect delays. SDF files can also contain other information in addition to simulation timing, but these need not concern Verilog simulation. The timing values in SDF files usually come from ASIC delay calculation tools that take advantage of connectivity, technology, and layout geometry information.

Verilog backannotation is the process by which timing values from the SDF file update specify path delays, specparam values, timing constraint values, and interconnect delays.

All this information is covered further in IEEE Std 1497-1999 [B2].

\subsection*{16.1 The SDF annotator}

The term SDF Annotator refers to any tool capable of backannotating SDF data to a Verilog simulator. It shall report a warning for any data it is unable to annotate.

An SDF file can contain many constructs which are not related to specify path delays, specparam values, timing check constraint values, or interconnect delays. An example is any construct in the TIMINGENV section of the SDF file. All constructs unrelated to Verilog timing shall be ignored without any warnings issued.

Any Verilog timing value for which the SDF file does not provide a value shall not be modified during the backannotation process, and its pre-backannotation value shall be unchanged.

\subsection*{16.2 Mapping of SDF constructs to Verilog}

SDF timing values appear within a CELL declaration, which can contain one or more of DELAY, TIMINGCHECK and LABEL sections. The DELAY section contains propagation delay values for specify paths and interconnect delays. The TIMINGCHECK section contains timing check constraint values. The LABEL section contains new values for specparams. Backannotation into Verilog is done by matching SDF constructs to the corresponding Verilog declarations, then replacing the existing Verilog timing values with those from the SDF file.

\subsection*{16.2.1 Mapping of SDF delay constructs to Verilog declarations}

When annotating DELAY constructs that are not interconnect delays (covered in 16.2.3), the SDF annotator looks for specify paths where the names and conditions match. When annotating TIMINGCHECK constructs, the SDF annotator looks for timing checks of the same type where the names and conditions match. Table 62 shows which Verilog structures can be annotated by each SDF construct in the DELAY section.

Table 62-Mapping of SDF delay constructs to Verilog declarations
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ SDF Construct } & \multicolumn{1}{c|}{ Verilog annotated structure } \\
\hline (PATHPULSE... & Conditional and non-conditional specify path pulse limits \\
\hline (PATHPULSEPERCENT... & Conditional and non-conditional specify path pulse limits \\
\hline (IOPATH... & Conditional and non-conditional specify path delays/pulse limits \\
\hline (IOPATH (RETAIN... & \begin{tabular}{l} 
Conditional and non-conditional specify path delays/pulse limits, \\
RETAIN ignored without warning
\end{tabular} \\
\hline (COND (IOPATH... & Conditional specify path delays/pulse limits \\
\hline (COND (IOPATH (RETAIN... & \begin{tabular}{l} 
Conditional specify path delays/pulse limits, RETAIN ignored without \\
warning
\end{tabular} \\
\hline (CONDELSE (IOPATH... & ifnone \\
\hline (CONDELSE (IOPATH (RETAIN... & \begin{tabular}{l} 
ifnone, RETAIN ignored without warning \\
\hline (DEVICE... \\
\hline (DEVICE port_instance...
\end{tabular} \begin{tabular}{l} 
All specify paths to module outputs. If no specify paths, all primitives \\
driving module outputs.
\end{tabular} \\
\hline & \begin{tabular}{l} 
If port_instance is a module instance, all specify paths to module \\
outputs. If no specify paths, all primitives driving module outputs. If \\
port_instance is a module instance output, all specify paths to that \\
module output. If no specify path, all primitives driving that module \\
output.
\end{tabular} \\
\hline
\end{tabular}

In this example the source SDF signal sel matches the source Verilog signal, and the destination SDF signal zout also matches the destination Verilog signal, and so the rise/fall times of 1.3 and 1.7 are annotated to the specify path.

SDF file:
(IOPATH sel zout (1.3) (1.7))

Verilog specify path:
(sel => zout) = 0;

A conditional IOPATH delay between two ports shall annotate only to Verilog specify paths between those same two ports with the same condition. In this example the rise/fall times of 1.3 and 1.7 are annotated only to the second specify path.

SDF file:
(COND mode (IOPATH sel zout (1.3) (1.7)))

Verilog specify paths:
```

if (!mode) (sel => zout) = 0;
if (mode) (sel => zout) = 0;

```

A non-conditional IOPATH delay between two ports shall annotate to all Verilog specify paths between those same two ports. In this example the rise/fall times of 1.3 and 1.7 are annotated to both specify paths.

SDF file:
(IOPATH sel zout (1.3) (1.7))

Verilog specify paths:
```

if (!mode) (sel => zout) = 0;
if (mode) (sel => zout) = 0;

```

\subsection*{16.2.2 Mapping of SDF timing check constructs to Verilog}

Table 63 shows which Verilog timing checks are annotated to by each type of SDF timing check. v1 is the first value of a timing check, v 2 is the second value, while x indicates no value is annotated.

Table 63-Mapping of SDF timing check constructs to Verilog
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ SDF Timing Check } & \multicolumn{1}{c|}{ Annotated Verilog Timing checks } \\
\hline (SETUP v1... & \$setup(v1), \$setuphold(v1,x) \\
\hline (HOLD v1... & \$hold(v1), \$setuphold(x,v1) \\
\hline (SETUPHOLD v1 v2... & \$setup(v1), \$hold(v2), \$setuphold(v1,v2) \\
\hline (RECOVERY v1... & \$recovery(v1), \$recrem(v1,x) \\
\hline (REMOVAL v1... & \$removal(v1), \$recrem(x,v1) \\
\hline (RECREM v1 v2... & \$recovery(v1), \$removal(v2), \$recrem(v1,v2) \\
\hline (SKEW v1... & \$skew(v1) \\
\hline (TIMESKEW v1...* & \$timeskew(v1) \\
\hline (FULLSKEW v1 v2... & \$fullskew(v1,v2) \\
\hline (WIDTH v1... & \$width(v1,x) \\
\hline (PERIOD v1... & \$period(v1) \\
\hline (NOCHANGE v1 v2... & \$nochange(v1,v2) \\
\hline
\end{tabular}
*Not part of current SDF standard Not usually implemented in Verilog simulators

The reference and data signals of timing checks can have logical condition expressions and edges associated with them. An SDF timing check with no conditions or edges on any of its signals shall match all corresponding Verilog timing checks regardless of whether conditions are present or not. In this example the SDF timing check shall annotate to all the Verilog timing checks:

SDF file:
```

(SETUPHOLD data clk (3) (4))

```

Verilog timing checks:
```

\$setuphold (posedge clk\&\&\& mode, data, 1, 1, ntfr);
\$setuphold (negedge clk\&\&\&!mode, data, 1, 1, ntfr);

```

When conditions and/or edges are associated with the signals in an SDF timing check, then they shall match those in any corresponding Verilog timing check before annotation shall happen. In this example the SDF timing check shall annotate to the first Verilog timing check, but not the second:

SDF file:
```

(SETUPHOLD data (posedge clk) (3) (4))

```

Verilog timing checks:
```

\$setuphold (posedge clk\&\&\& mode, data, 1, 1, ntfr); // Annotated
\$setuphold (negedge clk\&\&\&!mode, data, 1, 1, ntfr); // Not annotated

```

Here, the SDF timing check shall not annotate to any of the Verilog timing checks:
SDF file:
```

(SETUPHOLD data (COND !mode (posedge clk)) (3) (4))

```

Verilog timing checks:
```

\$setuphold (posedge clk\&\&\& mode, data, 1, 1, ntfr); // Not annotated
\$setuphold (negedge clk\&\&\&!mode, data, 1, 1, ntfr); // Not annotated

```

\subsection*{16.2.3 SDF annotation of specparams}

The SDF LABEL construct annotates to specparams. Any expression containing one or more specparams is reevaluated when annotated to from an SDF file.

This example shows SDF LABEL constructs annotating to specparams in a Verilog module. The specparams are used in procedural delays to control when the clock transitions. The SDF LABEL construct annotates the values of dhigh and dlow, thereby setting the period and duty cycle of the clock.

SDF file:
(LABEL
(ABSOLUTE
(dhigh 60)
(dlow 40)))

Verilog file:
```

module clock(clk);
output clk;
reg clk;
specparam dhigh=0, dlow=0;
initial clk = 0;
always
begin
\#dhigh clk = 1; // Clock remains low for time dlow
// before transitioning to 1
\#dlow clk = 0; // Clock remains high for time dhigh
// before transitioning to 0
end;
endmodule

```

This example shows a specparam in an expression of a specify path. The SDF LABEL construct can be used

In a Verilog module, a reference to an annotated port, wherever it occurs, whether in \$monitor and \$display statements or in expressions, shall provide the delayed signal value. A reference to the source shall yield the undelayed signal value, while a reference to the load shall yield the delayed signal value. In general, references to the signal value hierarchically before the load shall yield the undelayed signal value, while references to the signal at or hierarchically after the load shall yield the delayed signal value. An annotation to a hierarchical port shall affect all connected ports at higher or lower hierarchical levels, depending on the direction of annotation. An annotation from a source port shall be interpreted as being from all sources hierarchically higher or lower than that source port.

Up-hierarchy annotations shall be properly handled. This situation arises when the load is hierarchically above the source. The delay to all ports hierarchically above the load or which connect to the net at points hierarchically above the load is the same as the delay to that load.

Down-hierarchy annotation shall also be properly handled. This situation arises when the source is hierarchically above the load. The delay to the load is interpreted as being from all ports at or above the source or which connect to the net at points hierarchically above the source.

Hierarchically overlapping annotations are permitted. This occurs when annotations to or from the same port take place at different hierarchical levels, and therefore do not correspond to the same hierarchical subset of ports. In this example, the first INTERCONNECT statement annotates to all ports of the net that are at or hierarchically within i53/selmode, while the second annotates to a smaller subset of ports, only those at or hierarchically within i53/u21/in:
```

(INTERCONNECT i14/u5/out i53/selmode (1.43) (2.17))
(INTERCONNECT i14/u5/out i53/u21/in (1.58) (1.92))

```

Overlapping annotations can occur in many different ways, particularly on multi-source/multi-load nets, and SDF annotation shall properly resolve all the interactions.

\subsection*{16.3 Multiple annotations}

SDF annotation is an ordered process. The constructs from the SDF file are annotated in their order of occurrence. This means that annotation of an SDF construct can be changed by annotation of a subsequent construct that either modifies (INCREMENT) or overwrites (ABSOLUTE) it. These do not have to be the same construct. This example first annotates pulse limits to an IOPATH, then annotates the entire IOPATH, thereby overwriting the pulse limits that were just annotated:
```

(DELAY
(ABSOLUTE
(PATHPULSE A Z (2.1) (3.4))
(IOPATH A Z (3.5) (6.1))

```

Overwriting the pulse limits can be avoided by using empty parentheses to hold the current values of the pulse limits:
```

(DELAY
(ABSOLUTE
(PATHPULSE A Z (2.1) (3.4))
(IOPATH A Z ((3.5) () ()) ((6.1) () ()) )

```

The above annotation can be simplified into a single statement like this:
```

(DELAY
(ABSOLUTE
(IOPATH A Z ((3.5) (2.1) (3.4)) ((6.1) (2.1) (3.4)) )

```

A PORT annotation followed by an INTERCONNECT annotation to the same load shall cause only the delay from the INTERCONNECT source to be affected. For this net with three sources and a single load, the delay from all sources except i13/out remains 6:
```

(DELAY
(ABSOLUTE
(PORT i15/in (6))
(INTERCONNECT i13/out i15/in (5))

```

An INTERCONNECT annotation followed by a PORT annotation shall cause the INTERCONNECT annotation to be overwritten. Here, the delays from all sources to the load shall become 6 .
```

(DELAY
(ABSOLUTE
(INTERCONNECT i13/out i15/in (5))
(PORT i15/in (6))

```

\subsection*{16.4 Multiple SDF files}

More than one SDF file can be annotated. Each call to the \$sdf_annotate task annotates the design with timing information from an SDF file. Annotated values either modify (INCREMENT) or overwrite (ABSOLUTE) values from earlier SDF files. Different regions of a design can be annotated from different SDF files by specifying the region s hierarchy scope as the second argument to \$sdf_annotate.

\section*{16.5 ulse limit annotation}

For SDF annotation of delays (not timing constraints), the default values annotated for pulse limits shall be calculated using the percentage settings for the reject and error limits. By default these limits are \(100 \%\), but they can be modified through invocation options. For example, assuming invocation options have set the reject limit to \(40 \%\) and the error limit to \(80 \%\), this SDF construct shall annotate a delay of 5 , a reject limit of 2 , and an error limit of 4:
(DELAY
(ABSOLUTE
(IOPATH A Z (5))

Given that the specify path delay was originally 0 , this annotation results in a delay of 5 and pulse limits of 0 :
```

(DELAY
(ABSOLUTE
(IOPATH A Z ((5) () ()) )

```

Annotations in INCREMENT mode can result in pulse limits less than 0 , in which case they shall be adjusted to 0 . For example, if the specify path pulse limits were both 3 , this annotation results in a 0 value for both pulse limits:
```

(DELAY
(INCREMENT
(IOPATH A Z (() (-4) (-5)) )

```

There are two SDF constructs that annotate only to pulse limits, PATHPULSE and PATHPULSEPERCENT. They do not affect the delay. When PATHPULSE sets the pulse limits to values greater than the delay Verilog shall exhibit the same behavior as if the pulse limits had been set equal to the delay.

\subsection*{16.6 SDF to Verilog delay value mapping}

Verilog specify paths and interconnects can have unique delays for up to twelve state transitions (see 14.3.1). All other constructs, such as gate primitives and continuous assignments, can have only three state transition delays (see 7.14).

For Verilog specify path and interconnect delays, the number of transition delay values provided by SDF might be less than twelve.

Table 65 shows how fewer than twelve SDF delays are extended to be twelve delays. The Verilog transition types are shown down the left hand side, while the number of SDF delays provided is shown across the top. The SDF values are given the names v1 through v12.

Table 65-SDF to Verilog delay value mapping
\begin{tabular}{|c|c|c|c|c|c|}
\hline \multirow{2}{*}{Verilog transition} & \multicolumn{5}{|c|}{Number of SDF delay values provided} \\
\hline & 1 value & 2 values & 3 values & 6 values & 12 values \\
\hline \(0->1\) & v1 & v1 & v1 & v1 & v1 \\
\hline \(1->0\) & v1 & v2 & v2 & v2 & v2 \\
\hline \(0->\mathrm{z}\) & v1 & v1 & v3 & v3 & v3 \\
\hline z -> 1 & v1 & v1 & v1 & v4 & v4 \\
\hline \(1->\mathrm{z}\) & v1 & v2 & v3 & v5 & v5 \\
\hline z -> 0 & v1 & v2 & v2 & v6 & v6 \\
\hline \(0->\mathrm{x}\) & v1 & v1 & \(\min (\mathrm{v} 1, \mathrm{v} 3)\) & \(\min (\mathrm{v} 1, \mathrm{v} 3)\) & v7 \\
\hline x -> 1 & v1 & v1 & v1 & \(\max (\mathrm{v} 1, \mathrm{v} 4)\) & v8 \\
\hline \(1->\mathrm{x}\) & v1 & v2 & \(\min (\mathrm{v} 2, \mathrm{v} 3)\) & \(\min (\mathrm{v} 2, \mathrm{v} 5)\) & v9 \\
\hline x -> 0 & v1 & v2 & v2 & \(\max (\mathrm{v} 2, \mathrm{v} 6)\) & v10 \\
\hline x -> z & v1 & \(\max (\mathrm{v} 1, \mathrm{v} 2)\) & v3 & \(\max (\mathrm{v} 3, \mathrm{v} 5)\) & v11 \\
\hline z -> x & v1 & \(\min (\mathrm{v} 1, \mathrm{v} 2)\) & \(\min (\mathrm{v} 1, \mathrm{v} 2)\) & \(\min (\mathrm{v} 4, \mathrm{v} 6)\) & v12 \\
\hline
\end{tabular}

For other delays that can have at most three values, the expansion of less than three SDF delays into three Verilog delays is covered in Table 39. More than three SDF delays are reduced to three Verilog delays by simply ignoring the extra delays. The delay to the X -state is created from the minimum of the other three delays.

\section*{17. System tasks and functions}

This clause describes system tasks and functions that are considered part of the Verilog HDL. These system tasks and functions are divided into ten categories as follows:
\begin{tabular}{|c|c|c|c|}
\hline Display tasks & [17.1] & PLA modeling tasks & [17.5] \\
\hline \$display & \$strobe & \$async\$and\$array & \$async\$and\$plane \\
\hline \$displayb & \$strobeb & \$async\$nand\$array & \$async\$nandSplane \\
\hline \$display & \$strobeh & ¢\$or\$array & - \\
\hline \$displayo & \$strobeo & \$async\$nor\$array & \$async\$nor\$plane \\
\hline \$monitor & \$write & \$sync\$and\$array & \$sync\$and\$plane \\
\hline \$monitorb & \$writeb & \$sync\$nand\$array & \$sync\$nand\$plane \\
\hline \$monitorh & \$writeh & \$sync\$or\$array & \$sync\$or\$plane \\
\hline \$monitoro
\$monitoroff & \$writeo & \$sync\$nor\$array & \$sync\$nor\$plane \\
\hline \$monitoroff & Snonitoron & & \\
\hline File I/O tasks & [17.2] & Stochastic analysis task & ks [17.6] \\
\hline \$fclose & \$fopen & \$q_initialize & \$q_add \\
\hline \$fdisplay & \$fstrobe & \$q_remove & \$q_full \\
\hline \$fdisplayb & \$fstrobeb & \$q_exam & \\
\hline \$fdisplayh & \$fstrobeh & & \\
\hline \$fdisplayo & \$fstrobeo & & \\
\hline \$fgetc & \$ungetc & Simulation time functio & ons [17.7] \\
\hline \$fflush & \$ferror & Srealtime & \$stime \\
\hline \$fgets & \$rewind & Stime & \\
\hline \$fmonitor & \$fwrite & & \\
\hline \$fmonitorb & \$fwriteb & & \\
\hline \$fmonitorh & \$fwriteh & Conversion functions & [17.8] \\
\hline \$fmonitoro & \$fwriteo & \$bitstoreal & \\
\hline \$readmemb & \$readmemh & \$itor & Sitoi \\
\hline \$swrite & \$swriteb & & \\
\hline \$swriteo & \$swriteh & & \$unsigned \\
\hline \$sformat & \$sdf_annotate & & \\
\hline \$fscanf & \$sscanf & Probabilistic distributi & on functions [17.9] \\
\hline \$fread & \$ftell & & \\
\hline \$fseek & & \begin{tabular}{l}
Sdist_chi_square \\
\$dist_exponential
\end{tabular} & \$dist_erlang \$dist_normal \\
\hline Timescale tasks & [17.3] & \$dist_poisson & \$dist_t \\
\hline \$printtimescale & \$timeformat & \$dist_uniform & \$random \\
\hline Simulation control tasks & [17.4] & Command line input & [17.10] \\
\hline \$finish & \$stop & \$test\$plusargs & \$value\$plusargs \\
\hline
\end{tabular}

These utility tasks and functions provide some broadly useful capabilities. The following clauses describe the behavior of these tasks and functions. Additional tasks for value change dump (VCD) are described in Clause 18.

\subsection*{17.1 Display system tasks}

The display group of system tasks are divided into three categories: the display and write tasks, strobed monitoring tasks, and continuous monitoring tasks.

\subsection*{17.1.1 The display and write tasks}
```

display_tasks ::=
display_task_name ( list_of_arguments );
display_task_name ::=
\$display | Sdisplayb | \$displayo | Sdisplayh
|wwrite | $writeb |$writeo | \$writeh

```

\section*{Syntax 17-1 Syntax for \$display and \$write system tasks}

These are the main system task routines for displaying information. The two sets of tasks are identical except that \$display automatically adds a newline character to the end of its output, whereas the \$write task does not.

The \(\$\) display and \(\$\) write tasks display their arguments in the same order as they appear in the argument list. Each argument can be a quoted string, an expression that returns a value, or a null argument.

The contents of string arguments are output literally except when certain escape sequences are inserted to display special characters or to specify the display format for a subsequent expression.

Escape sequences are inserted into a string in three ways:
The special character \(\backslash\) indicates that the character to follow is a literal or nonprintable character (see Table 66).
The special character \% indicates that the next character should be interpreted as a format specification that establishes the display format for a subsequent expression argument (see Table 67). For each \% character, with the exception of \%m that appears in a string, a corresponding expression argument shall be supplied after the string.
The special character string \(\% \%\) indicates the display of the percent sign character \(\%\) (see Table 66).
Any null argument produces a single space character in the display. (A null argument is characterized by two adjacent commas in the argument list.)

The \$display task, when invoked without arguments, simply prints a newline character. A \$write task supplied without parameters prints nothing at all.

\subsection*{17.1.1.1 Escape sequences for special characters}

The escape sequences given in Table 66, when included in a string argument, cause special characters to be displayed.

Table 66-Escape sequences for printing special characters
\begin{tabular}{|l|l|}
\hline Argument & \multicolumn{1}{c|}{ Description } \\
\hline\(\backslash n\) & The newline character \\
\hline\(\backslash t\) & The tab character \\
\hline\(\backslash \\
) & The \(\backslash\) character \\
\hline\(\backslash "\) & The " character \\
\hline\(\backslash d d d\) & A character specified by 1 to 3 octal digits \\
\hline\(\% \%\) & The \% character \\
\hline
\end{tabular}

\section*{Example:}
```

module disp;
initial begin
\$display("<br>\t<br>\n\"\123");
end
endmodule
Simulating this example shall display the following:
\ \
"S

```

\subsection*{17.1.1.2 Format specifications}

Table 67 shows the escape sequences used for format specifications. Each escape sequence, when included in a string argument, specifies the display format for a subsequent expression. For each \% character (except \%m) that appears in a string, a corresponding expression shall follow the string in the argument list. The value of the expression replaces the format specification when the string is displayed.

Any expression argument that has no corresponding format specification is displayed using the default decimal format in \$display and \$write, binary format in \$displayb and \$writeb, octal format in \$displayo and \$writeo, and hexadecimal format in \$displayh and \$writeh.

Table 67-Escape sequences for format specifications
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ Argument } & \multicolumn{1}{|c|}{ Description } \\
\hline\(\%\) h or \%H & Display in hexadecimal format \\
\hline\(\% \mathrm{~d}\) or \%D & Display in decimal format \\
\hline\(\%\) or \%O & Display in octal format \\
\hline\(\% \mathrm{~b}\) or \%B & Display in binary format \\
\hline\(\% \mathrm{c}\) or \%C & Display in ASCII character format \\
\hline\(\% 1\) or \(\% \mathrm{~L}\) & Display library binding information \\
\hline\(\% \mathrm{v}\) or \%V & Display net signal strength \\
\hline
\end{tabular}

\section*{Table 67-Escape sequences for format specifications (continued)}
\begin{tabular}{|l|l|}
\hline\(\% \mathrm{~m}\) or \%M & Display hierarchical name \\
\hline\(\%\) s or \%S & Display as a string \\
\hline\(\%\) or \%T & Display in current time format \\
\hline\(\%\) or or U & Unformatted 2 value data \\
\hline\(\% \mathrm{z}\) or \%Z & Unformatted 4 value data \\
\hline
\end{tabular}

The formatting specification \(\% 1\) (or \(\% \mathrm{~L}\) ) is defined for displaying the library information of the specific module. This information shall be displayed as "library. cell" corresponding to the library name the current module instance was extracted from and the cell name of the current module instance. See Clause 13 for information on libraries and configuring designs.

The formatting specification \(\% \mathrm{u}(\mathrm{or} \% \mathrm{U}\) ) is defined for writing data without formatting (binary values). The application shall transfer the 2 value binary representation of the specified data to the output stream. This escape sequence can be used with any of the existing display system tasks, although \$ fwrite should be the preferred one to use. Any unknown or high-impedance bits in the source shall be treated as zero. This formatting specifier is intended to be used to support transferring data to and from external programs that have no concept of \(x\) and \(z\). Applications that require preservation of \(x\) and \(z\) are encouraged to use the \(\% z\) I/O format specification.

The data shall be written to the file in the native endian format of the underlying system (i.e., in the same endian order as if the PLI was used, and the C language write (2) system call was used). The data shall be written in units of 32 bits with the word containing the LSB written first.

NOTE For POSIX applications: It might be necessary to open files for unformatted I/O with the \(\mathrm{wb}, \mathrm{wb}+\), or \(\mathrm{w}+\mathrm{b}\) specifiers, to avoid the systems implementation of I/O altering patterns in the unformatted stream that match special characters.

The formatting specification \(\% \mathrm{z}\) ( or \(\% \mathrm{Z}\) ) is defined for writing data without formatting (binary values). The application shall transfer the 4 value binary representation of the specified data to the output stream. This escape sequence can be used with any of the existing display system tasks, although \$fwrite should be the preferred one to use.

This formatting specifier is intended to be used to support transferring data to and from external programs that recognize and support the concept of \(x\) and \(z\). Applications that do not require the preservation of \(x\) and \(z\) are encouraged to use the \(\% u\) I/O format specification.

The data shall be written to the file in the native endian format of the underlying system (i.e., in the same endian order as if the PLI was used, and the data were in a s_vpi_vecval structure (See 27.14, Figure 179), and the C language write (2) system call was used to write the structure to disk). The data shall be written in units of 32 bits with the structure containing the LSB written first.

NOTE For POSIX applications: It might be necessary to open files for unformatted I/O with the \(\mathrm{wb}, \mathrm{wb}+\mathrm{or} \mathrm{w}+\mathrm{b}\) specifiers, to avoid the systems implementation of I/O altering patterns in the unformatted stream that match special characters.

The format specifications in Table 68 are used with real numbers and have the full formatting capabilities available in the C language. For example, the format specification \(\% 10.3 \mathrm{~g}\) specifies a minimum field width of 10 with 3 fractional digits.

Table 68-Format specifications for real numbers
\begin{tabular}{|l|l|}
\hline Argument & \multicolumn{1}{c|}{ Description } \\
\hline \%e or \%E & Display real in an exponential format \\
\hline \%f or \%F & Display real in a decimal format \\
\hline \%g or \%G & \begin{tabular}{l} 
Display real in exponential or decimal format, which- \\
ever format results in the shorter printed output
\end{tabular} \\
\hline
\end{tabular}

The net signal strength, hierarchical name, and string format specifications are described in 17.1.1.5 through 17.1.1.7.

The \%t format specification works with the \$timeformat system task to specify a uniform time unit, time precision, and format for reporting timing information from various modules that use different time units and precisions. The \$timeformat task is described in 17.3.2.

\section*{Example:}
```

module disp;
reg [31:0] rval;
pulldown (pd);
initial begin
rval = 101;
\$display("rval = %h hex %d decimal",rval,rval);
\$display("rval = %o octal\nrval = %b bin",rval,rval);
\$display("rval has %c ascii character value",rval);
\$display("pd strength value is %v",pd);
\$display("current scope is %m");
\$display("%s is ascii value for 101",101);
\$display("simulation time is %t", \$time);
end
endmodule

```

Simulating this example shall display the following:
```

rval = 00000065 hex 101 decimal
rval = 00000000145 octal
rval = 00000000000000000000000001100101 bin
rval has e ascii character value
pd strength value is StX
current scope is disp
e is ascii value for 101

```

\subsection*{17.1.1.3 Size of displayed data}

For expression arguments, the values written to the output file (or terminal) are sized automatically.
For example, the result of a 12-bit expression would be allocated three characters when displayed in hexadecimal format and four characters when displayed in decimal format, since the largest possible value for the expression is FFF (hexadecimal) and 4095 (decimal).

When displaying decimal values, leading zeros are suppressed and replaced by spaces. In other radices, leading zeros are always displayed.

The automatic sizing of displayed data can be overridden by inserting a zero between the \% character and the letter that indicates the radix, as shown in the following example.
```

\$display("d=\%0h $a=\% 0 h ", ~ d a t a, ~ a d d r) ;$

```

Example:
```

module printval;
reg [11:0] r1;
initial begin
r1 = 10;
\$display( "Printing with maximum size - :%d: :%h:", r1,r1 );
\$display( "Printing with minimum size - :%0d: :%Oh:", r1,r1 );
end
endmodule
Printing with maximum size - : 10: :00a:
Printing with minimum size - :10: :a:

```

In this example, the result of a 12-bit expression is displayed. The first call to \$display uses the standard format specifier syntax and produces results requiring four and three columns for the decimal and hexadecimal radices, respectively. The second \$display call uses the \(\% 0\) form of the format specifier syntax and produces results requiring two columns and one column, respectively.

\subsection*{17.1.1.4 Unknown and high impedance values}

When the result of an expression contains an unknown or high impedance value, the following rules apply to displaying that value.

In decimal (\%d) format
If all bits are at the unknown value, a single lowercase x character is displayed.
If all bits are at the high impedance value, a single lowercase \(\quad \mathrm{z}\) character is displayed.
If some, but not all, bits are at the unknown value, the uppercase \(X\) character is displayed.
If some, but not all, bits are at the high impedance value, the uppercase Z character is displayed.
Decimal numerals always appear right-justified in a fixed-width field.
In hexadecimal (\%h) and octal (\%) formats
Each group of 4 bits is represented as a single hexadecimal digit; each group of 3 bits is represented as a single octal digit.
If all bits in a group are at the unknown value, a lowercase x is displayed for that digit.
If all bits in a group are at a high impedance state, a lowercase \(z\) is printed for that digit.
If some, but not all, bits in a group are unknown, an uppercase X is displayed for that digit.
If some, but not all, bits in a group are at a high impedance state, then an uppercase Z is displayed for that digit.

In binary ( \(\% \mathrm{~b}\) ) format, each bit is printed separately using the characters \(0,1, \mathrm{x}\), and z .

\section*{Example:}
\begin{tabular}{|c|c|}
\hline STATEMENT & RESULT \\
\hline \$display("\%d", 1'bx); & x \\
\hline \$display("\%h", 14'bx01010); & xxXa \\
\hline \$display("\%h \%o", \(12 \prime \mathrm{~b} 001 \mathrm{xxx101x01}\),l
\(\left.12^{\prime} \mathrm{b} 001 \mathrm{xxx} 101 \mathrm{x} 01\right) ;\) & XXX 1x5X \\
\hline
\end{tabular}

\subsection*{17.1.1.5 Strength format}

The \(\% v\) format specification is used to display the strength of scalar nets. For each \(\% v\) specification that appears in a string, a corresponding scalar reference shall follow the string in the argument list.

The strength of a scalar net is reported in a three-character format. The first two characters indicate the strength. The third character indicates the current logic value of the scalar and can be any one of the values given in Table 69.

\section*{Table 69-Logic value component of strength format}
\begin{tabular}{|l|l|}
\hline Argument & \multicolumn{1}{c|}{ Description } \\
\hline 0 & For a logic 0 value \\
\hline 1 & For a logic 1 value \\
\hline X & For an unknown value \\
\hline Z & For a high impedance value \\
\hline L & For a logic 0 or high impedance value \\
\hline H & For a logic 1 or high impedance value \\
\hline
\end{tabular}

The first two characters the strength characters are either a two-letter mnemonic or a pair of decimal digits. Usually, a mnemonic is used to indicate strength information; however, in less typical cases, a pair of decimal digits can be used to indicate a range of strength levels. Table 70 shows the mnemonics used to represent the various strength levels.

Table 70-Mnemonics for strength levels
\begin{tabular}{|l|l|l|}
\hline \multicolumn{1}{|c|}{ Mnemonic } & Strength name & \multicolumn{1}{c|}{ Strength level } \\
\hline Su & Supply drive & 7 \\
\hline St & Strong drive & 6 \\
\hline Pu & Pull drive & 5 \\
\hline La & Large capacitor & 4 \\
\hline We & Weak drive & 3 \\
\hline Me & Medium capacitor & 2 \\
\hline Sm & Small capacitor & 1 \\
\hline Hi & High impedance & 0 \\
\hline
\end{tabular}

Note that there are four driving strengths and three charge storage strengths. The driving strengths are associated with gate outputs and continuous assignment outputs. The charge storage strengths are associated with the trireg type net. (See Clause 7 for strength modeling.)

For the logic values 0 and 1 , a mnemonic is used when there is no range of strengths in the signal. Otherwise, the logic value is preceded by two decimal digits, which indicate the maximum and minimum strength levels.

For the unknown value, a mnemonic is used when both the 0 and 1 strength components are at the same strength level. Otherwise, the unknown value X is preceded by two decimal digits, which indicate the 0 and 1 strength levels respectively.

The high impedance strength cannot have a known logic value; the only logic value allowed for this level is Z.

For the values L and H , a mnemonic is always used to indicate the strength level.

\section*{Examples:}
```

always
\#15 $display($time,,"group=%b signals=%v %v %v",{s1,s2,s3}, s1,
s2, s3);

```

The example below shows the output that might result from such a call, while Table 71 explains the various strength formats that appear in the output.

0 group=111 signals=St1 Pu1 St1
15 group=011 signals=Pu0 Pu1 St1
30 group=0xz signals=520 PuH Hiz
45 group=0xx signals=Pu0 65X StX
group=000 signals=Me0 St0 St0

Table 71-Explanation of strength formats
\begin{tabular}{|l|l|}
\hline Argument & \multicolumn{1}{c|}{ Description } \\
\hline St 1 & Means a strong driving 1 value \\
\hline Pu 0 & Means a pull driving 0 value \\
\hline HiZ & Means the high-impedance state \\
\hline Me 0 & Means a 0 charge storage of medium capacitor strength \\
\hline StX & Means a strong driving unknown value \\
\hline PuH & Means a pull driving strength of 1 or high-impedance value \\
\hline 65 X & \begin{tabular}{l} 
Means an unknown value with a strong driving 0 component \\
and a pull driving 1 component
\end{tabular} \\
\hline 520 & \begin{tabular}{l} 
Means an 0 value with a range of possible strength from pull \\
driving to medium capacitor
\end{tabular} \\
\hline
\end{tabular}

\subsection*{17.1.1.6 Hierarchical name format}

The \%m format specifier does not accept an argument. Instead, it causes the display task to print the hierarchical name of the module, task, function, or named block that invokes the system task containing the format specifier. This is useful when there are many instances of the module that calls the system task. One obvious application is timing check messages in a flip-flop or latch module; the \(\% \mathrm{~m}\) format specifier shall pinpoint the module instance responsible for generating the timing check message.

\subsection*{17.1.1.7 String format}

The \(\% s\) format specifier is used to print ASCII codes as characters. For each \%s specification that appears in a string, a corresponding parameter shall follow the string in the argument list. The associated argument is interpreted as a sequence of 8 -bit hexadecimal ASCII codes, with each 8 bits representing a single character. If the argument is a variable, its value should be right-justified so that the rightmost bit of the value is the least-significant bit of the last character in the string. No termination character or value is required at the end of a string, and leading zeros are never printed.

\subsection*{17.1.2 Strobed monitoring}
```

strobe_tasks ::=
strobe_task_name ( list_of_arguments );
strobe_task_name ::=
\$strobe | \$strobeb | \$strobeo | \$strobeh

```

\section*{Syntax 17-2 Syntax for \$strobe system tasks}

The system task \$strobe provides the ability to display simulation data at a selected time. That time is the end of the current simulation time, when all the simulation events that have occurred for that simulation time, just before simulation time is advanced. The arguments for this task are specified in exactly the same manner as for the \$display system task including the use of escape sequences for special characters and format specifications (see 17.1.1).

\section*{Example:}
forever @(negedge clock)
\$strobe ("At time \%d, data is \%h",\$time,data);
In this example, \$strobe writes the time and data information to the standard output and the log file at each negative edge of the clock. The action shall occur just before simulation time is advanced and after all other events at that time have occurred, so that the data written is sure to be the correct data for that simulation time.

\subsection*{17.1.3 Continuous monitoring}
```

monitor_tasks ::=
monitor_task_name [( list_of_arguments )] ;
| \$monitoron ;
| \$monitoroff;
monitor_task_name ::=

```
\$monitor | \$monitorb | \$monitoro | \$monitorh

\section*{Syntax 17-3 Syntax for \$monitor system tasks}

The \$monitor task provides the ability to monitor and display the values of any variables or expressions specified as arguments to the task. The arguments for this task are specified in exactly the same manner as for the \$display system task including the use of escape sequences for special characters and format specifications (see 17.1.1).

When a \$monitor task is invoked with one or more arguments, the simulator sets up a mechanism whereby each time a variable or an expression in the argument list changes value with the exception of the \$time, \$stime or \$realtime system functions the entire argument list is displayed at the end of the time step as if reported by the \$display task. If two or more arguments change value at the same time, only one display is produced that shows the new values.

Only one \$monitor display list can be active at any one time; however, a new \$monitor task with a new display list can be issued any number of times during simulation.

The \$monitoron and \$monitoroff tasks control a monitor flag that enables and disables the monitoring. Use \$monitoroff to turn off the flag and disable monitoring. The \$monitoron system task can be used to turn on the flag so that monitoring is enabled and the most recent call to \$monitor can resume its display. A call to \$monitoron shall produce a display immediately after it is invoked, regardless of whether a value change has taken place; this is used to establish the initial values at the beginning of a monitoring session. By default, the monitor flag is turned on at the beginning of simulation.

\subsection*{17.2 File input-output system tasks and functions}

The system tasks and functions for file-based operations are divided into three categories:
Functions and tasks that open and close files
Tasks that output values into files
Tasks that output values into variables
Tasks and functions that read values from files and load into variables or memories

\subsection*{17.2.1 pening and closing files}
```

file_open_function ::=
integer multi_channel_descriptor = \$fopen ( " file_name " );
| integer fd = \$fopen ( " file_name ", type );
file_close_task ::=
\$fclose ( multi_channel_descriptor );
| \$close (fd);

```

Syntax 17-4 Syntax for \$fopen and \$fclose system tasks

The function \$fopen opens the file specified as the filename argument and returns either a 32 bit multi channel descriptor, or a 32 bit file descriptor, determined by the absence or presence of the type argument.
filename is a character string, or a reg containing a character string that names the file to be opened.
type is a character string, or a reg containing a character string of one of the following forms in the table below, which indicates how the file should be opened. If type is omitted, the file is opened for writing, and a multi channel descriptor mod is returned. If type is supplied, the file is opened as specified by the value of type, and a file descriptor fd is returned.

The multi channel descriptor mcd is a 32 bit reg in which a single bit is set indicating which file is opened. The least significant bit (bit 0 ) of a mcd always refers to the standard output. Output is directed to two or more files opened with multi channel descriptors by bitwise oring together their mcds and writing to the resultant value.

The most significant bit (bit 32) of a multi channel descriptor is reserved, and shall always be cleared, limiting an implementation to at most 31 files opened for output via multi channel descriptors.

The file descriptor \(f d\) is a 32 bit value. The most significant bit (bit 32) of a \(f d\) is reserved, and shall always be set; this allows implementations of the file input and output functions to determine how the file was opened. The remaining bits hold a small number indicating what file is opened. Three file descriptors are pre opened; they are STDIN, STDOUT and STDERR, which have the values 32 'h8000_0000, \(32^{\prime} \mathrm{h} 8000 \_0001\) and \(32^{\prime} \mathrm{h} 8000 \_0002\), respectively. STDIN is pre opened for reading, and STDOUT and STDERR \(\overline{\text { are pre opened for append. }}\)

Unlike multi channel descriptors, file descriptors can not be combined via bitwise or in order to direct output to multiple files. Instead, files are opened via file descriptor for input, output, input and output, as well as for append operations, based on the value of type, according to the following table:

Table 72-Types for file descriptors
\begin{tabular}{|l|l|}
\hline Argument & \multicolumn{1}{c|}{ Description } \\
"r" or "rb" & open for reading \\
"w" or "wb" & truncate to zero length or create for writing \\
"a" or "ab" & append; open for writing at end of file, or create for writing \\
"r+", "r+b", or "rb+" & open for update (reading and writing) \\
" \(w+", ~ " w+b "\), or "wb+" & truncate or create for update \\
"a+", "a+b", or "ab+" & append; open or create for update at end-of-file \\
\hline
\end{tabular}

If a file can not be opened (either the file doesn't exist, and the type specified is "r", "rb", "r+", "r+b", or "rb+", or the permissions do not allow the file to be opened at that path, a zero is returned for either the mcd or the fd. Applications can call \$ferror to determine the cause of the most recent error (see 17.2.7).

The " b " in the above types exists to distinguish binary files from text files. Many systems (such as Unix) make no distinction between binary and text files, and on these systems the " b " is ignored. However, some systems (such as machines running NT or Windows) perform data mappings on certain binary values written to and read from files that are opened for text access.

The \(\mathbf{\$ f c l o s e}\) system tasks closes the file specified by fd or closes the file(s) specified by the multi channel descriptor med. No further output to or input from any file descriptor(s) closed by \(\$ \mathbf{f c l o s e}\) is allowed. Active \$fmonitor and/or \$fstrobe operations on a file descriptor or multi channel descriptor are implicitly cancelled by an \$fclose operation. The \$fopen function shall reuse channels that have been closed.

NOTE The number of simultaneous input and output channels that can be open at any one time is dependent on the operating system. Some operating systems do not support opening files for update.

\subsection*{17.2.2 File output system tasks}
```

file_output_tasks ::=
file_output_task_name ( multi_channel_descriptor, list_of_arguments );
| file_output_task_name (fd, list_of_arguments );
file_output_task_name ::=
\$fdisplay | \$fdisplayb | \$fdisplayh | \$fdisplayo
\$fwrite | \$fwriteb | \$fwriteh | \$fwriteo
\$fstrobe |fstrobeb | \$fstrobeh | \$fstrobeo
\$fmonitor | \$fmonitorb | \$fmonitorh | \$fmonitoro

```
    Syntax 17-5 Syntax for file output system tasks

Each of the four formatted display tasks \$display, \$write, \$monitor, and \$strobe has a counterpart that writes to specific files as opposed to the standard output. These counterpart tasks \$fdisplay, \$fwrite, \$fmonitor, and \$fstrobe accept the same type of arguments as the tasks upon which they are based, with one exception: The first parameter shall be either a multi channel descriptor or a file descriptor, which indicates where to direct the file output. Multi channel descriptors are described in detail in 17.2.1. A multichannel descriptor is either a variable or the result of an expression that takes the form of a 32-bit unsigned integer value.

The \$fstrobe and \$fmonitor system tasks work just like their counterparts, \$strobe and \$monitor, except that they write to files using the multi channel descriptor for control. Unlike \$monitor, any number of \$fmonitor tasks can be set up to be simultaneously active. However, there is no counterpart to \$monitoron and \$monitoroff tasks. The task \$fclose is used to cancel an active \$fstrobe or \$fmonitor task.

\section*{Example:}

This example shows how to set up multi channel descriptors. In this example, three different channels are opened using the \$fopen function. The three multi channel descriptors that are returned by the function are then combined in a bit-wise or operation and assigned to the integer variable messages. The messages variable can then be used as the first parameter in a file output task to direct output to all three channels at once. To create a descriptor that directs output to the standard output as well, the messages variable is a bit-wise logical or with the constant 1 , which effectively enables channel 0 .
```

integer
messages, broadcast,
cpu_chann, alu_chann, mem_chann;
initial begin
cpu_chann = \$fopen("cpu.dat");
if (cpu_chann == 0) \$finish;
alu_chann = \$fopen("alu.dat");
if (alu_chann == 0) \$finish;
mem_chānn = \$fopen("mem.dat");
if (mem_chann == 0) \$finish;
messages = cpu_chann | alu_chann | mem_chann;
// broadcast includes standard output
broadcast = 1 | messages;
end
endmodule

```

The following file output tasks show how the channels opened in the preceding example might be used:
```

$fdisplay( broadcast, "system reset at time %d",$time );
\$fdisplay( messages, "Error occurred on address bus",
" at time %d, address = %h", \$time, address );
forever @(posedge clock)
\$fdisplay( alu_chann, "acc= %h f=%h a=%h b=%h", acc, f, a, b );

```

\subsection*{17.2.3 Formatting data to a string}
```

string_output_tasks ::=
string_output_tasks_name ( output_reg, list_of_arguments );
string_output_task_name ::=
\$swrite | \$swriteb | \$swriteh | \$swriteo
variable_format_string_output_task ::=
\$sformat ( output_reg, format, list_of_arguments );

```
                    Syntax 17-6 Syntax for formatting data tasks

The syntax for the string output system tasks is
```

\$swrite(output_reg, list_of_arguments);
\$sformat(output_reg, format_string, list_of_arguments);
length = \$sformat(output_reg, format_string, list_of_arguments);

```

The \$swrite family of tasks are based on the \$fwrite family of tasks, and accept the same type of arguments as the tasks upon which they are based, with one exception: The first parameter to \$swrite shall be a reg variable to which the resulting string shall be written, instead of a variable specifying the file to which to write the resulting string.

The variable output_reg is assigned using the Verilog s string assignment to variable rules, as specified in 4.2.3.

The system task \$sformat is similar to the system task \$swrite, with a one major difference.
Unlike the display and write family of output system tasks, \$sformat always interprets its second argument, and only its second argument as a format string. This format argument can be a static string, such as "data is \(\% \mathrm{~d}\) ", or can be a reg variable whose content is interpreted as the format string. No other arguments are interpreted as format strings. \$sformat supports all the format specifiers supported by \$display, as documented in 17.1.1.2.

The remaining arguments to \(\$\) sformat are processed using any format specifiers in the format_string, until all such format specifiers are used up. If not enough arguments are supplied for the format specifiers, or too many are supplied, then the application shall issue a warning, and continue execution. The application, if possible, can statically determine a mismatch in format specifiers and number of arguments, and issue a compile time error message.

NOTE If the format_string is a reg, it might not be possible to determine its value at compile time.
The variable output_reg is assigned using the Verilog s string assignment to variable rules, as specified in 4.2.3.

\subsection*{17.2.4 eading data from a file}

Files opened using file descriptors can be read from only if they were opened with either the \(r\) or \(r+\) type values. See 17.2.1 for more information about opening files.

\subsection*{17.2.4.1 eading a character at a time}
\[
c=\$ \text { fgetc }(f d) ;
\]

Read a byte from the file specified by fd. If an error occurs reading from the file, then \(c\) is set to EOF ( -1 ). Define the width of the reg to be wider than 8 bits so that a return value from \$fgetc of EOF ( -1 ) can be differentiated from the character code \(0 \times\) xFF. Applications can call \(\$\) ferror to determine the cause of the most recent error (see 17.2.7).
```

code = \$ungetc ( c, fd );

```

Insert the character specified by c into the buffer specified by file descriptor \(f d\). The character \(c\) shall be returned by the next \$fgetc call on that file descriptor. The file itself is unchanged. Note that the features of the underlying implementation of fileio on the host system limits the number of characters that can be pushed back onto a stream. Note also that operations like \$fseek might erase any pushed back characters. If an error occurs pushing a character onto a file descriptor, then code is set to EOF. Otherwise code is set to zero. Applications can call \$ferror to determine the cause of the most recent error (see 17.2.7).

\subsection*{17.2.4.2 eading a line at a time}
```

integer code = \$fgets ( str, fd );

```

Read characters from the file specified by fd into the reg str until either str is filled, or a newline character is read and transferred to str, or an end-of-file condition is encountered. If str is not an integral number of bytes in length, the most significant partial byte is not used in order to determine the size.

If an error occurs reading from the file, then code is set to zero. Otherwise the number of characters read is returned in code. Applications can call \$ferror to determine the cause of the most recent error (see below).

\subsection*{17.2.4.3 eading formatted data}
```

integer code = \$fscanf ( fd, format, args );
integer code = \$sscanf ( str, format, args );

```
\$fscanf reads from the files specified by the file descriptor \(f d\).
\$sscanf reads from the reg str.
Both functions read characters, interpret them according to a format, and store the results. Both expect as arguments a control string, format, and a set of arguments specifying where to place the results. If there are insufficient arguments for the format, the behavior is undefined. If the format is exhausted while arguments remain, the excess arguments are ignored.

If an argument is too small to hold the converted input, then in general, the least significant bits are transferred. Arguments of any length that is supported by Verilog can be used. However if the destination is a real or realtime then the value \(+\operatorname{Inf}\) ( or -Inf) is transferred. The format can be a string constant or a reg containing a string constant. The string contains conversion specifications, which direct the conversion of input into the arguments. The control string can contain
a) White-space characters (blanks, tabs, new-lines, or form-feeds) that, except in one case described below, cause input to be read up to the next non-white-space character.
b) An ordinary character (not \%) that must match the next character of the input stream.
c) Conversion specifications consisting of the character \% an optional assignment suppression character *, a decimal digit string that specifies an optional numerical maximum field width, and a conversion code.

A conversion specification directs the conversion of the next input field; the result is placed in the variable specified in the corresponding argument unless assignment suppression was indicated by the character *; in this case no argument shall be supplied.

The suppression of assignment provides a way of describing an input field that is to be skipped. An input field is defined as a string of non-space characters; it extends to the next inappropriate character or until the maximum field width, if one is specified, is exhausted. For all descriptors except the character c, white space leading an input field is ignored.
\(\% \quad\) A single \(\%\) is expected in the input at this point; no assignment is done.
b Matches a binary number, consisting of a sequence from the set \(0,1, \mathrm{X}, \mathrm{x}, \mathrm{Z}, \mathrm{z}\), ? and _.
o Matches a octal number, consisting of a sequence of characters from the set 0,1,2,3,4,5,6,7,X,x,Z,z,? and _.
d Matches an optionally signed decimal number, consisting of the optional sign from the set + or - , followed by a sequence of characters from the set \(0,1,2,3,4,5,6,7,8,9\) and, , or a single value from the set \(\mathrm{x}, \mathrm{X}, \mathrm{z}, \mathrm{Z}\), ? .
\(h\) or \(x\) Matches a hexadecimal number, consisting of a sequence of characters from the set \(0,1,2,3,4,5,6,7,8,9, a, A, b, B, c, C, d, D, e, E, f, F, x, X, z, Z, ?\) and .
fe or g Matches a floating point number. The format of a floating point number is an optional sign (either + or -), followed by a string of digits from the set \(0,1,2,3,4,5,6,7,8,9\) optionally containing a decimal point character (.), then an optional exponent part including e or E followed by an optional sign, followed by a string of digits from the set \(0,1,2,3,4,5,6,7,8,9\).
\(v \quad\) Matches a net signal strength, consisting of three character sequence as specified in 17.1.1.5. This conversion is not extremely useful, as strength values are really only usefully assigned to nets and \(\$\) fscanf can only assign values to regs (if assigned to regs, the values are converted to the 4 value equivalent).
t Matches a floating point number. The format of a floating point number is an optional sign (either + or - ), followed by a string of digits from the set \(0,1,2,3,4,5,6,7,8,9\) optionally containing a decimal point character (.), then an optional exponent part including e or E followed by an optional sign, followed by a string of digits from the set \(0,1,2,3,4,5,6,7,8,9\). The value matched is then scaled and rounded according to the current time scale as set by \$timeformat. For example, if the timescale is 'timescale \(1 \mathrm{~ns} / 100 \mathrm{ps}\) and the time format is \$timeformat \((-3,2, " \mathrm{~ms}\) ", 10 );, then a value read with \$sscanf("10.345", "\%t", t) would return 10350000.0.
c Matches a single character, whose 8 bit ASCII value is returned.
s Matches a string, which is a sequence of non white space characters.
u Matches unformatted (binary) data. The application shall transfer sufficient data from the input to fill the target reg. Typically the data is obtained from a matching \$fwrite ("\%u",data), or from an external application written in another programming language such as C , Perl or FORTRAN.

The application shall transfer the 2 value binary data from the input stream to the destination reg, expanding the data to the four value format. This escape sequence can be used with any of the existing input system tasks, although \(\$ \mathrm{fscanf}\) should be the preferred one to use. As the input data can not represent x or z , it is not possible to obtain an x or z in the result reg. This formatting specifier is intended to be used to support transferring data to and from external programs that have no concept of \(x\) and \(z\).

Applications that require preservation of \(x\) and \(z\) are encouraged to use the \(\% z I / O\) format specification.

The data shall be read from the file in the native endian format of the underlying system (i.e., in the same endian order as if the PLI was used, and the C language read ( 2 ) system call was used).

For POSIX applications: It might be necessary to open files for unformatted I/O with the "rb", "rb+" or " \(\mathrm{r}+\mathrm{b}\) " specifiers, to avoid the systems implementation of I/O altering patterns in the unformatted stream that match special characters.
\(\mathrm{z} \quad\) The formatting specification \(\% \mathrm{z}(\) or \(\% \mathrm{Z})\) is defined for reading data without formatting (binary values). The application shall transfer the 4 value binary representation of the specified data from the input stream to the destination reg. This escape sequence can be used with any of the existing input system tasks, although \(\$ \mathrm{fscanf}\) should be the preferred one to use.

This formatting specifier is intended to be used to support transferring data to and from external programs that recognize and support the concept of \(x\) and \(z\). Applications that do not require the preservation of \(x\) and \(z\) are encouraged to use the \(\% u I / O\) format specification.

The data shall be read from the file in the native endian format of the underlying system (i.e., in the same endian order as if the PLI was used, and the data were in a s_vpi_vecval structure (See 27.14, Figure 27-8), and the C language read(2) system call was used to read the data from disk).

For POSIX applications: It might be necessary to open files for unformatted I/O with the "rb", "rb+" or " \(\mathrm{r}+\mathrm{b}\) " specifiers, to avoid the systems implementation of I/O altering patterns in the unformatted stream that match special characters.
\(\mathrm{m} \quad\) Returns the current hierarchical path as a string. Does not read data from the input file or str argument. If an invalid conversion character follows the \(\%\), the results of the operation are implementation dependent.

If the format string, or the str argument to \$sscanf contains unknown bits ( x or z ) then the system task shall return EOF.

If end-of-file is encountered during input, conversion is terminated. If end-of-file occurs before any characters matching the current directive have been read (other than leading white space, where permitted), execution of the current directive terminates with an input failure; otherwise, unless execution of the current directive is terminated with a matching failure, execution of the following directive (if any) is terminated with an input failure.

If conversion terminates on a conflicting input character, the offending input character is left unread in the input stream. Trailing white space (including new-line characters) is left unread unless matched by a directive. The success of literal matches and suppressed assignments is not directly determinable.

The number of successfully matched and assigned input items is returned in code; this number can be 0 in the event of an early matching failure between an input character and the control string. If the input ends before the first matching failure or conversion, EOF is returned. Applications can call \$ferror to determine the cause of the most recent error (see below).

\subsection*{17.2.4.4 eading binary data}
```

integer code = \$fread( myreg, fd);
integer code = \$fread( mem, fd);
integer code = \$fread( mem, fd, start);
integer code = \$fread( mem, fd, start, count);
integer code = \$fread( mem, fd, , count);

```

Read a binary data from the file specified by fd into the reg myreg or the memory mem.
start is an optional argument. If present, start shall be used as the address of the first element in the memory to be loaded. If not present the lowest numbered location in the memory shall be used.
count is an optional argument. If present, count shall be the maximum number of locations in mem that shall be loaded. If not supplied the memory shall be filled with what data is available.
start and count are ignored if \$fread is loading a reg.
If no addressing information is specified within the system task, and no address specifications appear within the data file, then the default start address is the lowest address given in the declaration of the memory. Consecutive words are loaded towards the highest address until either the memory is full or the data file is completely read. If the start address is specified in the task without the finish address, then loading starts at the specified start address and continues towards the highest address given in the declaration of the memory.
start is the address in the memory. For start \(=12\) and the memory up[10:20], the first data would be loaded at up[12]. For the memory down[20:10], the first location loaded would be down[12], then down[13].

The data in the file shall be read byte by byte to fulfill the request. An 8 -bit wide memory is loaded using one byte per memory word, while a 9-bit wide memory is loaded using 2 bytes per memory word. The data is read from the file in a big endian manner; the first byte read is used to fill the most significant location in the memory element. If the memory width is not evenly divisible by \(8(8,16,24,32)\), not all data in the file is loaded into memory because of truncation.

The data loaded from the file is taken as "two value" data. A bit set in the data is interpreted as a 1 , and bit not set is interpreted as a 0 . It is not possible to read a value of \(x\) or \(z\) using \$fread.

If an error occurs reading from the file, then code is set to zero. Otherwise the number of characters read is returned in code. Applications can call \$ferror to determine the cause of the most recent error (see 17.2.7).

Note that there is not a "binary" mode and a "ASCII" mode; one can freely intermingle binary and formatted read commands from the same file.

\subsection*{17.2.5 File positioning}
```

integer pos = \$ftell ( fd );

```

Returns in pos the offset from the beginning of the file of the current byte of the file \(f d\) which shall be read or written by a subsequent operation on that file descriptor.

This value can be used by subsequent \(\$\) fseek calls to reposition the file to this point. Note that any repositioning shall cancel any \$ungetc operations. If an error occurs EOF is returned. Applications can call \$ferror to determine the cause of the most recent error (see 17.2.7).
```

code = \$fseek ( fd, offset, operation );
code = \$rewind ( fd );

```

Sets the position of the next input or output operation on the file specified by fd. The new position is at the signed distance offset bytes from the beginning, from the current position, or from the end of the file, according to an operation value of 0,1 and 2 as follows:

0 set position equal to offset bytes
1 set position to current location plus offset
2 set position to EOF plus offset
\$rewind is equivalent to \(\$\) fseek ( \(\mathrm{fd}, 0,0\) );
Repositioning the current file position with \$fseek or \$rewind shall cancel any \$ungetc operations.
\(\mathbf{\$ f s e e k}()\) allows the file position indicator to be set beyond the end of the existing data in the file. If data is later written at this point, subsequent reads of data in the gap shall return zero until data is actually written into the gap. \$fseek, by itself, does not extend the size of the file.

When a file is opened for append (that is, when type is "a", or " \(\mathrm{a}+\) "), it is impossible to overwrite information already in the file. \$fseek can be used to reposition the file pointer to any position in the file, but when output is written to the file, the current file pointer is disregarded. All output is written at the end of the file and causes the file pointer to be repositioned at the end of the output.

If an error occurs repositioning the file, then code is set to -1 . Otherwise code is set to 0 . Applications can call \$ferror to determine the cause of the most recent error (see 17.2.7).

\subsection*{17.2.6 Flushing output}
```

\$fflush ( mcd );
\$fflush ( fd );
\$fflush ( );

```

Writes any buffered output to the file(s) specified by mcd, the file specified by fd or if \$fflush is invoked with no arguments, writes any buffered output to all open files.

\subsection*{17.2.7 I/ error status}

Should any error be detected by one of the fileio routines, an error code is returned. Often this is sufficient for normal operation; (i.e., if the opening of a optional configuration file fails, the application typically would simply continue using default values.) However sometimes it is useful to obtain more information about the error for correct application operation. In this case the \(\$\) ferror function can be used:
```

integer errno = \$ferror ( fd, str );

```

A string description of type of error encountered by the most recent file I/O operation is written into str which should be at least 640 bits wide. The integral value of the error code is returned in errno. If the most recent operation did not result in an error, then the value returned shall be zero, and the reg str shall be cleared.

\subsection*{17.2.8 Loading memory data from a file}


Two system tasks \$readmemb and \$readmemh read and load data from a specified text file into a specified memory. Either task can be executed at any time during simulation. The text file to be read shall contain only the following:

White space (spaces, new lines, tabs, and form-feeds)
Comments (both types of comment are allowed)
Binary or hexadecimal numbers
The numbers shall have neither the length nor the base format specified. For \$readmemb, each number shall be binary. For \$readmemh, the numbers shall be hexadecimal. The unknown value ( x or X ), the high impedance value ( \(\mathbf{z}\) or \(Z\) ), and the underscore (_) can be used in specifying a number as in a Verilog HDL source description. White space and/or comments shall be used to separate the numbers.

In the following discussion, the term address refers to an index into the array that models the memory.
As the file is read, each number encountered is assigned to a successive word element of the memory. Addressing is controlled both by specifying start and/or finish addresses in the system task invocation and by specifying addresses in the data file.

When addresses appear in the data file, the format is an at character ( @) followed by a hexadecimal number as follows:
@hh... h

Both uppercase and lowercase digits are allowed in the number. No white space is allowed between the @ and the number. As many address specifications as needed within the data file can be used. When the system task encounters an address specification, it loads subsequent data starting at that memory address.

If no addressing information is specified within the system task, and no address specifications appear within the data file, then the default start address shall be the left-hand address given in the declaration of the memory. Consecutive words shall be loaded until either the memory is full or the data file is completely read. If the start address is specified in the task without the finish address, then loading shall start at the specified start address and shall continue towards the right-hand address given in the declaration of the memory.

If both start and finish addresses are specified as parameters to the task, then loading shall begin at the start address and shall continue toward the finish address, regardless of how the addresses are specified in the memory declaration.

When addressing information is specified both in the system task and in the data file, the addresses in the data file shall be within the address range specified by the system task parameters; otherwise, an error message is issued and the load operation is terminated.

A warning message shall be issued if the number of data words in the file differs from the number of words in the range implied by the start through finish addresses.

\section*{Example:}
```

reg [7:0] mem[1:256];

```

Given this declaration, each of the following statements load data into mem in a different manner:
```

initial \$readmemh("mem.data", mem);
initial \$readmemh("mem.data", mem, 16);
initial \$readmemh("mem.data", mem, 128, 1);

```

The first statement loads up the memory at simulation time 0 starting at the memory address 1 . The second statement begins loading at address 16 and continue on towards address 256. For the third and final statement, loading begins at address 128 and continue down towards address 1 .

In the third case, when loading is complete, a final check is performed to ensure that exactly 128 numbers are contained in the file. If the check fails, a warning message is issued.

\subsection*{17.2.9 Loading timing data from an SDF file}

The syntax for the \$sdf_annotate system task is shown in Syntax 17-8.
```

sdf_annotate_task ::=
\$sdf_annotate ("sdf_file" [, [ module_instance ] [, [ "config_file" ]
[,[ "log_file"][,["mtm_spec" ]
[,["scale_factors" ][,[ "scale_type" ]]]]]]]);

```

\section*{Syntax 17-8 Syntax for \$sdf_annotate system task}

The \$sdf_annotate system task reads timing data from an SDF file into a specified region of the design.
\(s d f\) file is a character string, or a reg containing a character string naming the file to be opened.
module_instance is an optional argument specifying the scope to which to annotate the information in the SDF file. The SDF annotator uses the hierarchy level of the specified instance for running the annotation. Array indices are permitted. If the module_instance not specified, the SDF Annotator uses the module containing the call to the \$sdf_annotate system task as the module_instance for annotation.
config_file is an optional character string argument providing the name of a configuration file. Information in this file can be used to provide detailed control over many aspects of annotation.
log_file is an optional character string argument providing the name of the log file used during SDF annotation. Each individual annotation of timing data from the SDF file results in an entry in the \(\log\) file.
mtm_spec is an optional character string argument specifying which member of the min/typ/max triples shall be annotated. The legal values for this string are described in Table 73. This overrides any MTM_SPEC keywords in the configuration file.

Table 73-mtm spec argument
\begin{tabular}{|l|l|}
\hline Keyword & \multicolumn{1}{c|}{ Description } \\
\hline \hline MAXIMUM & Annotate the maximum value \\
\hline MINIMUM & Annotate the minimum value \\
\hline TOOL_CONTROL (default) & \begin{tabular}{l} 
Annotate the value as selected by \\
the simulator
\end{tabular} \\
\hline TYPICAL & Annotate the typical value \\
\hline
\end{tabular}
scale_factors
scale_type
is an optional character string argument specifying the scale factors to be used while annotating timing values. For example, "1.6:1.4:1.2" causes minimum values to be multiplied by 1.6 , typical values by 1.4 , and maximum values by 1.2 . The default values are 1.0:1.0:1.0. The scale_factors argument overrides any SCALE_FACTORS keywords in the configuration file.
is an optional character string argument specifying how the scale factors should be applied to the min/typ/max triples. The legal values for this string are shown in Table 74. This overrides any SCALE_TYPE keywords in the configuration file.

Table 74-scale type argument
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ Keyword } & \multicolumn{1}{c|}{ Description } \\
\hline FROM_MAXIMUM & Apply scale factors to maximum value \\
\hline FROM_MINIMUM & Apply scale factors to minimum value \\
\hline FROM_MTM (default) & \begin{tabular}{l} 
Apply scale factors to min/typ/max \\
values
\end{tabular} \\
\hline FROM_TYPICAL & Apply scale factors to typical value \\
\hline
\end{tabular}

\subsection*{17.3 Timescale system tasks}

The following system tasks display and set timescale information:
a) Sprinttimescale
b) \$timeformat

\subsection*{17.3.1 \$printtimescale}

The \$printtimescale system task displays the time unit and precision for a particular module. The syntax for the system task is shown in Syntax 17-9.
```

printtimescale_task ::=
\$printtimescale [ ( hierarchical_identifier ) ] ;

```

\section*{Syntax 17-9 Syntax for \$printtimescale}

This system task can be specified with or without an argument.
When no argument is specified, \$printtimescale displays the time unit and precision of the module that is the current scope.
When an argument is specified, \$printtimescale displays the time unit and precision of the module passed to it.

The timescale information shall appear in the following format:
Time scale of (module_name) is unit / precision
Example:
```

'timescale 1 ms / 1 us
module a_dat;
initial
\$printtimescale(b_dat.c1);
endmodule
'timescale 10 fs / 1 fs
module b_dat;
c_dat c1 ();
endmodule
'timescale 1 ns / 1 ns
module c_dat;
•
•
endmodule

```

In this example, module a dat invokes the \$printtimescale system task to display timescale information about another module c_dat, which is instantiated in module b_dat.

The information about c_dat shall be displayed in the following format:
```

Time scale of (b_dat.cl) is 1ns / 1ns

```

\subsection*{17.3.2 \$timeformat}

The syntax for \$timeformat system task is shown in Syntax 17-10.
        \$timeformat[( units_number, precision_number, suffix_string, minimum_field_width )];

\section*{Syntax 17-10 Syntax for \$timeformat}

The \$timeformat system task performs the following two functions:
It specifies how the \(\% t\) format specification reports time information for the \(\$\) write, \$display, \$strobe, \$monitor, \$fwrite, \$fdisplay, \$fstrobe, and \$fmonitor group of system tasks.
It specifies the time unit for delays entered interactively.
The units number argument shall be an integer in the range from 0 to -15 . This argument represents the time unit as shown in Table 75.

Table 75-\$timeformat units_number arguments
\begin{tabular}{|l|l|l|l|}
\hline Unit number & Time unit & Unit number & \multicolumn{1}{c|}{ Time unit } \\
\hline 0 & 1 s & -8 & 10 ns \\
\hline-1 & 100 ms & -9 & 1 ns \\
\hline-2 & 10 ms & -10 & 100 ps \\
\hline-3 & 1 ms & -11 & 10 ps \\
\hline-4 & 100 us & -12 & 1 ps \\
\hline-5 & 10 us & -13 & 100 fs \\
\hline-6 & 1 us & -14 & 10 fs \\
\hline-7 & 100 ns & -15 & 1 fs \\
\hline
\end{tabular}

The \$timeformat system task performs the following two operations:
It sets the time unit for all later-entered delays entered interactively.
It sets the time unit, precision number, suffix string, and minimum field width for all \(\% t\) formats specified in all modules that follow in the source description until another \$timeformat system task is invoked.

The default \$timeformat system task arguments are given in Table 76.
Table 76-\$timeformat default value for arguments
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ Argument } & \multicolumn{1}{c|}{ Default } \\
\hline units_number & \begin{tabular}{l} 
The smallest time precision argument of all the 'timescale com- \\
piler directives in the source description
\end{tabular} \\
\hline precision_number & 0 \\
\hline suffix_string & A null character string \\
\hline minimum_field_width & 20 \\
\hline
\end{tabular}

\section*{Example:}

The following example shows the use of \%t with the \(\$ \mathbf{t i m e f o r m a t ~ s y s t e m ~ t a s k ~ t o ~ s p e c i f y ~ a ~ u n i f o r m ~ t i m e ~}\) unit, time precision, and format for timing information.
```

'timescale 1 ms / 1 ns
module cntrl;
initial
\$timeformat(-9, 5, " ns", 10);
endmodule
'timescale 1 fs / 1 fs
module a1_dat;
reg in1;
integer file;
buf \#10000000 (o1,in1);
initial begin
file = \$fopen("a1.dat");
\#00000000 \$fmonitor(file,"%m: %t in1=%d o1=%h", $realtime,in1,o1);
    #10000000 in1 = 0;
    #10000000 in1 = 1;
end
endmodule
'timescale 1 ps / 1 ps
module a2_dat;
reg in2;
integer file2;
buf #10000 (o2,in2);
initial begin
    file2=$fopen("a2.dat");
\#00000 $fmonitor(file2,"%m: %t in2=%d o2=%h",$realtime,in2,o2);
\#10000 in2 = 0;
\#10000 in2 = 1;
end
endmodule

```

The contents of file a1. dat are as follows:
```

a1_dat: 0.00000 ns in1= x o1=x
a1_dat: 10.00000 ns in1= 0 ol=x
a1_dat: 20.00000 ns in1= 1 ol=0
a1_dat: 30.00000 ns in1= 1 o1=1

```

The contents of file a2.dat are as follows:
```

a2_dat: 0.00000 ns in2=x o2=x
a2 dat: 10.00000 ns in2=0 o2=x
a2_dat: 20.00000 ns in2=1 o2=0
a2_dat: 30.00000 ns in2=1 o2=1

```

In this example, the times of events written to the files by the \$fmonitor system task in modules a1_dat and a2_dat are reported as multiples of 1 ns even though the time units for these modules are \(\overline{1}\) fs and 1 ps respectively because the first argument of the \$timeformat system task is -9 and the \(\% \mathrm{t}\) format specification is included in the arguments to \$fmonitor. This time information is reported after the module names with five fractional digits, followed by an ns character string in a space wide enough for 10 ASCII characters.

\subsection*{17.4 Simulation control system tasks}

There are two simulation control system tasks:
a) \(\$\) finish
b) \(\$\) stop

\subsection*{17.4.1 \$finish}

Syntax 17-11 shows the syntax for \(\mathbf{\$ f i n i s h}\) system task.
```

finish_task ::=
\$finish[(n)];

```

\section*{Syntax 17-11 Syntax for \$finish}

The \$finish system task simply makes the simulator exit and pass control back to the host operating system. If an expression is supplied to this task, then its value ( 0,1, or 2 ) determines the diagnostic messages that are printed before the prompt is issued. If no argument is supplied, then a value of 1 is taken as the default.

Table 77—Diagnostics for \$finish
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ Parameter value } & \multicolumn{1}{c|}{ Diagnostic message } \\
\hline 0 & Prints nothing \\
\hline 1 & Prints simulation time and location \\
\hline 2 & \begin{tabular}{l} 
Prints simulation time, location, and statistics about the memory \\
and CPU time used in simulation
\end{tabular} \\
\hline
\end{tabular}

\subsection*{17.4.2 \$stop}

The syntax for \$stop system task is shown in Syntax 17-12.
```

stop_task ::=
\$stop [ ( n )] ;

```

Syntax 17-12 Syntax for \$stop

The \$stop system task causes simulation to be suspended. This task takes an optional expression argument ( 0,1 , or 2 ) that determines what type of diagnostic message is printed. The amount of diagnostic messages output increases with the value of the optional argument passed to \$stop.

\subsection*{17.5 LA modeling system tasks}

The modeling of PLA devices is provided in the Verilog HDL by a group of system tasks. This clause describes the syntax and use of these system tasks and the formats of the logic array personality file. The syntax for PLA modeling system task is shown in Syntax 17-13.
```

pla_system_task ::=
$array_type$logic\$format ( memory_type, input_terms , output_terms );
array_type ::=
sync async
logic ::=
and or| nand nor
format ::=
array |plane
input_terms ::=
expression
output_terms ::=
variable_lvalue

```

\section*{Syntax 17-13 Syntax for PLA modeling system task}

NOTE The input terms can be nets or variables whereas the output terms shall only be variables.
The PLA syntax allows for the system tasks as shown in Table 78.
Table 78-PLA modeling system tasks
\begin{tabular}{|l|l|l|l|}
\hline \$async\$and\$array & \$sync\$and\$array & \$async\$and\$plane & \$sync\$and\$plane \\
\hline \begin{tabular}{l} 
\$async\$nand\$arra \\
y
\end{tabular} & \$sync\$nand\$array & \begin{tabular}{l} 
\$async\$nand\$plan \\
e
\end{tabular} & \begin{tabular}{l} 
\$sync\$nand\$plan \\
e
\end{tabular} \\
\hline \$async\$or\$array & \$sync\$or\$array & \$async\$or\$plane & \$sync\$or\$plane \\
\hline \$async\$nor\$array & \$sync\$nor\$array & \$async\$nor\$plane & \$sync\$nor\$plane \\
\hline
\end{tabular}

\subsection*{17.5.1 Array types}

The modeling of both synchronous and asynchronous arrays is provided by the PLA system tasks. The synchronous forms control the time at which the logic array shall be evaluated and the outputs shall be updated. For the asynchronous forms, the evaluations are automatically performed whenever an input term changes value or any word in the personality memory is changed.

For both the synchronous and asynchronous forms, the output terms are updated without any delay.

\section*{Examples:}

An example of an asynchronous system call is as follows:
```

wire a1, a2, a3, a4, a5, a6, a7;
reg b1, b2, b3;
wire [1:7] awire;
reg [1:3] breg;

```
\$async\$and\$array(mem, \(\{a 1, a 2, a 3, a 4, a 5, a 6, a 7\},\{b 1, b 2, b 3\})\);
or
\$async\$and\$array(mem, awire, breg);

An example of a synchronous system call is as follows:
```

$sync$or\$plane(mem,{a1,a2,a3,a4,a5,a6,a7}, {b1,b2,b3});

```

\subsection*{17.5.2 Array logic types}

The logic arrays are modeled with and, or, nand, and nor logic planes. This applies to all array types and formats.

\section*{Examples:}

An example of a nor plane system call is as follows:
\[
\text { \$async\$nor\$plane (mem, }\{\mathrm{a} 1, \mathrm{a} 2, \mathrm{a} 3, \mathrm{a} 4, \mathrm{a} 5, \mathrm{a} 6, \mathrm{a} 7\},\{\mathrm{b} 1, \mathrm{~b} 2, \mathrm{~b} 3\}) \text {; }
\]

An example of a nand plane system call is as follows:
\$sync\$nand\$plane(mem, \(\{a 1, a 2, a 3, a 4, a 5, a 6, a 7\},\{b 1, b 2, b 3\})\);

\subsection*{17.5.3 Logic array personality declaration and loading}

The logic array personality is declared as an array of regs that is as wide as the number of input terms and as deep as the number of output terms.

The personality of the logic array is normally loaded into the memory from a text data file using the system tasks \$readmemb or \$readmemh. Alternatively, the personality data can be written directly into the memory using the procedural assignment statements. PLA personalities can be changed dynamically at any time during simulation simply by changing the contents of the memory. The new personality shall be reflected on the outputs of the logic array at the next evaluation.

\section*{Example:}

The following example shows a logic array with \(n\) input terms and \(m\) output terms.
```

reg [1:n] mem[1:m];

```

NOTE Put PLA input terms, output terms, and memory in ascending order, as shown in examples in this clause.

\subsection*{17.5.4 Logic array personality formats}

Two separate personality formats are supported by the Verilog HDL and are differentiated by using either an array system call or a plane system call. The array system call allows for a 1 or 0 in the memory that has been declared. A 1 means take the input value and a 0 means do not take the input value.

The plane system call complies with the University of California at Berkeley format for Espresso. Each bit of the data stored in the array has the following meaning:

0 Take the complemented input value
1 Take the true input value
\(x \quad\) Take the worst case of the input value
\(z \quad\) Don \(t\)-care; the input value is of no significance
? \(\quad\) Same as \(z\)

Examples:
Example 1 The following example illustrates an array with logic equations:
```

b1 = a1 \& a2
b2 = a3 \& a4 \& a5
b3 = a5 \& a6 \& a7

```

The PLA personality is as follows:
```

1100000 in mem[1]
0011100 in mem[2]
0000111 in mem[3]

```

The module for the PLA is as follows:
```

module async_array(a1,a2,a3,a4,a5,a6,a7,b1,b2,b3);
input a1, a2, a3, a4, a5, a6, a7 ;
output b1, b2, b3;
reg [1:7] mem[1:3]; // memory declaration for array personality
reg b1, b2, b3;
initial begin
// setup the personality from the file array.dat
\$readmemb("array.dat", mem);
// setup an asynchronous logic array with the input
// and output terms expressed as concatenations
$async$and\$array(mem,{a1,a2,a3,a4,a5,a6,a7},{b1,b2,b3});
end
endmodule

```

Where the file array. dat contains the binary data for the PLA personality:
1100000
0011100
0000111

A synchronous version of this example has the following description:
```

module sync_array(a1,a2,a3,a4,a5,a6,a7,b1,b2,b3,clk);
input a1, a2, a3, a4, a5, a6, a7, clk;
output b1, b2, b3;
reg [1:7] mem[1:3]; // memory declaration
reg b1, b2, b3;
initial begin
// setup the personality
\$readmemb("array.dat", mem);
// setup a synchronous logic array to be evaluated
// when a positive edge on the clock occurs
forever@(posedge clk)
\$async\$and\$array(mem, $\{\mathrm{a} 1, \mathrm{a} 2, \mathrm{a} 3, \mathrm{a} 4, \mathrm{a} 5, \mathrm{a} 6, \mathrm{a} 7\},\{\mathrm{b} 1, \mathrm{~b} 2, \mathrm{~b} 3\}$ );
end
endmodule

```

Example 2 An example of the usage of the plane format tasks follows. The logical function of this PLA is shown first, followed by the PLA personality in the new format, the Verilog HDL description using the \$async\$and\$plane system task, and finally the result of running the simulation.

The logical function of the PLA is as follows:
```

b[1] = a[1] \& ~a[2];
b[2] = a[3];
b[3] = ~a[1] \& ~a[3];
b[4] = 1;

```

The PLA personality is as follows:
```

3'b10?
3'b??1
3'b0?0
3'b???

```
```

module pla;
'define rows 4
'define cols 3
reg [1:`cols] a, mem[1:`rows];
reg [1:`rows] b;
initial begin
// PLA system call
$async$and\$plane(mem,a [1:3],b[1:4]);
mem[1] = 3'b10?;
mem[2] = 3'b??1;
mem[3] = 3'b0?0;
mem[4] = 3'b???;
// stimulus and display
\#10 a = 3'b111;
\#10 \$displayb(a, " -> ", b);
\#10 a = 3'b000;
\#10 \$displayb(a, " -> ", b);
\#10 a = 3'bxxx;
\#10 \$displayb(a, " -> " , b);
\#10 a = 3'b101;
\#10 \$displayb(a, " -> " , b);
end
endmodule

```

The output is as follows:
```

111 -> 0101
000 -> 0011
xxx -> xxx1
101 -> 1101

```

\subsection*{17.6 Stochastic analysis tasks}

This clause describes a set of system tasks and functions that manage queues and generate random numbers with specific distributions. These tasks facilitate implementation of stochastic queueing models.

The set of tasks and functions that create and manage queues follow:
\$q_initialize (q_id, q_type, max_length, status) ;
\$q_add (q_id, job_id, inform_id, status) ;
\$q_remove (q_id, job_id, inform_id, status) ;
\$q_full (q_id, status) ;
\$q_exam (q_id, q_stat_code, q_stat_value, status) ;

\subsection*{17.6.1 \$q_initialize}

The \(\$ q\) initialize system task creates new queues. The q_id parameter is an integer input that shall uniquely identify the new queue. The q_type parameter is an integer input. The value of the q_type parameter specifies the type of the queue as shown in Table 79.

\section*{Table 79-Types of queues of \$q_type values}
\begin{tabular}{|l|c|}
\hline q_type value & Type of queue \\
\hline 1 & first-in, first-out \\
\hline 2 & last-in, first-out \\
\hline
\end{tabular}

The maximum length parameter is an integer input that specifies the maximum number of entries allowed on the queue. The success or failure of the creation of the queue is returned as an integer value in status. The error conditions and corresponding values of status are described in Table 79.

\subsection*{17.6.2 \$q_add}

The \(\mathbf{\$ q}\) _add system task places an entry on a queue. The q_id parameter is an integer input that indicates to which queue to add the entry. The job_id parameter is an integer input that identifies the job.

The inform_id parameter is an integer input that is associated with the queue entry. Its meaning is userdefined. For example, inform_id parameter can represent execution time for an entry in a CPU model. The status parameter reports on the success of the operation or error conditions as described in Table 79.

\subsection*{17.6.3 \$q_remove}

The \(\$ q\) _remove system task receives an entry from a queue. The q_id parameter is an integer input that indicates from which queue to remove. The job_id parameter is an integer output that identifies the entry being removed. The inform_id parameter is an integer output that the queue manager stored during \(\mathbf{\$ q}\) _add. Its meaning is user-defined. The status parameter reports on the success of the operation or error conditions as described in Table 79.

\subsection*{17.6.4 \$q_full}

The \(\mathbf{\$ q}\) _full system function checks whether there is room for another entry on a queue. It returns 0 when the queue is not full and 1 when the queue is full.

\subsection*{17.6.5 \$q_exam}

The \(\$ q\) _exam system task provides statistical information about activity at the queue q_id. It returns a value in q_stat_value depending on the information requested in q_stat_code. The values of q_stat_code and the corresponding information returned in q_stat_value are described in Table 80.

Table 80-Parameter values for \$q_exam system task
\begin{tabular}{|l|l|}
\hline \begin{tabular}{c} 
Value requested in \\
q_stat_code
\end{tabular} & \multicolumn{1}{|c|}{\begin{tabular}{c} 
Information received back \\
from q_stat_value
\end{tabular}} \\
\hline 1 & Current queue length \\
\hline 2 & Mean interarrival time \\
\hline 3 & Maximum queue length \\
\hline 4 & Shortest wait time ever \\
\hline 5 & Longest wait time for jobs still in the queue \\
\hline 6 & Average wait time in the queue \\
\hline
\end{tabular}

\subsection*{17.6.6 Status codes}

All of the queue management tasks and functions return an output status parameter. The status parameter values and corresponding information are described in Table 81.

Table 81-Status parameter values
\begin{tabular}{|l|l|}
\hline \begin{tabular}{c} 
Status parameter \\
values
\end{tabular} & \multicolumn{1}{c|}{ What it means } \\
\hline 0 & OK \\
\hline 1 & Queue full, cannot add \\
\hline 2 & Undefined q_id \\
\hline 3 & Queue empty, cannot remove \\
\hline 4 & Unsupported queue type, cannot create queue \\
\hline 5 & Specified length <=0, cannot create queue \\
\hline 6 & Duplicate q_id, cannot create queue \\
\hline 7 & Not enough memory, cannot create queue \\
\hline
\end{tabular}

\subsection*{17.7 Simulation time system functions}

The following system functions provide access to current simulation time:
\[
\text { \$time } \quad \text { \$stime } \quad \text { \$realtime }
\]

\subsection*{17.7.1 \$time}

The syntax for \$time system function is shown in Syntax 17-14.
```

time_function ::=
\$time

```

Syntax 17-14 Syntax for \$time

The \$time system function returns an integer that is a 64-bit time, scaled to the timescale unit of the module that invoked it.

\section*{Example:}

In this example, the reg set is assigned the value 0 at simulation time 16 ns , and the value 1 at simulation time 32 ns . Note that these times do not match the times reported by \(\$\) time. The time values returned by the \$time system function are determined by the following steps:
a) The simulation times 16 ns and 32 ns are scaled to 1.6 and 3.2 because the time unit for the module is 10 ns , so time values reported by this module are multiples of 10 ns .
b) The value 1.6 is rounded to 2 , and 3.2 is rounded to 3 because the \(\$\) time system function returns

The \$realtime system function returns a real number time that, like \$time, is scaled to the time unit of the module that invoked it.

Example:
```

'timescale 10 ns / 1 ns
module test;
reg set;
parameter p = 1.55;
initial begin
$monitor($realtime, , "set=", set);
\#p set = 0;
\#p set = 1;
end
endmodule
// The output from this example is as follows:
// 0 set=x
// 1.6 set=0
// 3.2 set=1

```

In this example, the event times in the reg set are multiples of 10 ns because 10 ns is the time unit of the module. They are real numbers because \$realtime returns a real number.

\subsection*{17.8 Conversion functions}

The following functions handle real values:
```

integer \$rtoi(real_val) ;
real Sitor(int_val) ;
[63:0] Srealtobits(real_val) ;
real \$bitstoreal(bit_val) ;

```
\$rtoi converts real values to integers by truncating the real value (for example, 123.45 becomes 123)
\$itor converts integers to real values (for example, 123 becomes 123.0)
\$realtobits passes bit patterns across module ports; converts from a real number to the 64-bit representation (vector) of that real number
\$bitstoreal is the reverse of \$realtobits; converts from the bit pattern to a real number.
The real numbers accepted or generated by these functions shall conform to the IEEE Std 754-1985 [B1] representation of the real number. The conversion shall round the result to the nearest valid representation.

\section*{Example:}

The following example shows how the \$realtobits and \$bitstoreal functions are used in port connections:
```

module driver (net_r);
output net_r;
real r;
wire [64:1] net_r = \$realtobits(r);
endmodule
module receiver (net_r);
input net_r;
wire [64:1] net_r;
real r;
initial assign r = \$bitstoreal(net_r);
endmodule

```

See 4.5 for a description of \$signed and \$unsigned.

\section*{17.9 robabilistic distribution functions}

There are a set of random number generators that return integer values distributed according to standard probabilistic functions.

\subsection*{17.9.1 \$random function}

The syntax for the system function \$random is shown in Syntax 17-17.
```

random_function ::=
\$random [( seed)];

```

\section*{Syntax 17-17 Syntax for \$random}

The system function \$random provides a mechanism for generating random numbers. The function returns a new 32 -bit random number each time it is called. The random number is a signed integer; it can be positive or negative. For further information on probabilistic random number generators, see 17.9.2.

The seed parameter controls the numbers that \$random returns such that different seeds generate different random streams. The seed parameter shall be either a reg, an integer, or a time variable. The seed value should be assigned to this variable prior to calling \$random.

\section*{Examples:}

Example 1 Where b is greater than 0 , the expression (\$random \% b) gives a number in the following range:
\([(-b+1):(b-1)]\). The following code fragment shows an example of random number generation between
-59 and 59:
```

reg [23:0] rand;
rand = \$random % 60;

```

Example 2 The following example shows how adding the concatenation operator to the preceding example gives rand a positive value from 0 to 59 .
reg [23:0] rand;
rand \(=\{\$\) random \(\}\) \% 60;
17.9.2 \$dist_ functions
```

dist_functions ::=
\$dist_uniform ( seed, start, end );
\$dist_normal ( seed, mean, standard_deviation );
\$dist_exponential ( seed, mean );
\$dist_poisson ( seed, mean );
\$dist_chi_square ( seed, degree_of_freedom );
\$dist_t ( seed, degree_of_freedom );
| \$dist_erlang ( seed, k_stage, mean );

```

\section*{Syntax 17-18 Syntax for the probabilistic distribution functions}

All parameters to the system functions are integer values. For the exponential, poisson, chisquare, \(t\), and erlang functions, the parameters mean, degree of freedom, and \(k\) _stage shall be greater than 0 .

Each of these functions returns a pseudo-random number whose characteristics are described by the function name. That is, \$dist_uniform returns random numbers uniformly distributed in the interval specified by its parameters.

For each system function, the seed parameter is an in-out parameter; that is, a value is passed to the function and a different value is returned. The system functions shall always return the same value given the same seed. This facilitates debugging by making the operation of the system repeatable. The argument for the seed parameter should be an integer variable that is initialized by the user and only updated by the system function. This ensures the desired distribution is achieved.

In the \$dist_uniform function, the start and end parameters are integer inputs that bound the values returned. The start value should be smaller than the end value.

The mean parameter, used by \$dist_normal, \$dist_exponential, \$dist_poisson, and \$dist_erlang, is an integer input that causes the average value returned by the function to approach the value specified.

The standard deviation parameter used with the \$dist_normal function is an integer input that helps determine the shape of the density function. Larger numbers for standard deviation spread the returned values over a wider range.

The degree of freedom parameter used with the \$dist_chi_square and \$dist_t functions is an integer input that helps determine the shape of the density function. Larger numbers spread the returned values over a wider range.

\subsection*{17.9.3 Algorithm for probabilistic distribution functions}

Table 82 shows the Verilog probabilistic distribution functions listed with their corresponding C functions.
Table 82-Verilog to C function cross-listing
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ Verilog function } & \multicolumn{1}{c|}{ C function } \\
\hline \$dist_uniform & rtl_dist_uniform \\
\hline \$dist_normal & rtl_dist_normal \\
\hline \$dist_exponential & rtl_dist_exponential \\
\hline \$dist_poisson & rtl_dist_poisson \\
\hline \$dist_chi_square & rtl_dist_chi_square \\
\hline \$dist_t & rtl_dist_t \\
\hline \$dist_erlang & rtl_dist_erlang \\
\hline \$random & \begin{tabular}{l} 
rtl_dist_uniform( seed, \\
LONG_MIN, LONG_MAX )
\end{tabular} \\
\hline
\end{tabular}

The algorithm for these functions is defined by the following C code.
```

/*

* Algorithm for probabilistic distribution functions.
* 
* IEEE Std 1364-2000 Verilog Hardware Description Language (HDL).
*/
\#include <limits.h>
static double uniform( long *seed, long start, long end );
static double normal( long *seed, long mean, long deviation);
static double exponential( long *seed, long mean);
static long poisson( long *seed, long mean);
static double chi_square( long *seed, long deg_of_free);
static double t( long *seed, long deg_of_free);
static double erlangian( long *seed, long k, long mean);
long
rtl_dist_chi_square( seed, df )
long *seed;
long df;
{
double r;
long i;
if(df>0)
{
r=chi_square(seed,df);
if (r>=0)
{
i=(long)(r+0.5);

```
```

        }
        else
            {
                r = -r;
                i=(long)(r+0.5);
                i = -i;
            }
        }
        else
        {
            print_error("WARNING: Chi_square distribution must have
    positive
degree of freedom\n");
i=0;
}
return (i);
}
long
rtl_dist_erlang( seed, k, mean )
long *seed;
long k, mean;
{
double r;
long i;
if(k>0)
{
r=erlangian(seed,k,mean);
if(r>=0)
{
i=(long)(r+0.5);
}
else
{
r = -r;
i=(long)(r+0.5);
i = -i;
}
}
else
{
print_error("WARNING: k-stage erlangian distribution must
have
positive k\n");
i=0;
}
return (i);
}
long
rtl_dist_exponential( seed, mean )

```
```

        long *seed;
        long mean;
    {
double r;
long i;
if(mean>0)
{
r=exponential(seed,mean);
if(r>=0)
{
i=(long)(r+0.5);
}
else
{
r = -r;
i=(long)(r+0.5);
i = -i;
}
}
else
{
print_error("WARNING: Exponential distribution must have a
positive mean\n");
i=0;
}
return (i);
}
long
rtl_dist_normal( seed, mean, sd )
long *seed;
long mean, sd;
{
double r;
long i;
r=normal(seed,mean,sd);
if(r>=0)
{
i=(long)(r+0.5);
}
else
{
r = -r;
i=(long)(r+0.5);
i = -i;
}
return (i);
}

```
```

long
rtl_dist_poisson( seed, mean )
long *seed;
long mean;
{
long i;
if (mean>0)
{
i=poisson(seed,mean);
}
else
{
print_error("WARNING: Poisson distribution must have a
positive
mean\n");
i=0;
}
return (i);
}
long
rtl_dist_t( seed, df )
long *seed;
long df;
{
double r;
long i;
if(df>0)
{
r=t(seed,df);
if(r>=0)
{
i=(long)(r+0.5);
}
else
{
r = -r;
i=(long)(r+0.5);
i = -i;
}
}
else
{
print_error("WARNING: t distribution must have positive
degree
of freedom\n");
i=0;
}
return (i);
}
long
rtl_dist_uniform(seed, start, end)
long *seed;

```
```

    long start, end;
    {
double r;
long i;
if (start >= end) return(start);
if (end != LONG_MAX)
{
end++;
r = uniform( seed, start, end );
if (r >= 0)
{
i = (long) r;
}
else
{
i = (long) (r-1);
}
if (i<start) i = start;
if (i>=end) i = end-1;
}
else if (start!=LONG_MIN)
{
start--;
r = uniform( seed, start, end) + 1.0;
if (r>=0)
{
i = (long) r;
}
else
{
i = (long) (r-1);
}
if (i<=start) i = start+1;
if (i>end) i = end;
}
else
{
r =(uniform(seed,start,end)+2147483648.0)/
4294967295.0=;
r = r*4294967296.0-2147483648.0;
if (r>=0)
{
i = (long) r;
}
else
{
i = (long) (r-1);
}
}
return (i);
}

```
```

static double
uniform( seed, start, end )
long *seed, start, end;
{
union u_s
{
float s;
unsigned stemp;
} u;
double d = 0.00000011920928955078125;
double a,b,c;
if ((*seed) == 0)
*seed = 259341593;
if (start >= end)
{
a = 0.0;
b = 2147483647.0;
}
else
{
a = (double) start;
b = (double) end;
}
*seed = 69069 * (*seed) + 1;
u.stemp = *seed;
/*
* This relies on IEEE floating point format
*/
u.stemp = (u.stemp >> 9) 0x3f800000;
c = (double) u.s;
c = c+(c*d);
c = ((b - a) * (c - 1.0)) + a;
return (c);
}
static double
normal(seed,mean,deviation)
long *seed,mean,deviation;
{
double v1,v2,s;
double log(), sqrt();
s = 1.0;
while((s >= 1.0) || (s == 0.0))
{
v1 = uniform(seed,-1,1);

```
```

        v2 = uniform(seed,-1,1);
        s = v1 * v1 + v2 * v2;
    }
    s = v1 * sqrt(-2.0 * log(s) / s);
    v1 = (double) deviation;
    v2 = (double) mean;
    return(s * v1 + v2);
    }
static double
exponential(seed,mean)
long *seed,mean;
{
double log(),n;
n = uniform(seed,0,1);
if(n != 0)
{
n = -log(n) * mean;
}
return(n);
}
static long
poisson(seed,mean)
long *seed,mean;
{
long n;
double p,q;
double exp();
n = 0;
q = -(double)mean;
p = exp(q);
q = uniform(seed,0,1);
while(p < q)
{
n++;
q = uniform(seed,0,1) * q;
}
return(n);
}
static double
chi_square(seed,deg_of_free)
long *seed,deg_of_free;
{
double x;
long k;
if(deg_of_free % 2)
{
x = normal(seed,0,1);
x = x * x;
}
else
{
x = 0.0;

```
```

    }
    double log(),n;
    n = uniform(seed,0,1);
    if(n != 0)
    {
        n = - log(n) * mean;
    }
    return(n);
    }
static double
t(seed,deg_of_free)
long *seed,deg_of_free;
{
double sqrt(),x;
double chi2 = chi_square(seed,deg_of_free);
double div = chi2 / (double)deg_of_free;
double root = sqrt(div);
x = normal(seed,0,1) / root;
return(x);
}
static double
erlangian(seed,k,mean)
long *seed,k,mean;
{
double x,log(),a,b;
long i;
x=1.0;
for(i=1;i<=k;i++)
{
x = x * uniform(seed,0,1);
}
a=(double)mean;
b=(double)k;
x= -a* log(x)/b;
return(x);
}

```

\subsection*{17.10 Command line input}

An alternative to reading a file to obtain information for use in the simulation is specifying information with the command to invoke the simulator. This information is in the form of a optional argument provided to the simulation. These arguments are visually distinguished from other simulator arguments by the starting with the plus \((+)\) character.

These arguments, referred to below as plusargs, are accessible through the system functions described in 17.10.1 and 17.10.2.

\subsection*{17.10.1 \(\$\) test \(\$\) plusargs (string)}

This system function searches the list of plusargs for the provided string. The plusargs present on the command line are searched in the order provided. If the prefix of one of the supplied plusargs matches all characters in the provided string, a non-zero integer is returned. If no plusarg from the command line matches the string provided, the integer value zero (0) is returned.

\section*{Examples:}

Run simulator with command: +HELLO

The Verilog code is:
```

initial begin
if ($test$plusargs("HELLO")) $display("Hello argument found.")
    if ($test\$plusargs("HE")) $display("The HE subset string is
detected.");
    if ($test\$plusargs("H")) $display("Argument starting with H
found.");
    if ($test$plusargs("HELLO_HERE"))$display("Long argument.");
if ($test$plusargs("HI")) $display("Simple greeting.");
    if ($test\$plusargs("LO")) \$display("Does not match.");
end

```

This would produce the following output:
```

Hello argument found.
The HE subset string is detected.
Argument starting with H found.

```

\subsection*{17.10.2 \$value\$plusargs (user_string, variable)}

This system function searches the list of plusargs (like the \$test\$plusargs system function) for a user specified plusarg string. The string is specified in the first argument to the system function as either a string or a register which is interpreted as a string. If the string is found, the remainder of the string is converted to the type specified in the user_string and the resulting value stored in the variable provided. If a string is found, the function returns a non-zero integer. If no string is found matching, the function returns the integer value zero and the variable provided is not modified. No warnings shall be generated when the function returns zero (0).

The user_string shall be of the form: "plusarg_string format_string". The format strings are the same as the \$display system tasks. These are the only valid ones (upper and lower case as well as a leading 0 forms are valid):
\begin{tabular}{ll}
\(\% \mathrm{~d}\) & decimal conversion \\
\(\% \mathrm{o}\) & octal conversion \\
\(\% \mathrm{~h}\) & hexadecimal conversion \\
\(\% \mathrm{~b}\) & binary conversion \\
\(\% \mathrm{e}\) & real exponential conversion \\
\(\% \mathrm{f}\) & real decimal conversion \\
\(\% \mathrm{~g}\) & real decimal or exponential conversion \\
\(\% \mathrm{~S}\) & string (no conversion)
\end{tabular}

The first string, from the list of plusargs provided to the simuator, which matches the plusarg_string portion of the user_string specified shall be the plusarg string available for conversion. The remainder string of the matching plusarg (the remainder is the part of the plusarg string after the portion which matches the users
plusarg_string) shall be converted from a string into the format indicated by the format string and stored in the variable provided. If there is no remaining string, the value stored into the variable shall either be a zero \((0)\) or an empty string value.

If the size of the variable is larger than the value after conversion, the value stored is zero (0) padded to the width of the variable. If the variable can not contain the value after conversion, the value shall be truncated. If the value is negative, the value shall be considered larger than the variable provided. If characters exist in the string available for conversion, which are illegal for the specified conversion, the variable shall be written with the value 'bx.

Examples:
```

+FINISH=10000 +TESTNAME=this_test +FREQ+5.6666 +FREQUENCY +TEST12
// Get clock to terminate simulation if specified.
real frequency;
reg 8*32:1 testname;
integer stop_clock;
if ($value$plusargs("FINISH=%d", stop_clock))
begin
repeat (stop_clock) @(posedge clk);
$finish;
    end
// Get testname from plusarg.
if ($value\$plusargs("TESTNAME=%s", testname))
begin
$display("Running test %0s.", testname);
    startTest();
    end
// Get frequency from command line; set default if not specified.
if (!$value\$plusargs("FREQ+%0F", frequency))
frequency = 8.33333; // 166MHz;

```
```

forever

```
forever
begin
begin
    #frequency clk = 0;
    #frequency clk = 0;
    #frequency clk = 1;
    #frequency clk = 1;
end
end
reg [64*8:1] pstring;
reg [64*8:1] pstring;
pstring = "+TEST%d";
pstring = "+TEST%d";
if ($value$plusargs(pstring, test[31:0))
if ($value$plusargs(pstring, test[31:0))
    begin
    begin
    $display("Running test number %0d.", test);
    $display("Running test number %0d.", test);
    startTest();
    startTest();
    end
```

    end
    ```

This code would have the following effects:
The variable test would get the value 'd12.
The variable stop_clock obtains the value 10000 .
The variable testname obtains the value this_test.
The variable frequency obtains the value 5.6666; note the final plusarg +FREQUENCY does not affect the value of the variable frequency.

The output is:
```

Running test this_test.
Running test number 12.

```

\section*{18. Value change dump (VCD) files}

A value change dump (VCD) file contains information about value changes on selected variables in the design stored by value change dump system tasks. Two types of VCD files exist:
a) Four state: to represent variable changes in \(0,1, x\), and \(z\) with no strength information.
b) Extended: to represent variable changes in all states and strength information.

This clause describes how to generate both types of VCD files and their format.

\subsection*{18.1 Creating the four state value change dump file}

The steps involved in creating the four state VCD file are listed below and illustrated in Figure 51.
a) Insert the VCD system tasks in the Verilog source file to define the dump file name and to specify the variables to be dumped.
b) Run the simulation.


Figure 51-Creating the four state VCD file
A VCD file is an ASCII file which contains header information, variable definitions, and the value changes for all variables specified in the task calls.

Several system tasks can be inserted in the source description to create and control the VCD file.

\subsection*{18.1.1 Specifying the name of the dump file (\$dumpfile)}

The \$dumpfile task shall be used to specify the name of the VCD file. The syntax for the task is given in Syntax 18-1.
```

dumpfile_task ::=
\$dumpfile(filename);

```

Syntax 18-1 Syntax for \$dumpfile task

The filename syntax is given in Syntax 18-2.
\(\square\)
filename ::=
literal_string
| variable
expression

Syntax 18-2 Syntax for filename

The filename is optional and defaults to the literal string dump. vcd if not specified.

\section*{Example:}
initial Sdumpfile ("module1.dump") ;

\subsection*{18.1.2 Specifying the variables to be dumped (\$dumpvars)}

The \$dumpvars task shall be used to list which variables to dump into the file specified by \$dumpfile. The \$dumpvars task can be invoked as often as desired throughout the model (for example, within various blocks), but the execution of all the \$dumpvars tasks shall be at the same simulation time.

The \$dumpvars task can be used with or without arguments. The syntax for the \$dumpvars task is given in Syntax 18-3.
```

dumpvars_task ::= (Not in the Annex A BNF)
\$dumpvars;
| \$dumpvars (levels [, list_of_modules_or_variables ] );
list_of_modules_or_variables ::= (Not in the Annex A BNF)
module_or_variable {, module_or_variable }
module_or_variable ::=
module_identifier
| variable_identifier

```

\section*{Syntax 18-3 Syntax for \$dumpvars task}

When invoked with no arguments, \$dumpvars dumps all the variables in the model to the VCD file.

When the \$dumpvars task is specified with arguments, the first argument indicates how many levels of the hierarchy below each specified module instance to dump to the VCD file. Subsequent arguments specify which scopes of the model to dump to the VCD file. These arguments can specify entire modules or individual variables within a module.

Setting the first argument to 0 causes a dump of all variables in the specified module and in all module instances below the specified module. The argument 0 applies only to subsequent arguments which specify module instances, and not to individual variables.

\section*{Examples:}

\section*{Example 1}
\$dumpvars (1, top);
Because the first argument is a 1 , this invocation dumps all variables within the module top; it does not dump variables in any of the modules instantiated by module top.

\section*{Example 2}
\$dumpvars (0, top);
In this example, the \$dumpvars task shall dump all variables in the module top and in all module instances below module top in the hierarchy.

Example 3 This example shows how the \$dumpvars task can specify both modules and individual variables:
\$dumpvars (0, top.mod1, top.mod2.net1);
This call shall dump all variables in module mod1 and in all module instances below mod1, along with variable net 1 in module mod2. The argument 0 applies only to the module instance top.mod1 and not to the individual variable top.mod2. net1.

\subsection*{18.1.3 Stopping and resuming the dump (\$dumpoff/\$dumpon)}

Executing the \$dumpvars task causes the value change dumping to start at the end of the current simulation time unit. To suspend the dump, the \$dumpoff task can be invoked. To resume the dump, the \$dumpon task can be invoked. The syntax of these two tasks is given in Syntax 18-4.
```

dumpoff_task ::=
\$dumpoff;
dumpon_task ::=
\$dumpon;

```

\section*{Syntax 18-4 Syntax for \$dumpoff and \$dumpon tasks}

When the \$dumpoff task is executed, a checkpoint is made in which every selected variable is dumped as an \(x\) value. When the \$dumpon task is later executed, each variable is dumped with its value at that time. In the interval between \$dumpoff and \$dumpon, no value changes are dumped.

The \$dumpoff and \$dumpon tasks provide the mechanism to control the simulation period during which the dump shall take place.

\section*{Example:}
```

initial begin
\#10 \$dumpvars( . . . );
\#200 \$dumpoff;
\#800 Sdumpon;
\#900 \$dumpoff;
end

```

This example starts the value change dump after 10 time units, stops it 200 time units later (at time 210), restarts it again 800 time units later (at time 1010), and stops it 900 time units later (at time 1910).

\subsection*{18.1.4 Generating a checkpoint (\$dumpall)}

The \$dumpall task creates a checkpoint in the VCD file which shows the current value of all selected variables. The syntax is given in Syntax 18-5.
```

dumpall_task ::=
\$dumpall;

```

\section*{Syntax 18-5 Syntax for \$dumpall task}

When dumping is enabled, the value change dumper records the values of the variables which change during each time increment. Values of variables which do not change during a time increment are not dumped.

\subsection*{18.1.5 Limiting the size of the dump file (\$dumplimit)}

The \$dumplimit task can be used to set the size of the VCD file. The syntax for this task is given in Syntax 18-6.
```

dumplimit_task ::=
\$dumplimit (filesize);

```

Syntax 18-6 Syntax fro \$dumplimit task

The filesize argument which specifies the maximum size of the VCD file in bytes. When the size of VCD file reaches this number of bytes, the dumping stops and a comment is inserted in the VCD file indicating the dump limit was reached.

\subsection*{18.1.6 eading the dump file during simulation (\$dumpflush)}

The \$dumpflush task can be used to empty the VCD file buffer of the operating system to ensure all the data in that buffer is stored in the VCD file. After executing a \$dumpflush task, dumping is resumed as before so no value changes are lost. The syntax for the task is given in Syntax 18-7.
```

dumpflush_task ::=
\$dumpflush ;

```

\section*{Syntax 18-7 Syntax for \$dumpflush task}

A common application is to call \$dumpflush to update the dump file so an application program can read the VCD file during a simulation.

\section*{Examples:}

Example 1 This example shows how the \$dumpflush task can be used in a Verilog HDL source file:

\section*{initial begin}

\section*{\$dumpvars;}
.

\section*{\$dumpflush ;}
\$(applications program) ;
end

Example 2 The following is a simple source description example to produce a VCD file.

In this example, the name of the dump file is verilog. dump. It dumps value changes for all variables in the model. Dumping begins when an event do_dump occurs. The dumping continues for 500 clock cycles, then stops and waits for the event do_dump to be triggered again. At every 10000 time steps, the current values of all VCD variables are dumped.
```

module dump;
event do dump;
initial \$dumpfile ("verilog.dump");
initial @do_dump
\$dumpvars; //dump variables in the design
always @do_dump //to begin the dump at event do_dump
begin
\$dumpon; //no effect the first time through
repeat (500) @(posedge clock); //dump for 500 cycles
\$dumpoff; //stop the dump
end
initial @(do dump)
forever \#10000 \$dumpall; //checkpoint all variables
endmodule

```

\subsection*{18.2 Format of the four state VCD file}

The dump file is structured in a free format. White space is used to separate commands and to make the file easily readable by a text editor.

\subsection*{18.2.1 Syntax of the four state VCD file}

The syntax of the four state VCD file is given in Syntax 18-8.
```

value_change_dump_definitions ::=
\{ declaration_command \}\{ simulation_command \}
declaration_command ::=
declaration_keyword
[ command_text ]
Send
simulation_command ::=
simulation_keyword \{ value_change \} Send
\$comment [ comment_text] Send
simulation_time
| value_change
declaration_keyword ::=
\$comment | Sdate | Senddefinitions | \$scope | Stimescale | Supscope
| \$var | \$version
simulation_keyword ::=
\$dumpall | \$dumpoff | Sdumpon | Sdumpvars
simulation_time ::=
\# decimal_number
value_change ::=
scalar_value_change
| vector_value_change
scalar_value_change ::=
value identifier_code
value ::=
$\mathbf{0}|\mathbf{1}| \mathbf{x}|\mathbf{X}| \mathbf{z} \mid \mathbf{Z}$
vector_value_change ::=
b binary_number identifier_code
B binary_number identifier_code
$\mathbf{r}$ real_number identifier_code
$\mathbf{R}$ real_number identifier_code
identifier_code ::=
\{ ASCII character $\}$

```

\section*{Syntax 18-8 Syntax of the output four state VCD file}

The VCD file starts with header information giving the date, the version number of the simulator used for the simulation, and the timescale used. Next, the file contains definitions of the scope and type of variables being dumped, followed by the actual value changes at each simulation time increment. Only the variables which change value during a time increment are listed.

The simulation time recorded in VCD file is the absolute value of the simulation time for the changes in variable values which follow.

Value changes for real variables are specified by real numbers.Value changes for all other variables are specified in binary format by \(0,1, x\), or \(z\) values. Strength information and memories are not dumped.

A real number is dumped using a \(\% .16 \mathrm{~g}\) printf() format. This preserves the precision of that number by outputting all 53 bits in the mantissa of a 64 -bit IEEE Std 754-1985 [B1] double-precision number. Application programs can read a real number using a \(\% g\) format to \(\operatorname{scanf}()\).

The value change dumper generates character identifier codes to represent variables. The identifier code is a code composed of the printable characters which are in the ASCII character set from! to \(\sim\) (decimal 33 to 126).

\section*{NOTES:}
1) The VCD format does not support a mechanism to dump part of a vector. For example, bits 8 to 15 (8:15) of a 16-bit vector cannot be dumped in VCD file; instead, the entire vector \((0: 15)\) has to be dumped. In addition, expressions, such as \(\mathrm{a}+\mathrm{b}\), cannot be dumped in the VCD file.
2) Data in the VCD file is case sensitive.

\subsection*{18.2.2 Formats of variable values}

Variables can be either scalars or vectors. Each type is dumped in its own format. Dumps of value changes to scalar variables shall not have any white space between the value and the identifier code.

Dumps of value changes to vectors shall not have any white space between the base letter and the value digits, but they shall have one white space between the value digits and the identifier code.

The output format for each value is right-justified. Vector values appear in the shortest form possible: redundant bit values which result from left-extending values to fill a particular vector size are eliminated.

The rules for left-extending vector values are given in Table 83.
Table 83-Rules for left-extending vector values
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ When the value is } & VCD left-extends with \\
\hline 1 & 0 \\
\hline 0 & 0 \\
\hline\(Z\) & \(Z\) \\
\hline\(X\) & \(X\) \\
\hline
\end{tabular}

Table 84 shows how the VCD can shorten values.
Table 84-How the VCD can shorten values
\begin{tabular}{|l|l|l|}
\hline \multicolumn{1}{|c|}{ The binary value } & \multicolumn{1}{|c|}{\begin{tabular}{c} 
Extends to fill a \\
4-bit reg as
\end{tabular}} & \multicolumn{1}{c|}{\begin{tabular}{c} 
Appears in the \\
VCD file as
\end{tabular}} \\
\hline 10 & 0010 & b 10 \\
\hline X 10 & XX 10 & bX 10 \\
\hline ZX 0 & ZZX 0 & bZX 0 \\
\hline 0 X 10 & 0 X 10 & b 0 X 10 \\
\hline
\end{tabular}

Events are dumped in the same format as scalars; for example, \(1^{*} \%\). For events, however, the value ( 1 in this example) is irrelevant. Only the identifier code ( \(* \%\) in this example) is significant. It appears in the VCD file as a marker to indicate the event was triggered during the time step.

\section*{Examples:}
```

1*@ No space between the value 1 and the identifier code *@
b1100x01z (k No space between the b and 1100x01z,
but a space between b1100x01z and (k

```

\subsection*{18.2.3 Description of keyword commands}

The general information in the VCD file is presented as a series of sections surrounded by keywords. Keyword commands provide a means of inserting information in the VCD file. Keyword commands can be inserted either by the dumper or manually.

This sub clause deals with the keyword commands given in Table 85.
Table 85-Keyword commands
\begin{tabular}{|l|l|l|}
\hline \multicolumn{2}{|c|}{ Declaration keywords } & Simulation keywords \\
\hline \$comment & \$timescale & \$dumpall \\
\hline \$date & \$upscope & \$dumpoff \\
\hline \begin{tabular}{l} 
Senddefini- \\
tions
\end{tabular} & \$var & \$dumpon \\
\hline \$scope & \$version & \$dumpvars \\
\hline
\end{tabular}

\subsection*{18.2.3.1 \$comment}

The \$comment section provides a means of inserting a comment in the VCD file. The syntax for the section is given in Syntax 18-9.
```

vcd_declaration_comment ::=
\$comment comment_text \$end

```

Syntax 18-9 Syntax for \$comment section

\section*{Examples:}
```

\$comment This is a single-line comment \$end
\$comment This is a
multiple-line comment
\$end

```

\subsection*{18.2.3.2 \$date}

The \$date section indicates the date on which the VCD file was generated.The syntax for the section is given in Syntax 18-10.

> vcd_declaration_date \(::=\) \$date date_text \$end

\section*{Syntax 18-10 Syntax for \$date section}

\section*{Example:}
\$date
June 25, 1989 09:24:35
\$end

\subsection*{18.2.3.3 \$enddefinitions}

The \$enddefinitions section marks the end of the header information and definitions. The syntax for the section is given in Syntax 18-11.
```

vcd_declaration_enddefinitions ::=
\$enddefinitions \$end

```

\section*{Syntax 18-11 Syntax for \$enddefinitions section}

\subsection*{18.2.3.4 \$scope}

The \$scope section defines the scope of the variables being dumped. The syntax for the section is given in Syntax 18-12.
\(\square\)

\section*{Syntax 18-12 Syntax for \$scope section}

The scope type indicates one of the following scopes:
\begin{tabular}{ll} 
module & Top-level module and module instances \\
task & Tasks \\
function & Functions \\
begin & Named sequential blocks \\
fork & Named parallel blocks
\end{tabular}

\section*{Example:}
```

\$scope
module top
\$end

```

\subsection*{18.2.3.5 \$timescale}

The \$timescale keyword specifies what timescale was used for the simulation. The syntax for the keyword is given in Syntax 18-13.
```

vcd_declaration_timescale ::=
\$timescale time_number time_unit \$end
time_number ::=
1| 10 100
time_unit ::=
s|\mathbf{ms}|\mathbf{us}|\mathbf{ns}|\mathbf{ps}|\mathbf{fs}

```

Syntax 18-13 Syntax for \$timescale

Example:
\[
\text { \$timescale } 10 \text { ns \$end }
\]

\subsection*{18.2.3.6 \$upscope}

The \$upscope section indicates a change of scope to the next higher level in the design hierarchy. The syntax for the section is given in Syntax 18-14.
```

vcd_declaration_upscope ::=
\$upscope \$end

```

\section*{Syntax 18-14 Syntax for \$upscope section}

\subsection*{18.2.3.7 \$version}

The \$version section indicates which version of the VCD writer was used to produce the VCD file and the \$dumpfile system task used to create the file. If a variable or an expression was used to specify the filename within \$dumpfile, the unevaluated variable or expression literal shall appear in the \$version string. The syntax for the \$version section is given in Syntax 18-15.
```

vcd_declaration_version ::=
\$version version_text system_task \$end

```

Syntax 18-15 Syntax for \$version section

\section*{Example:}

\section*{\$version}

VERILOG-SIMULATOR 1.0a
\$du]!E2R\&gVr|mnl>gHRJ4X5E252Rf6yG23SyA2le5EU235y572R]!EAy66A52Rf6E2pyps1|mnl>g

\subsection*{18.2.3.9 \$dumpall}

The \$dumpall keyword specifies current values of all variables dumped. The syntax for the keyword is given in Syntax 18-17.
```

vcd_simulation_dumpall ::=
\$dumpall { value_changes } Send

```

Syntax 18-17 Syntax for \$dumpall keyword

Example:
\$dumpall \(\quad 1^{*}\) © \(x^{\star} \# \quad 0 * \$ \quad b x \quad(k \quad\) Send

\subsection*{18.2.3.10 \$dumpoff}

The \$dumpoff keyword indicates all variables dumped with X values. The syntax for the keyword is given in Syntax 18-18.
```

vcd_simulation_dumpoff ::=
\$dumpoff { value_changes } Send

```

\section*{Syntax 18-18 Syntax for \$dumpoff keyword}

Example:
\[
\text { Sdumpoff } \quad x^{\star} @ x^{\star} \# \quad x^{\star} \$ \quad b x \quad(k) \text { Send }
\]

\subsection*{18.2.3.11 \$dumpon}

The \$dumpon keyword indicates resumption of dumping and lists current values of all variables dumped. The syntax for the keyword is given in Syntax 18-19.
```

vcd_simulation_dumpon ::=
\$dumpon { value_changes } Send

```

\section*{Syntax 18-19 Syntax for \$dumpon keyword}

Example:
\[
\begin{array}{lllllll}
\text { Sdumpon } & x^{\star} @ & 0 \star \# & x^{\star} \$ & \text { b1 } & (k) & \text { Send }
\end{array}
\]

\subsection*{18.2.3.12 \$dumpvars}

The section beginning with \(\$\) dumpvars keyword lists initial values of all variables dumped. The syntax for the keyword is given in Syntax 18-20.
\(\square\)
vcd_simulation_dumpvars ::=
\$dumpvars \{ value_changes \} Send

Syntax 18-20 Syntax for \$dumpvars keyword

Example:
\$dumpvars \(\quad x^{\star} @ \quad z^{\star} \$\) b0 (k \$end

\subsection*{18.2.4 Four state VCD file format example}

The following example illustrates the format of the four state VCD file.
```

\$date June 26, 1989 10:05:41
\$end
\$version VERILOG-SIMULATOR 1.0a
\$end
\$timescale 1 ns
\$end
\$scope module top \$end
\$scope module m1 \$end
\$var trireg 1 *@ net1 \$end
\$var trireg 1 * \# net2 \$end
\$var trireg 1 * \$ net3 \$end
\$upscope \$end
\$scope task t1 \$end
\$var reg 32 (k accumulator[ 31:0] \$end
\$var integer 32 { 2 index \$end
\$upscope \$end
\$upscope \$end
\$enddefinitions \$end
\$comment
Note: \$dumpvars was executed at time ' \#500'.
All initial values are dumped at this time.
\$end
\#500
\$dumpvars
x*@
x* \#
x* \$
bx (k
bx { 2
\$end
\#505
0* @
1* \#
1* \$
b10zx1110x11100 (k
b1111000101z01x { 2
\#510
0* \$
\#520
1* \$
\#530
0* \$
bz (k
\#535
\$dumpall 0*@ 1* \# 0* \$

```
( Continued in right column )
(Continued from left column)
bz (k
b1111000101z01x \{ 2
\$end
\#540
1* \$
\#1000
\$dumpoff
x* @
\(x^{*}\) \#
x* \$
bx (k
bx \{ 2
Send
\#2000
\$dumpon
z* @
1* \#
0* \(\$\)
b0 (k
bx \{ 2
\$end
\#2010
1* \$

\subsection*{18.3 Creating the extended value change dump file}

The steps involved in creating the extended VCD file are listed below and illustrated in Figure 52.


Figure 52-Creating the extended VCD file
a) Insert the extended VCD system tasks in the Verilog source file to define the dump file name and to specify the variables to be dumped.
b) Run the simulation.

The four state VCD file rules and syntax apply to the extended VCD file unless otherwise stated in this section.

\subsection*{18.3.1 Specifying the dumpfile name and the ports to be dumped (\$dumpports)}

The \$dumpports task shall be used to specify the name of the VCD file and the ports to be dumped. The syntax for the task is given in Syntax 18-21.
```

dumpports_task ::=
\$dumpports ( scope_list, file_pathname ) ;
scope_list ::=
module_identifier \{, module_identfier \}
file_pathname ::=
literal_string
| variable
expression

```

\section*{Syntax 18-21 Syntax for \$dumpports task}

Where the arguments are optional and are defined as:
scope_list one or more module identifiers. Only modules are allowed (not variables). If more than one module_identifier is specified, they shall be separated by a comma. Pathnames to modules are allowed, using the period hierarchy separator. Literal strings are not allowed for the module_identifier.

If no scope_list value is provided, the scope shall be the module from which \$dumpports is called.
file_pathname can be a double quoted pathname (literal string), a reg type variable, or an expression which denotes the file which shall contain the port VCD information. If no file_pathname is provided, the file shall be written to the current working directory with the name dumpports.vcd. If that file already exists, it shall be silently overwritten. All file writing checks shall be made by the simulator (write rights, correct pathname, etc.) and appropriate errors or warnings issued.

The following rules apply to the use of the \$dumpports system task:

All the ports in the model from the point of the \$dumpports call are considered primary I/O pins and shall be included in the VCD file. However, any ports which exist in instantiations below scope_list are not dumped.
If no arguments are specified for the task, \$dumpports; and \$dumpports() are allowed. In both of these cases, the default values for the arguments shall be used.
If the first argument is null, a comma shall be used before specifying the second argument in the argument list.
Each scope specified in the scope_list shall be unique. If multiple calls to \$dumpports are specified, the scope_list values in these calls shall also be unique.
The \$dumpports task can be used in source code which also contains the \$dumpvars task.
When \$dumpports executes, the associated value change dumping shall start at the end of the current simulation time unit.
The \$dumpports task can be invoked multiple times throughout the model, but the execution of all \$dumpports tasks shall be at the same simulation time. Specifying the same file_pathname multiple times is not allowed.

\subsection*{18.3.2 Stopping and resuming the dump (\$dumpportsoff/\$dumpportson)}

The \$dumpportsoff and \$dumpportson system tasks provide a means to control the simulation period for dumping port values. The syntax for these system tasks is given in Syntax 18-22.


Syntax 18-22 Syntax for \$dumpportsoff and \$dumpportson system tasks

The file_pathname argument can be a double quoted pathname (literal string), a reg type variable, or an expression which denotes the file_pathname specified in the \$dumpports system task.
\$dumpportsoff. When this task is executed, a checkpoint is made in the file_pathname where each specified port is dumped with an X value. Port values are no longer dumped from that simulation time forward. If file_pathname is not specified, all dumping to files opened by \$dumpports calls shall be suspended.
\$dumpportson. When this task is executed, all ports specified by the associated \$dumpports call shall have their values dumped. This system task is typically used to resume dumping after the execution of \$dumpportsoff. If file_pathname is not specified, dumping shall resume for all files specified by \$dumpports calls, if dumping to those files was stopped.

If \$dumpportson is executed while ports are already being dumped to file_pathname, the system task is ignored. If \$dumpportsoff is executed while port dumping is already suspended for file_pathname, the system task is ignored.

\subsection*{18.3.3 Generating a checkpoint (\$dumpportsall)}

The \$dumpportsall system task creates a checkpoint in the VCD file which shows the value of all selected ports at that time in the simulation, regardless of whether the port values have changed since the last timestep. The syntax for this system task is given in Syntax 18-23.
```

dumpportsall_task ::=
\$dumpportsall ( file_pathname) ;
file_pathname ::=
literal_string
| variable
| expression

```

\section*{Syntax 18-23 Syntax for \$dumpportsall system task}

The file_pathname argument can be a double quoted pathname (literal string), a reg type variable, or an expression which denotes the file_pathname specified in the \$dumpports system task.

If the file_pathname is not specified, checkpointing occurs for all files opened by calls to \$dumpports.

\subsection*{18.3.4 Limiting the size of the dump file (\$dumpportslimit)}

The \$dumpportslimit system task allows control of the VCD file size. The syntax for this system task is given in Syntax 18-24.
dumpportslimit_task ::=
\$dumpportslimit (filesize, file_pathname) ;
file_size ::=
integer
file_pathname ::=
literal_string
| variable
expression

\section*{Syntax 18-24 Syntax for \$dumpportslimit system task}

The filesize argument is required and it specifies the maximum size in bytes for the associated file_pathname. When this filesize is reached, the dumping stops and a comment is inserted into file_pathname indicating the size limit was attained.

The file_pathname argument can be a double quoted pathname (literal string), a reg type variable, or an expression which denotes the file_pathname specified in the \$dumpports system task.

If the file_pathname is not specified, the filesize limit applies to all files opened for dumping due to calls to \$dumpports.

\subsection*{18.3.5 eading the dump file during simulation (\$dumpportsflush)}

To facilitate performance, simulators often buffer VCD output and write to the file at intervals, instead of line by line. The \$dumpportsflush system task writes all port values to the associated file, clearing a simulator s VCD buffer.

The syntax for this system task is given in Syntax 18-25.
\(\square\)
dumpportsflush_task ::=
    \$dumpportsflush ( file_pathname ) ;
file_pathname ::=
            literal_string
    | variable
    | expression

\section*{Syntax 18-25 Syntax for \$dumpportsflush system task}

The file_pathname argument can be a double quoted pathname (literal string), a reg type variable, or an expression which denotes the file_pathname specified in the \$dumpports system task.

If the file_pathname is not specified, the VCD buffers shall be flushed for all files opened by calls to \$dumpports.

\subsection*{18.3.6 Description of keyword commands}

The general information in the extended VCD file is presented as a series of sections surrounded by keywords. Keyword commands provide a means of inserting information in the extended VCD file. Keyword commands can be inserted either by the dumper or manually. Extended VCD provides one additional keyword command to that of the four state VCD.

\subsection*{18.3.6.1 \$vcdclose}

The \$vedclose keyword indicates the final simulation time at the time the extended VCD file is closed. This allows accurate recording of the end simulation time, regardless of the state of signal changes, in order to assist parsers which require this information. The syntax for the keyword is given in Syntax 18-26.
```

vcdclose_task ::=
\$vcdclose final_simulation_time \$end

```

Syntax 18-26 Syntax for \$vcdclose keyword

Example:
Svedclose \#13000 \$end

\subsection*{18.3.7 General rules for extended VCD system tasks}

For each extended VCD system task, the following rules apply:
If a file_pathname is specified which does not match a file_pathname specified in a \$dumpports call, the control task shall be ignored.
If no arguments are specified for the tasks which have only optional arguments, the system task name can be used with no arguments or the name followed by () can be specified. For example: \$dumpportsflush; or \$dumpportsflush(). In both of these cases, the default actions for the arguments shall be executed.

\subsection*{18.4 Format of the extended VCD file}

The format of the extended VCD file is similar to that of the four state VCD file, as it is also structured in a free format. White space is used to separate commands and to make the file easily readable by a text editor.

\subsection*{18.4.1 Syntax of the extended VCD file}

The syntax of the extended VCD file is given in Syntax 18-27. A four state VCD construct name which matches an extended VCD construct shall be considered equivalent, except if preceded by an *.
```

value_change_dump_definitions ::={declaration_command} {simulation_command}
declaration_command ::= declaration_keyword [command_text] \$end
simulation_command ::= (Not in the Annex A BNF)
simulation_keyword { value_change } \$end
| \$comment [comment_text] \$end
simulation_time
value_change

* declaration_keyword ::=
\$comment | \$date | \$enddefinitions | \$scope | \$timescale | \$upscope | \$var
| \$vcdclose | \$version
command_text ::=
comment_text | close_text | date_section | scope_section | timescale_section
| var_section | version_section
* simulation_keyword ::= \$dumpports | \$dumpportsoff | \$dumpportson |
\$dumpportsall
simulation_time ::= \#decimal_number
value_change ::= value identifier_code
value ::= pport_value 0_strength_component l_strength_component
port_value ::= input_value | output_value | unknown_direction_value
input_value ::=\mathbf{D}|\mathbf{U}|\mathbf{N}|\mathbf{Z}|\mathbf{d}|\mathbf{u}
output_value ::=\mathbf{L}|\mathbf{H}|\mathbf{X | T | l |h}
unknown_direction_value ::=\mathbf{0}|\mathbf{1}|\mathrm{ ? |F | A | a | B | b | C c |f}
strength_component ::=0 1 1 | 2 | 3 | 4 5 5 6 | 7
* identifier_code ::=<{integer }
comment_text ::= {ASCII_character }
close_text ::= final_simulation_time
date_section ::= date_text
date_text :: = day month date time year
scope_section ::= scope_type scope_identifier
* scope_type ::= module
timescale_section ::= number time_unit
number ::=1 | 10 | 100
time_unit ::= fs | ps | ns | us |ms | s
var_section ::= var_type size identifier_code reference
* var_type ::= port
* size ::= \mathbf{1}|}\mathrm{ vector_index
vector_index ::= [ msb_index : lsb_index ]
index ::= decimal_number
* reference ::= port_identifier
identifier ::= {printable_ASCII_character}
version_section ::= version_text
* version_text ::= version_identifier {dumpports_command}
dumpports_command ::=
\$dumpports (scope_identifier, string_literal
variable
expression )

```

The extended VCD file starts with header information giving the date, the version number of the simulator used for the simulation, and the timescale used. Next, the file contains definitions of the scope of the ports being dumped, followed by the actual value changes at each simulation time increment. Only the ports which change value during a time increment are listed.

The simulation time recorded in the extended VCD file is the absolute value of the simulation time for the changes in port values which follow.

Value changes for all ports are specified in binary format by \(0,1, x\), or \(z\) values and include strength information.

A real number is dumped using \(\mathrm{a} \% .16 \mathrm{~g}\) printf() format. This preserves the precision of that number by outputting all 53 bits in the mantissa of a 64-bit IEEE Std 754-1985 [B1] double-precision number. Application programs can read a real number using a \(\% \mathrm{~g}\) format to \(\operatorname{scanf}()\).

\section*{NOTES:}
1) The extended VCD format does not support a mechanism to dump part of a vector. For example, bits 8 to 15 (8:15) of a 16-bit vector cannot be dumped in VCD file; instead, the entire vector \((0: 15)\) has to be dumped. In addition, expressions, such as \(\mathrm{a}+\mathrm{b}\), cannot be dumped in the VCD file.
2) Data in the extended VCD file is case sensitive.

\subsection*{18.4.2 Extended VCD node information}

The node information section (also referred to as the variable definitions section) is affected by the \$dumpports task as Syntax 18-28 shows.
```

\$var var_type size < identifier_code reference \$end
var_type ::=
port
size ::=
1
| vector_index
vector_index ::=
[msb_index : lsb_index]
index ::=
decimal_number
identifier_code ::=
integer
reference ::=
port_identifier

```

The constructs are defined as:
var_type the keyword port. No other keyword is allowed.
size a decimal number indicating the number of bits in the port. If the port is a single bit, the value shall be 1 . If the port is a bus, the actual index is printed. The msb indicates the most significant index; \(l s b\) the least significant index.
identifier_code an integer preceded by < which starts at zero and ascends in one unit increments for each port, in the order found in the module declaration.
reference identifier indicating the port name.
Example:
```

module test_device(count_out, carry, data, reset)
output count_out, carry ;
input [ 0:3] data;
input reset;
initial
begin
\$dumpports(testbench.DUT, "testoutput.vcd");
end

```

This example produces the following node information in the VCD file:
\begin{tabular}{lcccc} 
\$scope module & testbench. DUT & \$end & & \\
\$var port & 1 & \(<0\) & count_out & \$end \\
\$var port & \(1<1\) & carry & \$end \\
\$var port & {\([0: 3]<2\)} & data & \$end \\
\$var port & \(1<3\) & reset & \$end \\
\$upscope \$end & & &
\end{tabular}

At least one space shall separate each syntactical element. However, the formatting of the information is the choice of the simulator vendor. All four state VCD syntax rules for the vector_index apply.

If the vector_index appears in the port declaration, this shall be the index dumped. If the vector_index is not in the port declaration, the vector_index in the net or reg declaration matching the port name shall be dumped. If no vector_index is found, the port is considered scalar (1 bit wide).

Concatenated ports shall appear in the extended VCD file as separate entries.

\section*{Example:}
module addbit (\{ A, b\} , ci, sum, co);
input \(A, b, C i ;\)
output sum, co;
. . .

The VCD file output looks like:
```

\$scope module addbit \$end
\$var port 1 <0 A \$end
\$var port 1 < | b \$end
\$var port 1 <2 ci \$end
\$enddefinitions \$end

```

\subsection*{18.4.3 Value changes}

The value change section of the VCD file is also affected by \$dumpports, as Syntax 18-29 shows.


Where the constructs are defined as:
p key character which indicates a port. There is no space between the \(\mathbf{p}\) and the port_value.
port_value
state character (described below).
0_strength_component one of the 8 Verilog strengths which indicates the strength 0 specification for the port.
1_strength_component one of the 8 Verilog strengths which indicates the strength1 specification for the port.

The Verilog strength values are (append keyword with 0 or 1 as appropriate for the strength component):
\(0 \quad\) highz
1 small

2 medium
3 weak
4 large
5 pull
6 strong
7 supply
identifier_code the integer preceded by the \(<\) character as defined in the \(\$\) var construct for the port.

\subsection*{18.4.3.1 State characters}

The following state information is listed in terms of input values from a test fixture, the output values of the device under test (DUT), and the states representing unknown direction:

INPUT (TESTFIXTURE)
\begin{tabular}{ll} 
D & low \\
U & high \\
N & unknown \\
Z & three-state \\
d & low (two or more drivers active) \\
u & high (two or more drivers active)
\end{tabular}

\section*{OUTPUT (DUT)}
\begin{tabular}{ll} 
L & low \\
H & high \\
X & unknown (don't care) \\
T & three-state \\
l & low (two or more drivers active) \\
h & high (two or more drivers active)
\end{tabular}

\section*{UNKNOWN DIRECTION}

0 low (both input and output are active with 0 value)
1 high (both input and output are active with 1 value)
? unknown
F three-state (input and output unconnected)
A unknown (input 0 and output 1)
a unknown (input 0 and output \(X\) )
B unknown (input 1 and output 0 )
b unknown (input 1 and output X )
C unknown (input \(X\) and output 0 )
C unknown (input \(X\) and output 1)
f unknown (input and output three-stated)

\subsection*{18.4.3.2 Drivers}

Where drivers are considered only in terms of primitives, continuous assignments, and procedural continuous assignments. Value \(0 / 1\) means both input and output are active with value \(0 / 1.0\) and 1 are conflict states. The following rules apply to conflicts:

If both input and output are driving the same value with the same range of strength, then this is a conflict. The resolved value is \(0 / 1\) and the strength is the stronger of the two.
If the input is driving a strong strength (range) and the output is driving a weak strength (range), the resolved value is \(d / u\) and the strength is the strength of the input.
If the input is driving a weak strength (range) and the output is driving a strong strength (range), then the resolved value is \(1 / \mathrm{h}\) and the strength is the strength of the output.

Where range is:
Strength supply 7 to 5 (large) - strong strength
Strength 4 to 1 - weak strength

\subsection*{18.4.4 Extended VCD file format example}

The following example illustrates the format of the extended VCD file.

A module declaration:
```

module adder(data0, data1, data2, data3, carry, as, rdn, reset,
test,
write);
inout data0, data1, data2, data3;
output carry;
input as, rdn, reset, test, write;

```

And the resulting VCD fragment:


\section*{19. Compiler directives}

All Verilog compiler directives are preceded by the (') character. This character is called accent grave. It is different from the character ( ), which is the single quote character. The scope of compiler directives extends from the point where it is processed, across all files processed, to the point where another compiler directive supersedes it or the processing completes.

This clause describes the following compiler directives:
\begin{tabular}{lr} 
'celldefine & {\([19.1]\)} \\
'default_nettype & {\([19.2]\)} \\
'define & {\([19.3]\)} \\
'else & {\([19.4]\)} \\
'elsif & {\([19.4]\)} \\
'endcelldefine & {\([19.1]\)} \\
'endif & {\([19.4]\)} \\
'ifdef & {\([19.4]\)} \\
'ifndef & {\([19.4]\)} \\
'include & {\([19.5]\)} \\
'line & {\([19.7]\)} \\
'nounconnected_drive & {\([19.9]\)} \\
'resetall & {\([19.6]\)} \\
'timescale & {\([19.8]\)} \\
'unconnected_drive & {\([19.9]\)} \\
'undef & {\([19.3]\)}
\end{tabular}

\section*{19.1 'celldefine and 'endcelldefine}

The directives 'celldefine and 'endcelldefine tag modules as cell modules. Cells are used by certain PLI routines for applications, such as delay calculations. It is advisable to pair each 'celldefine with an 'endcelldefine. More than one of these pairs may appear in a single source description.

These directives may appear anywhere in the source description, but it is recommended that the directives be specified outside the module definition.

The 'resetall directive includes the effects of a 'endcelldefine directive.

\section*{19.2 'default_nettype}

The directive 'default_nettype controls the net type created for implicit net declarations (see 3.5). It can be used only outside of module definitions. It affects all modules that follow the directive, even across source file boundaries. Multiple 'default_nettype directives are allowed. The latest occurrence of this directive in the source controls the type of nets that will be implicitly declared. Syntax 19-1 contains the syntax of the directive.
```

default_nettype_compiler_directive $::=$
'default_nettype net_type
net_type $::=$ wire $\mid$ tri $\mid$ tri0 $\mid$ wand $\mid$ triand $\mid$ wor $\mid$ trior $\mid$ trireg | none

```

Syntax 19-1 Syntax for default nettype compiler directive

When no 'default_nettype directive is present or if the 'resetall directive is specified, implicit nets are of type wire. When the 'default_nettype is set to none, all nets must be explicitly declared. If a net is not explicitly declared, an error is generated.

\section*{19.3 'define and 'undef}

A text macro substitution facility has been provided so that meaningful names can be used to represent commonly used pieces of text. For example, in the situation where a constant number is repetitively used throughout a description, a text macro would be useful in that only one place in the source description would need to be altered if the value of the constant needed to be changed.

The text macro facility is not affected by the compiler directive 'resetall.

\subsection*{19.3.1 define}

The directive 'define creates a macro for text substitution. This directive can be used both inside and outside module definitions. After a text macro is defined, it can be used in the source description by using the (') character, followed by the macro name. The compiler shall substitute the text of the macro for the string 'macro_name. All compiler directives shall be considered predefined macro names; it shall be illegal to redefine a compiler directive as a macro name.

A text macro can be defined with arguments. This allows the macro to be customized for each use individually.

The syntax for text macro definitions is given in Syntax 19-2.
```

text_macro_definition ::=
'define text_macro_name macro_text
text_macro_name ::=
text_macro_identifier [ ( list_of_formal_arguments ) ]
list_of_formal_arguments ::=
formal_argument_identifier \{, formal_argument_identifier \}
text_macro_identifier $::=$ (From Annex A - A.9.3)
simple_identifier

```

\section*{Syntax 19-2 Syntax for text macro definition}

The macro text can be any arbitrary text specified on the same line as the text macro name. If more than one line is necessary to specify the text, the newline shall be preceded by a backslash \((\backslash)\). The first newline not preceded by a backslash shall end the macro text. The newline preceded by a backslash shall be replaced in the expanded macro with a newline (but without the preceding backslash character).

When formal arguments are used to define a text macro, the scope of the formal argument shall extend up to the end of the macro text. A formal argument can be used in the macro text in the same manner as an identifier.

If a one-line comment (that is, a comment specified with the characters //) is included in the text, then the comment shall not become part of the substituted text. The macro text can be blank, in which case the text macro is defined to be empty, and no text is substituted when the macro is used.

The syntax for using a text macro is given in Syntax 19-3.
```

text_macro_usage ::=
'text_macro_identifier [( list_of_actual_arguments )]
list_of_actual_arguments ::=
actual_argument {, actual_argument }
actual_argument ::=
expression

```

\section*{Syntax 19-3 Syntax for text macro usage}

For a macro without arguments, the text shall be substituted as is for every occurrence of 'text_macro_name. However, a text macro with one or more arguments shall be expanded by substituting each formal argument with the expression used as the actual argument in the macro usage.

Once a text macro name has been defined, it can be used anywhere in a source description; that is, there are no scope restrictions. Text macros can be defined and used interactively. The text macro name shall be a simple identifier.

The text specified for macro text shall not be split across the following lexical tokens:
Comments
Numbers
Strings
Identifiers
Keywords
Operators

\section*{Examples:}
```

'define wordsize 8
reg [ 1:`wordsize] data;
//define a nand with variable delay
'define var_nand(dly) nand \#dly
'var_nand(2) g121 (q21, n10, n11);
'var_nand(5) g122 (q22, n10, n11);

```

The following is illegal syntax because it is split across a string:
```

'define first_half "start of string
\$display(`firs̄t_half end of string");

```

\section*{NOTES:}
1) Each actual argument is substituted for the corresponding formal argument literally. Therefore, when an expression is used as an actual argument, the expression will be substituted in its entirety. This may cause an expression to be evaluated more than once if the formal argument was used more than once in the macro text. For example,
```

'define max (a,b) ((a) > (b) ? (a) : (b))
n = 'max (p+q, r+s) ;

```
will expand as
```

    n = ((p+q) > (r+s)) ? (p+q) : (r+s) ;
    Here, the larger of the two expressions p + q and r + s will be
    evaluated
twice.

```
2) The word define is known as a compiler directive keyword, and it is not part of the normal set of keywords. Thus, normal identifiers in a Verilog HDL source description can be the same as compiler directive keywords (although this is not recommended). The following problems should be considered:
a) Text macro names may not be the same as compiler directive keywords.
b) Text macro names can re-use names being used as ordinary identifiers. For example, signal_name and signal_name are different.
c) Redefinition of text macros is allowed; the latest definition of a particular text macro read by the compiler prevails when the macro name is encountered in the source text.
3) The macro text can contain usages of other text macros. Such usages shall be substituted after the original macro is substituted, not when it is defined. It shall be an error for a macro to expand directly or indirectly to text containing another usage of itself (a recursive macro).

\subsection*{19.3.2 'undef}

The directive 'undef shall undefine a previously defined text macro. An attempt to undefine a text macro that was not previously defined using a 'define compiler directive can result in a warning. The syntax for 'undef compiler directive is given in Syntax 19-4.

\section*{Syntax 19-4 Syntax for undef compiler directive}

An undefined text macro has no value, just as if it had never been defined.

\section*{19.4 'ifdef, ‘else, ‘elsif, ‘endif, 'ifndef}

These conditional compilation compiler directives are used to include optionally lines of a Verilog HDL source description during compilation. The 'ifdef compiler directive checks for the definition of a text macro name. If the text macro name is defined, then the lines following the 'ifdef directive are included. If the text_macro_name is not defined and an 'else directive exists, then this source is compiled. The 'ifndef compiler directive checks for the definition of a text macro name. If the text macro name is not defined, then the lines following the 'ifndef directive are included. If the text_macro_name is defined and an 'else directive exists, then this source is compiled.

If the 'elsif directive exists (instead of the 'else) the compiler checks for the definition of the text macro name. If the name exists the lines following the 'elsif directive are included. The 'elsif directive is equivalent to the compiler directive sequence 'else 'ifdef ... 'endif. This directive does not need a corresponding 'endif directive. This directive must be preceded by an 'ifdef or 'ifndef directive.

These directives may appear anywhere in the source description.

Situations where the 'ifdef, 'else, 'elsif, 'endif, and 'ifndef compiler directives may be useful include:
Selecting different representations of a module such as behavioral, structural, or switch level Choosing different timing or structural information
Selecting different stimulus for a given run
The 'ifdef, 'else, 'elsif, 'endif, and 'ifndef compiler directives have the syntax shown in Syntax 19-5.
```

conditional_compilation_directive ::=
ifdef_directive
| ifndef_directive
ifdef_directive ::=
'ifdef text_macro_identifier
ifdef_group_of_lines
{ 'elsif text_macro_identifier elsif_group_of_lines }
[ 'else else_group_of_lines ]
'endif
ifndef_directive ::=
'ifndef text_macro_identifier
ifndef_group_of_lines
{'elsif text_macro_identifier elsif_group_of_lines }
[ 'else else_group_of_lines ]
'endif

```

\section*{Syntax 19-5 Syntax for conditional compilation directives}

The text_macro_identifier is a Verilog HDL simple_identifier. The ifdef_group_of_lines, ifndef_group_ōflines, elsif_group_of_lines and the else_group_of_lines are parts of a Verilog \(\overline{\mathrm{H}} \mathrm{L}\) - source description. The 'else and 'elsif compiler directives and all of the groups of lines are optional.

The 'ifdef, 'else, 'elsif, and 'endif compiler directives work together in the following manner:
When an 'ifdef is encountered, the ifdef text macro identifier is tested to see if it is defined as a text macro name using 'define within the Verilog HDL source description.
If the the ifdef text macro identifier is defined, the ifdef group of lines is compiled as part of the description and if there are 'else or 'elsif compiler directives, these compiler directives and corresponding groups of lines are ignored.
If the ifdef text macro identifier has not been defined, the ifdef group of lines is ignored.
If there is an 'elsif compiler directive, the elsif text macro identifier is tested to see if it is defined as a text macro name using 'define within the Verilog HDL source description.
If the elsif text macro identifier is defined, the elsif group of lines is compiled as part of the description and if there are other 'elsif or 'else compiler directives, the other 'elsif or 'else directives and corresponding groups of lines are ignored.
If the first elsif text macro identifier has not been defined, the first elsif group of lines is ignored.
If there are multiple 'elsif compiler directives, they are evaluated like the first 'elsif compiler directive in the order they are written in the Verilog HDL source description.
If there is an 'else compiler directive, the else group of lines is compiled as part of the description.

Although the names of compiler directives are contained in the same name space as text macro names, the names of compiler directives are considered not to be defined by 'ifdef, 'ifndef, and 'elsif.

The 'ifndef, 'else, 'elsif, and 'endif compiler directives work together in the following manner:
When an 'ifndef is encountered, the ifndef text macro identifier is tested to see if it is defined as a text macro name using 'define within the Verilog HDL source description.
If the ifndef text macro identifier is not defined, the ifndef group of lines is compiled as part of the description and if there are 'else or 'elsif compiler directives, these compiler directives and corresponding groups of lines are ignored.
If the ifndef text macro identifier is defined, the ifndef group of lines is ignored.
If there is an 'elsif compiler directive, the elsif text macro identifier is tested to see if it is defined as a text macro name using 'define within the Verilog HDL source description.
If the elsif text macro identifier is defined, the elsif group of lines is compiled as part of the description and if there are other 'elsif or 'else compiler directives, the other 'elsif or 'else directives and corresponding groups of lines are ignored.
If the first elsif text macro identifier has not been defined, the first elsif group of lines is ignored.
If there are multiple 'elsif compiler directives, they are evaluated like the first 'elsif compiler directive in the order they are written in the Verilog HDL source description.
If there is an 'else compiler directive, the else group of lines is compiled as part of the description.
Although the names of compiler directives are contained in the same name space as text macro names, the names of compiler directives are considered not to be defined by 'ifdef, 'ifndef, and 'elseif.

Nesting of 'ifdef, 'ifndef, 'else, 'elsif, and 'endif compiler directives shall be permitted.
NOTE Any group of lines that the compiler ignores still has to follow the Verilog HDL lexical conventions for white space, comments, numbers, strings, identifiers, keywords, and operators.

\section*{Examples:}

Example 1 The example below shows a simple usage of an 'ifdef directive for conditional compilation. If the identifier behavioral is defined, a continuous net assignment will be compiled in; otherwise, an and gate will be instantiated.
```

module and op ( $\mathrm{a}, \mathrm{b}, \mathrm{c}$ );
output $a$;
input $\mathrm{b}, \mathrm{c}$;
'ifdef behavioral
wire $\mathrm{a}=\mathrm{b}$ \& c ;
'else
and $\mathrm{al}(\mathrm{a}, \mathrm{b}, \mathrm{c})$;
'endif
endmodule

```

Example 2 The following example shows usage of nested conditional compilation directives.
```

module test(out);
output out;
'define wow
'define nest_one
'define second_nest
'define nest two
'ifdef wow
initial \$display("wow is defined");
'ifdef nest one
initial \$display("nest_one is defined");
'ifdef nest two
initial \$display("nest_two is defined");
'else
initial \$display("nest_two is not defined");
'endif
'else
initial \$display("nest_one is not defined");
'endif
'else
initial \$display("wow is not defined");
'ifdef second_nest
initial \$display("nest_two is defined");
'else
initial \$display("nest_two is not defined");
'endif
'endif
endmodule

```

Example 3 The following example shows usage of chained nested conditional compilation directives.
```

module test;
'ifdef first block
'ifndef second_nest
initial \$display("first_block is defined");
'else
initial \$display("first_block and second_nest defined");
'endif
'elsif second block
initial \$display("second_block defined, first_block is not");
'else
'ifndef last_result
initial \$display("first_block, second_block, last_result
not defined.");
'elsif real_last
initial \$display("first block, second block not defined,
last_result and real_last defined.");
'else
initial \$display("Only last_result defined!");
'endif
'endif
endmodule

```

\section*{19.5 'include}

The file inclusion ('include) compiler directive is used to insert the entire contents of a source file in another file during compilation. The result is as though the contents of the included source file appear in place of the 'include compiler directive. The 'include compiler directive can be used to include global or commonly used definitions and tasks without encapsulating repeated code within module boundaries.

Advantages of using the 'include compiler directive include the following:

Providing an integral part of configuration management
Improving the organization of Verilog HDL source descriptions
Facilitating the maintenance of Verilog HDL source descriptions
The syntax for the 'include compiler directive is given in Syntax 19-6.
```

include_compiler_directive ::=
'include "filename"

```

\section*{Syntax 19-6 Syntax for include compiler directive}

The compiler directive 'include can be specified anywhere within the Verilog HDL description. The filename is the name of the file to be included in the source file. The filename can be a full or relative path name.

Only white space or a comment may appear on the same line as the 'include compiler directive.
A file included in the source using the 'include compiler directive may contain other 'include compiler directives. The number of nesting levels for included files shall be finite.

\section*{Examples:}

Examples of legal comments for the 'include compiler directive are as follows:
```

'include "parts/count.v"
'include "fileB"
'include "fileB" // including fileB

```

NOTE Implementations may limit the maximum number of levels to which include files can be nested, but the limit shall be at least 15 .

\section*{19.6 'resetall}

When 'resetall compiler directive is encountered during compilation, all compiler directives are set to the default values. This is useful for ensuring that only those directives that are desired in compiling a particular source file are active.

The recommended usage is to place 'resetall at the beginning of each source text file, followed immediately by the directives desired in the file.

\section*{19.7 'line}

The compiler is expected to maintain the current line and the filename of the file being compiled. The line number ('line) compiler directive is used to reset the current line number and filename of the current file to the line number and filename presented. This can be used to reflect the location in an original file; if the actual source file has been modified by addition or reduction of lines. After specifying the new line number or file name, the compiler can correctly refer to the original source file location. For example error messages, source code debugging, etc. can direct the user to the actual original line.

The syntax for the 'line compiler directive is given in Syntax 19-7.
```

line_compiler_directive ::=
`line number "filename" level

```

\section*{Syntax 19-7 Syntax for line compiler directive}

The directive can be specified anywhere within the Verilog HDL source description. The number parameter is the new line number of the next line. The filename parameter is the new name of the file. The filename can be a full or relative path name. The level parameter indicates whether an include file has been entered (value is 1 ), an include file is exited (value is 2 ), or neither has been done (value is 0 ).

The results of this directive are not affected by the compiler directive 'resetall. As the compiler processes the remainder of the file and new files, the line number shall be incremented as each line is read and the filename shall be updated to the new current file being processed. When beginning to read include files, the current line and filename shall be stored for restoration at the termination of the include file. The updated line number and filename information shall be available for PLI access. The mechanism of library searching is not affected by the effects of the 'line compiler directive.

\section*{19.8 'timescale}

This directive specifies the time unit and time precision of the modules that follow it. The time unit is the unit of measurement for time values such as the simulation time and delay values.

To use modules with different time units in the same design, the following timescale constructs are useful:
The 'timescale compiler directive to specify the unit of measurement for time and precision of time in the modules in the design
The \$printtimescale system task to display the time unit and precision of a module
The \$time and \$realtime system functions, the \$timeformat system task, and the \%t format specification to specify how time information is reported

The 'timescale compiler directive specifies the unit of measurement for time and delay values and the degree of accuracy for delays in all modules that follow this directive until another 'timescale compiler directive is read. If there is no 'timescale specified or it has been reset by a 'resetall directive, the time unit and precision are simulator specific. It shall be an error if some modules have a 'timescale specified and others do not.

The syntax for the 'timescale directive is given in Syntax 19-8.

\section*{Syntax 19-8 Syntax for timescale compiler directive}

The time_unit argument specifies the unit of measurement for times and delays.
The time precision argument specifies how delay values are rounded before being used in simulation. The values used are accurate to within the unit of time specified here, even if there is a smaller time_precision argument elsewhere in the design. The smallest time_precision argument of all the 'timescale compiler directives in the design determines the precision of the time unit of the simulation.

The time_precision argument shall be at least as precise as the time_unit argument; it cannot specify a longer unit of time than time_unit.

The integers in these arguments specify an order of magnitude for the size of the value; the valid integers are 1,10 , and 100 . The character strings represent units of measurement; the valid character strings are \(\mathbf{s}, \mathbf{m s}\), us, ns, ps, and fs.

The units of measurement specified by these character strings are given in Table 86.
Table 86-Arguments of time_precision
\begin{tabular}{|l|l|}
\hline \begin{tabular}{c} 
Character \\
string
\end{tabular} & Unit of measurement \\
\hline s & seconds \\
\hline ms & milliseconds \\
\hline us & microseconds \\
\hline ns & nanoseconds \\
\hline ps & picoseconds \\
\hline fs & femtoseconds \\
\hline
\end{tabular}

\section*{Examples:}

The following example shows how this directive is used:
'timescale 1 ns / 1 ps
Here, all time values in the modules that follow the directive are multiples of 1 ns because the time_unit argument is 1 ns . Delays are rounded to real numbers with three decimal places or precise to within one thousandth of a nanosecond because the time_precision argument is 1 ps , or one thousandth of a nanosecond.

Consider the following example:
'timescale 10 us / 100 ns

The time values in the modules that follow this directive are multiples of 10 us because the time unit argument is 10 us. Delays are rounded to within one tenth of a microsecond because the time_precision argument is 100 ns , or one tenth of a microsecond.

The following example shows a 'timescale directive in the context of a module:
```

'timescale 10 ns / 1 ns
module test;
reg set;
parameter d = 1.55;
initial begin
\#d set = 0;
\#d set = 1;
end
endmodule

```

The 'timescale \(10 \mathrm{~ns} / 1 \mathrm{~ns}\) compiler directive specifies that the time unit for module test is 10 ns . As a result, the time values in the module are multiples of 10 ns , rounded to the nearest 1 ns and, therefore, the value stored in parameter \(d\) is scaled to a delay of 16 ns . This means that the value 0 is assigned to reg set at simulation time \(16 \mathrm{~ns}(1.6 \times 10 \mathrm{~ns})\), and the value 1 at simulation time 32 ns .

Parameter d retains its value no matter what timescale is in effect.

These simulation times are determined as follows:
a) The value of parameter \(d\) is rounded from 1.55 to 1.6 according to the time precision.
b) The time unit of the module is 10 ns , and the precision is 1 ns , so the delay of parameter d is scaled from 1.6 to 16 .
c) The assignment of 0 to reg set is scheduled at simulation time 16 ns and the assignment of 1 at simulation time 32 ns . The time values are not rounded when the assignments are scheduled.

\section*{19.9 'unconnected_drive and 'nounconnected_drive}

All unconnected input ports of a module appearing between the directives 'unconnected_drive and 'nounconnected_drive are pulled up or pulled down instead of the normal default.

The directive 'unconnected_drive takes one of two arguments pull1 or pull0. When pull1 is specified, all unconnected input ports are automatically pulled up. When pull0 is specified, unconnected ports are pulled down. These directives shall be specified in pairs, and outside of the module declarations.

The 'resetall directive includes the effects of a 'endcelldefine directive.

\section*{20. PLI overview}

\subsection*{20.1 LI purpose and history (informative)}

Clause 20 through Clause 27 and Annex E through Annex G describe the C language procedural interface standard and interface mechanisms that are part of the Verilog HDL. This procedural interface, known as the Programming Language Interface, or PLI, provides a means for Verilog HDL users to access and modify data in an instantiated Verilog HDL data structure dynamically. An instantiated Verilog HDL data structure is the result of compiling Verilog HDL source descriptions and generating the hierarchy modeled by module instances, primitive instances, and other Verilog HDL constructs that represent scope. The PLI procedural interface provides a library of C language functions that can directly access data within an instantiated Verilog HDL data structure.

A few of the many possible applications for the PLI procedural interface are:
C language delay calculators for Verilog model libraries that can dynamically scan the data structure of a Verilog software product and then dynamically modify the delays of each instance of models from the library
C language applications that dynamically read test vectors or other data from a file and pass the data into a Verilog software product
Custom graphical waveform and debugging environments for Verilog software products
Source code decompilers that can generate Verilog HDL source code from the compiled data structure of a Verilog software product
Simulation models written in the C language and dynamically linked into Verilog HDL simulations
Interfaces to actual hardware, such as a hardware modeler, that dynamically interact with simulations
This document standardizes the Verilog PLI that has been in use since the mid-1980s. This standard comprises three primary generations of the Verilog PLI.
a) Task/function routines, called \(T F\) routines, make up the first generation of the PLI. These routines, most of which start with the characters \(\mathbf{t f}_{-}\), are primarily used for operations involving user-defined task/function arguments, along with utility functions, such as setting up call-back mechanisms and writing data to output devices. The TF routines are sometimes referred to as utility routines
b) Access routines, called \(A C C\) routines, form the second generation of the PLI. These routines, which all start with the characters acc_, provide an object-oriented access directly into a Verilog HDL structural description. ACC routines are used to access and modify information, such as delay values and logic values on a wide variety of objects that exist in a Verilog HDL description. There is some overlap in functionality between ACC routines and TF routines.
c) Verilog Procedural Interface routines, called VPI routines, are the third generation of the PLI. These routines, all of which start with the characters vpi_, provide an object-oriented access for both Verilog HDL structural and behavioral objects. The VPI routines are a superset of the functionality of the TF routines and ACC routines.

\subsection*{20.2 User-defined system task or function names}

A user-defined system task or function name is the name that will be used within a Verilog HDL source file to invoke specific PLI applications. The name shall adhere to the following rules:

The first character of the name shall be the dollar sign character (\$)
The remaining characters shall be letters, digits, the underscore character ( \(\_\)) or the dollar character (\$)
Uppercase and lowercase letters shall be considered to be unique the name is case sensitive
The name can be any size, and all characters are significant

\subsection*{20.3 User-defined system task or function types}

The type of a user-defined system task or function determines how a PLI application is called from the Verilog HDL source code. The types are:

A user task can be used in the same places a Verilog HDL task can be used (refer to 10.2). A userdefined system task can read and modify the arguments of the task, but does not return any value.
A user function can be used in the same places a Verilog HDL function can be used (refer to 10.3). A user-defined system function can read and modify the arguments of the function, and shall return a scalar or vector value. The bit width of the return value shall be determined by a user-supplied sizetf application (see 21.1.1).
A user real-function can be used in the same places a Verilog HDL function can be used (refer to 10.3). A user-defined system real-function can read and modify the arguments of the function, and will return a double-precision floating point value.

\section*{20.4 verriding built-in system task and function names}

Clause 17 defines a number of built-in system tasks and functions that are part of the Verilog language. In addition, software products can include other built-in system tasks and functions specific to the product. These built-in system task and function names begin with the dollar sign character (\$) just as user-defined system task and function names.

If a user-provided PLI application is associated with the same name as a built-in system task or function (using the PLI interface mechanism), the user-provided C application shall override the built-in system task/ function, replacing its functionality with that of the user-provided C application. For example, a user could write a random number generator as a PLI application and then associate the application with the name \$random, thereby overriding the built-in \$random function with the user s application.

Verilog timing checks, such as \$setup, are not system tasks, and cannot be overridden.
The system functions \$signed and \$unsigned can be overridden. These system functions are unique in the Verilog HDL, in that the return width is based on the width of their argument. If overridden, the PLI version shall have the same return width for all instances of the system function. The PLI return width is defined by the PLI sizetf routine.

\subsection*{20.5 User-supplied LI applications}

User-supplied PLI applications are C language functions that utilize the library of PLI C functions to access and interact dynamically with Verilog HDL software implementations as the Verilog HDL source code is executed.

These PLI applications are not independent C programs. They are C functions, which are linked into a software product, and become part of the product. This allows the PLI application to be called when the userdefined system task or function \(\$\) name is compiled or executed in the Verilog HDL source code.

\subsection*{20.6 LI interface mechanism}

The PLI interface mechanism provides a means to have PLI applications called for various reasons when the associated system task or function \$ name is encountered in the Verilog HDL source description. For example, when a Verilog HDL simulator first compiles the Verilog HDL source description, a specific PLI application can be called that performs syntax checking to ensure the user-defined system task or function is being used correctly. Then, as simulation is executing, a different PLI application can be called to perform the operations required by the PLI application. Other PLI applications can be automatically called by the simulator for miscellaneous reasons, such as the end of a simulation time step or a logic value change on a specific signal.

The PLI provides two interface mechanisms:

The TF and ACC interface mechanism is an older interface, which can be used to associate PLI applications which use routines from the ACC and TF function libraries. This interface mechanism is described in Clause 21.
The VPI interface mechanism is a newer interface, which can be used to associate PLI applications which use routines from the VPI function libraries. This interface mechanism is described in Clause 26.

Instances of system tasks and functions which are defined using the TF and ACC interface mechanism can only be accessed using the TF and ACC function libraries. Instances of system tasks and functions which are defined using the VPI interface mechanism can only be accessed using the VPI function library.

\subsection*{20.7 User-defined system task and function arguments}

When a user-defined system task or function is used in a Verilog HDL source file, it can have arguments that can be used by the PLI applications associated with the system task or function. In the following example, the user-defined system task \$get_vector has two arguments:
```

\$get_vector("test_vector.pat", input_bus);

```

The arguments to a system task or function are referred to as task/function arguments (often abbreviated as tfargs). These arguments are not the same as C language arguments. When the PLI applications associated with a user-defined system task or function are called, the task/function arguments are not passed to the PLI application. Instead, a number of PLI routines are provided that allow the PLI applications to read and write to the task/function arguments. Refer to the sections on ACC routines, TF routines and VPI routines for information on specific routines that work with task/function arguments.

\subsection*{20.8 LI include files}

The libraries of PLI functions are defined in C include files, which are a normative part of the 1364 standard. These files also define constants, structures, and other data used by the library of PLI routines and the interface mechanisms. The files are acc_user.h (listed in Annex E), veriuser.h (listed in Annex F) and vpi_user.h (listed in Annex G).

> PLI applications that use the TF routines shall include the file veriuser.h.
> PLI applications that use the ACC routines shall include the file acc_user.h.
> PLI applications that use the VPI routines shall include the file vpi_user.h.

\subsection*{20.9 LI Memory estrictions}

Memory allocated by the PLI routines is not to be modified by the user, with the exception of the value storage returned by the PLI routines \(t f\) _exprinfo() and \(t f\) _nodeinfo(), as defined in 25.15 and 25.35.

\section*{21. LI TF and ACC interface mechanism}

The interface mechanism described in this section provides a means for users to link applications based on PLI task/function (TF) routines and access (ACC) routines to Verilog software products. Through the interface mechanism, a user can:

Specify a user-defined system task or function name that can be included in Verilog HDL source descriptions; the user-defined system task or function name shall begin with a dollar sign (\$), such as \$get_vector
Provide one or more PLI C applications to be called by a software product (such as a logic simulator)
Define which PLI C applications are to be called and when the applications should be called when the user-defined system task or function name is encountered in the Verilog HDL source description
Define whether the PLI applications should be treated as functions (which return a value) or tasks (analogous to subroutines in other programming languages)
Define a data argument to be passed to the PLI applications each time they are called
NOTE The PLI interface mechanism described in this section does not apply to applications that use the Verilog Procedural Interface (VPI) routines; these routines use the VPI registry mechanism described in Clause 26 and Clause 27.

\subsection*{21.1 User-supplied LI applications}

User-supplied PLI applications are C language functions that utilize the library of PLI C functions to access and interact dynamically with Verilog HDL software implementations as the Verilog HDL source code is executed.

These PLI applications are not independent C programs. They are C functions, which are linked into a software product, and become part of the product. This allows the PLI application to be called when the userdefined system task or function \(\$\) name is compiled or executed in the Verilog HDL source code.

The PLI interface mechanism for TF and ACC routines provides five classes of user-supplied PLI applications: checktf applications, sizetf applications, calltf applications, misctf applications, and consumer applications. The sizetf, checktf, calltf, and misctf routines are called during specific periods during processing. The purpose of each of the PLI application classes is explained in the following subsections.

\subsection*{21.1.1 The sizetf class of LI applications}

A sizetf PLI application can be used in conjunction with user-defined system functions. A function shall return a value, and software products that execute the system function may need to determine how many bits wide that return value shall be. The sizetf application may be called early in the process, prior to a complete instantiation of the design. As a result, access to objects may be limited at this time. Each sizetf function shall be called at most once. It shall be called if its associated system function appears in the design. The value returned by the sizetf function shall be the number of bits that the calltf routine shall provide as the return value for the system function. If no sizetf application is specified for a user-defined system function, the function shall return 32-bits. The sizetf application shall not be called for user-defined system tasks or real-functions.

\subsection*{21.1.2 The checktf class of LI applications}

A checktf PLI application shall be called when the user-defined system task or function name is encountered during parsing or compiling the Verilog HDL source code. This application is typically used to check the correctness of any arguments used with the system task in the Verilog HDL source code. The checktf PLI application shall be called one time for each instance of a system task or function in the source description. Providing a checktf application is optional, but it is recommended that any arguments used with the system task or function be checked for correctness to avoid problems when the calltf or other PLI applications read and perform operations on the arguments. The checktf shall be called at the earliest possible time after all
simulation data structures required by the PLI are available. Generally this means after the design is fully instantiated, but no simulation events have occurred. By the time the checktf application is called, all PLI routines can be used without concern for the state of the process, with the exception of setting the return value of user defined system functions. The return value of a user defined system function can only be set during a calltf application.

\subsection*{21.1.3 The calltf class of LI applications}

A calltf PLI application shall be called each time the associated user-defined system task or function is executed within the Verilog HDL source code. For example, the following Verilog loop would call the PLI calltf application that is associated with the \$get_vector user-defined system task name 1024 times:
```

for (i = 1; i <= 1024; i = i + 1)
@(posedge clk) \$get_vector("test_vector.pat", input_bus);

```

In this example, the user-supplied PLI calltf application might read a test vector from a file called test_vector.pat (the first task/function argument), perhaps manipulate the vector to put it in a proper format for Verilog, and then assign the vector value to the second task/function argument called input_bus.

\subsection*{21.1.4 The misctf class of LI applications}

A misctf PLI application shall be called by a Verilog software product for miscellaneous reasons while the Verilog HDL source description is being executed. Among these reasons can be the end of a simulation time step, a logic value change on a user-defined system task/function argument, or the execution of the \$stop and \$finish built-in system tasks. When the software product calls the misctf PLI application, it shall pass in a reason argument, which can be used within the misctf application to determine why the application was called. The reason argument shall be a predefined integer constant. Table 87 and Table 88 list the reasons for which the misctf application can be called.

For most reasons, the misctf routine will not be called until the instance of the system task has been executed (at which point the calltf routine is called). The following reasons are exceptions, and will be called for each instance of the system task in the design regardless of whether or not it has been executed:
```

reason_endofcompile
reason_save
reason_startofsave
reason_restart
reason_endofreset
reason_reset

```

\subsection*{21.1.5 The consumer class of LI applications}

A consumer PLI application shall be called through a PLI callback mechanism referred to as the Value Change Link (VCL). Using the VCL, another PLI application, typically the calltf application, can place VCL flags on objects within the Verilog HDL data structure, such as a specific net. Whenever an object with a VCL flag changes value during a simulation, the consumer PLI application shall be called and passed information about the change.

\subsection*{21.2 Associating LI applications to a class and system task/function name}

Each user-provided PLI application is a standard C language function that makes use of the library of PLI functions. These user-provided PLI applications shall be associated with both the class of application (such as calltf or checktf) and the user-defined system task or function \$ name. In addition, the user-defined name shall be declared as either a system task or a system function.

For the TF and ACC interface mechanism, the method of associating PLI applications with a class and system task/function name is not defined as part of this standard. Each software product vendor shall define an association mechanism specific to their product. Refer to the documentation provided by the vendor for instructions on associating PLI applications to classes and system task/function names and then linking the PLI applications into the software products of the vendor.

\subsection*{21.3 LI application arguments}

When the calltf, checktf, and sizetf PLI applications are called by a Verilog software implementation, they shall be passed two C arguments, data and reason, in that order. When the misctf application is called, it shall be passed three C arguments, data, reason, and paramvc, in that order. These arguments are defined in more detail in the following subsections.

\subsection*{21.3.1 The data C argument}

The data C argument shall be an integer value. The value is defined by the user at the time the PLI applications are associated with a user-defined system task/function name. This value can be used to allow several different system task/function names to use the same calltf, checktf, sizetf, or misctf applications. To do this, each system task/function name would be associated with the same PLI applications, but each would have a different value for the user-defined data argument. When a PLI application is called, it can then check the value of the data argument to determine which system task/function name was used to call the application.

\subsection*{21.3.2 The reason \(C\) argument}

The reason C argument shall be a predefined integer constant that is passed to the calltf, checktf, sizetf, and misctf applications each time the applications are called. Generally, the calltf, checktf, and sizetf applications do not need to check the reason argument, since these applications can only be called under specific circumstances. The misctf application, however, can be called for a wide variety of reasons, and therefore it should always examine the reason argument to determine why the application was called. The value for the reason argument is defined in the PLI include file veriuser.h. The reason constant that is passed is based on the class of the PLI application, as follows:

The calltf application is passed the reason constant reason_calltf.
The checktf application is passed the reason constant reason_checktf.
The sizetf application is passed the reason constant reason_sizetf.
The misctf application is passed one of the constants listed in Table 87. Software implementations can define additional reason constants to be passed to the misctf application. Table 88 lists some common reason constants that can be available in some software implementations.

Table 87-(normative) Predefined misctf reason constants
\begin{tabular}{|c|l|}
\hline Integer constant & \multicolumn{1}{c|}{ Reason } \\
\hline reason_endofcompile & end of Verilog source compilation/start of execution \\
\hline reason_paramvc & \begin{tabular}{l} 
a change of value on a user-defined system task or function \\
argument
\end{tabular} \\
\hline reason_synch & end of a time step flagged by tf_synchronize() \\
\hline reason_rosynch & end of a time step flagged by tf_rosynchronize() \\
\hline reason_reactiviate & a simulation event scheduled by tf_setdelay() \\
\hline reason_finish & the \(\$\) finish() built-in system task executed \\
\hline
\end{tabular}

Table 88-(informative) Additional misctf reason constants
\begin{tabular}{|c|l|}
\hline Integer constant & \multicolumn{1}{c|}{ Reason } \\
\hline reason_paramdre & \begin{tabular}{l} 
a value change on the driver of a user-defined system task or \\
function argument
\end{tabular} \\
\hline reason_force & \begin{tabular}{l} 
execution of a procedural force or procedural continuous \\
assignment on any net, reg, integer variable, time variable or real \\
variable
\end{tabular} \\
\hline reason_release & \begin{tabular}{l} 
execution of a procedural release or procedural deassign on any \\
net, reg, integer variable, time variable or real variable
\end{tabular} \\
\hline reason_disable & \begin{tabular}{l} 
execution of a procedural disable statement
\end{tabular} \\
\hline reason_interactive & execution of the \$stop() built-in system task \\
\hline reason_scope & execution of the \$scope() built-in system task \\
\hline reason_startofsave & start of execution of the \$save() built-in system task \\
\hline reason_save & completion of execution of the \$save() built-in system task \\
\hline reason_restart & execution of the \$restart() built-in system task \\
\hline reason_reset & start of execution of the \$reset() built-in system task \\
\hline reason_endofreset & completion of execution of the \$reset() built-in system task \\
\hline
\end{tabular}

\subsection*{21.3.3 The paramvc C argument}

The paramvc C argument shall be an integer value passed to the misctf application. The value of paramve shall indicate which task/function argument changed value when the misctf application was called back after activating the utility routine \(\mathbf{t f}\) _asynchon(). This routine shall cause the misctf application to be called with a reason argument of reason_paramve or reason_paramdrc. Task/function argument index numbering shall proceed from left to right, with the left-most argument being number 1.

\section*{22. Using ACC routines}

This clause presents a general discussion of how and why to use PLI ACC routines. Clause 23 defines the ACC routine syntax, listed in alphabetical order.

\subsection*{22.1 ACC routine definition}

ACC routines are \(C\) programming language functions that provide procedural access to information within the Verilog HDL.

ACC routines perform one of two functions:
a) Read data about particular objects in the Verilog HDL description directly from internal data structures.
b) Write new information about certain objects in the Verilog HDL description into the internal data structures.

ACC routines shall read information about the following objects:
Module instances
Module ports
Module or data paths
Intermodule paths
Top-level modules
Primitive instances
Primitive terminals
Nets
Reg variables
Parameters
Specparams
Timing checks
Named events
Integer, real, and time variables
ACC routines shall read and write information on the following objects:
Intermodule path delays
Module path delays
Module input port delays (MIPDs)
Primitive instance delays
Timing check limits
Reg variable logic values
Net logic values (force/release only)
Integer, real and time variable values
Sequential UDP logic values

\subsection*{22.2 The handle data type}

A handle is a predefined data type that is a pointer to a specific object in the design hierarchy. Each handle conveys information to ACC routines about a unique instance of an accessible object information about the type of the object, plus how and where to find data about the object.

Most ACC routines require a handle argument to indicate the objects about which they need to read or write information. The PLI provides two categories of ACC routines that return handles for objects: handle routines, which begin with the prefix acc_handle_, and next routines, which begin with the prefix acc_next_. Refer to 22.4.2 for a discussion of handle routines and 22.4.3 for more information about next routines.

Handles shall be passed to and from ACC routines through handle variables. To declare a handle variable, the keyword handle (all lowercase) shall be used, followed by the variable name, as in this example:
handle net_handle;

After declaring a handle variable, it can be passed to any ACC routine that requires a handle argument or be used to receive a handle returned by an ACC routine. The following C language code fragment uses the variable net handle to store the handle returned by the ACC routine acc_handle_object():
```

handle net_handle;
net_handle = acc_handle_object("top.mod1.w3");

```

\subsection*{22.3 Using ACC routines}

\subsection*{22.3.1 Header files}

The header file acc_user. h shall be included in any C language source file containing an application program that calls ACC routines. The acc_user. h file is listed in Annex E.

\subsection*{22.3.2 Initializing ACC routines}

The ACC routine acc_initialize() shall initialize the environment for ACC routines and shall be called from the C language application program before the program invokes any other ACC routines.

\subsection*{22.3.3 Exiting ACC routines}

Before exiting a C language application program that calls ACC routines, the ACC routine acc_close() should be called. This routine shall reset ACC routine configuration parameters back to their defaults, and it shall also free memory allocated by the ACC routines.

\subsection*{22.4 List of ACC routines by major category}

The ACC routines are divided into the following major categories:
Fetch routines
Handle routines
Next routines
Modify routines
\(V C L\) routines
Miscellaneous routines
This section contains a summary list of each major category. The ACC routines sorted by the types of objects they work with are listed in 22.5 . Clause 23 presents an alphabetical list of all ACC routines, with their functions, syntax, and usage.

\subsection*{22.4.1 Fetch routines}

Fetch routines shall return a variety of information about different objects in the design hierarchy. The name of each routine begins with the prefix acc_fetch_ and indicates the type of information desired. For example, acc_fetch_fullname() retrieves the full hierarchical path name for any named object, while acc_fetch_paramval() retrieves the value of a parameter or specparam.

Table 89-List of fetch routines
\begin{tabular}{|c|c|}
\hline ACC routine & Description \\
\hline acc_fetch_arge() & Get the number of invocation command line arguments \\
\hline acc_fetch_argv() & Get the invocation command line arguments \\
\hline acc_fetch_attribute() & Get the value of a Verilog parameter or specparam as a double \\
\hline acc_fetch_attribute_int() & Get the value of a Verilog parameter or specparam as an integer \\
\hline acc_fetch_attribute_str() & Get the value of a Verilog parameter or specparam as a string \\
\hline acc_fetch_defname() & Get the definition name of a module or primitive \\
\hline acc_fetch_delay_mode() & Get the delay mode of a module instance \\
\hline acc_fetch_delays() & Get the existing delays for a primitive, module path, timing check, intermodule path, or module input port \\
\hline acc_fetch_direction() & Get the direction of a module port or primitive terminal \\
\hline acc_fetch_edge() & Get the edge specifier of a module path input terminal \\
\hline acc_fetch_fullname() & Get the full hierarchical name of an object \\
\hline acc_fetch_fulltype() & Get the full type description of an object as a predefined integer constant \\
\hline acc_fetch_index() & Get the index number of a port or terminal \\
\hline acc_fetch_itfarg() & Get the value of an instance of a system task/function argument as a double \\
\hline acc_fetch_itfarg_int() & Get the value of an instance of a system task/function argument as an integer \\
\hline acc_fetch_itfarg_str) & Get the value of an instance of a system task/function argument as a string \\
\hline acc_fetch_location() & Get the location of an object in a Verilog source file \\
\hline acc_fetch_name() & Get the local name of an object \\
\hline acc_fetch_paramtype() & Get the data type of a parameter or specparam \\
\hline acc_fetch_paramval() & Get the value of a parameter or specparam \\
\hline acc_fetch_polarity() & Get the polarity of a module path or data path \\
\hline acc_fetch_precision() & Get the simulation time precision \\
\hline acc_fetch_pulsere() & Get the current pulse handling values of a module path, intermodule path or module input port \\
\hline acc_fetch_range() & Get the range of a vector \\
\hline acc_fetch_size() & Get the bit size of a vector or port \\
\hline acc_fetch_tfarg() & Get the value of a system task/function argument as a double \\
\hline acc_fetch_tfarg_int() & Get the value of a system task/function argument as an integer \\
\hline acc_fetch_tfarg_str() & Get the value of a system task/function argument as a string \\
\hline acc_fetch_timescale_info() & Get the timescale information for an object \\
\hline acc_fetch_type() & Get the general type classification of an object as an integer constant \\
\hline acc_fetch_type_str) & Get the string representation of a type or fulltype integer constant \\
\hline acc_fetch_value() & Get the logic or strength value of a net, reg, integer variable, time variable or real variable \\
\hline
\end{tabular}

\subsection*{22.4.2 Handle routines}

Handle routines can return handles to a variety of objects in the design hierarchy. The name of each routine begins with the prefix acc_handle_ and indicates the type of handle desired. For example, acc_handle_object() retrieves a handle for a named object, while acc_handle_conn() retrieves a handle for a net connected to a particular terminal. Each handle routine shall return a handle to an object. This handle can, in turn, be passed as an argument to other ACC routines.

Table 90-List of handle routines
\begin{tabular}{|c|c|}
\hline ACC routine & Description \\
\hline acc_handle_by_name() & Get the handle to any named object \\
\hline acc_handle_condition() & Get the handle to the condition of a module path, data path, or timing check \\
\hline acc_handle_conn() & Get the handle to the net connected to a primitive, path, or timing check terminal \\
\hline acc_handle_datapath() & Get the handle to a data path \\
\hline acc_handle_hiconn() & Get the handle to the hierarchically higher net connected to a module port bit \\
\hline acc_handle_interactive_scope() & Get the handle to the current simulation interactive scope \\
\hline acc_handle_itfarg() & Get the handle to an argument of a specific system task/function instance \\
\hline acc_handle_loconn() & Get the handle to the hierarchically lower net connected to a module port bit \\
\hline acc_handle_modpath() & Get the handle to a module path \\
\hline acc_handle_notifier() & Get the handle to the notifier argument of a timing check \\
\hline acc_handle_object() & Get the handle to any named object \\
\hline acc_handle_parent() & Get the handle to the parent of an object \\
\hline acc_handle_path() & Get the handle to an intermodule path \\
\hline acc_handle_pathin() & Get the handle to the first net connected to a module path source \\
\hline acc_handle_pathout() & Get the handle to the first net connected to a module path destination \\
\hline acc_handle_port() & Get the handle to a module port based on the port index \\
\hline acc_handle_scope() & Get the handle to the scope containing an object \\
\hline acc_handle_simulated_net() & Get the handle to the net associated with a collapsed net \\
\hline acc_handle_tchk() & Get the handle to a timing check \\
\hline acc_handle_tchkarg10 & Get the handle to the first argument of a timing check \\
\hline acc_handle_tchkarg20 & Get the handle to the second argument of a timing check \\
\hline acc_handle_terminal() & Get the handle to a terminal of a primitive based on the terminal index \\
\hline acc_handle_tfarg() & Get the handle to the object named in a system task/function argument \\
\hline acc_handle_tfinst() & Get the handle to the current instance of a system task/function \\
\hline
\end{tabular}

\subsection*{22.4.3 Next routines}

When used inside a C loop construct, next routines shall find each object of a given type that is related to a particular reference object in the design hierarchy. The name of each routine begins with the prefix acc_next_ and indicates the type of object desired, known as the target object. For example, acc_next_net() retrieves each net in a module, while acc_next_driver() retrieves each terminal driving a net. Each call to a next routine returns a handle to the object it finds.

Most next routines require two arguments:
The first argument shall be a handle to a reference object.
The second argument shall be a handle that indicates whether to retrieve the first or next target object.

The reference object shall indicate where the next routine shall look for the target object. The target object is the type of object to be returned by a next routine.

Table 91 summarizes how next routines shall find each target object associated with a given reference object.

Table 91-How next routines use the target object argument
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ When } & \multicolumn{1}{c|}{ A next routine shall return } \\
\hline the target object is null & a handle to the first target object related to the reference object \\
\hline the target object is a handle to the last target object returned & a handle to the next target object related to the reference object \\
\hline no target objects remain for the reference object & a null handle \\
\hline no target objects are found initially for the reference object & a null handle \\
\hline an error occurs & a null handle \\
\hline
\end{tabular}

Each call to a next routine shall return only one handle. Therefore, to retrieve all target objects for a particular reference object, the following process can be used:
a) Chose an appropriate ACC routine to retrieve the handle of the desired reference object.
b) Set the target object handle variable to null. When a next routine is called with a null target handle, it shall return the first target associated with the reference.
c) Call the next routine, assigning the return value to the same variable as the target object argument. This automatically updates the target object argument to point to the last object found.
d) Place the next routine call inside a C while loop that terminates when the loop control value is null. When a next routine cannot access any more target objects, it shall return a null.

NOTE Most next routines can return objects in an arbitrary order. However, certain next routines shall return objects in a defined order, as noted in the description of the routine in Clause 23.

The following example, display_net_names, uses a next routine to display the names of all nets in a module.
```

\#include "acc_user.h"
display_net_names()
{
handle module_handle;
handle net_handle;
/*initialize environment for access routines*/
acc_initialize();
/*get handle for module* /
module_handle = acc_handle_tfarg(1);
/*display names of all nets in the module*/
net_handle = null;
whi\overline{le( net_handle = acc_next_net( module_handle, net_handle ) )}
io_printf("Net name ís: %\overline{s}\n", acc_fe\overline{tch_fullname(net_handle) );}
acc_close();
}

```

Table 92-List of next routines
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ ACC routine } & \\
\hline acc_next() & Get handles to all objects of a set of types \\
\hline acc_next_bit() & Get handles to all bits of a port or vector \\
\hline acc_next_cell() & Get handles to all cell modules in the current hierarchy and below \\
\hline acc_next_cell_load() & Get handles to all cell loads on a net \\
\hline acc_next_child() & Get handles to all module instances within a module \\
\hline acc_next_driver() & Get handles to all primitive terminals that drive a net \\
\hline acc_next_hiconn() & Get handles to all nets connected hierarchically higher to a module port \\
\hline acc_next_input() & Get handles to all input terminals of a module path or data path \\
\hline acc_next_load() & Get handles to all primitive terminals driven by a net \\
\hline acc_next_loconn() & Get handles to all nets connected hierarchically lower to a module port \\
\hline acc_next_modpath() & Get handles to all paths in a module \\
\hline acc_next_net() & Get handles to all nets in a module \\
\hline acc_next_output() & Get handles to all output terminals of a module path or data path \\
\hline acc_next_parameter() & Get handles to all parameters in a module \\
\hline acc_next_port() & Get handles to all ports of a module or connected to a net \\
\hline acc_next_portout() & Get handles to all output ports of a module \\
\hline acc_next_primitive() & Get handles to all primitive instances in a module \\
\hline acc_next_scope() & Get handles to all hierarchy scopes within a scope \\
\hline acc_next_specparam() & Get handles to all specify block parameters in a module \\
\hline acc_next_tchk() & Get handles to all timing checks in a module \\
\hline acc_next_terminal() & Get handles to all terminals of a primitive \\
\hline acc_next_topmod() & Get handles to all top-level modules \\
\hline & \\
\hline
\end{tabular}

\subsection*{22.4.4 Modify routines}

Modify routines shall alter the values of a variety of objects in the design hierarchy. Table 93 lists the types of values that shall be modified for particular objects.

Table 93-Values that can be modified
\begin{tabular}{|c|c|}
\hline Modify routines alter & For these objects \\
\hline Delay values & Primitives \\
& Module paths \\
& Intermodule paths \\
Module input ports \\
Timing checks \\
\hline Logic values & Variable data types \\
& Net data types \\
& Sequential UDPs \\
\hline Pulse handling values & Module paths \\
& Intermodule paths \\
& Module input ports \\
\hline
\end{tabular}

Table 94—List of modify routines
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ ACC routine } & \multicolumn{1}{c|}{ Description } \\
\hline acc_append_delays() & \begin{tabular}{l} 
Add delays to existing delays on primitives, module paths, timing checks, \\
intermodule paths, and module input ports
\end{tabular} \\
\hline acc_append_pulsere() & \begin{tabular}{l} 
Add to existing pulse control values of module paths, intermodule paths and \\
module input ports
\end{tabular} \\
\hline acc_replace_delays() & \begin{tabular}{l} 
Replace existing delays on primitives, module paths, timing checks, inter- \\
module paths and module input ports
\end{tabular} \\
\hline acc_replace_pulsere() & \begin{tabular}{l} 
Replace existing values on pulse control values of module paths, intermod- \\
ule paths and module input ports
\end{tabular} \\
\hline acc_set_pulsere() & \begin{tabular}{l} 
Set the pulse control values for a module path, intermodule path or module \\
input port as a percentage of the delay
\end{tabular} \\
\hline acc_set_value() & \begin{tabular}{l} 
Set and propagate a logic value onto a reg, integer variable, time variable, \\
real variable or sequential UDP; \\
continuously assign/deassign a reg or variable; \\
force/release a net or reg or variable
\end{tabular} \\
\hline
\end{tabular}

More details on using the acc_append_delays() and acc_replace_delays() ACC routines are provided in 22.8.

\subsection*{22.4.5 Miscellaneous routines}

Miscellaneous routines shall perform a variety of operations, such as initializing and configuring the ACC routine environment.

Table 95-List of miscellaneous routines
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ ACC routine } & \multicolumn{1}{c|}{ Description } \\
\hline acc_close() & Close ACC routine environment \\
\hline acc_collect() & Collect an array of handles for a reference object \\
\hline acc_compare_handles() & Determine if two handles are for the same object \\
\hline acc_configure() & Set the ACC routine environment parameters \\
\hline acc_count() & Count the number of objects related to a reference object \\
\hline acc_free() & Free up memory allocated by acc_collect() \\
\hline acc_initialize() & Initialize the ACC routine environment \\
\hline acc_object_in_typelist() & Determine if an object matches a set of types, fulltypes, or special properties \\
\hline acc_object_of_type() & Determine if an object matches a specific type, fulltype, or special property \\
\hline acc_product_type() & Get the type of software product being used \\
\hline acc_product_version() & Get the version of software product being used \\
\hline acc_release_object() & Release memory allocated by acc_next_input() or acc_next_output() \\
\hline acc_reset_buffer() & Reset the string buffer \\
\hline acc_set_interactive_scope() & Set the interactive scope of a software implementation \\
\hline acc_set_scope() & Set the scope used by acc_handle_object() \\
\hline acc_version() & Get the version of the ACC routines being used \\
\hline
\end{tabular}

\subsection*{22.4.6 VCL routines}

The VCL shall allow a PLI application to monitor simulation value changes of selected objects. It consists of two ACC routines that instruct a Verilog simulator to start or stop informing an application when an object changes value. How the VCL routine is used is discussed in 22.10.

Table 96-List of VCL routines
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ ACC routine } & \multicolumn{1}{c|}{ Description } \\
\hline acc_vel_add () & Add a value change callback on an object \\
\hline acc_vel_delete () & Remove a value change callback \\
\hline
\end{tabular}

\subsection*{22.5 Accessible objects}

ACC routines shall access information about the following objects:
Module instances
Module ports
Individual bits of a port
Module or data paths
Intermodule paths

Top-level modules
Primitive instances
Primitive terminals
Nets (scalars, vectors, and bit- or part-selects of vectors)
Regs (scalars, vectors, and bit- or part-selects of vectors)
Integer variables (and bit- or part-selects of integers)
Real and time variables
Named events
Parameters
Specparams
Timing checks
Timing check terminals
User-Defined system task/function arguments

The following tables summarize the operations that can be performed for each of the above object types.

\subsection*{22.5.1 ACC routines that operate on module instances}

Table 97-Operations on module instances
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & \multicolumn{1}{c|}{ Use } \\
\hline \begin{tabular}{l} 
Obtain handles for module instances tagged as cells within a hierarchi- \\
cal scope and below
\end{tabular} & acc_next_cell() \\
\hline \begin{tabular}{l} 
Obtain handles for module instances within a particular module \\
instance
\end{tabular} & acc_next_child() \\
\hline Obtain a handle to the parent (the module that contains the instance) & acc_handle_parent() \\
\hline Get the instance name & acc_fetch_name() \\
\hline Get the full hierarchical name & acc_fetch_fullname() \\
\hline Get the module definition name & acc_fetch_defname() \\
\hline \begin{tabular}{l} 
Get the fulltype of a module instance (cell instance, module instance, or \\
top-level module)
\end{tabular} & acc_fetch_fulltype() \\
\hline \begin{tabular}{l} 
Get the delay mode of a module instance (none, zero, unit, distributed, \\
or path)
\end{tabular} & acc_fetch_delay_mode() \\
\hline Get timescale information for a module instance & acc_fetch_timescale_info() \\
\hline
\end{tabular}

\subsection*{22.5.2 ACC routines that operate on module ports}

\section*{Table 98-Operations on module ports}
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & \multicolumn{1}{c|}{ Use } \\
\hline Obtain handles for ports of a module instance & acc_next_port() \\
\hline Obtain handles for output ports of a module instance & acc_next_portout() \\
\hline Obtain a handle for a particular port & acc_handle_port() \\
\hline \begin{tabular}{l} 
Obtain a handle to the parent (the module instance that contains the \\
port)
\end{tabular} & acc_handle_parent() \\
\hline Obtain handles to hierarchically higher-connected nets & acc_next_hiconn() \\
\hline Obtain handles to hierarchically lower-connected nets & acc_next_loconn() \\
\hline \begin{tabular}{l} 
Obtain a handle to the hierarchically higher-connected net of a scalar \\
module port or bit of a vector port
\end{tabular} & acc_handle_hiconn() \\
\hline \begin{tabular}{l} 
Obtain a handle to the hierarchically lower-connected net of a scalar \\
module port or bit of a vector port
\end{tabular} & acc_handle_loconn() \\
\hline Get the instance name & acc_fetch_name() \\
\hline Get the full hierarchical name & acc_fetch_fullname() \\
\hline Get the port direction & acc_fetch_direction() \\
\hline Get the port index number & acc_fetch_index() \\
\hline Get the fulltype of a module port & acc_fetch_fulltype() \\
\hline Add VCL value change callback monitors & acc_vcl_add() \\
\hline Delete VCL value change callback monitors & acc_vcl_delete() \\
\hline Read Module Input Port Delay (MIPD) & acc_fetch_delays() \\
\hline Append to existing MIPD & acc_append_delays() \\
\hline Replace existing MIPD & \\
\hline
\end{tabular}

\subsection*{22.5.3 ACC routines that operate on bits of a port}

Table 99-Operations on bits of a port
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Obtain handles for bits of a module port & acc_next_bit() \\
\hline Obtain a handle to the port from a port bit & acc_handle_parent() \\
\hline Get the port name & acc_fetch_name () \\
\hline Get the full hierarchical name & acc_fetch_fullname() \\
\hline Get the fulltype of a port s bit & acc_fetch_fulltype() \\
\hline Read Module Input Port Delay (MIPD) & acc_fetch_delays() \\
\hline Append to existing MIPD & acc_append_delays() \\
\hline Replace existing MIPD & acc_replace_delays() \\
\hline Add VCL value change callback monitors & acc_vcl_add() \\
\hline Delete VCL value change callback monitors & acc_vcl_delete() \\
\hline
\end{tabular}

\subsection*{22.5.4 ACC routines that operate on module paths or data paths}

Table 100-Operations on module paths and data paths
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & \multicolumn{1}{c|}{ Use } \\
\hline Obtain handles for module paths within a scope & acc_next_modpath() \\
\hline Obtain a handle to the first connected net & \begin{tabular}{c} 
acc_handle_pathin() \\
acc_handle_pathout()
\end{tabular} \\
\hline Obtain a handle to a module path & acc_handle_modpath() \\
\hline Obtain a handle to a datapath & acc_handle_datapath() \\
\hline Obtain a handle to a conditional expression for a path & acc_handle_condition() \\
\hline Obtain handles for input terminals of a module path or data path & acc_next_input() \\
\hline Obtain handles for output terminals of a module path or data path & acc_next_output) \\
\hline Get the path name & acc_fetch_name() \\
\hline Get the full hierarchical name & acc_fetch_fullname() \\
\hline Get the polarity of a path & acc_fetch_polarity() \\
\hline Get the edge specified for a path terminal & acc_fetch_edge() \\
\hline Read path delays & acc_fetch_delays() \\
\hline Append to existing path delays & acc_append_delays() \\
\hline Replace existing path delays & acc_replace_delays() \\
\hline Read path pulse handling & acc_fetch_pulsere() \\
\hline Append to existing path pulse control values & acc_append_pulsere() \\
\hline Replace existing path pulse control values & acc_replace_pulsere() \\
\hline Specify path pulse control values & acc_set_pulsere() \\
\hline Free memory allocated by acc_next_input() or acc_next_output() & acc_release_object() \\
\hline
\end{tabular}

\subsection*{22.5.5 ACC routines that operate on intermodule paths}

Table 101-Operations on intermodule paths
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Obtain a handle for an intermodule path & acc_handle_path() \\
\hline Get the fulltype of an intermodule path & acc_fetch_fulltype() \\
\hline Read intermodule path delays & acc_fetch_delays() \\
\hline Modify intermodule path delays & acc_replace_delays() \\
\hline Read intermodule path pulse control values & acc_fetch_pulsere() \\
\hline Append to existing intermodule path pulse control values & acc_append_pulsere() \\
\hline Replace existing intermodule path pulse control values & acc_replace_pulsere() \\
\hline Specify intermodule path pulse control values & acc_set_pulsere() \\
\hline
\end{tabular}

\subsection*{22.5.6 ACC routines that operate on top-level modules}

Table 102-Operations on top-level modules
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Obtain handles for top-level modules in a design & \begin{tabular}{c} 
acc_next_topmod() \\
acc_next_child()
\end{tabular} \\
\hline Get the module name & \begin{tabular}{c} 
acc_fetch_name () \\
acc_fetch_fullname() \\
acc_fetch_defname()
\end{tabular} \\
\hline
\end{tabular}

\subsection*{22.5.7 ACC routines that operate on primitive instances}

Table 103-Operations on primitive instances
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Obtain handles for primitive instances within a module instance & acc_next_primitive() \\
\hline Obtain a handle to the parent (the module that contains the primitive) & acc_handle_parent() \\
\hline Get the instance name & acc_fetch_name() \\
\hline Get the full hierarchical name & acc_fetch_fullname() \\
\hline Get the definition name & acc_fetch_defname() \\
\hline Get the primitive fulltype & acc_fetch_fulltype() \\
\hline Read delays & acc_fetch_delays() \\
\hline Append to existing primitive delays & acc_append_delays() \\
\hline Replace existing primitive delays & acc_replace_delays() \\
\hline
\end{tabular}

\subsection*{22.5.8 ACC routines that operate on primitive terminals}

Table 104-Operations on primitive terminals
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Obtain handles for terminals of a primitive instance & acc_next_terminal() \\
\hline Obtain a handle to the net connected to the terminal & acc_handle_conn() \\
\hline \begin{tabular}{l} 
Obtain a handle to the parent (primitive instance containing the \\
terminal)
\end{tabular} & acc_handle_parent() \\
\hline Get the direction (input, output, inout) & acc_fetch_direction() \\
\hline Get the terminal index number & acc_fetch_index() \\
\hline Get the fulltype & acc_fetch_fulltype() \\
\hline Add VCL value change callback monitors & acc_vcl_add() \\
\hline Delete VCL value change callback monitors & acc_vcl_delete() \\
\hline
\end{tabular}

\subsection*{22.5.9 ACC routines that operate on nets}

Table 105-Operations on nets
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Obtain handles for nets within a module instance & acc_next_net() \\
\hline Obtain handles for nets within a module instance & acc_next() \\
\hline Obtain a handle to the parent (the module instance that contains the net) & acc_handle_parent() \\
\hline Determine if net is scalar, vector, collapsed, or expanded & acc_object_of_type() \\
\hline Obtain handles to bits of a vector net & acc_next_bit() \\
\hline Obtain handles to driving terminals of the net & acc_next_driver() \\
\hline Obtain handles to load terminals of the net & acc_next_load() \\
\hline \begin{tabular}{l} 
Obtain handles to connected load terminals; only one per driven cell \\
port
\end{tabular} & acc_next_cell_load() \\
\hline Obtain a handle to the simulated net of a collapsed net & acc_handle_simulated_net() \\
\hline Get the net name & acc_fetch_name() \\
\hline Get the full hierarchical name & acc_fetch_fullname() \\
\hline Get the net vector size & acc_fetch_size() \\
\hline Get the msb and lsb vector range & acc_fetch_range() \\
\hline Get the net fulltype & acc_fetch_fulltype() \\
\hline Get the net logic or strength value & acc_fetch_value() \\
\hline Force or release the net logic value & acc_set_value() \\
\hline Add VCL value change callback monitors & acc_vcl_add() \\
\hline Delete VCL value change callback monitors & acc_vcl_delete() \\
\hline
\end{tabular}

\subsection*{22.5.10 ACC routines that operate on reg types}

Table 106-Operations on reg types
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Obtain handles to regs within a given scope & acc_next() \\
\hline Obtain handles to bits of a vector reg & acc_next_bit() \\
\hline Obtain a handle to the parent (module instance containing the reg) & acc_handle_parent() \\
\hline Obtain handles to load terminals of the reg & acc_next_load() \\
\hline Determine if reg is a scalar or a vector & acc_object_of_type() \\
\hline Get the reg name & acc_fetch_name() \\
\hline Get the full hierarchical name & acc_fetch_fullname() \\
\hline Get the reg size & acc_fetch_size() \\
\hline Get the msb and lsb vector range & acc_fetch_range() \\
\hline Get the reg value & acc_fetch_value() \\
\hline Set the reg value & acc_set_value() \\
\hline Add VCL value change callback monitors & acc_vcl_add \()\) \\
\hline Delete VCL value change callback monitors & acc_vcl_delete() \\
\hline
\end{tabular}

\subsection*{22.5.11 ACC routines that operate on integer, real, and time variables}

Table 107-Operations on integer, real, and time variables
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Obtain handles to variables within a given scope & acc_next() \\
\hline Obtain a handle to the parent (module instance containing the variable) & acc_handle_parent() \\
\hline Get the variable name & acc_fetch_name() \\
\hline Get the full hierarchical name & acc_fetch_fullname() \\
\hline Get the variable value & acc_fetch_value() \\
\hline Set the variable value & acc_set_value() \\
\hline Add VCL value change callback monitors & acc_vcl_add() \\
\hline Delete VCL value change callback monitors & acc_vcl_delete() \\
\hline
\end{tabular}

\subsection*{22.5.12 ACC routines that operate on named events}

Table 108-Operations on named events
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Obtain handles to named events within a given scope & acc_next() \\
\hline \begin{tabular}{l} 
Obtain a handle to the parent (module instance containing the named \\
event)
\end{tabular} & acc_handle_parent() \\
\hline Get the named-event name & acc_fetch_name() \\
\hline Get the full hierarchical name & acc_fetch_fullname() \\
\hline Add VCL value change callback monitors & acc_vcl_add() \\
\hline Delete VCL value change callback monitors & acc_vcl_delete() \\
\hline
\end{tabular}

\subsection*{22.5.13 ACC routines that operate on parameters and specparams}

Table 109-Operations on parameters and specparams
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & \multicolumn{1}{c|}{ Use } \\
\hline Obtain handles for parameters within a module instance & acc_next_parameter() \\
\hline Obtain handles for specparams within a module instance & acc_next_specparam() \\
\hline \(\begin{array}{l}\text { Obtain a handle to the parent (the module instance that contains the } \\
\text { parameter) }\end{array}\) & acc_handle_parent() \\
\hline Get the parameter or specparam name & acc_fetch_name() \\
\hline Get the full hierarchical name & \(\begin{array}{c}\text { acc_fetch_fullname() }\end{array}\) \\
\hline Get the parameter value data type (integer, floating point, string) \\
acc_fetch_fulltype()
\end{tabular}\(]\)\begin{tabular}{l} 
acc_fetch_paramval() \\
\hline Get the value of a parameter \\
\hline Get the attribute value of a parameter defined with an attribute name \\
\begin{tabular}{l} 
acc_fetch_attribute() \\
acc_fetch_attribute_int() \\
acc_fetch_attribute_str()
\end{tabular} \\
\hline
\end{tabular}

\subsection*{22.5.14 ACC routines that operate on timing checks}

Table 110-Operations on timing checks
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Obtain handles for timing checks within a module instance & acc_next_tchk() \\
\hline Obtain a handle to a specific timing check & acc_handle_tchk() \\
\hline Obtain handles to all timing check terminals & acc_next_input() \\
\hline Free memory allocated by acc_next_input() & acc_release_object() \\
\hline Obtain a handle to a timing check terminal & \begin{tabular}{c} 
acc_handle_tchkarg10 \\
acc_handle_tchkarg2()
\end{tabular} \\
\hline Get the timing check fulltype & acc_fetch_fulltype() \\
\hline Get a timing check limit & acc_fetch_delays() \\
\hline Append to an existing timing check limit & acc_append_delays() \\
\hline Replace to an existing timing check limit & acc_replace_delays() \\
\hline
\end{tabular}

\subsection*{22.5.15 ACC routines that operate on timing check terminals}

Table 111-Operations on timing check terminals
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Obtain a handle to the net attached to timing check terminals & acc_handle_conn() \\
\hline Obtain a handle to the condition on a timing check terminal & acc_handle_condition() \\
\hline Get edge information on a timing check terminal & acc_fetch_edge() \\
\hline
\end{tabular}

\subsection*{22.5.16 ACC routines that operate on user-defined system task/function arguments}

Table 112-Operations on user-defined system task/function arguments
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Obtain a handle for an object named in a task/function argument & \begin{tabular}{c} 
acc_handle_tfarg() \\
acc_handle_itfarg()
\end{tabular} \\
\hline Get the value of a task/function argument as a double & \begin{tabular}{c} 
acc_fetch_tfarg() \\
acc_fetch_itfarg()
\end{tabular} \\
\hline Get the value of a task/function argument as an integer & \begin{tabular}{c} 
acc_fetch_tfarg_int() \\
acc_fetch_itfarg_int()
\end{tabular} \\
\hline Get the value of a task/function argument as a string pointer & \begin{tabular}{l} 
acc_fetch_tfarg_str() \\
acc_fetch_itfarg_str()
\end{tabular} \\
\hline
\end{tabular}

\subsection*{22.6 ACC routine types and fulltypes}

Many objects in the Verilog HDL can have both a type and a fulltype associated with them. A type shall be a general classification of an object, whereas a fulltype shall be a specific classification. The type and fulltype for a given object can be different constants, or they can be the same constant. For example, an and logic gate has a type of accPrimitive and a fulltype of accAndPrimitive. The type and fulltype are predefined integer constants in the file acc_user.h. Several ACC routines either return a type or fulltype value, or use a type or fulltype value as an argument. Table 113 lists all type and fulltype constants that shall be supported by ACC routines, listed alphabetically by the type name.

Table 113-List of all predefined type and fulltype constants
\begin{tabular}{|l|l|l|}
\hline \multicolumn{1}{|c|}{ type constant } & \multicolumn{1}{c|}{ fulltype constant } & \multicolumn{1}{c|}{ Description } \\
\hline accConstant & accConstant & Object is a constant \\
\hline accDataPath & accDataPath & Object is a data path in a path delay \\
\hline accFunction & accFunction & Object is a Verilog HDL function \\
\hline accIntegerVar & accIntegerVar & Object is declared as an integer data type \\
\hline accModPath & accModPath & Object is a module path \\
\hline \multirow{3}{*}{ accModule } & accModuleInstance & Object is a module instance \\
\cline { 2 - 3 } & accCellInstance & a cell is a module instance that has been defined as \\
\hline & accTopModule & Object is a top-level module \\
\hline \multirow{3}{*}{ accNamedEvent } & accNamedEvent & Object is declared as an event data type \\
\hline
\end{tabular}

Table 113-List of all predefined type and fulltype constants (continued)
\begin{tabular}{|c|c|c|}
\hline type constant & fulltype constant & Description \\
\hline \multirow[t]{11}{*}{accNet} & accSupply0 & Object is declared as a supply0 net data type \\
\hline & accSupply 1 & Object is declared as a supply1 net data type \\
\hline & accTri & Object is declared as a tri net data type \\
\hline & accTriand & Object is declared as a triand net data type \\
\hline & accTrior & Object is declared as a trior net data type \\
\hline & accTrireg & Object is declared as a trireg net data type \\
\hline & accTri0 & Object is declared as a tri0 net data type \\
\hline & accTri1 & Object is declared as a tri1 net data type \\
\hline & accWand & Object is declared as a wand net data type \\
\hline & accWire & Object is declared as a wire net data type \\
\hline & accWor & Object is declared as a wor net data type \\
\hline accNetBit & accNetBit & Object is a bit-select of a net data type \\
\hline accoperator & accOperator & Object is a Verilog HDL operator \\
\hline \multirow[t]{3}{*}{accParameter} & accIntegerParam & Object is a parameter with an integer value \\
\hline & accRealParam & Object is a parameter with a real value \\
\hline & accStringParam & Object is a parameter with a string value \\
\hline accPartSelect & accPartSelect & Object is a part-select of a vector \\
\hline \multirow[t]{2}{*}{accPathTerminal} & accPathInput & Object is an input terminal of a module path \\
\hline & accPathOutput & Object is an output terminal of a module path \\
\hline \multirow[t]{5}{*}{accPort} & accConcatPort & Object is a module port concatenation \\
\hline & accScalarPort & Object is a scalar module port \\
\hline & accBitSelectPort & \begin{tabular}{l}
Object is a bit-select of a module port (e.g.: \\
module (.a[ 1], .a[ 0], ...); \\
input [1:0] a;
\end{tabular} \\
\hline & accPartSelectPort & Object is a part-select of a module port (e.g.: module (.a[3:2], .a[1:0], ...); input [ 3:0] a; \\
\hline & accVectorPort & Object is a vector module port \\
\hline accPortBit & accPortBit & Object is a bit of a module port \\
\hline
\end{tabular}

Table 113-List of all predefined type and fulltype constants (continued)
\begin{tabular}{|c|c|c|}
\hline type constant & fulltype constant & Description \\
\hline \multirow[t]{28}{*}{accPrimitive} & accAndGate & Object is an and primitive \\
\hline & accBufGate & Object is a buf primitive \\
\hline & accBufif0Gate & Object is a bufif0 primitive \\
\hline & accBufif1Gate & Object is a bufif1 primitive \\
\hline & accCmosGate & Object is a cmos primitive \\
\hline & accCombPrim & Object is a combinational logic UDP \\
\hline & accNandGate & Object is a nand primitive \\
\hline & accNmosGate & Object is an nmos primitive \\
\hline & accNorGate & Object is a nor primitive \\
\hline & accNotGate & Object is a not primitive \\
\hline & accNotif0Gate & Object is a notif0 primitive \\
\hline & accNotif1Gate & Object is a notif1 primitive \\
\hline & accOrGate & Object is an or primitive \\
\hline & accPmosGate & Object is a pmos primitive \\
\hline & accPulldownGate & Object is a pulldown primitive \\
\hline & accPullupGate & Object is a pullup primitive \\
\hline & accRemosGate & Object is an remos primitive \\
\hline & accRnmosGate & Object is an rnmos primitive \\
\hline & accRpmosGate & Object is an rpmos primitive \\
\hline & accRtranGate & Object is an rtran primitive \\
\hline & accRtranif0Gate & Object is an rtranif0 primitive \\
\hline & accRtranif1Gate & Object is an rtranif1 primitive \\
\hline & accSeqPrim & Object is a sequential logic UDP \\
\hline & accTranGate & Object is a tran primitive \\
\hline & accTranif0Gate & Object is a tranif0 primitive \\
\hline & accTranif1Gate & Object is a tranif1 primitive \\
\hline & accXnorGate & Object is an xnor primitive \\
\hline & accXorGate & Object is an xor primitive \\
\hline accRealVar & accRealVar & Object is declared as a real data type \\
\hline accReg & accReg & Object is declared as a reg data type \\
\hline accRegBit & accRegBit & Object is a bit-select of a reg data type \\
\hline \multirow[t]{3}{*}{accSpecparam} & accIntegerParam & Object is a specparam with an integer value \\
\hline & accRealParam & Object is a specparam with a real value \\
\hline & accStringParam & Object is a specparam with a string value \\
\hline
\end{tabular}

Table 113-List of all predefined type and fulltype constants (continued)
\begin{tabular}{|l|l|l|}
\hline \multicolumn{1}{|c|}{ type constant } & \multicolumn{1}{c|}{ fulltype constant } & \\
\hline \multirow{3}{*}{ accStatement } & accStatement & Object is a procedural statement \\
\cline { 2 - 3 } & accNamedBeginStat & Object is a named begin statement \\
\cline { 2 - 3 } & accNamedForkStat & Object is a named fork statement \\
\hline accSystemTask & accSystemTask & Object is a built-in system task \\
\hline \multirow{2}{*}{ accSystemFunction } & accSystemFunction & \begin{tabular}{l} 
Object is a built-in system function with a scalar or \\
vector return
\end{tabular} \\
\hline \multirow{3}{*}{ accSystemRealFunction } & accSystemRealFunction & \begin{tabular}{l} 
Object is a built-in system function with a real \\
value return
\end{tabular} \\
\hline accTask & accTask & Object is a Verilog HDL task \\
\hline \multirow{3}{*}{ accTchk } & accHold & Object is a \$hold timing check \\
\cline { 2 - 4 } & accNochange & Object is a \$nochange timing check \\
\cline { 2 - 4 } & accPeriod & Object is a \$period timing check \\
\cline { 2 - 4 } & accRecovery & Object is a \$recovery timing check \\
\hline \multirow{4}{*}{ accWirePath } & accSetup & Object is a \$setup timing check \\
\hline \multirow{4}{*}{ accUserRealFunction } & accSetuphold & Object is a \$setuphold timing check \\
\hline accTimeVar & accSkew & Object is a \$skew timing check \\
\hline accUserTask & accWidth & Object is an intermodule path (from a module out- \\
put to a module input)
\end{tabular}

\subsection*{22.7 Error handling}

When an ACC routine detects an error, it shall perform the following operations:
a) Set the global error flag acc_error_flag to non-zero
b) Display an error message at run time to the output channel of the software product which invoked the PLI application
c) Return an exception value

When an ACC routine is called, it automatically resets acc_error_flag to 0 .

\subsection*{22.7.1 Suppressing error messages}

By default, ACC routines shall display error messages. Error messages can be suppressed using the ACC routine acc_configure() to set the configuration parameter accDisplayErrors to "false".

\subsection*{22.7.2 Enabling warnings}

By default, ACC routines shall not display warning messages. To enable warning messages, use the ACC routine acc_configure() to set the configuration parameter accDisplayWarnings to "true".

\subsection*{22.7.3 Testing for errors}

If automatic error reporting is suppressed, error handling can be performed by checking the acc_error_flag explicitly after calling an ACC routine. This procedure is described in Figure 53.


Figure 53-Using acc_error_flag to detect errors

\subsection*{22.7.4 Example}

The following example shows a C language application that performs error checking for ACC routines. This example uses acc_configure() to suppress automatic error reporting. Instead, it checks acc_error_flag explicitly and displays its own specialized error message.
```

\#include "acc_user.h"
check_new_timing()
{
handle gate_handle;
/* initialize and configure access routines */
acc_initialize();
/* suppress error reporting by access routines */
acc_configure( accDisplayErrors, "false" );
/* check type of first argument, the object */
gate_handle = acc_handle_tfarg( 1 );
/* check for valid argument */
if (acc_error_flag)
tf_error("Cannot derive handle from argument\n");
else
/* argument is valid */
/* make sure it is a primitive */
if ( acc_fetch_type(gate_handle) != accPrimitive )
tf_error("Invalid argument type:not a primitive\n");
acc_close();

```
\}

\subsection*{22.7.5 Exception values}

ACC routines shall return one of three exception values when an error occurs, unless specified differently in the syntax of a specific ACC routine.

Table 114-Exception values returned by ACC routines on errors
\begin{tabular}{|c|c|}
\hline When routine returns & The exception value shall be \\
\hline PLI_INT32 & 0 \\
\hline double values & 0.0 \\
\hline pointers or handles & null \\
\hline bool (boolean) values & false \\
\hline
\end{tabular}

Because ACC routines can return valid values that are the same as exception values, the only definitive way to detect errors explicitly is to check acc_error_flag.

Note that null and false are predefined constants, declared in acc_user.h.

\section*{22.8 eading and writing delay values}

This section explains how ACC routines that read and modify delays are used. The ACC routines acc_fetch_delays(), acc_replace_delays(), and acc_append_delays() can read or modify delay values in a Verilog software implementation data structure. Refer to Clause 23 for the complete syntax of each of these routines.

\subsection*{22.8.1 Number of delays for Verilog HDL objects}

There are a variety of objects in a Verilog HDL source description that can model delays. These objects can

Table 116-Configuration parameters for delay ACC routines
\begin{tabular}{|c|l|}
\hline Configuration parameter & \multicolumn{1}{c|}{ Description } \\
\hline accMinTypMaxDelays & \begin{tabular}{l} 
When "false", each delay shall be represented by one value. \\
When "true", each delay shall be represented by three delay values, rep- \\
resenting minimum, typical, maximum, respectively. \\
The default shall be "false".
\end{tabular} \\
\hline accToHiZDelay & \begin{tabular}{l} 
When set to "average", "max" or "min", the delay modify ACC rou- \\
tines shall calculate the toZ delay for 3-state primitives, or for path and input \\
port objects when accPathDelayCount is set to 2. When set to \\
"from user", the toZ delay shall not be calculated. \\
The default is "from_user". This parameter shall be ignored when \\
accMinTypMaxDelays is set to "true".
\end{tabular} \\
\hline accPathDelayCount & \begin{tabular}{l} 
Sets the number of delay arguments to be used by the ACC routines for mod- \\
ule path, intermodule path and module input port delays. Shall be set to " \(1 "\), \\
" 2 ", " 2 ", " 6 ", or " 12 ". The default shall be " 6 ".
\end{tabular} \\
\hline
\end{tabular}

\subsection*{22.8.3 Determining the number of arguments for ACC delay routines}

The ACC routines acc_fetch_delays(), acc_replace_delays(), and acc_append_delays() shall require a different number of arguments based on

The type of object handle
The setting of configuration parameters
The following subsections discuss how these factors affect the number of arguments for delay ACC routines.

\subsection*{22.8.3.1 Single delay value mode}

When the configuration parameter accMinTypMaxDelays is "false" (the default), a single value shall be used for each delay transition. In this mode, the routines acc_fetch_delays(), acc_replace_delays(), and acc_append_delays() shall require each delay value as a separate argument. For acc_replace_delays() and acc_append_delays(), the arguments shall be a literal value of type double or variables of type double. For acc_fetch_delays(), the arguments shall be pointers to variables of type double.

The number of arguments required is determined by the type of object handle passed to the delay ACC routine, as shown in Table 117.

Table 117-Number of delay arguments in single delay mode
\begin{tabular}{|c|c|l|}
\hline Object handle type & Configuration parameters & \multicolumn{1}{c|}{ Number and order of delay arguments } \\
\hline Timing check & & \(\mathbf{1}\) argument: timing check limit \\
\hline 2-state primitive & & \(\mathbf{2}\) arguments: rise, fall transitions \\
\hline \multirow{2}{*}{ 3-state primitive } & \begin{tabular}{c} 
accToHiZDelay set to \\
"min", "max", or "average"
\end{tabular} & \begin{tabular}{c}
\(\mathbf{2}\) arguments: rise, fall transitions \\
(toZ delay is calculated; see Section 22.8.3.3)
\end{tabular} \\
\cline { 2 - 3 } & \begin{tabular}{c} 
accToHiZDelay set to \\
"from_user"
\end{tabular} & \(\mathbf{3}\) arguments: rise, fall, toZ transitions \\
\hline
\end{tabular}

Table 117-Number of delay arguments in single delay mode (continued)
\begin{tabular}{|c|c|c|}
\hline Object handle type & Configuration parameters & Number and order of delay arguments \\
\hline \multirow{5}{*}{Module paths Intermodule paths Module ports Module port bits} & accPathDelayCount set to " 1 " & 1 argument: all transitions \\
\hline & accPathDelayCount set to " 2 " & \(\mathbf{2}\) arguments: rise, fall transitions \\
\hline & accPathDelayCount set to " 3 " & 3 arguments: rise, fall, toZ transitions \\
\hline & accPathDelayCount set to "6" & 6 arguments:
\[
0->1,1->0,0->Z, Z->1,1->Z, Z->0
\] \\
\hline & accPathDelayCount set to "12" & 12 arguments:
\[
\begin{aligned}
& 0->1, \quad 1->0,0->z, \quad z->1, \quad 1->z, \quad z->0 \\
& 0->x, X->1,1->x, x->0, x->Z, Z->x
\end{aligned}
\] \\
\hline
\end{tabular}

\subsection*{22.8.3.2 Min:typ:max delay value mode}

When the configuration parameter accMinTypMaxDelays is "true", a three-value set shall be used for each delay transition. In this mode, the routines acc_fetch_delays(), acc_replace_delays(), and acc_append_delays() shall require the delay argument to be a pointer of an array of variables of type double. The number of elements placed into or read from the array shall be determined by the type of object handle passed to the delay ACC routine, as shown in Table 118.

Table 118-Number of delay elements in min:typ:max delay mode
\begin{tabular}{|c|c|c|}
\hline Object handle type & Configuration parameters & Size and order of the delay array \\
\hline Timing check & & ```
3 elements:
    array[ 0] = min limit
    array[ 1] = typ limit
    array[ 2] = max limit
``` \\
\hline \begin{tabular}{l}
2-state primitive \\
3-state primitive
\end{tabular} & & \begin{tabular}{l}
9 elements: \\
array[ 0] = min rise delay \\
array[ 1] = typ rise delay \\
array[ 2] = max rise delay \\
array[ 3] = min fall delay \\
array[ 4] = typ fall delay \\
array[5] = max fall delay \\
array[ 6] = min toz delay \\
array[ 7] = typ toz delay \\
array[ 8] = max toz delay \\
(an array of at least 9 elements shall be declared, even if toZ delays are not used by the object)
\end{tabular} \\
\hline
\end{tabular}

Table 118-Number of delay elements in min:typ:max delay mode (continued)
\begin{tabular}{|c|c|c|}
\hline Object handle type & Configuration parameters & Size and order of the delay array \\
\hline \multirow{4}{*}{\begin{tabular}{l}
Module path Intermodule paths \\
Module ports Module port bits
\end{tabular}} & accPathDelayCount set to " 1 " & ```
3 elements:
    array[ 0] = min delay
    array[ 1] = typ delay
    array[ 2] = max delay
``` \\
\hline & accPathDelayCount set to "2" & \begin{tabular}{l}
6 elements: \\
array[ 0] = min rise delay \\
array[ 1] = typ rise delay \\
array[ 2] = max rise delay \\
array[ 3] = min fall delay \\
array[ 4] = typ fall delay \\
array[ 5] = max fall delay
\end{tabular} \\
\hline & accPathDelayCount is set to " 3 " & \[
\begin{aligned}
9 \text { elements: } & \\
\text { array[ 0] } & =\text { min rise delay } \\
\text { array[ 1] } & =\text { typ rise delay } \\
\text { array[ 2] } & =\text { max rise delay } \\
\text { array[ 3] } & \text { min fall delay } \\
\text { array[ 4] } & =\text { typ fall delay } \\
\text { array[ 5] } & =\text { max fall delay } \\
\text { array[ 6] } & =\text { min toz delay } \\
\text { array[ 7] } & =\text { typ toz delay } \\
\text { array[ 8] } & =\text { max toz delay }
\end{aligned}
\] \\
\hline & accPathDelayCount set to "6" & \begin{tabular}{l}
18 elements: \\
array[ 0] \(=\min 0->1\) delay \\
array[ 1] = typ 0->1 delay \\
array[ 2] \(=\max 0->1\) delay \\
array[ 3] \(=\min 1->0\) delay \\
array[ 4] = typ 1->0 delay \\
array[ 5] \(=\max 1->0\) delay \\
array[ 6] \(=\min 0->Z\) delay \\
array[ 7] = typ 0->Z delay \\
array[ 8] \(=\max 0->Z\) delay \\
array[ 9] \(=\) min \(\mathrm{Z}->1\) delay \\
array[ 10] = typ Z->1 delay \\
array[ 11] = max \(\mathrm{Z}->1\) delay \\
array[ 12] \(=\min 1->z\) delay \\
array[ 13] = typ 1->z delay \\
array[ 14] \(=\max 1->Z\) delay \\
array[ 15] \(=\) min \(Z->0\) delay \\
array[ 16] = typ Z->0 delay \\
array[ 17] \(=\max \mathrm{Z}->0\) delay
\end{tabular} \\
\hline
\end{tabular}

Table 118-Number of delay elements in min:typ:max delay mode (continued)
\begin{tabular}{|c|c|c|}
\hline Object handle type & Configuration parameters & Size and order of the delay array \\
\hline Module path (continued) & accPathDelayCount set to " 12 " & 36 elements: \\
\hline
\end{tabular}

\subsection*{22.8.3.3 Calculating turn-off delays from rise and fall delays}

In single delay mode (accMinTypMaxDelays set to "false"), the routines acc_replace_delays() and acc_append_delays() can be instructed to calculate automatically the turn-off delays from rise and fall delays. How the calculation shall be performed is controlled by the configuration parameter accToHiZDelay, as shown in Table 119.

Table 119-Configuring accToHiZDelay to determine the toZ delay
\begin{tabular}{|c|l|}
\hline Configuration of accToHiZDelay & \multicolumn{1}{c|}{ Value of the toZ delay } \\
\hline "average" & \begin{tabular}{l} 
The toZ turn-off delay shall be the average of the \\
rise and fall delays.
\end{tabular} \\
\hline "min" & \begin{tabular}{l} 
The toZ turn-off delay shall be the smaller of the \\
rise and fall delays.
\end{tabular} \\
\hline "max" & \begin{tabular}{l} 
The toZ turn-off delay shall be the larger of the rise \\
and fall delays.
\end{tabular} \\
\hline \begin{tabular}{c} 
"from_user" \\
(the default)
\end{tabular} & \begin{tabular}{l} 
The toZ turn-off delay shall be set to the value \\
passed as a user-supplied argument.
\end{tabular} \\
\hline
\end{tabular}

\subsection*{22.9 String handling}

\subsection*{22.9.1 ACC routines share an internal string buffer}

ACC routines that return pointers to strings can share an internal buffer to store string values. These routines shall return a pointer to the location in the buffer that contains the first character of the string, as illustrated in Figure 54. In this example, mod_name points to the location in the buffer where top.m1 (the name of the module associated with module_handle) is stored.


Figure 54-How ACC routines store strings in the internal buffer

\subsection*{22.9.2 String buffer reset}

ACC routines shall place strings at the next available sequential location in the string buffer, which stores at least 4096 characters. If there is not enough room to store an entire string starting at the next location, a condition known as buffer reset shall occur.

When buffer reset occurs, ACC routines shall place the next string starting at the beginning of the buffer, overwriting data already stored there. The result can be a loss of data, as illustrated in Figure 55.


Figure 55—Buffer reset causes data in the string buffer to be overwritten

\subsection*{22.9.2.1 The buffer reset warning}

ACC routines shall issue a warning whenever the internal string buffer resets. To view the warning message, the configuration parameter accDisplayWarnings shall be set to "true", using the ACC routine acc_configure().

\subsection*{22.9.3 reserving string values}

Applications that use strings immediately for example, to print names of objects do not need to be concerned about overwrites after a string buffer reset. Applications that have to preserve string values while calling other ACC routines that write to the string buffer should preserve the string value before it is overwritten. To preserve a string value, the C routine strcpy can be used to copy the string to a local character array.

\subsection*{22.9.4 Example of preserving string values}

The following example code illustrates preserving string values. If the module in this example contains many cells, one of the calls to acc_fetch_name() could eventually overwrite the module name in the string buffer with a cell name. To preserve the module name, strcpy is used to store it locally in an array called mod_name.
```

nclude "acc_user.h"
id display_cells_in_module(mod)
ndle mod;
handle cell;
char *mod_name;
PLI_BYTE8 *temp;
/* save the module name in local buffer mod_name */
temp $=$ acc_fetch_fullname (mod);
storage the size of the full
mod_name $=($ char*) malloc ((strlen ((char*)temp) +1$)$ * sizeof(PLI_BYTE8));

strcpy (mod_name, | strar saves the full module |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| name in mod_name |

    cell = null;
    while (cell = acc_next_cell( mod, cell ) )
        io_printf( "\%s.\%s\n", mod_name, acc_fetch_name( cell ) );
    free(mod_name);
    ```

\subsection*{22.10 Using VCL ACC routines}

The VCL routines add or delete value change monitors on a specified object. If a value change monitor is placed on an object, then whenever the object changes logic value or strength, a PLI consumer routine shall be called.

The ACC routine acc_vcl_add() adds a value change monitor on an object. The arguments for acc_vcl_add() specify

A handle to an object in the Verilog HDL structure
The name of a consumer routine
A user_data value
A VCL reason_flag
The following example illustrates the usage of acc_vcl_add().
```

acc_vcl_add(net, netmon_consumer, net_name, vcl_verilog_logic);

```

The purpose of each of these arguments is described in the following paragraphs. Refer to 23.97 for the full syntax and usage of acc_vel_add() and its arguments.

The handle argument shall be a handle to any object type in the list in 22.10.1.
The consumer routine argument shall be the name of a C application that shall be called for the reasons specified by the reason_flag, such as a logic value change. When a consumer routine is called, it shall be passed a pointer to a C record, called vc_record. This record shall contain information about the object, including the simulation time of the change and the new logic value of the object. The vc_record is defined in the file acc_user. h and is listed in Figure 56.

The user_data argument shall be a PLI_BYTE8 pointer. The value of the user_data argument shall be passed to the consumer routine as part of the vc_record. The user_data argument can be used to pass a single value to the consumer routine, or it can be used to pass a pointer to information. For example, the name of the object could be stored in a global character string array, and a pointer to that array could be passed as the user_data argument. The consumer routine could then have access to the object name. Another example is to allocate memory for a user-defined structure with several values that need to be passed to the consumer routine. A pointer to the memory for the user-defined structure is then passed as the user_data argument. Note that the user_data argument is defined as a PLI_BYTE8 pointer; therefore, any other data type should be cast to a PLI_BYTE8 pointer.

The VCL reason_flag argument is one of two predefined constants that sets up the VCL callback mechanism to call the consumer routine under specific circumstances. The constant vcl_verilog_logic sets up the VCL to call the consumer routine whenever the monitored object changes logic value. The constant vcl_verilog_strength sets up the VCL to call the consumer routine when the monitored object changes logic value or logic strength.

An object can have any number of VCL monitors associated with it, as long as each monitor is unique in some way. VCL monitors can be deleted using the ACC routine acc_vcl_delete().

\subsection*{22.10.1 VCL objects}

The VCL shall monitor value changes for the following objects:
Scalar variables and bit-selects of vector variables
Scalar nets, unexpanded vector nets, and bit-selects of expanded vector nets
Integer, real, and time variables
Module ports
Primitive output or inout terminals
Named events
Note Adding a value change link to a module port is equivalent to adding a value change link to the loconn of the port. The vc_reason returned shall be based on the loconn of the port.

\subsection*{22.10.2 The VCL record definition}

Each time a consumer routine is called, it shall be passed a pointer to a record structure called vc_record. This structure shall contain information about the most recent change that occurred on the monitored object. The vc_record structure is defined in acc_user. h and is listed in Figure 56.
```

typedef struct t_vc_record
{
PLI_INT32 vc_reason;
PLI_INT32 vC_hightime;
PLI_INT32 vc_lowtime;
PLI_BYTE8 *user_data;
union
{
PLI_UBYTE8 logic_value;
double real_value;
handle vector_handle;
s_strengths strengths_s;
} out_value;
s_vc_record, *p_vc_record;

```

Figure 56-The VCL s_vc_record structure
The \(v c\) _reason field of vc _record shall contain a predefined integer constant that shall describe what type of change occurred. The constants that can be passed in the vc_reason field are described in Table 120.

Table 120—Predefined vc_reason constants
\begin{tabular}{|c|l|}
\hline Predefined vc_reason constant & \multicolumn{1}{c|}{ Description } \\
\hline logic_value_change & A scalar net or bit-select of a vector net changed logic value. \\
\hline strength_value_change & A scalar net or bit-select of a vector net changed logic value or strength. \\
\hline vector_value_change & A vector net or part-select of a vector net changed logic value. \\
\hline sregister_value_change & A scalar reg changed logic value. \\
\hline vregister_value_change & A vector reg or part-select of a vector reg changed logic value. \\
\hline integer_value_change & An integer variable changed value. \\
\hline real_value_change & A real variable changed value. \\
\hline time_value_change & A time variable changed value. \\
\hline event_value_change & A named event occured. \\
\hline
\end{tabular}

The \(v c_{\text {_hightime }}\) and \(v c_{-}\)lowtime fields of vc_record shall be 32-bit integers that shall contain the simulation time in the simulator's time units during which the change occurred, as follows:


The user_data field of vc_record shall be a PLI_BYTE8 pointer, and it shall contain the value specified as the user_data argument in the acc_vcl_add() ACC routine.

The out_value field of vc_record shall be a union of several data types. Only one data type shall be passed in the structure, based on the reason the callback occurred, as shown Table 121.

Table 121-Predefined out_value constants
\begin{tabular}{|c|c|l|}
\hline If vc_reason is & \begin{tabular}{c} 
The out_value shall \\
be a type of
\end{tabular} & \multicolumn{1}{|c|}{ Description } \\
\hline logic_value_change & PLI_UBYTE8 & \begin{tabular}{l} 
A predefined constant, from the following: vcl0 vcl1 \\
vclX vclx vclZ vclz
\end{tabular} \\
\hline strength_value_change & \begin{tabular}{c} 
s_strengths \\
structure
\end{tabular} & \begin{tabular}{l} 
A structure with logic and strength, as shown in Figure \\
57
\end{tabular} \\
\hline vector_value_change & handle & A handle to a vector net or part-select of a vector net \\
\hline sregister_value_change & PLI_UBYTE8 & \begin{tabular}{l} 
A predefined constant, from the following: vcl0 vcl1 \\
vclX vclx vclZ vclz
\end{tabular} \\
\hline vregister_value_change & handle & A handle to a vector reg or part-select of a vector reg \\
\hline integer_value_change & handle & A handle to an integer variable \\
\hline real_value_change & double & The value of a real variable \\
\hline time_value_change & handle & A handle to a time variable \\
\hline event_value_change & none & Event types have no value \\
\hline
\end{tabular}

When the vc_reason field of the vc_record is strength_value_change, the s_strengths structure fields of the out_value field of vc_recor \(\bar{d}\) shall contain the value. This structure shall contain three fields, as shown in Figure 57.
```

typedef struct t strengths
{
PLI_UBYTE8 logic_value;
PLI_UBYTE8 strength1;
PLI_UBYTE8 strength2;
} s_strengths, *p_strengths;

```

Figure 57-The VCL s_strengths structure
The values of the s_strengths structure fields are defined in Table 122.
Table 122—Predefined out_value constants
\begin{tabular}{|c|c|c|}
\hline s_strengths field & C data type & Description \\
\hline logic_value & PLI_UBYTE8 & A predefined constant, from the following: vel0 vel1 velX velx velZ velz \\
\hline strength1 strength2 & PLI_UBYTE8 & \begin{tabular}{l}
A predefined constant, from the following: \\
velSupply velWeak \\
velStrong vclMedium \\
vclPull velSmall \\
vclLarge velHighZ
\end{tabular} \\
\hline
\end{tabular}

The strength1 and strength2 fields of the s_strengths structure can represent
a) A known strength when strength1 and strength2 contain the same value, the signal strength shall be that value.
b) An ambiguous strength with a known logic_value when strength 1 and strength 2 contain different values and the logic_value contains either vcl0 or vel1, the signal strength shall be an ambiguous strength, where the strength1 value shall be the maximum possible strength and strength 2 shall be the minimum possible strength.
c) An ambiguous strength with an unknown logic_value when strength1 and strength2 contain different values and the logic_value contains vclX, the signal strength shall be an ambiguous strength, where the strength 1 value shall be the logic 1 component and strength 2 shall be the logic 0 component.

\subsection*{22.10.3 Effects of acc_initialize() and acc_close() on VCL consumer routines}

The ACC routines acc_initialize() and acc_close() shall reset all configuration parameters set by the routine acc_configure() back to default values. Care should be taken to ensure that the VCL consumer routine does not depend on any configuration parameters, as these parameters might not have the same value when a VCL callback occurs. Refer to 23.6 on acc_configure() for a list of routines that are affected by configuration parameters.

\subsection*{22.10.4 An example of using VCL ACC routines}

The following example contains three PLI routines: a checktf application, a calltf application, and a consumer routine. The example is based on the checktf and calltf applications both being associated with two user-defined system tasks, using the PLI interface mechanism described in Clause 21.
```

\$net_monitor(<net_name>,<net_name>, ...);
\$net_monitor_off(<<net_name>,<<net_name>, ...);

```

The checktf application, netmon_checktf, is shown below. This application performs syntax checking on instances of the user-defined system tasks to ensure there is at least one argument and that the arguments are valid net names.
```

PLI_INT32 netmon_checktf()
{
int i;
PLI_INT32 arg_cnt = tf_nump();
/* initialize the environment for access routines */
acc_initialize();
/* check number and type of task/function arguments */
if (arg_cnt == 0)
tf_error("$net_monitor[ _off] must have at least one argument");
    else
        for (i = 1; i <= arg_cnt; i++)
            if (acc_fetch_type\overline{(acc_handle_tfarg(i)) != accNet) {}
                tf_error("$\overline{net_monitōr[ _off] arg %d is not a net type",i);}
}
acc_close();
return(0);
}

```

The calltf application, netmon_calltf, follows. This application gets a handle to each task function argument and either adds or deletes a VCL monitor on the net. The application checks the data C argument associated with each system task name to determine whether the application was called by
\$net_monitor or \$net_monitor_off.
```

PLI INT32 netmon calltf(data)
PLI_INT32 data;
{
handle net;
PLI_INT32 netmon_consumer();
PLI_INT32 tfnum;
\#define ADD 0 /* data value associated with \$net_monitor */
\#define DELETE 1 /* data value associated with \$net_monitor_off */
/* initialize the environment for access routines */
acc initialize();
switch (data) /* see which system task name called this application
{
case ADD: /* called by \$net_monitor */
/* add a VCL flag to each net in the task/function argument list
tfnum = 1;
while ((net = acc_handle_tfarg(tfnum++)) != null)
{
/* add a VCL monitor; pass net pointer as user_data argument*/
acc_vcl_add(net, netmon_consumer, (PLI_BYTE8*)net,
vcl_verilog_logic);
}
break;
case DELETE: /* called by \$net_monitor_off */
tfnum = 1;
while ((net = acc_handle_tfarg(tfnum++)) != null)
{
/* delete the VCL monitor * /
acc_vcl_delete(net, netmon_consumer, (PLI_BYTE8*) net, vcl_veri:
}
break;
}
acc_close();
}

```

The consumer routine, netmon consumer, is shown in the following example. The consumer routine is called by the VCL callback mechanism. Since the checktf application only permits net data types to be used, the consumer routine only needs to check for scalar and vector net value changes when it is called. In this example, it is assumed that \$net_monitor is associated with a data value of 0 , and \$net_monitor_off is associated with a data value of 1 . Refer to 21.3.1 for a description of associating data values.
```

LI INT32 netmon consumer(vc record)
vc record vc record; /* record type passed to consumer routine */
PLI_BYTE8 net_value;
char value;
handle vector_value;
/* check reason VCL call-back occurred */
switch (vc_record->vc_reason)
{
case logic_value_change : /* scalar signal changed logic value */
{
net_value = vc_record->out_value.logic_value;
/* \overline{convert logíc value cons̄tant to a character for printing */}
switch (net_value)
{
case vcl0 : value = '0'; break;
case vcl1 : value = '1'; break;
case vclX : value = 'X'; break;
case vclZ : value = 'Z'; break;
}
io_printf("%d : %s = %c\n",
vc_record->vc_lowtime,
acc_fetch_name((handle) vc_record->user_data),
value);
break;
}
case vector_value_change :/* vector signal changed logic value */
{
vector_value = vc_record->out_value.vector_handle;
io_printf("%d : %s = %s\n",
vc_record->vc_lowtime,
ac\overline{c}_fetch_namē((handle) vc record->user data),
acc_fetch_value(vector_value, "%b",NULL) );
break;
}
}

```

\section*{23. ACC routine definitions}

This clause describes the PLI access (ACC) routines, explaining their function, syntax, and usage. The routines are listed in alphabetical order.

The following conventions are used in the definitions of the PLI routines described in Clause 23, Clause 25, and Clause 27.

Synopsis: A brief description of the PLI routine functionality, intended to be used as a quick reference when searching for PLI routines to perform specific tasks.

Syntax: The exact name of the PLI routine and the order of the arguments passed to the routine.
Returns: The definition of the value returned when the PLI routine is called, along with a brief description of what the value represents. The return definition contains the fields

Type: The data type of the C value that is returned. The data type is either a standard ANSI C type or a special type defined within the PLI.
Description: A brief description of what the value represents.
Arguments: The definition of the arguments passed with a call to the PLI routine. The argument definition contains the fields

Type: The data type of the C values that are passed as arguments. The data type is either a standard ANSI C type, or a special type defined within the PLI.
Name: The name of the argument used in the Syntax definition.
Description: A brief description of what the value represents.
All arguments shall be considered mandatory unless specifically noted in the definition of the PLI routine. Two tags are used to indicate arguments that may not be required:

Conditional: Arguments tagged as conditional shall be required only if a previous argument is set to a specific value, or if a call to another PLI routine has configured the PLI to require the arguments. The PLI routine definition explains when conditional arguments are required.
Optional: Arguments tagged as optional may have default values within the PLI, but they may be required if a previous argument is set to a specific value, or if a call to another PLI routine has configured the PLI to require the arguments. The PLI routine definition explains the default values and when optional arguments are required.

Related routines: A list of PLI routines that are typically used with, or provide similar functionality to, the PLI routine being defined. This list is provided as a convenience to facilitate finding information in this standard. It is not intended to be all-inclusive, and it does not imply that the related routines have to be used.

\section*{23.1 acc_append_delays()}

\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|r|}{acc_append_delays() for min:typ:max delays (accMinTypMaxDelays set to true )} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Add min:typ:max delay values to existing delay values for primitives, module paths, intermodule paths, timing checks or module input ports; the delay values are contained in an array.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_append_delays (object_handle, array_ptr)} \\
\hline \multicolumn{3}{|c|}{Type} & Description \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{1 if successful; 0 if an error is encountered} \\
\hline \multicolumn{2}{|r|}{Type} & Type Name & Descrip \\
\hline \multirow[t]{2}{*}{Arguments:} & handle & object_handle & Handle of a primitive, module ing check, module input port or \\
\hline & double address & array_ptr & Pointer to array of min:typ:ma the size of the array depends on setting of accPathDelayCoun \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_fetch_delays() to retrieve an objects delay values Use acc_replace_delays() to replace an object s delay values Use acc_configure() to set accPathDelayCount and accMinTypMaxDelays} \\
\hline
\end{tabular}

The ACC routine acc_append_delays() shall work differently depending on how the configuration parameter accMinTypMaxDelays is set. When this parameter is set to false, a single delay per transition shall be assumed, and delays shall be passed as individual arguments. For this single delay mode, the first syntax table in this section shall apply.

When accMinTypMaxDelays is set to true, acc_append_delays() shall pass one or more sets of minimum:typical:maximum delays contained in an array, rather than single delays passed as individual arguments. For this min:typ:max delay mode, the second syntax table in this section shall apply.

The number of delay values appended by acc_append_delays( () shall be determined by the type of object and the setting of configuration parameters. Refer to 22.8 for a description of how the number of delay values are determined.

The acc_append_delays() routine shall write delays in the timescale of the module that contains the object_handle.

When altering the delay via acc_append_delays() the value of the reject/error region will not be affected unless they exceed the value of the delay. If the reject/error limits exceed the delay they will be truncated down to the new delay limit.

The example shown in Figure 58 is an example of backannotation. It reads new delay values from a file called primdelay.dat and uses acc_append_delays() to add them to the current delays on a gate. The format of the file is shown below.

Figure 58-Using acc_append_delays() in single delay value mode
The example shown in Figure 59 shows how to append min:typ:max delays for a 2 -state primitive (no highimpedance state). The C application follows these steps:
a) Declares an array of nine double-precision floating-point values to hold three sets of min:typ:max values, one set each for rising transitions, falling transitions, and transitions to Z .
b) Sets the configuration parameter accMinTypMaxDelays to true to instruct acc_append_delays() to write delays in min:typ:max format.
c) Calls acc_append_delays() with a valid primitive handle and the array pointer.

Since the primitive to be used in this example does not have a high-impedance state, acc_append_delays() automatically appends just the rise and fall delay value sets. The last three array elements for the toZ delay values are not used. However, even though the last three array elements are not used with a 2 -state primitive,

For this example, the C application, append_mintypmax_delays, is associated through the ACC interface mechanism with a user-defined system task called \$appendprimdelays. A primitive with no Z state and new delay values are passed as task/function arguments to \$appendprimdelays as follows:

\$appendprimdelays( g1, 3.0, 5.0, 6.7, 2.4, 8.1, 9.1 );

```

\#include "acc_user.h"
PLI_INT32 append_mintypmax_delays()
{
handle prim;
double delay_array[9];
int i;
acc_configure(accMinTypMaxDelays, "true");
/* get handle for primitive * /
prim = acc_handle_tfarg(1);
/* store new delay values in array * /
for (i = 0; i < 9; i++)
delay_array[ i] = acc_fetch_tfarg(i+2);
/* append min:typ:max delays */
acc_append_delays (prim, delay_array);
}

```

Figure 59-Using acc_append_delays() in min:typ:max mode

\section*{23.2 acc_append_pulsere()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_append_pulsere ()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Add delays to existing pulse handling reject_limit and e_limit for a module path, intermodule path or module input port.} \\
\hline Syntax: & \multicolumn{3}{|l|}{\[
\begin{array}{r}
\text { acc_append_pulsere(object,r1,e1, r2,e2, r3,e3, r4,e4, r5,e5, r6,e6, } \\
r 7, \mathrm{e} 7 \mathrm{r} 8, \mathrm{e} 8, \mathrm{r} 9, \mathrm{e} 9, \mathrm{r} 10, \mathrm{e} 10, \mathrm{r} 11, \mathrm{e} 11, \mathrm{r} 12, \mathrm{e} 12)
\end{array}
\]} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{1 if successful; 0 if an error is encountered} \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Description \\
\hline & handle & object & Handle of module path, intermod port \\
\hline & double & r1...r12 & reject_limit values; the number of by accPathDelayCount \\
\hline & double & e1...e12 & e_limit values; the number of arg accPathDelayCount \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_fetch_pulsere() to get current pulse handling values Use acc_replace_pulsere() to replace existing pulse handling values Use acc_set_pulsere() to set pulse handling values as a percentage of the path delay Use acc_configure() to set accPathDelayCount} \\
\hline
\end{tabular}

The ACC routine acc_append_pulsere() shall add to an existing pulse handling reject_limit value and e_limit value for a module path, intermodule path and module input port. The reject_limit and e_limit values are used to control how pulses are propagated through paths.

A pulse is defined as two transitions that occur in a shorter period of time than the delay. Pulse control values determine whether a pulse should be rejected, propagated through to the output, or considered an error. The pulse control values consist of a reject_limit and an e_limit pair of values, where

The reject_limit shall set a threshold for determining when to reject a pulse any pulse less than the reject_limit shall not propagate.
The e_limit shall set a threshold for determining when a pulse is considered to be an error any pulse less than the e_limit and greater than or equal to the reject_limit shall propagate a logic x . A pulse that is greater than or equal to the e_limit shall propagate.

Table 123 illustrates the relationship between the reject_limit and the e_limit.
Table 123-Pulse control example
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ When } & \multicolumn{1}{c|}{ The pulse shall be } \\
\hline \begin{tabular}{l} 
reject_limit \(=10.5\) \\
e_limit \(=22.6\)
\end{tabular} & Rejected if \(<10.5\) \\
& An error if \(>=10.5\) and \(<22.6\) \\
& Passed if \(>=22.6\) \\
\hline
\end{tabular}

The following rules shall apply when specifying pulse handling values:
a) The value of reject_limit shall be less than or equal to the value of e_limit.
b) The reject_limit and e_limit shall not be greater than the delay.

If any of the limits do not meet the above rules, they shall be truncated.
The number of pulse control values that acc_append_pulsere() sets shall be controlled using the ACC routine acc_configure() to set the delay count configuration parameter accPathDelayCount, as shown in Table 124.

Table 124-How the value of accPathDelayCount affects acc_append_pulsere()
\begin{tabular}{|c|c|}
\hline When accPathDelayCount is & acc_append_pulsere() shall write \\
\hline 1 & One pair of reject_limit and e_limit values: one pair for all transitions, rl and el \\
\hline 2 & Two pairs of reject_limit and e_limit values: one pair for rise transitions, rl and el one pair for fall transitions, r2 and e2 \\
\hline 3 & Three pairs of reject_limit and e_limit values: one pair for rise transitions, rl and el one pair for fall transitions, r2 and e2 one pair for turn-off transitions, r 3 and e 3 \\
\hline \[
\begin{gathered}
\mathbf{6} \\
\text { (the default) }
\end{gathered}
\] & Six pairs of reject_limit and e_limit values a different pair for each possible transition among 0,1 , and Z : one pair for \(0->1\) transitions, rl and e1 one pair for \(1->0\) transitions, r 2 and e 2 one pair for \(0->Z\) transitions, r 3 and e 3 one pair for \(\mathrm{Z}->1\) transitions, r 4 and e4 one pair for \(1->\mathrm{Z}\) transitions, r 5 and e5 one pair for Z->0 transitions, r6 and e6 \\
\hline 12 & Twelve pairs of reject_limit and e_limit values a different pair for each possible transition among \(0,1, \mathrm{X}\), and Z : one pair for \(0->1\) transitions, r 1 and e1 one pair for \(1->0\) transitions, r 2 and e 2 one pair for \(0->Z\) transitions, r 3 and e3 one pair for \(Z->1\) transitions, r 4 and e 4 one pair for \(1->Z\) transitions, \(r 5\) and e5 one pair for \(\mathrm{Z}->0\) transitions, r 6 and e6 one pair for \(0->\) X transitions, r 7 and e7 one pair for \(\mathrm{X}->1\) transitions, r 8 and e8 one pair for \(1->\mathrm{X}\) transitions, r 9 and e9 one pair for \(\mathrm{X}->0\) transitions, r10 and e10 one pair for \(\mathrm{X}->\mathrm{Z}\) transitions, r11 and e11 one pair for Z->X transitions, r12 and e12 \\
\hline
\end{tabular}

The minimum number of pairs of reject_limit and e_limit arguments to pass to acc_append_pulsere() has to equal the value of accPathDelayCount. Any unused reject_limit and e_limit argument pairs shall be ignored by acc_append_pulsere() and can be dropped from the argument list.

If accPathDelayCount is not set explicitly, it shall default to six; therefore, six pairs of pulse reject_limit and e_limit arguments have to be passed when acc_append_pulsere() is called. Note that the value assigned to accPathDelayCount also affects acc_append_delays(), acc_fetch_delays(), acc_replace_delays(), acc_fetch_pulsere(), and acc_replace_pulsere().

Pulse control values shall be appended using the timescale of the module that contains the object handle.

\section*{23.3 acc_close()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_close()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Free internal memory used by ACC routines; reset all configuration parameters to default values.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_close()} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & void & No return & \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Description \\
\hline & None & & \\
\hline Related routines: & \multicolumn{2}{|l|}{Use acc_initialize() to initialize the ACC routine environment} & \\
\hline
\end{tabular}

The ACC routine acc_close() shall free internal memory used by ACC routines and reset all configuration parameters to default values. No other ACC routines should be called after calling acc_close(); in particular, ACC routines that are affected by acc_configure() should not be called.

Potentially, multiple PLI applications running in the same simulation session can interfere with each other because they share the same set of configuration parameters. To guard against application interference, both acc_initialize() and acc_close() reset all configuration parameters to their default values.

The example shown in Figure 60 presents a C language routine that calls acc_close() before exiting.
```

include "acc_user.h"
void show_versions()
/*initialize environment for ACC routines*/
acc_initialize();
/*show version of ACC routines and simulator */
io_printf("Running %s with %s\n",acc_version(), acc_product_version() )
acc_close();

```

Figure 60-Using acc_close()

\section*{23.4 acc_collect()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_collect()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Obtain an array of handles for all objects related to a particular reference object; get the number of objects collected.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_collect(acc_next_routine_name, object_handle, number_of_objects)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & handle array address & \multicolumn{2}{|l|}{An address pointer to an array of handles of the objects collected} \\
\hline \multirow{4}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & pointer to acc_next_ routine & acc_next_routine_name & Actual name (unquoted) of the acc_next_ routine that finds the objects to be collected \\
\hline & handle & object_handle & Handle of the reference object for acc_next_ routine \\
\hline & PLI_INT32 * & number_of_objects & Integer pointer where the count of objects collected shall be written \\
\hline Related routines: & \multicolumn{3}{|l|}{\begin{tabular}{l}
All acc_next_ routines except acc_next_topmod() \\
Use acc_free() to free memory allocated by acc_collect()
\end{tabular}} \\
\hline
\end{tabular}

The ACC routine acc_collect() shall scan through a reference object, such as a module, and collect handles to all occurrences of a specific target object. The collection of handles shall be stored in an array, which can then be used by other ACC routines.

The object associated with object_handle shall be a valid type of handle for the reference object required by the acc_next routine to be called.

The routine acc_collect() should be used in the following situations:
To retrieve data that can be used more than once
Instead of using nested or concurrent calls to acc_next_loconn(), acc_next_hiconn(), acc_next_load(), and acc_next_cell_load() routines

Otherwise, it can be more efficient to use the an acc_next_ routine directly.
The routine acc_collect() shall allocate memory for the array of handles it returns. When the handles are no longer needed, the memory can be freed by calling the routine acc_free().

The ACC routine acc_next_topmod() does not work with acc_collect(). However, top-level modules can be collected by passing acc_next_child() with a null reference object argument. For example:
acc_collect(acc_next_child, null, \&count);

The example shown in Figure 61 presents a C language routine that uses acc_collect() to collect and display all nets in a module.
```

\#include "acc_user.h"
PLI_INT32 display_nets()
{
handle *list_of_nets, module_handle;
PLI_INT32 net_count, i;
/*initialize environment for ACC routines*/
acc_initialize();
/* get handle for the module* /
module_handle = acc_handle_tfarg(1);
/*collect all nets in the module*/
list_of_nets = acc_collect(acc_next_net, module_handle, \&net_count);
/*display names of net instances*/
for(i=0; i < net count; i++)
io_printf("Net name is: %s\n", acc_fetch_name(list_of_nets[ i]));
/* free memory used by array list_of_nets*/
acc_free(list_of_nets);
acc_close();
}

```

Figure 61-Using acc_collect()

\section*{23.5 acc_compare_handles()}
\begin{tabular}{||l|l|l||}
\hline \multicolumn{2}{|c|}{ acc_compare_handles() } \\
\hline Synopsis: & Determine if two handles refer to the same object. \\
\hline Syntax: & acc_compare_handles (handle1, handle2) \\
\hline \multirow{4}{*}{ Returns: } & \multicolumn{2}{c|}{ Type } \\
\cline { 2 - 4 } & PLI_INT32 & Description \\
\hline \multirow{3}{*}{ Arguments: } & true if handles refer to the same object; false if different objects \\
\cline { 2 - 4 } & handle & Name \\
\hline & handle & handle1 \\
\hline
\end{tabular}

The ACC routine acc_compare_handles() shall determine if two handles refer to the same object. In some cases, two different handles might reference the same object if each handle is retrieved in a different way for example, if an acc_next routine returns one handle and acc_handle_object() returns the other.

The \(\mathrm{C}==\) operator cannot be used to determine if two handles reference the same object.
```

if (handle1 == handle2) /* this does not work */

```

The example shown in Figure 62 uses acc_compare_handles() to determine if a primitive drives the specified output of a scalar port of a module.
```

\#include "acc_user.h"
PLI_INT32 prim_drives_scalar_port(prim, mod, port_num)
handle prim, mod;
PLI INT32 port num;
{
/* retrieve net connected to scalar port */
handle port = acc_handle_port(mod, port_num);
handle port_conn = acc_next_loconn(port, null);
/* retrieve net connected to primitive output */
handle out_term = acc_handle_terminal(prim, 0);
handle prim}_conn = ac\overline{c}_handl\overline{e_conn(out_term);
/* compare handles */
if (acc_compare_handles(port_conn, prim_conn) )
return(true);
else
return(false);
}

```

Figure 62-Using acc_compare_handles()

\section*{23.6 acc_configure()}


The ACC routine acc_configure() shall set parameters that control the operation of various ACC routines. Tables 125 through 134 describe each parameter and its set of values. Note that a call to either acc_initialize() or acc_close() shall set each configuration parameter back to its default value.

Table 125-accDefaultAttr0 configuration parameter
\begin{tabular}{|c|c|c|c|}
\hline \multirow{3}{*}{accDefaultAttr0} & Set of values & Effect & Default \\
\hline & "true" & acc_fetch_attribute() shall return zero when it does not find the attribute requested and shall ignore the default_value argument & \multirow[t]{2}{*}{"false"} \\
\hline & "false" & acc_fetch_attribute() shall return the value passed as the default_value argument when it does not find the attribute requested & \\
\hline
\end{tabular}

Table 126-accDevelopmentVersion configuration parameter
\begin{tabular}{|l|l|l|c|}
\hline \multirow{6}{*}{ accDevelopmentVersion } & \multicolumn{1}{|c|}{ Set of values } & \multicolumn{1}{c|}{ Effect } & \multicolumn{1}{c|}{ Default } \\
\cline { 2 - 4 } & \begin{tabular}{l} 
Quoted string of letters, \\
numbers, and the period \\
character that form a valid \\
PLI version, such as: \\
IEEE 1364 PLI
\end{tabular} & \begin{tabular}{l} 
None (can be used to document \\
which version of ACC routines was \\
used to develop a PLI application)
\end{tabular} & \begin{tabular}{l} 
Current version \\
of ACC routines
\end{tabular} \\
\hline \begin{tabular}{l} 
Software vendors can \\
define version strings spe- \\
cific to their products
\end{tabular} & & \\
\hline
\end{tabular}

Table 127-accDisplayErrors configuration parameter
\begin{tabular}{|c|c|l|c|}
\hline \multirow{4}{*}{ accDisplayErrors } & Set of values & \multicolumn{1}{|c|}{ Effect } & Default \\
\cline { 2 - 4 } & "true" & \begin{tabular}{l} 
ACC routines shall display error \\
messages
\end{tabular} & "true" \\
\cline { 2 - 3 } & "false" & \begin{tabular}{l} 
ACC routines shall not display error \\
messages
\end{tabular} & \\
\hline
\end{tabular}

Table 128-accDisplayWarnings configuration parameter
\begin{tabular}{|c|c|l|c|}
\hline \multirow{4}{*}{ accDisplayWarnings } & Set of values & \multicolumn{1}{|c|}{ Effect } & Default \\
\cline { 2 - 4 } & "true" & \begin{tabular}{l} 
ACC routines shall display warning mes- \\
sages
\end{tabular} & "false" \\
\cline { 2 - 3 } & "false" & \begin{tabular}{l} 
ACC routines shall not display warning \\
messages
\end{tabular} & \\
\hline
\end{tabular}

Table 129-accEnableArgs configuration parameter
\begin{tabular}{|c|c|c|c|}
\hline \multirow{7}{*}{accEnableArgs} & Set of values & Effect & Default \\
\hline & acc_handle_modpath & acc_handle_modpath() shall recognize its optional arguments & ```
no_acc_handle_modpath
    no_acc_handle_tchk
``` \\
\hline & no_acc_handle_modpath & acc_handle_modpath() shall ignore its optional arguments & no_acc_set_scope \\
\hline & acc_handle_tchk & acc_handle_tchk() shall recognize its optional arguments & \\
\hline & no_acc_handle_tchk & acc_handle_tchk() shall ignore its optional arguments & \\
\hline & acc_set_scope & acc_set_scope() shall recognize its optional arguments & \\
\hline & no_acc_set_scope & acc_set_scope() shall ignore its optional arguments & \\
\hline
\end{tabular}

Table 130-accMapToMipd configuration parameter
\begin{tabular}{|c|c|c|c|}
\hline \multirow{4}{*}{accMapToMipd} & Set of values & Effect & Default \\
\hline & "max" & acc_replace_delays() and acc_append_delays() shall map the longest intermodule path delay to the MIPD & "max" \\
\hline & "min" & acc_replace_delays() and acc_append_delays() shall map the shortest intermodule path delay to the MIPD & \\
\hline & "latest" & acc_replace_delays() and acc_append_delays() shall map the last intermodule path delay to the MIPD & \\
\hline
\end{tabular}

Table 131-accMinTypMaxDelays configuration parameter
\begin{tabular}{|c|c|l|c|}
\hline \multirow{5}{*}{ accMinTypMaxDelays } & Set of values & \multicolumn{1}{c|}{ Effect } & Default \\
\cline { 2 - 4 } & "true" & \begin{tabular}{l} 
acc_append_delays(), acc_fetch_delays(), \\
acc_replace_delays(0, ac_append_pulsere(), \\
acc_fetch_pulsere(), and acc_replace_pulsere() \\
shall use_min:typ:max delay sets
\end{tabular} & "false" \\
\cline { 3 - 4 } & "false" & \begin{tabular}{l} 
acc_append_delays(), acc_fetch_delays(), \\
acc_replace_delays(), acc_append_pulsere(), \\
acc_fetch_pulsere(), and acc_replace_pulsere() \\
shall use a single delay value
\end{tabular} & \\
\hline
\end{tabular}

Table 132-accPathDelayCount configuration parameter
\begin{tabular}{|c|c|c|c|}
\hline \multirow{6}{*}{accPathDelayCount} & Set of values & Effect & Default \\
\hline & "1" & ```
acc_append_delays(), acc_fetch_delays(),
acc_replace_delays(), acc_append_pulsere(),
acc_fetch_pulsere(), and
acc_replace_pulsere()
shall use 1 delay value or value set
``` & "6" \\
\hline & "2" & ```
acc_append_delays(), acc_fetch_delays(),
acc_replace_delays(), acc_append_pulsere(),
acc_fetch_pulsere(), and
acc_replace_pulsere()
shall use 2 delay values or value sets
``` & \\
\hline & \(" 3 "\) & ```
acc_append_delays(), acc_fetch_delays(),
acc_replace_delays(), acc_append_pulsere(),
acc_fetch_pulsere(), and
acc_replace_pulsere()
shall use 3 delay values or value sets
``` & \\
\hline & "6" & ```
acc_append_delays(), acc_fetch_delays(),
acc_replace_delays(), acc_append_pulsere(),
acc_fetch_pulsere(), and
acc_replace_pulsere()
shall use 6 delay values or value sets
``` & \\
\hline & "12" & ```
acc_append_delays(), acc_fetch_delays(),
acc_replace_delays(), acc_append_pulsere(),
acc_fetch_pulsere(), and
acc_replace_pulsere()
shall use 12 delay values or value sets
``` & \\
\hline
\end{tabular}

Table 133-accPathDelimStr configuration parameter
\begin{tabular}{|l|l|l|c|}
\hline & \multicolumn{1}{c|}{ Set of values } & \multicolumn{1}{c|}{ Effect } & Default \\
\cline { 2 - 4 } accPathDelimStr & \(\begin{array}{l}\text { Quoted string of } \\
\text { letters, numbers, \$ } \\
\text { or_ }\end{array}\) & \(\begin{array}{l}\text { acc_fetch_name(), acc_fetch_fullname(), } \\
\text { and acc_fetch_attribute() shall use the } \\
\text { string literal as the delimiter separating the } \\
\text { source and destination in module path } \\
\text { names }\end{array}\) & "\$"
\end{tabular}\(\}\)

Table 134-accToHiZDelay configuration parameter
\begin{tabular}{|c|c|c|c|}
\hline \multirow{5}{*}{accToHiZDelay} & Set of values & Effect & Default \\
\hline & "average" & acc_append_delays() and acc_replace_delays() shall derive turn-off delays from the average of the rise and fall delays & "from_user" \\
\hline & "max" & acc_append_delays() and acc_replace_delays() shall derive turn-off delays from the larger of the rise and fall delays & \\
\hline & "min" & acc_append_delays() and acc_replace_delays() shall derive turn-off delays from the smaller of the rise and fall delays & \\
\hline & "from_user" & acc_append_delays() and acc_replace_delays() shall derive turn-off delays from user-supplied argument(s) & \\
\hline
\end{tabular}

The example shown in Figure 63 presents a C language application that obtains the load capacitance of all scalar nets connected to the ports in a module. This application uses acc_configure() to direct acc_fetch_attribute() to return zero if a load capacitance is not found for a net; as a result, the third argument, default_value, can be dropped from the call to acc_fetch_attribute().
```

\#include "acc_user.h"
PLI_INT32 display_load_capacitance()
handle module_handle, port_handle, net_handle;
double cap_val;
/*initialize environment for ACC routines*/
acc_initialize();
/*configure acc_fetch_attribute to return 0 when it does not find*/
/* the attribute*/
acc_configure(accDefaultAttr0, "true");
/*get handle for module* /
module_handle = acc_handle_tfarg(1);
/*scan all ports in module; display load capacitance*/
port handle = null;
while(port_handle = acc_next_port(module_handle, port_handle) )
{
/*ports are scalar, so pass "null" to get single net connection*/
net_handle = acc_next_loconn(port_handle, null);
/*since accDefaultAttr0 is "true", drop default value argument*/
cap_val = acc_fetch_attribute (net_handle,"LoadCap_" );
if (!acc_error_flag) default_value
io_printf("Load capacitance of net \#%d = %1f\n",
acc_fetch_index(port_handle), cap_val);
}
acc_close();

```

Figure 63-Using acc_configure() to set accDefaultAttr0

The example shown in Figure 64 presents a C language application that displays the name of a module path. It uses acc_configure() to set accEnableArgs and, therefore, forces acc_handle_modpath() to ignore its null name arguments and recognize its optional handle arguments, src_handle and dst_handle.

The example shown in Figure 65 fetches the rise and fall delays of each path in a module and backannotates the maximum delay value as the delay for all transitions. The value of accPathDelayCount specifies the minimum number of arguments that have to be passed to routines that read or write delay values. By setting accPathDelayCount to the minimum number of arguments needed for acc_fetch_delays() and again for acc_replace_delays(), all unused arguments can be eliminated from each call.
```

\#include "acc_user.h"
PLI_INT32 set_path_delays()
{
handle mod_handle;
handle path_handle;
double rise_delay,fall_delay,max_delay;
/*initialize environment for ACC routines*/
acc_initialize();
/*get handle to module*/
mod_handle = acc_handle_tfarg(1);
/*fetch rise delays for all paths in module "top.m1"*/
path_handle = null;
while(path_handle = acc_next_modpath(mod_handle, path_handle) )
{
/*configure accPathDelayCount for rise and fall delays only*/
acc_configure(accPathDelayCount, "2");
acc_fetch_delays(path_handle, \&rise_delay, \&fall_delay); arguments are needed
/*find the maximum of the rise and fall delays*/
max_delay = (rise_delay > fall_delay) ? rise_delay : fall_delay;
/*configure accPathDelayCount to apply one delay for all transitions*/
acc_configure(accPathDelayCount, "1");
acc_replace_delays(path_handle, max_delay);
}
acc_close(); argument is needed
}

```

Figure 65-Using acc_configure() to set accPathDelayCount

The example shown in Figure 66 shows how accToHiZDelay is used to direct acc_replace_delays() to derive the turn-off delay for a Z-state primitive automatically as the smaller of its rise and fall delays.
```

\#include "acc_user.h"
PLI_INT32 set_buf_delays()
{
handle primitive_handle;
handle path_handle;
double added_rise, added_fall;
/*initialize environment for ACC routines*/
acc_initialize();
/*configure accToHiZDelay so acc_append_delays derives turn-off */
/* delay from the smaller of the rise and fall delays*/
acc_configure(accToHiZDelay, "min");
/*get handle to Z-state primitive*/
primitive_handle = acc_handle_tfarg(1);
/*get delay values*/
added_rise = tf_getrealp(2);
added_fall = tf_getrealp(3);
acc_append_delays(primitive_handle, added_rise, added_fall);
acc_close();
}

```

Figure 66-Using acc_configure() to set accToHiZDelay

\section*{23.7 acc_count()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_count()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Count the number of objects related to a particular reference object.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_count (acc_next_routine_name, object_handle)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{Number of objects} \\
\hline \multirow{3}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & pointer to an acc_next_ routine & acc_next_routine_name & Actual name (unquoted) of the acc_next_routine that finds the objects to be counted \\
\hline & handle & object_handle & Handle of the reference object for the acc_next_ routine \\
\hline Related routines: & \multicolumn{3}{|l|}{All acc_next_routines except acc_next_topmod()} \\
\hline
\end{tabular}

The ACC routine acc_count() shall find the number of objects that exist for a specific acc_next_ routine with a given reference object. The object associated with object_handle shall be a valid reference object for the type acc_next_routine to be called.

Note that the ACC routine acc_next_topmod() does not work with acc_count(). However, top-level modules can be counted using acc_next_child() with a null reference object argument. For example:
```

acc_count(acc_next_child, null);

```

The example shown in Figure 67 uses acc_count() to count the number of nets in a module.
```

include "acc_user.h"
'LI_INT32 count_nets()
handle module_handle;
PLI_INT32 number_of_nets;
/*initialize environment for ACC routines*/
acc_initialize();
/* get handle for module* /
module_handle = acc_handle_tfarg(1);
/* count and display number of nets in the module* /
number_of_nets = acc_count(acc_next_net, module_handle);
io_printf("number of nets = %d\n", \overline{number_of_nets);}
acc_close();

```

Figure 67-Using acc_count()

\section*{23.8 acc_fetch_argc()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_fetch_argc()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the number of command-line arguments supplied with a Verilog software tool invocation.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_fetch_argc()} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type Description} & \\
\hline & PLI_INT32 & \multicolumn{2}{|l|}{Number of command-line arguments} \\
\hline \multirow[b]{2}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & None & & \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_fetch_argv() to get a character string array of the invocation options} \\
\hline
\end{tabular}

The ACC routine acc_fetch_arge() shall obtain the number of command-line arguments given on a Verilog software product invocation command line.

The example shown in Figure 68 uses acc_fetch_arge() to determine the number of invocation arguments used.
```

Finclude "acc_user.h"
Finclude <strīng.h> /* string.h is implementation dependent */
?LI_BYTE8* my_scan_plusargs(str)
PLI_BYTE8 *str;
PLI_INT32 i;
int length = strlen(str);
PLI_BYTE8 * curStr;
PLI_BYTE8 **argv = acc_fetch_argv();
for (i = acc_fetch_argc()-1; i>0; i--)
{
curStr = argv[ i] ;
if ((curStr[ 0] == ' +') \&\& (!strncmp(curStr+1,str,length)))
{
PLI_BYTE8 * retVal;
length = strlen(\&(curStr[ length]) + 1);
retVal = (PLI_BYTE8 *)malloc(sizeof(PLI_BYTE8) * length);
strcpy(retVal, \&(curStr[ length]));
return(retVal);
}
}
return(null);

```

Figure 68-Using acc_fetch_argc()

\section*{23.9 acc_fetch_argv()}


The ACC routine acc_fetch_argv() shall obtain an array of character pointers that make up the commandline arguments.

The format of the argv array is that each pointer in the array shall point to a NULL terminated character array which contains the string located on the tool's invocation command line. There shall be argc entries in the argv array. The value in entry zero shall be the tool s name.

The argument following a -f argument shall contain a pointer to a NULL terminated array of pointers to characters. This new array shall contain the parsed contents of the file. The value in entry zero shall contain the name of the file. The remaining entries shall contain pointers to NULL terminated character arrays containing the different options in the file. The last entry in this array shall be a NULL. If one of the options is a \(-f\) then the next pointer shall behave the same as described above.

The example shown in Figure 69 uses acc_fetch_argv() to retrieve the invocation arguments used.
```

Finclude "acc_user.h"
Finclude <string.h> /* string.h is implementation dependent */
?LI_BYTE8* my_scan_plusargs(str)
PLI_BYTE8 *str;
PLI INT32 i;
int - length = strlen(str);
PLI_BYTE8 *curStr;
PLI_BYTE8 **argv = acc_fetch_argv();
for (i = acc_fetch_argc()-1; i>0; i--)
{
curStr = argv[ i] ;
if ((curStr[ 0] == ' +') \&\& (!strncmp(curStr+1,str,length)))
{
PLI_BYTE8 *retVal;
length = strlen(\&(curStr[ length]) + 1);
retVal = (PLI_BYTE8 *)malloc(sizeof(PLI_BYTE8) * length);
strcpy(retVal, \&(curStr[ length] ));
return(retVal);
}
}
return(null);

```

Figure 69-Using acc_fetch_argv()

\subsection*{23.10 acc_fetch_attribute()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_fetch_attribute()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the value of a parameter or specparam named as an attribute in the Verilog source description.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_fetch_attribute(object_handle, attribute_string, default_value)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & double & \multicolumn{2}{|l|}{Value of the parameter or specparam} \\
\hline \multicolumn{2}{|r|}{Type} & Name & Description \\
\hline \multirow[t]{2}{*}{Arguments:} & handle & object_handle & Handle of a named object \\
\hline & quoted string or PLI_BYTE8 * & attribute_string & Literal string or character string pointer with the attribute portion of the parameter or specparam declaration \\
\hline Optional & double & default_value & Double-precision value to be returned if the attribute is not found (depends on accDefaultAttr0) \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_fetch_attribute_int() to get an attribute value as an integer Use acc_fetch_attribute_str() to get an attribute value as a string Use acc_configure(accDefaultAttr0...) to set default value returned when attribute is not found Use acc_fetch_paramtype() to get the data type of the parameter value Use acc_fetch_paramval() to get parameters or specparam values not declared in attribute/object format} \\
\hline
\end{tabular}

The ACC routine acc_fetch_attribute() shall obtain the value of a parameter or specparam that is declared as an attribute in the Verilog HDL source description. The value shall be returned as a double.

Any parameter or specparam can be an attribute by naming it in one of the following ways:
As a general attribute associated with more than one object in the module where the parameter or specparam attribute is declared
As a specific attribute associated with a particular object in the module where the parameter or specparam attribute is declared

Each of these methods uses its own naming convention, as described in Table 135. For either convention, attribute_string shall name the attribute and shall be passed as the second argument to acc_fetch_attribute(). The object_name shall be the actual name of a design object in a Verilog HDL source description.

Table 135-Naming conventions for attributes
\begin{tabular}{|c|c|c|}
\hline For & Naming convention & Example \\
\hline A general attribute & \begin{tabular}{l}
attribute_string \\
A mnemonic name that describes the attribute
\end{tabular} & specparam DriveStrength\$ = 2.8; attribute_string is DriveStrength\$ \\
\hline A specific attribute associated with a particular object & \begin{tabular}{l}
attribute_string object_name \\
Concatenate a mnemonic name that describes the attribute with the name of the object
\end{tabular} & ```
    specparam DriveStrength$g1 = 2.8;
attribute_string is DriveStrength$
object_name is g1
``` \\
\hline
\end{tabular}

The ACC routine acc_fetch_attribute() shall identify module paths in terms of their sources and destinations in the following format:
\begin{tabular}{|l|l|l|}
\hline source & path_delimiter & destination \\
\hline
\end{tabular}

The acc_fetch_attribute() routine shall look for module path names in this format, and acc_fetch_name() and acc_fetch_fullname() shall return names of module paths in this format. Therefore, the same naming convention should be used when associating an attribute with a module path. Note that names of module paths with multiple sources or destinations shall be derived from the first source or destination only.

By default, the path_delimiter used in path names is the "\$" character. This default can be changed by using the ACC routine acc_configure() to set the delimiter parameter accPathDelimStr to another character string.

The examples in Table 136 show how to name module paths using different delimiter strings.
Table 136-Example module path names using delimiter strings
\begin{tabular}{|c|c|c|}
\hline For module path & If accPathDelimStr is & Then the module path name is \\
\hline \((\mathrm{a} \quad\) > q\()=10\); & "\$" & a\$q \\
\hline \((\mathrm{b}\) *> q1, q2) = 8; & "_\$_" & b_\$_q1 \\
\hline (d,e,f *> r,s) \(=8\); & "_" & d_r \\
\hline
\end{tabular}

The following example shows an attribute name for a particular module path object:
Given the module path: \(\quad(\mathrm{a}=>\mathrm{q})=10\);
An attribute name is: \(\quad\) specparam RiseStrength\$a\$q \(=20\);

In this example, the attribute_string is RiseStrength\$, the object_name is a\$q, and the path_delimiter is \$ (the default path delimiter).

The following flowchart illustrates how acc_fetch_attribute() shall work:


This flowchart shows that when acc_fetch_attribute() finds the attribute requested, it returns the value of the attribute as a double-precision floating-point number.
1) The routine shall first look for the attribute name that concatenates attribute_string with the name associated with object_handle. For example, to find an attribute InputLoad\$ for a net n1, acc_fetch_attribute() would search for InputLoad\$n1.
2) If acc_fetch_attribute() does not find the attribute associated with the object specified with object_handle, the routine shall then search for a name that matches attribute_string. Assume that, in the previous example, acc_fetch_attribute() does not find InputLoad\$n1. It would then look for InputLoad\$. Other variants of that name, such as InputLoad\$n3 or InputLoad\$n, shall not be considered matches.
3) Failing both search attempts, the routine acc_fetch_attribute() shall return a default value. The default value is controlled by using the ACC routine acc_configure() to set or reset the configuration parameter accDefaultAttr0 as shown in Table 137.

Table 137-Controlling the default value returned by acc_fetch_attribute()
\begin{tabular}{|c|l|}
\hline When accDefaultAttr0 is & \multicolumn{1}{c|}{ acc_fetch_attribute() shall return } \\
\hline true & \begin{tabular}{l} 
Zero when the attribute is not found; the \\
default_value argument can be dropped
\end{tabular} \\
\hline false & \begin{tabular}{l} 
The value passed as the default_value argument \\
when the attribute is not found
\end{tabular} \\
\hline
\end{tabular}

The example shown in Figure 70 presents a C language application that uses acc_fetch_attribute() to obtain the load capacitance of all scalar nets connected to the ports in a module. Note that acc_fetch_attribute() does not require its third argument, default_value, because acc_configure() is used to set accDefaultAttr0 to true.
```

include "acc_user.h"
LI_INT32 display_load_capacitance()
handle module_handle, port_handle, net_handle;
double cap_val;
/*initialize environment for ACC routines*/
acc_initialize();
/*configure acc_fetch_attribute to return 0 when it does not find*/
/*the attribute*/
acc_configure(accDefaultAttr0, "true");
/*get handle for module* /
module_handle = acc_handle_tfarg(1);
/*scan all ports in module; display load capacitance*/
port_handle = null;
while(port_handle = acc_next_port(module_handle, port_handle) )
{
/*ports are scalar, so pass "null" to get single net connection*/
net_handle = acc_next_loconn(port_handle, null);
/*since accDefaultAttr0 is "true", drop default_value argument*/
cap_val = acc_fetch_attribute (net_handle,"LoadCap_");
if (!acc_error_flag)
io_printf("Load capacitance of net \#%d = %1f\n",
acc_fetch_index(port_handle), cap_val);
}
acc_close();

```

Figure 70—Using acc_fetch_attribute()

\subsection*{23.11 acc_fetch_attribute_int()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_fetch_attribute_int()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the integer value of a parameter or specparam named as an attribute in the Verilog source description.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_fetch_attribute_int (object_handle, attribute_string, default_value)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{Value of the parameter or specparam} \\
\hline \multirow{3}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & handle & object_handle & Handle of a named object \\
\hline & quoted string or PLI_BYTE8 * & attribute_string & Literal string or character string pointer with the attribute portion of the parameter or specparam declaration \\
\hline Optional & PLI_INT32 & default_value & Integer value to be returned if the attribute is not found (depends on accDefaultAttr0) \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_fetch_attribute() to get an attribute value as a double Use acc_fetch_attribute_str() to get an attribute value as a string Use acc_configure(accDefaultAttr0...) to set default value returned when attribute is not found Use acc_fetch_paramtype() to get the data type of the parameter value Use acc_fetch_paramval() to get parameters or specparam values not declared in attribute/object format} \\
\hline
\end{tabular}

The ACC routine acc_fetch_attribute_int() shall obtain the value of a parameter or specparam that is declared as an attribute in the Verilog HDL source description. The value shall be returned as an integer.

Any parameter or specparam can be an attribute. Refer to 23.10 for a description of attribute naming and how attribute values are fetched.

\subsection*{23.12 acc_fetch_attribute_str()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_fetch_attribute_str()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the value of a parameter or specparam named as an attribute in the Verilog source description.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_fetch_attribute_str(object_handle, attribute_string, default_value)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_BYTE8 * & \multicolumn{2}{|l|}{Value of the parameter or specparam} \\
\hline \multirow{3}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & handle & object_handle & Handle of a named object \\
\hline & quoted string or PLI_BYTE8 * & attribute_string & Literal string or character string pointer with the attribute portion of the parameter or specparam declaration \\
\hline Optional & quoted string or PLI_BYTE8 * & default_value & Character string value to be returned if the attribute is not found (depends on accDefaultAttr0) \\
\hline Related routines: & \multicolumn{3}{|l|}{\begin{tabular}{l}
Use acc_fetch_attribute() to get an attribute value as a double \\
Use acc_fetch_attribute_int() to get an attribute value as an integer \\
Use acc_configure(accDefaultAttr0...) to set default value returned when attribute is not found \\
Use acc_fetch_paramtype() to get the data type of the parameter value \\
Use acc_fetch_paramval() to get parameters or specparam values not declared in attribute/object format
\end{tabular}} \\
\hline
\end{tabular}

The ACC routine acc_fetch_attribute_str() shall obtain the value of a parameter or specparam that is declared as an attribute in the Verilog HDL source description. The value shall be returned as a pointer to a character string. The return value for this routine is placed in the ACC internal string buffer. See 22.9 for explanation of strings in ACC routines.

Any parameter or specparam can be an attribute. Refer to 23.10 for a description of attribute naming and how attribute values are fetched.

\subsection*{23.13 acc_fetch_defname()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_fetch_defname()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the definition name of a module instance or primitive instance.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_fetch_defname(object_handle)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_BYTE8 * & \multicolumn{2}{|l|}{Pointer to a character string containing the definition name} \\
\hline \multirow[b]{2}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & handle & object_handle & Handle of the module instance or primitive instance \\
\hline Related routines & \multicolumn{3}{|l|}{Use acc_fetch_name() to display the instance name of an object} \\
\hline
\end{tabular}

The ACC routine acc_fetch_defname() shall obtain the definition name of a module instance or primitive instance. The definition name is the declared name of the object as opposed to the instance name of the object. In the illustration shown below, the definition name is "dff", and the instance name is " i 15 ".


The return value for this routine is placed in the ACC internal string buffer. See 22.9 for explanation of strings in ACC routines.

The example shown in Figure 71 presents a C language application that uses acc_fetch_defname() to display the definition names of all primitives in a module.
```

include "acc_user.h"
-oid get_primitive_definitions(module_handle)
.andle module_handle;
handle prim_handle;
/*get and display defining names of all primitives in the module*/
prim_handle = null;
while(prim_handle = acc_next_primitive(module_handle,prim_handle))
io_printf("primitive definition is %s\n",
acc_fetch_defname(prim_handle) );

```

Figure 71-Using acc_fetch_defname()

\subsection*{23.14 acc_fetch_delay_mode()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|r|}{acc_fetch_delay_mode()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the delay mode of a module instance.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_fetch_delay_mode (module_handle)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type Description} & \\
\hline & PLI_INT32 & A predefined integer co accDelayModeNone accDelayModePath & representing the delay mode of th accDelayModeZero accDelayModeDistrib \\
\hline & Type & Name & Descri \\
\hline Arguments: & handle & module_handle & Handle to a module instance \\
\hline
\end{tabular}

The ACC routine acc_fetch_delay_mode() shall return the delay mode of a module or cell instance. The delay mode determines how delays are stored for primitives and paths within the module or cell. The routine shall return one of the predefined constants given in Table 138.

Table 138—Predefined constants used by acc_fetch_delay_mode()
\begin{tabular}{|l|l|}
\hline Predefined constant & \multicolumn{1}{c|}{ Description } \\
\hline accDelayModeNone & No delay mode specified. \\
\hline accDelayModeZero & All primitive delays are zero; all path delays are ignored. \\
\hline accDelayModeUnit & All primitive delays are one; all path delays are ignored. \\
\hline accDelayModeDistrib & \begin{tabular}{l} 
If a logical path has both primitive delays and path delays spec- \\
ified, the primitive delays shall be used.
\end{tabular} \\
\hline
\end{tabular}

Table 138-Predefined constants used by acc_fetch_delay_mode() (continued)
\begin{tabular}{|c|l|}
\hline Predefined constant & \multicolumn{1}{c|}{ Description } \\
\hline accDelayModePath & \begin{tabular}{l} 
If a logical path has both primitive delays and path delays spec- \\
ified, the path delays shall be used.
\end{tabular} \\
\hline accDelayModeMTM & \begin{tabular}{l} 
If this property is true, Minimum:Typical:Maximum delay sets \\
for each transition are being stored; if this property is false, a \\
single delay for each transition is being stored.
\end{tabular} \\
\hline
\end{tabular}

Figure 72 uses acc_fetch_delay_mode() to retrieve the delay mode of all children of a specified module.
```

\#include "acc_user.h"
PLI_INT32 display_delay_mode()
{
handle mod, child;
/*reset environment for ACC routines*/
acc_initialize();
/*get module passed to user-defined system task*/
mod = acc_handle_tfarg(1);
/*find and display delay mode for each module instance*/
child = null;
while(child = acc_next_child(mod, child))
{
io_printf("Module %s set to: ",acc_fetch_fullname(child));
swītch(acc_fetch_delay_mode(child) )
{
case accDelayModePath:
io_printf(" path delay mode\n");
break;
case accDelayModeDistrib:
io_printf(" distributed delay mode\n");
break;
}
}
}

```

Figure 72—Using acc_fetch_delay_mode()

\subsection*{23.15 acc_fetch_delays()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|r|}{acc_fetch_delays() for single delay values (accMinTypMaxDelays set to false )} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get existing delays for primitives, module paths, timing checks, module input ports, and intermodule paths.} \\
\hline \begin{tabular}{l}
Syntax: \\
Primitives
\end{tabular} & \multicolumn{3}{|l|}{acc_fetch_delays(object_handle, rise_delay, fall_delay, z_delay)} \\
\hline Module paths Intermodule paths Ports or port bits & \multicolumn{3}{|l|}{```
acc_fetch_delays(object_handle,
    d1,d2,d3,d4,d5,d6,d7,d8,d9,d10,d11,d12)
```} \\
\hline Timing checks & \multicolumn{3}{|l|}{acc_fetch_delays (object_check_handle, limit)} \\
\hline & \multicolumn{2}{|l|}{Type} & Description \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{1 if successful; 0 if an error occurred} \\
\hline \multirow{3}{*}{Arguments:} & Type & Name & Description \\
\hline & handle & object_handle & Handle of a primitive, module path, timing check, module input port, bit of a module input port, or intermodule path \\
\hline & double * & rise_delay fall_delay & Rise and fall delay for 2-state primitive or 3-state primitive \\
\hline \multirow[t]{2}{*}{Conditional} & double * & z_delay & Turn-off (to Z) transition delay for 3-state primitives \\
\hline & double * & d1 & \begin{tabular}{l}
For module/intermodule paths and input ports/port bits: If accPathDelayCount is set to 1 : \\
delay for all transitions \\
If accPathDelayCount is set to 2 or 3: rise transition delay \\
If accPathDelayCount is set to 6 or 12 : \(\mathbf{0}->\mathbf{1}\) transition delay
\end{tabular} \\
\hline Conditional & double * & d2 & \begin{tabular}{l}
For module/intermodule paths and input ports/port bits: If accPathDelayCount is set to 2 or 3 : \\
fall transition delay \\
If accPathDelayCount is set to 6 or 12 : \\
1->0 transition delay
\end{tabular} \\
\hline Conditional & double * & d3 & \begin{tabular}{l}
For module/intermodule paths and input ports/port bits: If accPathDelayCount is set to 3 : \\
turn-off transition delay \\
If accPathDelayCount is set to 6 or 12 : \\
\(\mathbf{0}->\mathbf{Z}\) transition delay
\end{tabular} \\
\hline Conditional & double * & \[
\begin{aligned}
& \mathrm{d} 4 \\
& \mathrm{~d} 5 \\
& \mathrm{~d} 6
\end{aligned}
\] & \begin{tabular}{l}
For module/intermodule paths and input ports/port bits: If accPathDelayCount is set to 6 or 12 : \\
d4 4 is \(\mathbf{Z}->\mathbf{1}\) transition delay \\
d5 is \(\mathbf{1}->\mathbf{Z}\) transition delay \\
d6 is \(\mathbf{Z}->\mathbf{0}\) transition delay
\end{tabular} \\
\hline Conditional & double * & \begin{tabular}{l}
d7 \\
d8 \\
d9 \\
d10 \\
d11 \\
d12
\end{tabular} & \begin{tabular}{l}
For module/intermodule paths and input ports/port bits: If accPathDelayCount is set to 12 : \\
d7 is \(\mathbf{0}->\mathbf{X}\) transition delay \\
d8 is \(\mathbf{X}->\mathbf{1}\) transition delay \\
d9 is \(\mathbf{1}->\mathbf{X}\) transition delay \\
d 10 is \(\mathbf{X}->\mathbf{0}\) transition delay \\
d11 is \(\mathbf{X}->\mathbf{Z}\) transition delay \\
d12 is \(\mathbf{Z}->\mathbf{X}\) transition delay
\end{tabular} \\
\hline & double * & limit & Limit of timing check \\
\hline
\end{tabular}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|r|}{acc_fetch_delays() for min:typ:max delays (accMinTypMaxDelays set to true )} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get existing delay values for primitives, module paths, timing checks, module input ports, or intermodule paths; the delay values are contained in an array.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_fetch_delays (object_handle, array_ptr),} \\
\hline \multicolumn{3}{|c|}{Type} & Description \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{1 if successful; 0 if an error is encountered} \\
\hline \multicolumn{2}{|r|}{Type} & Name & Description \\
\hline \multirow[t]{2}{*}{Arguments:} & handle & object_handle & Handle of a primitive, module path, timing check, module input port, bit of a module input port, or intermodule path \\
\hline & double address & array_ptr & Pointer to array of min:typ:max delay values; the size of the array depends on the type of object and the setting of accPathDelayCount (see Section 22.8) \\
\hline
\end{tabular}

The ACC routine acc_fetch_delays() shall work differently depending on how the configuration parameter accMinTypMaxDelays is set. When this parameter is set to "false", a single delay per transition shall be assumed, and each delay shall be fetched into variables pointed to as individual arguments. For this single delay mode, the first syntax table in this section shall apply.

When accMinTypMaxDelays is set to "true", acc_fetch_delays() shall fetch one or more sets of minimum:typical:maximum delays into an array, rather than single delays fetched as individual arguments. For this min:typ:max delay mode, the second syntax table in this section shall apply.

The number of delay values that shall be fetched by acc_fetch_delays() is determined by the type of object and the setting of configuration parameters. Refer to 22.8 for a description of how the number of delay values is determined.

The ACC routine acc_fetch_delays() shall retrieve delays in the timescale of the module that contains the object_handle.

The example shown in Figure 73 presents a C language application that uses acc_fetch_delays() to retrieve the rise, fall, and turn-off delays of all paths through a module.
```

include "acc_user.h"
oid display_path_delays()
handle mod_handle;
handle path_handle;
double rise_delay,fall_delay,toz_delay;
/*initialize environment for ACC routines* /
acc_initialize();
/*set accPathDelayCount to return rise, fall and turn-off delays */
acc_configure(accPathDelayCount, "3");
/*get handle to module* /
mod_handle = acc_handle_tfarg(1);
/* fetch rise delays for all paths in module "top.m1"*/
path_handle = null;
while(path_handle = acc_next_modpath(mod_handle, path_handle) )
{
acc_fetch_delays (path_handle,
\&rise_delay,\&fall_delay,\&toz_delay);
/*display rise, fall and turn-off delays for each path*/
io_printf("For module path %s,delays are:\n",
acc_fetch_fullname(path_handle) );
io_printf("rise = %lf, fall = %lf, turn-off = %lf\n",
rise_delay,fall_delay,toz_delay);
}
acc_close();

```

Figure 73-Using acc_fetch_delays() in single delay mode
The example shown in Figure 74 is a C language code fragment of an application that shows how to fetch min:typ:max delays for the intermodule paths. The example follows these steps:
a) Declares an array of nine double-precision floating-point values as a buffer for storing three sets of min:typ:max values, one set each for rise, fall, and turn-off delays.
b) Sets the configuration parameter accMinTypMaxDelays to "true" to instruct acc_fetch_delays() to retrieve delays in min:typ:max format.
c) Calls acc_fetch_delays() with a valid intermodule path handle and the array pointer.
```

include "acc_user.h"
oid fetch_mintypmax_delays(port_output, port_input)
andle port_output, port_input;
handle intermod_path;
double delay_array[ 9] ;
intermod_path = acc_handle_path(port_output, port_input);
acc_fetch_delays(intermod_path, delay_array);
\&

```


Figure 74-Using acc_fetch_delays() in min:typ:max delay mode

\subsection*{23.16 acc_fetch_direction()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_fetch_direction()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the direction of a port or terminal.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_fetch_direction(object_handle)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type Description} & \\
\hline & PLI_INT32 & \multicolumn{2}{|l|}{A predefined integer constant representing the direction of a port or terminal accInput accOutput accInout accMixedIo} \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Description \\
\hline & handle & object_handle & Handle of a port or terminal \\
\hline
\end{tabular}

The ACC routine acc_fetch_direction() shall return a predefined integer constant indicating the direction of a module port or primitive terminal. The values returned are given in Table 139.

Table 139-The operation of acc_fetch_direction()
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ When direction is } & acc_fetch_direction() shall return \\
\hline Input only & accInput \\
\hline Output only & accOutput \\
\hline Bidirectional (input and output) & accInout \\
\hline \begin{tabular}{l} 
A concatenation of input ports and output \\
ports
\end{tabular} & accMixedIo \\
\hline
\end{tabular}

The example shown in Figure 75 presents a C language application that uses acc_fetch_direction() to determine whether or not a port is an input.
```

include "acc_user.h"
nt is_port_input(port_handle)
.andle port_handle;
PLI_INT32 direction;
direction = acc fetch direction(port handle);
if (direction == accInput || directiōn == accInout)
return(true);
else
return(false);

```

Figure 75-Using acc_fetch_direction()

\subsection*{23.17 acc_fetch_edge()}
\begin{tabular}{|c|c|c|c|c|}
\hline \multicolumn{5}{|c|}{acc_fetch_edge()} \\
\hline Synopsis: & \multicolumn{4}{|l|}{Get the edge specifier of a module path or timing check terminal.} \\
\hline Syntax: & \multicolumn{4}{|l|}{acc_fetch_edge(pathio_handle)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{4}{|l|}{Type Description} \\
\hline & PLI_INT32 & \multicolumn{3}{|l|}{A predefined integer constant representing the edge specifier of a path input or output terminal:} \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & \multicolumn{2}{|r|}{Description} \\
\hline & handle & pathio_handle & Handle to a ing check te & path input or \\
\hline
\end{tabular}

The ACC routine acc_fetch_edge() shall return a value that is a masked integer representing the edge specifier for a module path or timing check terminal.

Table 140 lists the predefined edge specifiers as they are specified in acc_user.h.
Table 140-Edge specifiers constants
\begin{tabular}{|c|c|c|}
\hline Edge type & Defined constant & Binary value \\
\hline None & accNoedge & 0 \\
\hline Positive edge \((0 \rightarrow 1,0 \rightarrow \mathrm{x}, \mathrm{x} \rightarrow 1)\) & accPosedge & 00001101 \\
\hline Negative edge \((1 \rightarrow 0,1 \rightarrow \mathrm{x}, \mathrm{x} \rightarrow 0)\) & accNegedge & 00110010 \\
\hline \(0 \rightarrow 1\) edge & accEdge01 & 00000001 \\
\hline \(1 \rightarrow 0\) edge & accEdge10 & 00000010 \\
\hline \(0 \rightarrow \mathrm{x}\) edge & accEdge0x & 00000100 \\
\hline \(\mathrm{x} \rightarrow 1\) edge & accEdgex 1 & 00001000 \\
\hline \(1 \rightarrow \mathrm{x}\) edge & accEdge1 x & 00010000 \\
\hline \(\mathrm{x} \rightarrow 0\) edge & accEdgex0 & 00100000 \\
\hline
\end{tabular}

The integer mask returned by acc_fetch_edge() is usually either accPosedge or accNegedge. Occasionally, however, the mask is a hybrid mix of specifiers that is equal to neither. The example shown in Figure 76 illustrates how to check for these hybrid edge specifiers. The value accNoEdge is returned if no edge is found.

The example takes a path input or output and returns the string corresponding to its edge specifier. It provides analogous functionality to that of acc_fetch_type_str() in that it returns a string corresponding to an integer value that represents a type.

This example first checks to see whether the returned mask is equal to accPosedge or accNegedge, which are the most likely cases. If it is not, the application does a bitwise AND with the returned mask and each of the other edge specifiers to find out which types of edges it contains. If an edge type is encoded in the returned mask, the corresponding edge type string suffix is appended to the string "accEdge".
```

?LI_BYTE8 * acc_fetch_edge_str(pathio)
randle pathio;
PLI_INT32 edge = acc_fetch_edge(pathio);
static PLI_BYTE8 edge_str[ 32] ;
if (! acc_error flag)
{
if (edge == accNoEdge)
strcpy(edge_str, "accNoEdge");
/* accPosedge == (accEdge01 \& accEdge0x \& accEdgex1) */
else if (edge == accPosEdge)
strcpy(edge_str, "accPosEdge");
/* accNegedge == (accEdge10 \& accEdge 1x \& accEdgex0) * /
else if (edge == accNegEdge)
strcpy(edge_str, "accNegEdge");
/* edge is neither posedge nor negedge, but some combination
of other edges */
else {
strcpy(edge_str, "accEdge");
if (edge \& accEdge01) strcat(edge_str, "_01");
if (edge \& accEdge10) strcat(edge_str, "_10");
if (edge \& accEdge0x) strcat(edge_str, "-0x");
if (edge \& accEdgex1) strcat(edge_str, "-x1");
if (edge \& accEdge1x) strcat(edge_str, "_1x");
if (edge \& accEdgex0) strcat(edge_str, "_x0");
}
return(edge str);
}
else
return(null);

```

Figure 76—Using acc_fetch_edge()

\subsection*{23.18 acc_fetch_fullname()}


The ACC routine acc_fetch_fullname() shall obtain the full hierarchical name of an object. The full hierarchical name is the name that uniquely identifies an object. In Figure 84, the top-level module, top1, contains module instance mod3, which contains net w4. In this example, the full hierarchical name of the net is top1.mod3.w4.

\section*{top1}


Figure 77-A design hierarchy; the fullname of net w4 is "top1.mod3.w4"

Table 141 lists the objects in a Verilog HDL description for which acc_fetch_fullname() shall return a name.

Table 141-Named objects supported by acc_fetch_fullname()
\begin{tabular}{|c|c|}
\hline Modules & Integer, time and real variables \\
\hline Module ports & Named events \\
\hline Module paths & Parameters \\
\hline Data paths & Specparams \\
\hline Primitives & Named blocks \\
\hline Nets & Verilog HDL tasks \\
\hline Regs or Variables & Verilog HDL functions \\
\hline
\end{tabular}

Module path names shall be derived from their sources and destinations in the following format:
\begin{tabular}{|l|l|l|}
\hline source & path_delimiter & destination \\
\hline
\end{tabular}

By default, the path_delimiter shall be the character \(\$\). However, the delimiter can be changed by using the ACC routine acc_configure() to set the delimiter parameter accPathDelimStr to another character string.

The following examples show names of paths within a top-level module m 3 , as returned by acc_fetch_fullname() when the path_delimiter is \(\$\). Note that names of module paths with multiple sources or destinations shall be derived from the first source and destination only.

Table 142-Module path names returned by acc_fetch_fullname()
\begin{tabular}{|c|c|}
\hline For paths in module \(\boldsymbol{m 3}\) & acc_fetch_fullname() returns a pointer to \\
\hline\((\mathrm{a}=>\mathrm{q})=10 ;\) & \(\mathbf{m 3 . a \$ q}\) \\
\hline\((\mathrm{b} *>\mathrm{q} 1, \mathrm{q} 2)=8 ;\) & \(\mathbf{m 3 . b \$ q 1}\) \\
\hline\((\mathrm{d}, \mathrm{e}, \mathrm{f} *>\mathrm{r}, \mathrm{s})=8 ;\) & \(\mathbf{m 3 . d \$ r}\) \\
\hline
\end{tabular}

If a Verilog software product creates default names for unnamed instances, acc_fetch_fullname() shall return the full hierarchical default name. Otherwise, the routine shall return null for unnamed instances.

Using acc_fetch_fullname() with a module port handle shall return the full hierarchical implicit name of the port.

The routine acc_fetch_fullname() shall store the returned string in a temporary buffer. To preserve the string for later use in an application, it should be copied to another variable (refer to 22.9).

In the example shown in Figure 78, the routine uses acc_fetch_fullname() to display the full hierarchical name of an object if the object is a net.
```

\#include "acc_user.h"
PLI_INT32 display_if_net(object_handle)
han\overline{d}le object hāndle;
{
/*get and display full name if object is a net*/
if (acc_fetch_type(object_handle) == accNet)
io_printf("Object is a net: %s\n",
acc_fetch_fullname(object_handle) );
else
io_printf("Object is not a net: %s\n",
acc_fetch_fullname(object_handle) );
}

```

Figure 78-Using acc_fetch_fullname()

\subsection*{23.19 acc_fetch_fulltype()}
\begin{tabular}{|c|c|c|c|c|}
\hline \multicolumn{5}{|c|}{acc_fetch_fulltype()} \\
\hline Synopsis: & \multicolumn{4}{|l|}{Get the fulltype of an object.} \\
\hline Syntax: & \multicolumn{4}{|l|}{acc_fetch_fulltype(object_handle)} \\
\hline \multicolumn{5}{|c|}{Type Description} \\
\hline Returns: & PLI_INT32 & \multicolumn{3}{|l|}{A predefined integer constant from the list shown in 22.6} \\
\hline & Type & Name & \multicolumn{2}{|r|}{Description} \\
\hline Arguments: & handle & object_handle & Handle of the object & \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_fetch_type() to get the general type classification of an object Use acc_fetch_type_str() to get the fulltype as a character string} & \\
\hline
\end{tabular}

The ACC routine acc_fetch_fulltype() shall return the fulltype of an object. The fulltype is a specific classification of a Verilog HDL object, represented as a predefined constant (defined in acc_user.h). Table 113 lists all of the fulltype constants that can be returned by acc_fetch_fulltype().

Many Verilog HDL objects have both a type and a fulltype. The type of an object is its general Verilog HDL type classification. The fulltype is the specific type of the object. The examples in Table 143 illustrate the difference between the type of an object and the fulltype of the same object for selected objects.

Table 143-The difference between the type and the fulltype of an object
\begin{tabular}{|c|c|c|}
\hline For a handle to & acc_fetch_type() shall return & acc_fetch_fulltype() shall return \\
\hline A setup timing check & accTchk & accSetup \\
\hline An and gate primitive & accPrimitive & accAndGate \\
\hline A sequential UDP & accPrimitive & accSeqPrim \\
\hline
\end{tabular}

The example shown in Figure 79 uses acc_fetch_fulltype() to find and display the fulltypes of timing checks. This application is called by a higher-level application, display_object_type, presented as the usage example for acc_fetch_type().
```

:include "acc_user.h"
'LI_INT32 display_timing_check_type(tchk_handle)
ıandle tchk_handle;
/*display timing check type*/
io_printf("Timing check is");
switch(acc_fetch_fulltype(tchk_handle) )
{
case accHold:
io_printf(" hold\n");
break;
case accNochange:
io_printf(" nochange\n");
break;
case accPeriod:
io_printf(" period\n");
break;
case accRecovery:
io_printf(" recovery\n");
break;
case accSetup:
io_printf(" setup\n");
break;
case accSkew:
io_printf(" skew\n");
break;
case accWidth:
io_printf(" width\n");
}

```

Figure 79—Using acc_fetch_fulltype() to display the fulltypes of timing checks

The example shown in Figure 80 uses acc_fetch_fulltype() to find and display the fulltypes of primitive objects passed as input arguments. This application is called by a higher-level application, display_object_type, presented as the usage example for acc_fetch_type().
```

!include "acc_user.h"
'LI_INT32 display_primitive_type(primitive_handle)
ıandle primitive_handle;
/*display primitive type*/
io printf("Primitive is");
switch(acc_fetch_fulltype(primitive_handle) )
{
case accAndGate:
io_printf(" and gate\n"); break;
case accBufGate:
io printf(" buf gate\n"); break;
case accBufif0Gate:case accBufif1Gate:
io_printf(" bufif gate\n"); break;
case accCmosGate:case accNmosGate:case accPmosGate:
case accRcmosGate:case accRnmosGate:case accRpmosGate:
io printf(" MOS or Cmos gate\n"); break;
case accCombPrim:
io_printf(" combinational UDP\n"); break;
case accSeqPrim:
io_printf(" sequential UDP\n"); break;
case accNotif0Gate:case accNotif1Gate:
io_printf(" notif gate\n"); break;
case accRtranGate:
io_printf(" rtran gate\n"); break;
case accRtranif0Gate:case accRtranif1Gate:
io_printf(" rtranif gate\n"); break;
case accNandGate:
io_printf(" nand gate\n"); break;
case accNorGate:
io_printf(" nor gate\n"); break;
case accNotGate:
io_printf(" not gate\n"); break;
case accOrGate:
io_printf(" or gate\n"); break;
case accPulldownGate:
io_printf(" pulldown gate\n"); break;
case accPullupGate:
io_printf(" pullup gate\n"); break;
case accXnorGate:
io_printf(" xnor gate\n"); break;
case accXorGate:
io_printf(" xor gate\n");
}

```

Figure 80-Using acc_fetch_fulltype() to display the fulltypes of primitives

\subsection*{23.20 acc_fetch_index()}
\begin{tabular}{|l|l|l||}
\hline \multicolumn{3}{|c|}{ acc_fetch_index() } \\
\hline Synopsis: & Get the index number for a port or terminal. \\
\hline Syntax: & acc_fetch_index (object_handle) \\
\hline \multicolumn{4}{|c|}{ Type } & Description \\
\cline { 2 - 4 } Returns: & PLI_INT32 & Integer index for a port or terminal, starting with zero \\
\hline & Type & \multicolumn{2}{c|}{ Name } & Description \\
\cline { 2 - 3 } Arguments: & handle & object_handle \\
\hline \hline
\end{tabular}

The ACC routine acc_fetch_index() shall return the index number for a module port or primitive terminal. Indices are integers that shall start at zero and increase from left to right.

The index of a port shall be its position in a module definition in the Verilog HDL source description. The index of a terminal shall be its position in a gate, switch, or UDP instance.

Table 144 shows how indices shall be derived.
Table 144—Deriving indices
\begin{tabular}{|c|c|}
\hline For & Indices are \\
\hline Terminals: nand g1(out, in1, in2); & 0 for terminal out \(\mathbf{1}\) for terminal in1 2 for terminal in2 \\
\hline \begin{tabular}{l}
Implicit ports: \\
module A(q, a, b);
\end{tabular} & 0 for port q 1 for port a 2 for port b \\
\hline ```
Explicit ports:
    module top;
        reg ra,rb;
        wire wq;
        explicit_port_mod epm1(.b(rb), .a(ra), .q(wq));
    endmodule
    module explicit_port_mod(q, a, b);
        input a, b;
        output q;
        nand (q, a, b);
    endmodule
``` & 0 for explicit port epm1.q 1 for explicit port epm1. a 2 for explicit port epm1.b \\
\hline
\end{tabular}

The example shown in Figure 81 presents a C language application that uses acc_fetch_index() to find and display the input ports of a module.
```

:include "acc_user.h"
'LI_INT32 display_inputs(module_handle)
tandle module_handle;
handle port_handle;
PLI_INT32 direction;
/*get handle for the module and each of its ports*/
port_handle = null;
whil\overline{e (port_handle = acc_next_port(module_handle, port_handle) )}
{
/*determine if port is an input*/
direction = acc_fetch_direction(port_handle);
/*give the index of eāch input port*/
if (direction == accInput)
io_printf("Port \#%d of %s is an input\n",
acc_fetch_index(port_handle),
acc_fetch_fullname(module handle) );
}

```

Figure 81—Using acc_fetch_index()

\subsection*{23.21 acc_fetch_location()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_fetch_location()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the location of an object in a Verilog-HDL source file.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_fetch_location(loc_p, object_handle)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type Description} & \\
\hline & PLI_INT32 & \multicolumn{2}{|l|}{1 if successful; 0 if an error is encountered} \\
\hline \multirow{3}{*}{Arguments:} & Type & Name & Description \\
\hline & p_location & loc_p & Pointer to a predefined location structure \\
\hline & handle & object_handle & Handle to an object \\
\hline
\end{tabular}

The ACC routine acc_fetch_location() shall return the file name and line number in the file for the specified object. The file name and line number shall be returned in an s_location data structure. This data structure is defined in acc_user. h , and listed in Figure 82.
```

typedef struct t_location
{
PLI_INT32 line_no;
PLI_BYTE8 * filēname;
} s_location, *p_location;

```

Figure 82-s_location data structure
filename field is a character pointer.
line_no field is a nonzero positive integer.
Table 141 lists the objects that shall be supported by acc_fetch_location().

Table 145-Objects supported by acc_fetch_location()
\begin{tabular}{|c|c|}
\hline Object type & Location returned \\
\hline Modules & Module instantiation line \\
\hline Module ports & Module definition \\
\hline Module paths & Module path line \\
\hline Data paths & Module path line \\
\hline Primitives & Instantiation line \\
\hline Explicit nets & Definition line \\
\hline Implicit nets & Line where first used \\
\hline Reg variables & Definition line \\
\hline Integer, time and real variables & Definition line \\
\hline Named events & Definition line \\
\hline Parameters & Definition line \\
\hline Specparams & Definition line \\
\hline Named blocks & Definition line \\
\hline Verilog HDL tasks & Definition line \\
\hline Verilog HDL functions & Definition line \\
\hline
\end{tabular}

The return value for filename is placed in the ACC internal string buffer. See 22.9 for an explanation of strings in ACC routines.

The example shown in Figure 83 uses acc_fetch_location() to print the file name and line number for an object.
```

.I_INT32 find_object_location (object)
handle object;
s_location s_loc;
p_location loc_p = \&s_loc;
acc_fetch_location(loc_p, object); /*get the filename and line_no*/
if (! acc_error_flag) /* On success */
io_printf ("Object located in file %s on line %d \n",
loc_p->filename, loc_p-> line_no);

```

Figure 83-Using acc_fetch_location()

\subsection*{23.22 acc_fetch_name()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_fetch_name()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the instance name of any named object or module path.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_fetch_name (object_handle)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type Description} & \\
\hline & PLI_BYTE8 * & \multicolumn{2}{|l|}{Character pointer to a string containing the instance name of the object} \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Description \\
\hline & handle & object_handle & Handle of the named object \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_fetch_fullname() to get the full hierarchical name of the object Use acc_fetch_defname() to get the definition name of the object Use acc_configure(accPathDelimStr...) to set the naming convention for module paths} \\
\hline
\end{tabular}

The ACC routine acc_fetch_name() shall obtain the name of an object. The name of an object is its lowestlevel name. In the following example, the top-level module, top1, contains module instance mod3, which contains net w 4 , as shown in Figure 84. In this example, the name of the net is w 4 .


Figure 84-A design hierarchy; the name of net w4 is "w4"
The return value for this routine is placed in the ACC internal string buffer. See 22.9 for an explanation of strings in ACC routines.

Table 141 lists the objects in a Verilog HDL description for which acc_fetch_name() shall return a name.
Table 146-Named objects supported by acc_fetch_name()
\begin{tabular}{|c|c|}
\hline Modules & Integer, time and real variables \\
\hline Module ports & Named events \\
\hline Module paths & Parameters \\
\hline Data paths & Specparams \\
\hline Primitives & Named blocks \\
\hline Nets & Verilog HDL tasks \\
\hline Regs or Variables & Verilog HDL functions \\
\hline
\end{tabular}

Module path names shall be derived from their sources and destinations in the following format:
\begin{tabular}{|l|l|l|}
\hline source & path_delimiter & destination \\
\hline
\end{tabular}

By default, the path_delimiter is the character \$. However, the delimiter can be changed by using the ACC routine acc_configure() to set the delimiter parameter accPathDelimStr to another character string.

Table 147 shows names of paths within a top-level module m 3 , as returned by acc_fetch_name() when the path_delimiter is \(\mathbf{\$}\). Note that names of module paths with multiple sources or destinations shall be derived from the first source and destination only.

Table 147-Module path names returned by acc_fetch_name()
\begin{tabular}{|c|c|}
\hline For paths in module \(\boldsymbol{m} \mathbf{3}\) & acc_fetch_name() returns a pointer to \\
\hline\((\mathrm{a}=>\mathrm{q})=10 ;\) & \(\mathbf{a \$ q}\) \\
\hline\((\mathrm{b} *>\mathrm{q} 1, \mathrm{q} 2)=8 ;\) & \(\mathbf{b} \$ \mathbf{q} \mathbf{1}\) \\
\hline\((\mathrm{~d}, \mathrm{e}, \mathrm{f} *>\mathrm{r}, \mathrm{s})=8 ;\) & \(\mathbf{d \$ r}\) \\
\hline
\end{tabular}

If a Verilog software implementation creates default names for unnamed instances, acc_fetch_name() shall return the default name. Otherwise, the routine shall return null for unnamed instances.

Using acc_fetch_name() with a module port handle shall return the implicit name of the port.
The following example uses acc_fetch_name() to display the names of top-level modules.
```

include "acc_user.h"
'LI_INT32 show_top_mods()
handle module_handle;
/*initialize environment for ACC routines*/
acc_initialize();
/*scan all top-level modules*/
io_printf("The top-level modules are:\n");
module_handle = null;
while (module_handle = acc_next_topmod(module_handle) )
io_printf("
acc_close();

```

Figure 85-Using acc_fetch_name()

\subsection*{23.23 acc_fetch_paramtype()}


The ACC routine acc_fetch_paramtype() shall return an integer constant that represents the data type of a value that has been assigned to a parameter or specparam.

Figure 86 uses acc_fetch_paramtype() to display the values of all parameters within a module.
```

nclude "acc_user.h"
I_INT32 print_parameter_values()
andle module_handle, param_handle;
*initialize environment for ACC routines*/
cc_initialize();
odule_handle = acc_handle_tfarg(1);
aram_handle = null;
while(param_handle = acc_next_parameter(module_handle,param_handle) )
{
io_printf("Parameter %s has value: ",acc_fetch_fullname(param_handle))
switch(acc_fetch_paramtype (param_handle) )
{
case accRealParam:
io_printf("%lf\n", acc_fetch_paramval(param_handle) ); break;
case accIntegerParam:
io_printf("%d\n",
(int)acc_fetch_paramval(param_handle) ); break;
case accStrin
io_printf("%s\n",
(char*)(int)acc_fetch_paramval(param_handle) ); break;
}
}
cc_close();

```

Figure 86-Using acc_fetch_paramtype()

\subsection*{23.24 acc_fetch_paramval()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|r|}{acc_fetch_paramval()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the value of a parameter or specparam.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_fetch_paramval (parameter_handle)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|r|}{Type Description} & \\
\hline & double & \multicolumn{2}{|l|}{The value of a parameter or specparam} \\
\hline & Type & Name & \\
\hline Arguments: & handle & parameter_handle & Handle to a param \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_fetch_paramtype() to retrieve the data type of a parameter Use acc_next_parameter() to scan all parameters within a module Use acc_next_specparam() to scan all specparams within a module} \\
\hline
\end{tabular}

The ACC routine acc_fetch_paramtype() shall return the value stored in a parameter or specparam. The value shall be returned as a double-precision floating-point number.

A parameter value can be stored as one of three data types:
A double-precision floating-point number
An integer value
A string

Therefore, it can be necessary to call acc_fetch_paramtype() to determine the data type of the parameter value, as shown in the example in Figure 87.

The routine acc_fetch_paramval() returns values as type double. The values can be converted back to integers or character pointers using the C language cast mechanism, as shown in Table 148. Note that some C language compilers do not allow casting a double-precision value directly to a character pointer; it is therefore necessary to use a two-step cast to first convert the double value to an integer and then convert the integer to a character pointer.

If a character string is returned, it is placed in the ACC internal string buffer. See 22.9 for explanation of strings in ACC routines.

Table 148-Casting acc_fetch_paramval() return values
\begin{tabular}{|c|l|}
\hline To convert to & \multicolumn{1}{c|}{ Follow these steps } \\
\hline Integer & \begin{tabular}{l} 
Cast the return value to the integer data type using the C language cast operator \\
(int): \\
int_val= (int) acc_fetch_paramval (. . .) ;
\end{tabular} \\
\hline String & \begin{tabular}{l} 
Cast the return value to a character pointer using the C language cast operators \\
(char*)(int): \\
str_ptr=(char*)(int) acc_fetch_paramval ( . . .) ;
\end{tabular} \\
\hline
\end{tabular}

The example shown in Figure 87 presents a C language application, print_parameter_values, that uses acc_fetch_paramtype() to display the values of all parameters within a module.
```

nclude "acc_user.h"
I_INT32 print_parameter_values()
handle module handle;
handle param_handle;
/*initialize environment for ACC routines*/
acc_initialize();
/*get handle for module* /
module_handle = acc_handle_tfarg(1);
/*scan all parameters in the module and display their values*/
/* according to type*/
param_handle = null;
while(param_handle = acc_next_parameter(module_handle,param_handle) )
{
io_printf("Parameter %s has value: ",acc_fetch_fullname(param_handle)
switch(acc_fetch_paramtype(param_handle) )
{
case accRealParam:
io_printf("%lf\n", acc_fetch_paramval(param_handle) );
break;
case accIntegerParam:
io_printf("%d\n", (int)acc_fetch_paramval(param_handle) );
break;
case accStringParam:
io_printf("%s\n",
(char*) (int) acc_fetch_paramval (param_handle) );
break;
}
}
acc_close();
two-step cast

```

Figure 87-Using acc_fetch_paramval()

\subsection*{23.25 acc_fetch_polarity()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_fetch_polarity()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the polarity of a path.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_fetch_polarity (path_handle)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type \(\quad\) Description} & \\
\hline & PLI_INT32 & \multicolumn{2}{|l|}{A predefined integer constant representing the polarity of a path: accPositive accNegative accUnknown} \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Description \\
\hline & handle & path_handle & Handle to a module path or data path \\
\hline
\end{tabular}

The ACC routine acc_fetch_polarity() shall return an integer constant that represents the polarity of the specified path. The polarity of a path describes how a signal transition at its source propagates to its destination in the absence of logic simulation events. The return value shall be one of the predefined integer constant polarity types listed in Table 149.

Table 149-Polarity types returned by acc_fetch_polarity()
\begin{tabular}{|c|l|}
\hline Integer constant & \multicolumn{1}{c|}{ Description } \\
\hline accPositive & \begin{tabular}{l} 
A rise at the source causes a rise at the destination. \\
A fall at the source causes a fall at the destination.
\end{tabular} \\
\hline accNegative & \begin{tabular}{l} 
A rise at the source causes a fall at the destination. \\
A fall at the source causes a rise at the destination.
\end{tabular} \\
\hline accUnknown & \begin{tabular}{l} 
Unpredictable; a rise or fall at the source causes either a rise or fall \\
at the destination.
\end{tabular} \\
\hline
\end{tabular}

The example shown in Figure 88 takes a path argument and returns the string corresponding to its polarity.
```

PLI_BYTE8 * fetch_polarity_str(path)
{
switch (acc_fetch_polarity(path)) {
case accPositive: return("accPositive");
case accNegative: return("accNegative");
case accUnknown: return("accUnknown");
default: return(null);
}
}

```

Figure 88-Using acc_fetch_polarity()

\subsection*{23.26 acc_fetch_precision()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_fetch_precision()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the smallest time precision argument specified in all `timescale compiler directives in a given design.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_fetch_precision()} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{An integer value that represents a time precision} \\
\hline \multirow[b]{2}{*}{Arguments:} & \multicolumn{2}{|l|}{Type \(\quad\) Name} & Description \\
\hline & None & & \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_fetch_timescale_info() to get the timescale and precision of a specific object} \\
\hline
\end{tabular}

The ACC routine acc_fetch_precision() shall return the smallest time precision argument specified in all -timescale compiler directives for a given design. The value returned shall be the order of magnitude of one second, as shown in Table 150.

Table 150-Value returned by acc_fetch_precision()
\begin{tabular}{|c|c|}
\hline Integer value returned & \begin{tabular}{c} 
Simulation time precision \\
represented
\end{tabular} \\
\hline 2 & 100 s \\
\hline 1 & 10 s \\
\hline 0 & 1 s \\
\hline-1 & 100 ms \\
\hline-2 & 10 ms \\
\hline-3 & 1 ms \\
\hline-4 & 100 s \\
\hline-5 & 10 s \\
\hline-6 & 1 s \\
\hline-7 & 100 ns \\
\hline-8 & 10 ns \\
\hline-9 & 1 ns \\
\hline-10 & 100 ps \\
\hline-11 & 10 ps \\
\hline-12 & 1 ps \\
\hline-13 & 100 fs \\
\hline-14 & 10 fs \\
\hline-15 & 1 fs \\
\hline
\end{tabular}

If there are no `timescale compiler directives specified for a design, acc_fetch_precision() shall return a value of \(0(1 \mathrm{~s})\).

\subsection*{23.27 acc_fetch_pulsere()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_fetch_pulsere()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get current pulse handling reject_limit and e_limit for a module path, intermodule path or module input port.} \\
\hline Syntax: & \multicolumn{3}{|l|}{\[
\begin{array}{r}
\text { acc_fetch_pulsere (object,r1,e1, r2,e2, r3,e3, r4,e4, r5,e5, r6,e6, } \\
r 7, e 7, r 8, e 8, r 9, e 9, r 10, e 10, ~ r 11, e 11, ~ r 12, e 12)
\end{array}
\]} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type Description} & \\
\hline & PLI_INT32 & \multicolumn{2}{|l|}{1 if successful; 0 if an error is encountered} \\
\hline \multirow{4}{*}{Arguments:} & Type & Name & Descriptio \\
\hline & handle & object & Handle of module path, intermodul port \\
\hline & double * & r1...r12 & reject_limit values; the number of by accPathDelayCount \\
\hline & double * & e1...e12 & e_limit values; the number of argu accPathDelayCount \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_append_pulsere() to add to the existing pulse handling values Use acc_replace_pulsere() to replace existing pulse handling values Use acc_set_pulsere() to set pulse handling values as a percentage of the path delay Use acc_configure() to set accPathDelayCount} \\
\hline
\end{tabular}

The ACC routine acc_fetch_pulsere() shall obtain the current values controlling how pulses are propagated through a module path, intermodule path or module input port.

A pulse is defined as two transitions that occur in a shorter period of time than the delay. Pulse control values determine whether a pulse should be rejected, propagated through to the output, or considered an error. The pulse control values consist of a reject_limit and an e_limit pair of values, where

The reject_limit shall set a threshold for determining when to reject a pulse any pulse less than the reject_limit shall not propagate
The e_limit shall set a threshold for determining when a pulse is an error any pulse less than the e_limit and greater than or equal to the reject_limit shall propagate a logic \(x\)
A pulse that is greater than or equal to the e_limit shall propagate
Table 151 illustrates the relationship between the reject_limit and the e_limit.
Table 151-Pulse control example
\begin{tabular}{|c|l|}
\hline When & \multicolumn{1}{c|}{ The pulse shall be } \\
\hline \begin{tabular}{c} 
reject_limit \(=10.5\) \\
e_limit \(=22.6\)
\end{tabular} & \begin{tabular}{l} 
Rejected if \(<10.5\) \\
\\
\\
\\
\\
\\
An error if \(>=10.5\) and \(<22.6\) \\
\\
\end{tabular} \\
\hline
\end{tabular}

The number of pulse control values that acc_fetch_pulsere() shall retrieve is controlled using the ACC routine acc_configure() to set the delay count configuration parameter accPathDelayCount, as shown in Table 152.

Table 152-How the accPathDelayCount affects acc_fetch_pulsere()
\begin{tabular}{|c|c|}
\hline When accPathDelayCount is & acc_fetch_pulsere() shall retrieve \\
\hline 1 & One pair of reject_limit and e_limit values: one pair for all transitions, rl and e 1 \\
\hline 2 & Two pairs of reject_limit and e_limit values: one pair for rise transitions, rl and el one pair for fall transitions, r 2 and e2 \\
\hline 3 & Three pairs of reject_limit and e_limit values: one pair for rise transitions, rl and el one pair for fall transitions, r 2 and e2 one pair for turn-off transitions, r3 and e3 \\
\hline \[
\begin{gathered}
\mathbf{6} \\
\text { (the default) }
\end{gathered}
\] & Six pairs of reject_limit and e_limit values a different pair for each possible transition among 0,1 , and Z : one pair for \(0->1\) transitions, rl and e1 one pair for \(1->0\) transitions, r 2 and e 2 one pair for \(0->\mathrm{Z}\) transitions, r 3 and e3 one pair for \(\mathrm{Z}->1\) transitions, r 4 and e4 one pair for \(1->Z\) transitions, r 5 and e5 one pair for Z->0 transitions, r6 and e6 \\
\hline 12 & Twelve pairs of reject_limit and e_limit values a different pair for each possible transition among \(0,1, \mathrm{X}\), and Z : one pair for \(0->1\) transitions, rl and el one pair for \(1->0\) transitions, r 2 and e 2 one pair for \(0->Z\) transitions, r 3 and e3 one pair for \(\mathrm{Z}->1\) transitions, r 4 and e4 one pair for \(1->Z\) transitions, r 5 and e5 one pair for \(\mathrm{Z}->0\) transitions, r 6 and e6 one pair for \(0->\mathrm{X}\) transitions, r 7 and e7 one pair for \(\mathrm{X}->1\) transitions, r 8 and e8 one pair for \(1->\) X transitions, r 9 and e9 one pair for \(\mathrm{X}->0\) transitions, r10 and e10 one pair for \(\mathrm{X}->\mathrm{Z}\) transitions, r11 and e11 one pair for Z->X transitions, r12 and e12 \\
\hline
\end{tabular}

The minimum number of pairs of reject_limit and e_limit arguments to pass to acc_fetch_pulsere() shall equal the value of accPathDelayCount. Any unused reject_limit and e_limit argument pairs shall be ignored by acc_fetch_pulsere() and can be dropped from the argument list.

If accPathDelayCount is not set explicitly, it shall default to 6 , and therefore six pairs of pulse reject_limit and e_limit arguments have to be used when acc_fetch_pulsere() is called. Note that the value assigned to accPathDelayCount also affects acc_append_delays(), acc_fetch_delays(), acc_replace_delays(), acc_append_pulsere(), and acc_replace_pulsere().

Pulse control values shall be retrieved using the timescale of the module that contains the object handle.
The example shown in Figure 89 shows how an application, get_pulsevals, uses acc_fetch_pulsere() to retrieve rise and fall pulse handling values of paths listed in a file called path. dat. The format of the file is shown in the following diagram.

```

include <stdio.h>
include "acc user.h"
define NAME SIZE 256
LI_INT32 get_pulsevals()
FILE *infile;
PLI_BYTE8 mod_name[ NAME_SIZE] ;
PLI-BYTE8 pat\overline{hin name[ N\overline{AME SIZE], pathout name[ NAME SIZE] ;}}\mathbf{N}=\mp@code{M}
handle mod, path;
double rise_reject_limit,rise_e_limit,fall_reject_limit,fall_e_limit;
/*initialize environment for ACC routines*/
acc_initialize();
/* set accPathDelayCount to return two pairs of pulse handling values,*
/* one each for rise and fall transitions*/
acc_configure(accPathDelayCount, "2");
/*read all module path specifications from file "path.dat"*/
infile = fopen("path.dat", "r");
while(fscanf(infile, "%s %s %s"
mod_name,pathin_name,pathout_name)!=EOF)
{
mod=acc_handle_object(mod_name);
path=acc_handle_modpath(mod,pathin_name,pathout_name);
if(acc_fetch_pulsere(path,
\&rise_reject_limit,\&rise_e_limit,
\&fall_reject_limit, \&fall_e_limit))
{
io_printf("rise reject limit = %lf, rise e limit = %lf\n",
rise_reject_limit, rise_e_limit);
io_printf("fa\overline{l}}\mathrm{ rejec̄t limit = %\̄f, fall e limit = %lf\n",
fall_reject_limit, fall_e_limit);
}
}
acc_close();

```

Figure 89—Using acc_fetch_pulsere()

\subsection*{23.28 acc_fetch_range()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_fetch_range()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the most significant bit and least significant bit range values for a vector.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_fetch_range(vector_handle, msb, lsb)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{Zero if successful; nonzero upon error} \\
\hline \multirow{4}{*}{Arguments:} & Type & Name & Description \\
\hline & handle & vector_handle & Handle to a vector net or reg \\
\hline & PLI_INT32 * & msb & Pointer to an integer variable to hold the most significant bit of vector_handle \\
\hline & PLI_INT32 * & 1sb & Pointer to an integer variable to hold the least significant bit of vector_handle \\
\hline Related routines & \multicolumn{3}{|l|}{Use acc_fetch_size() to get the number of bits in a vector} \\
\hline
\end{tabular}

The ACC routine acc_fetch_range() shall obtain the most significant bit (msb) and least significant bit (lsb) numbers of a vector.

The \(m s b\) shall be the left range element, while the \(l s b\) shall be the right range element in the Verilog HDL source code.

The example shown in Figure 90 takes a handle to a module instance as its input. It then uses acc_fetch_range() to display the name and range of each vector net found in the module as: <name>[ <msb>: <lsb>].
```

PLI_INT32 display_vector_nets()
{
handle mod = acc_handle_tfarg(1);
handle net;
PLI_INT32 msb, lsb;
io_printf ("Vector nets in module %s:\n:",
acc_fetch_fullname (mod));
net = null;
while (net = acc_next_net(mod, net))
if (acc_object_of_type(net, accVector))
{
acc_fetch_range(net, \&msb, \&lsb);
io_printf(") %s[ %d:%d]\n",
acc_fetch_name(net), msb, lsb);
}
}

```

Figure 90—Using acc_fetch_range()

\subsection*{23.29 acc_fetch_size()}
\begin{tabular}{||l|l|l||}
\hline \multicolumn{4}{|c|}{ acc_fetch_size() } \\
\hline Synopsis: & Get the bit size of a net, reg, integer, time, real or port. \\
\hline Syntax: & acc_fetch_size (object_handle) \\
\hline \multirow{3}{|c|}{ Type } & Description \\
\cline { 2 - 3 } Returns: & PLI_INT32 & Number of bits in the net, reg, integer, time, real or port \\
\hline \multirow{3}{|c|}{ Type } & \multicolumn{2}{c|}{ Name } \\
\cline { 2 - 3 } Arguments: & handle & object_handle \\
\hline
\end{tabular}

The ACC routine acc_fetch_size() shall return the number of bits of a net, reg, integer, time, real or port.

The example shown in Figure 91 uses acc_fetch_size() to display the size of a vector net.
```

include "acc_user.h"
'LI_INT32 display_vector_size()
handle net_handle;
PLI_INT32 siz\overline{e}_in_bits;
/* reset environment for ACC routines */
acc_initialize();
/*get first argument passed to user-defined system task*/
/* associated with this routine*/
net_handle = acc_handle_tfarg(1);
/*if net is a vector, find and display its size in bits*/
if (acc_object_of_type(net_handle, accVector) )
{
size_in_bits = acc_fetch_size(net_handle);
io_printf("Net %s is a vector of size %d\n",
acc_fetch_fullname(net_handle),size_in_bits);
}
else
io_printf("Net %s is not a vector net\n",
acc_fetch_fullname(net_handle) );

```

Figure 91-Using acc_fetch_size()

\subsection*{23.30 acc_fetch_tfarg(), acc_fetch_itfarg()}


The ACC routine acc_fetch_tfarg() shall return the value of arguments passed to the current instance of a user-defined system task or function. The ACC routine acc_fetch_itfarg() shall return the value of arguments passed to a specific instance of a user-defined system task or function, using a handle to the task or function. The value is returned as a double-precision floating-point number.

Argument numbers shall start at one and increase from left to right in the order that they appear in the system task or function call.

If an argument number is passed in that is out of range for the number of arguments in the user-defined system task/function call, acc_fetch_tfarg() and acc_fetch_itfarg() shall return a value of 0.0 , and generate a warning message if warnings are enabled. Note that the acc_error_flag is not set for an out-of-range index number.

If a user-defined system task/function argument that does not represent a valued object is referenced, acc_fetch_tfarg() and acc_fetch_itfarg() shall return a value of 0.0 and generate a warning message if warnings are enabled. Literal numbers, nets, regs, integer variables, and real variables all have values. Objects such as module instance names do not have a value. Note that the acc_error_flag is not set when a nonvalued argument is referenced.

The routine acc_fetch_tfarg() returns values as type double. The routines acc_fetch_tfarg_int() and acc_fetch_tfarg_str() return values as integers or string pointers, respectively. The value returned by acc_fetch_tfarg() can also be converted to integers or character pointers using the C language cast mechanism, as shown in Table 153. Note that some C language compilers do not allow casting a double-precision value directly to a character pointer; it is therefore necessary to use a two-step cast to first convert the double value to an integer and then convert the integer to a character pointer. If a character string is returned, it is placed in the ACC internal string buffer. See 22.9 for explanation of strings in ACC routines.

Table 153-Casting acc_fetch_tfarg() return values
\begin{tabular}{|c|l|}
\hline To convert to & \multicolumn{1}{c|}{ Follow these steps } \\
\hline Integer & \begin{tabular}{l} 
Cast the return value to the integer data type using the C language cast oper- \\
ator (PLI_INT32): \\
int_val=(PLI_INT32) acc_fetch_tfarg ( . . .); ;
\end{tabular} \\
\hline String & \begin{tabular}{l} 
Cast the return value to a character pointer using the C language cast opera- \\
tors (char*)(int): \\
str_ptr= (char*)(int) acc_fetch_tfarg ( . . .);
\end{tabular} \\
\hline
\end{tabular}

The example shown in Figure 92 uses acc_fetch_tfarg(), acc_fetch_tfarg_int(), and acc_fetch_tfarg_str() to return the value of the first argument of a user-defined system task or function.
```

\#include "acc_user.h"
\#include "veriuser.h"
?LI_INT32 display_arg_value()
PLI_INT32 arg_type;
/*initialize environment for ACC routines*/
acc_initialize();
/* check type of argument* /
io_printf("Argument value is ");
switch(tf_typep(1) )
{
case tf_readonlyreal:
case tf_readwritereal:
io_printf("%1f\n", acc_fetch_tfarg(1) );
break;
case tf_readonly:
case tf_readwrite:
io_printf("%d\n", acc_fetch_tfarg_int(1) );
break;
case tf_string:

```

```

            break; - - pointer to a
        default:
            io_printf("Error in argument specification\n");
            break;
    }
    acc_close();
    ```

Figure 92—Using acc_fetch_tfarg(), acc_fetch_tfarg_int(), and acc_fetch_tfarg_str()

\subsection*{23.31 acc_fetch_tfarg_int(), acc_fetch_itfarg_int()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_fetch_tfarg_int(), acc_fetch_itfarg_int()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the value of the specified argument of the system task or function associated with the PLI application; the value is returned as an integer number.} \\
\hline Syntax: & \multicolumn{3}{|l|}{```
acc_fetch_tfarg_int(argument_number)
acc_fetch_itfarg_int(argument_number, tfinst)
```} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{The value of the task/function argument, returned as an integer number} \\
\hline \multicolumn{3}{|r|}{Type Name} & Description \\
\hline Arguments: & PLI_INT32 & argument_number & Integer number that references the system task or function argument by its position in the argument list \\
\hline & handle & tfinst & Handle to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_fetch_tfarg() or acc_fetch_itfarg() to get the task/function argument value as a double Use acc_fetch_tfarg_str() or acc_fetch_itfarg_str() to get the task/function argument value as a string Use acc_handle_tfinst() to get a handle to a specific instance of a user-defined system task or function} \\
\hline
\end{tabular}

The ACC routine acc_fetch_tfarg_int() shall return the value of arguments passed to the current userdefined system task or function. The ACC routine acc_fetch_itfarg_int() shall return the value of arguments passed to a specific instance of a user-defined system task and function, using a handle to the task or function. The value is returned as an integer number.

Argument numbers shall start at one and increase from left to right in the order that they appear in the system task or function call.

If an argument number is passed in that is out of range for the number of arguments in the user-defined system task/function call, acc_fetch_tfarg_int() and acc_fetch_itfarg_int() shall return a value of 0 and generate a warning message if warnings are enabled. Note that the acc_error_flag is not set for an out-ofrange index number.

If a user-defined system task/function argument that does not represent a valued object is referenced, acc_fetch_tfarg_int() and acc_fetch_itfarg_int() shall return a value of 0 and generate a warning message if warnings are enabled. Literal numbers, nets, regs, integer variables, and real variables all have values. Objects such as module instance names do not have a value. Note that the acc_error_flag is not set when a nonvalued argument is referenced.

If a user-defined task/function argument is a real value, the value is cast to a PLI_INT32 and returned as an integer. If the task/function argument is a string value, the string is copied into the ACC string buffer and the pointer to the string is cast to the type PLI_INT32 and returned as an integer.

Refer to Figure 92 for an example of using acc_fetch_tfarg_int().

\subsection*{23.32 acc_fetch_tfarg_str(), acc_fetch_itfarg_str()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_fetch_tfarg_str(), acc_fetch_itfarg_str()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the value of the specified argument of the system task or function associated with the PLI application; the value is returned as a pointer to a character string.} \\
\hline Syntax: & \multicolumn{3}{|l|}{\begin{tabular}{l}
acc_fetch_tfarg_str(argument_number) \\
acc_fetch_itfarg_str(argument_number, tfinst)
\end{tabular}} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_BYTE8 * & \multicolumn{2}{|l|}{The value of the task/function argument, returned as a pointer to a character string} \\
\hline \multicolumn{3}{|r|}{Type Name} & Description \\
\hline Arguments: & PLI_INT32 & argument_number & Integer number that references the system task or function argument by its position in the argument list \\
\hline & handle & tfinst & Handle to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_fetch_tfarg() or acc_fetch_itfarg() to get the task/function argument value as a double Use acc_fetch_tfarg_int() or acc_fetch_itfarg_int() to get the task/function argument value as an integer Use acc_handle_tfinst() to get a handle to a specific instance of a user-defined system task or function} \\
\hline
\end{tabular}

The ACC routine acc_fetch_tfarg_str() shall return the value of arguments passed to the current instance of a user-defined system task or function. The ACC routine acc_fetch_itfarg_str() shall return the value of arguments passed to a specific instance of a user-defined system task or function, using a handle to the task or function. The value shall be returned as a pointer to a character string. The return value for this routine is placed in the ACC internal string buffer. See 22.9 for explanation of strings in ACC routines.

Argument numbers shall start at one and increase from left to right in the order that they appear in the system task or function call.

If an argument number is passed in that is out of range for the number of arguments in the user-defined system task/function call, acc_fetch_tfarg_str() and acc_fetch_itfarg_str() shall return a value of null and generate a warning message if warnings are enabled. Note that the acc_error_flag is not set for an out-of-range index number.

If a user-defined system task/function argument that does not represent a valued object is referenced, acc_fetch_tfarg_str() and acc_fetch_itfarg_str() shall return a value of null and generate a warning message if warnings are enabled. Literal numbers, nets, regs, integer variables, and real variables all have values. Objects such as module instance names do not have a value. Note that the acc_error_flag is not set when a nonvalued argument is referenced.

If a user-defined task/function argument is a value, each 8 bits of the value are converted into its equivalent ASCII character.

Refer to Figure 92 for an example of using acc_fetch_tfarg_str().

\subsection*{23.33 acc_fetch_timescale_info()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_fetch_timescale_info()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get timescale information for an object or for an active \$timeformat system task invocation.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_fetch_timescale_info (object_handle, timescale_p)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & void & & \\
\hline \multirow{3}{*}{Arguments:} & Type & Name & Description \\
\hline & handle & object_handle & Handle of a module instance, module definition, PLI userdefined system task/function call, or null \\
\hline & p_timescale_info & timescale_p & Pointer to a variable defined as a s_timescale_info structure \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_fetch_precision() to fetch the smallest timescale precision in a design} \\
\hline
\end{tabular}

The ACC routine acc_fetch_timescale_info() shall obtain the timescale information for an object or for an active \$timeformat built-in system task invocation. The timescale returned shall be based on the type of object handle, as defined in Table 154.

Table 154-Return values from acc_fetch_timescale_info()
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ If the object_handle is } & \multicolumn{1}{c|}{ acc_fetch_timescale_info() shall return } \\
\hline \begin{tabular}{l} 
A handle to a module instance or module \\
definition
\end{tabular} & The timescale for the corresponding module definition \\
\hline \begin{tabular}{l} 
A handle to a user-defined system task or \\
function
\end{tabular} & \begin{tabular}{l} 
The timescale for the corresponding module definition \\
that represents the parent module instance of the object
\end{tabular} \\
\hline null & \begin{tabular}{l} 
The timescale for an active \$timeformat system task \\
invocation
\end{tabular} \\
\hline
\end{tabular}

The routine acc_fetch_timescale_info() shall return a value to an s_timescale_info structure pointed to by the timescale_p argument. This structure is declared in the file acc_user. h, as shown in Figure 82.
```

typedef struct t_timescale_info
{
PLI_INT16 unit;
PLI_INT16 precision;
s_timescale_info, *p_timescale_info;

```

Figure 93-s_timescale_info data structure
The term unit is a short integer that shall represent the timescale unit in all cases of object The term precision is a short integer that shall represent the timescale precision. In the case of a null object handle, precision shall be the number of decimal points specified in the active \(\$\) timeformat system task invocation.

The value returned for unit and precision shall be the order of magnitude of 1 s , as shown in Table 155.
Table 155-Value returned by acc_fetch_timescale_info()
\begin{tabular}{|c|c|}
\hline Integer value returned & Time unit r \\
\hline 2 & 100 s \\
\hline 1 & 10 s \\
\hline 0 & 1 s \\
\hline-1 & 100 ms \\
\hline-2 & 10 ms \\
\hline-3 & 1 ms \\
\hline-4 & 100 s \\
\hline-5 & 10 s \\
\hline-6 & 1 s \\
\hline-7 & 100 ns \\
\hline-8 & 10 ns \\
\hline-9 & 1 ns \\
\hline-10 & 100 ps \\
\hline-11 & 10 ps \\
\hline-12 & 1 ps \\
\hline-13 & 100 fs \\
\hline-14 & 10 fs \\
\hline-15 & 1 fs \\
\hline
\end{tabular}

For example, a call to
```

acc_fetch_timescale_info(obj, timescale_p)

```

Where obj is defined in a module that has 'timescale \(1 \mathrm{us} / 1 \mathrm{~ns}\) specified for its definition, shall return
```

timescale_p->unit: -6
timescale_p->precision: -9

```

\subsection*{23.34 acc_fetch_type()}


The ACC routine acc_fetch_type() shall return the type of an object. The type is a general classification of a Verilog HDL object, represented as a predefined constant (defined in acc_user.h). Refer to Table 113 for a list of all of the type constants that can be returned by acc_fetch_type().

Many Verilog HDL objects can have a type and a fulltype. The type of an object is its general Verilog HDL type classification. The fulltype is the specific type of the object. Table 143 illustrates the difference between the type of an object and the fulltype of the same object.

Table 156-The difference between the type and the fulltype of an object
\begin{tabular}{|c|c|c|}
\hline For a handle to & acc_fetch_type() shall return & acc_fetch_fulltype() shall return \\
\hline A setup timing check & accTchk & accSetup \\
\hline An and gate primitive & accPrimitive & accAndGate \\
\hline A sequential UDP & accPrimitive & accSeqPrim \\
\hline
\end{tabular}

The example shown in Figure 94 uses acc_fetch_type() to identify the type of an object (the functions display_primitive_type and display_timing_check_type used in this example are presented in the usage examples in 23.19).
```

include "acc_user.h"
'LI_INT32 display_object_type()
handle object_handle;
/*initialize environment for ACC routines*/
acc_initialize();
object_handle = acc_handle_tfarg(1);
/*display object type* /
switch(acc_fetch_type(object_handle) )
{
case accModule:
io_printf("Object is a module\n");
break;
case accNet:
io_printf("Object is a net\n");
break;
case accPath:
io_printf("Object is a module path\n");
break;
case accPort:
io_printf("Object is a module port\n");
break;
case accPrimitive:
display_primitive_type(object_handle);
break;
case accTchk:
display_timing_check_type(object_handle);
break;
case accTerminal:
io_printf("Object is a primitive terminal\n");
break;
}
acc_close();

```

Figure 94-Using acc_fetch_type()

\subsection*{23.35 acc_fetch_type_str()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_fetch_type_str()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get a string that indicates the type of its argument.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_fetch_type_str (type)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_BYTE8 * & \multicolumn{2}{|l|}{Pointer to a character string} \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_INT32 & type & A predefined integer constan or fulltype \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_fetch_type() to get the type of an object as an integer constant Use acc_fetch_fulltype() to get the fulltype of an object as an integer constant} \\
\hline
\end{tabular}

The ACC routine acc_fetch_type_str() shall return the character string that specifies the type of its argument. The argument passed to acc_fetch_type_str() should be an integer value returned from either acc_fetch_type() or acc_fetch_fulltype().

The return value for this routine is placed in the ACC internal string buffer. See 22.9 for explanation of strings in ACC routines.

In the example shown in Figure 95, a handle to an argument is passed to a C application. The application displays the name of the object and the type of the object.
```

\#include "acc_user.h"
PLI_INT32 display_object_type(object)
handle object;
{
PLI_INT32 type = acc_fetch type(object);
io_printf("Object %s is of type %s \n",
acc_fetch_fullname(object),
acc_fetch_type_str(type));
}

```

Figure 95—Using acc_fetch_type_str()
In this example, if the application is passed a handle to an object named top. param1, the application shall produce the following output:
```

Object top.param1 is of type accParameter

```

The output string, accParameter, is the name of the integer constant that represents the parameter type.

\subsection*{23.36 acc_fetch_value()}


The ACC routine acc_fetch_value() shall return logic simulation values for scalar or vector nets, reg, and integer, time and real variables; acc_fetch_value() shall return strength values for scalar nets and scalar regs only.

The routine acc_fetch_value() shall return the logic and strength values in one of two ways:
The value can be returned as a string
The value can be returned as an aval/bval pair in a predefined structure.

The return method used by acc_fetch_value() shall be controlled by the format_string argument, as shown in Table 157.

Table 157-How acc_fetch_value() returns values
\begin{tabular}{|c|c|c|}
\hline format_specifier & Return format & \multicolumn{2}{c}{ Description } \\
\hline\(\% \mathrm{~b}\) & binary & \begin{tabular}{l} 
Value shall be retrieved as a string, and a character pointer to the \\
string shall be returned
\end{tabular} \\
\hline\(\% \mathrm{~d}\) & decimal & \\
\hline\(\% \mathrm{~h}\) & hexadecimal & \\
\hline\(\% \mathrm{om}\) & octal & \multirow{3}{*}{} \\
\hline\(\% \mathrm{v}\) & strength & \multirow{2}{*}{\begin{tabular}{c} 
S_acc_value \\
structure
\end{tabular}}
\end{tabular} \begin{tabular}{l} 
Value shall be retrieved and placed in a structure variable pointed \\
to by the optional value argument
\end{tabular}

The string value returned shall have the same form as output from the formatted built-in system task \$display, in terms of value lengths and value characters used. The length shall be of arbitrary size, and unknown and high-impedance values shall be obtained. Note that strings are placed in a temporary buffer, and they should be preserved if not used immediately. Refer to 22.9 for details on preserving strings.

The \(\%\) v format shall return a three character string containing the strength code of a scalar net. Refer to 17.1.1.5 for the strength representations.

When a format_string of \(\% \%\) is specified, acc_fetch_value() shall retrieve the logic value and strength to a predefined structure, s_acc_value, which is defined in acc_user.h and is shown below [note that this structure definition is also used with the acc_set_value() routine].
```

typedef struct t setval value
{
PLI INT32 format;
union
{
PLI BYTE8 *str;
PLI-INT32 scalar;
PLI_INT32 integer;
double real;
p_acc_vecval vector;
} value;
} s_setval_value, *p_setval_value, s_acc_value, *p_acc_value;

```

Figure 96-s_acc_value structure
To use the \(\mathbf{\% \%} \%\) format_string to retrieve values to a structure requires the following steps:
a) A structure variable shall first be declared of type s_acc_value.
b) The format field of the structure has to be set to a predefined constant. The format controls which fields in the s_acc_value structure shall be used when acc_fetch_value() returns the value. The predefined constants for the format shall be one of the constants shown in Table 158.
c) The structure variable has to be passed as the third argument to acc_fetch_value().
d) The function return value from acc_fetch_value() should be ignored.

Table 158-Format constants for the s_acc_value structure
\begin{tabular}{|c|c|l|}
\hline Format constant & \begin{tabular}{c} 
acc_fetch_value() shall return the \\
value to the s_acc_value union field
\end{tabular} & \multicolumn{1}{c|}{ Description } \\
\hline accBinStrVal & str & value is retrieved in the same format as \%b \\
\hline accOctStrVal & str & value is retrieved in the same format as \%o \\
\hline accDecStrVal & str & value is retrieved in the same format as \%d \\
\hline accHexStrVal & str & value is retrieved in the same format as \%h \\
\hline accStringVal & str & \begin{tabular}{l} 
value is converted to a string, see Section 2.6 for a \\
description of Verilog strings
\end{tabular} \\
\hline accScalarVal & integer & \begin{tabular}{l} 
value is retrieved as one of the constants: acc0, \\
acc1, accZ or accX
\end{tabular} \\
\hline accIntVal & real & value is retrieved as a C integer \\
\hline accRealVal & vector & value is retrieved as a C double \\
\hline accVectorVal & & \begin{tabular}{l} 
value is represented as aval/bval pairs stored in an \\
array of s_acc_vecval structures
\end{tabular} \\
\hline
\end{tabular}

For example, calling acc_fetch_value() with the following setup would return a string in the value.str field. (This is essentially the same as using acc_fetch_value() with a \%b format string.)
```

s_acc_value value;
value.format = accBinStrVal;
(void)acc_fetch_value(Net, "%%", \&value);

```

If the format field for acc_fetch_value() is set to accVectorVal, then the value shall be placed in the record(s) pointed to by the value field. The value field shall be a pointer to an array of one or more s_acc_vecval structures. The s_acc_vecval structure is defined in the acc_user.h file and is listed in Figure \(\overline{9} 6\). The structure shall contain two integers: aval and bval. Each s_acc_vecval record shall represent 32 bits of a vector. The encoding for each bit value is shown in Table 159.
```

typedef struct t_acc_vecval
{
PLI_INT32 aval;
PLI_INT32 bval;
} s_acc_vecval, *p_acc_vecval;

```

Figure 97-s_acc_vecval structure

Table 159—Encoding of bits in the s_acc_vecval structure
\begin{tabular}{|c|c|c|}
\hline aval & bval & Value \\
\hline 0 & 0 & 0 \\
\hline 1 & 0 & 1 \\
\hline 0 & 1 & Z \\
\hline 1 & 1 & X \\
\hline
\end{tabular}

The array of s acc vecval structures shall contain a record for every 32 bits of the vector, plus a record for any remaining bits. If a vector has N bits, then there shall be ( \((\mathrm{N}-1) / 32)+1 \mathrm{~s} \_\)acc_vecval records. The routine acc_fetch_size() can be used to determine the value of N . The lsb of the vector shall be represented by the lsb of the first record of s_acc_vecval array. The 33rd bit of the vector shall be represented by the lsb of the second record of the array, and so on. See Figure 99 for an example of acc_fetch_value () used in this way.

Note that when using aval/bval pairs, the s_acc_value record and the appropriately sized s_acc_vecval array shall first be declared. Setting the second parameter to acc_fetch_value() to \%\% and the third parameter to null shall be an error.

The example application shown in Figure 98 uses acc_fetch_value() to retrieve the logic values of all nets in a module as strings.
```

include "acc_user.h"
LI_INT32 display_net_values()
handle mod, net;
/*initialize environment for ACC routines*/
acc_initialize();
/*get handle for module* /
mod = acc_handle_tfarg(1);
/*get all nets in the module and display their values*/
/* in binary format* /
net = null;
while(net = acc_next_net(mod, net))
io_printf("Net value: %s\n", acc_fetch_value(net,"%b", null));
acc_close();

```

Figure 98-Using acc_fetch_value() to retrieve the logic values as strings
The example in Figure 99 uses acc_fetch_value() to retrieve a value into a structure, and then prints the value. The example assumes the application, my_fetch_value, is called from the following user-defined system task:
```

\$my_fetch_value(R);

```
```

.nclude "acc_user.h"
sI_INT32 my_fetch_value()
handle reg = acc_handle_tfarg(1);
PLI_INT32 size = ((acc_fetch_size(reg) - 1) / 32) + 1;
s_acc_value value;
int indexl, min_size;
static PLI_BYTE8 table[ 4] = {' 0' ,' 1' ,' z' ,' x'} ;
static PLI_BYTE8 outString[ 33];
io_printf("The value of %s is ",acc_fetch_name(reg));
value.format = accVectorVal;
value.value.vector = (p_acc_vecval)malloc(size * sizeof(s_acc_vecval));
(void) acc_fetch_value(reg, "%%",\&value);
for (index1 = size - 1; index1 >= 0; index1--)
{
int index2;
PLI_INT32 abits = value.value.vector[ index1].aval;
PLI_INT32 bbits = value.value.vector[ index1].bval;
if (index1 == size - 1)
{
min_size = (acc_fetch_size(reg) % 32);
if (!min_size)
min_size-= 32;
}
else
min_size = 32;
outString[min_size] = '\0' ;
min_size--;
outString[min_size] = table[ ((bbits \& 1) << 1) | (abits \& 1)];
abits >>= 1;
for (index2 = min_size - 1; index2 >= 0; index2--)
{
outString[ index2] = table[ (bbits \& 2) | (abits \& 1)];
abits >>= 1;
bbits >>= 1;
}
io_printf("%s", outString);
}
io_printf("\n");
return(0);

```

Figure 99—Using acc_fetch_value() to retrieve values into a data structure

\subsection*{23.37 acc_free()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_free()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Frees memory allocated by acc_collect().} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_free (handle_array_pointer)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & void & \multicolumn{2}{|l|}{No return} \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Description \\
\hline & handle * & handle_array_pointer & Pointer to the array of handles allocated by acc_collect() \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_collect() to collect handles returned by acc_next_routines} \\
\hline
\end{tabular}

The ACC routine acc_free() shall deallocate memory that was allocated by the routine acc_collect().
The example shown in Figure 100 uses acc_free() to deallocate memory allocated by acc_collect() to collect handles to all nets in a module.
```

include "acc_user.h"
'LI_INT32 display_nets()
handle *list_of_nets, module_handle;
PLI_INT32 net_count, i;
/*initialize environment for ACC routines*/
acc_initialize();
/*get handle for module* /
module_handle = acc_handle_tfarg(1);
/* collect and display all nets in the module*/
list_of_nets = acc_collect(acc_next_net, module_handle, \&net_count);
for(i=0; i < net_count; i++)
io_printf("Net name is: %s\n", acc_fetch_name(list_of_nets[ i]));
/* free memory used by array list_of_nets*/
acc_free(list_of_nets);
acc_close();

```

Figure 100—Using acc_free()

\subsection*{23.38 acc_handle_by_name()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_handle_by_name()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the handle to any named object based on its name and scope.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_handle_by_name (object_name, scope_handle)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & handle & \multicolumn{2}{|l|}{A handle to the specified object} \\
\hline \multicolumn{2}{|r|}{Type} & Name & Description \\
\hline \multirow[t]{2}{*}{Arguments:} & quoted string or PLI_BYTE8 * & object_name & Literal name of an object or a character string pointer to the object name \\
\hline & handle & scope_handle & Handle to scope, or null \\
\hline \begin{tabular}{l}
Related \\
Routines
\end{tabular} & \multicolumn{3}{|l|}{Use acc_handle_object() to get a handle based on the local instance name of an object} \\
\hline
\end{tabular}

The ACC routine acc_handle_by_name() shall return the handle to any named object based on its specified name and scope. The routine can be used in two ways, as shown in Table 160.

Table 160-How acc_handle_by_name() works
\begin{tabular}{|c|l|}
\hline When the scope_handle is & \multicolumn{1}{c|}{ acc_handle_by_name() shall } \\
\hline A valid scope handle & Search for the object_name in the scope specified \\
\hline null & \begin{tabular}{l} 
Search for the object_name in the module containing the \\
current system task or function
\end{tabular} \\
\hline
\end{tabular}

The routine acc_handle_by_name() combines the functionality of acc_set_scope() and acc_handle_object(), making it possible to obtain handles for objects that are not in the local scope without having to first change scopes. Object searching shall conform to rules in 12.4 on hierarchical name referencing.

Table 161 lists the objects in a Verilog HDL description for which acc_handle_by_name() shall return a handle.

Table 161-Named objects supported by acc_handle_by_name()
\begin{tabular}{|c|c|}
\hline Modules & Parameters \\
\hline Primitives & Specparams \\
\hline Nets & Named blocks \\
\hline Regs & Verilog HDL tasks \\
\hline Integer, time and real variables & Verilog HDL functions \\
\hline Named events & \\
\hline
\end{tabular}

The routine acc_handle_by_name() does not return handles for module paths, intermodule paths, data paths, or ports. Use an appropriate acc_next_ or other ACC routines for these objects.

The example shown in Figure 101 uses acc_handle_by_name() to set the scope and get the handle to an object if the object is in the module.
```

include "acc_user.h"
'LI_INT32 is_net_in_module(module_handle, net_name)
andle module handle;
'LI_BYTE8 * net_name;
handle net_handle;
/* set scope to module and get handle for net */
net_handle = acc_handle_by_name(net_name, module_handle);
if (net_handle)
io_printf("Net %s found in module %s\n",
net_name,
acc_fetch_fullname(module_handle) );
else
io_printf("Net %s not found in module %s\n",
net_name,
acc_fetch_fullname(module_handle) );

```

Figure 101-Using acc_handle_by_name()
Note that in this example
```

net_handle = acc_handle_by_name(net_name, module_handle);

```
could also have been written as follows:
```

acc_set_scope(module_handle);
net_han\overline{d}le = acc_han\overline{d}le_object(net_name);

```

\subsection*{23.39 acc_handle_calling_mod_m}
\begin{tabular}{||l|l|l||}
\hline \multicolumn{4}{|l|}{ acc_handle_calling_mod_m } \\
\hline Synopsis: & \multicolumn{1}{l|}{\begin{tabular}{l} 
Get a handle to the module containing the instance of the user-defined system task or function that called the \\
PLI application.
\end{tabular}} \\
\hline Syntax: & acc_handle_calling_mod_m () \\
\hline \multirow{4}{|l|}{ Type } & Description \\
\cline { 2 - 3 } Returns: & handle & Handle to a module \\
\hline & Type & Name \\
Arguments: & None & Description \\
\hline
\end{tabular}

The ACC routine acc_handle_calling_mod_m shall return a handle to the module that contains the instance of the user-defined system task or function that called the PLI application.

\subsection*{23.40 acc_handle_condition()}
\begin{tabular}{||l|c|l||}
\hline \multicolumn{2}{|c|}{ acc_handle_condition() } \\
\hline Synopsis: & Get a handle to the conditional expression of a module path, data path, or timing check terminal. \\
\hline Syntax: & acc_handle_condition (path_handle) \\
\hline \multicolumn{3}{|c|}{ Type } \\
Returns: & handle & Handle to a conditional expression \\
\hline & Type & \multicolumn{2}{c|}{ Name } \\
\hline & handle & path_handle \\
Arguments: & \begin{tabular}{l} 
Handle to a module path, data path, or timing check \\
terminal
\end{tabular} \\
\hline
\end{tabular}

The ACC routine acc_handle_condition() shall return a handle to a conditional expression for the specified module path, data path, or timing check terminal. The routine shall return null when

The module path, data path, or timing check terminal has no condition specified The module path has an ifnone condition specified

To determine if a module path has an ifnone condition specified, use the ACC routine acc_object_of_type() to check for the property type of accModPathHasIfnone.

The example shown in Figure 102 provides functionality to see if a path is conditional, and, if it is, whether it is level-sensitive or edge-sensitive. The application assumes that the input is a valid handle to a module path.
```

int is_path_conditional(path)
if (acc_handle_condition(path) )
return(TRUE);
else
return(FALSE);
int is_level_sensitive(path)
int flag;
handle path_in = acc_next_input(path, null);
if (is_path_conditional(path) \&\& acc_fetch_edge(path_in))
flag = FALSE; /* path is edge-sensitive */
else
flag = TRUE; /* path is level_sensitive */
acc_release_object(path_in);
return (flag);

```

Figure 102-Using acc_handle_condition()

\subsection*{23.41 acc_handle_conn()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_handle_conn()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the handle to the net connected to a primitive terminal, path terminal, or timing check terminal.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_handle_conn(terminal_handle)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & handle & \multicolumn{2}{|l|}{Handle of a net} \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Description \\
\hline & handle & terminal_handle & Handle of the primitive terminal, path terminal, or timing check terminal \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_handle_terminal() or acc_next_terminal() to obtain a terminal_handle} \\
\hline
\end{tabular}

The ACC routine acc_handle_conn() shall return a handle to the net connected to a primitive terminal, path terminal, or timing check terminal. This handle can then be passed to other ACC routines to traverse a design hierarchy or to extract information about the design.

The example shown in Figure 103 displays the net connected to the output terminal of a gate.
```

include "acc_user.h"
'LI_INT32 display_driven_net()
handle gate_handle, terminal_handle, net_handle;
/*initialize environment for ACC routines*/
acc_initialize();
/*get handle for the gate*/
gate_handle = acc_handle_tfarg(1);
/*get handle for the gate's output terminal*/
terminal_handle = acc_handle_terminal(gate_handle, 0);
/*get handle for the net connected to the output terminal*/
net_handle = acc_handle_conn(terminal_handle);
/* display net name*/
io_printf("Gate %s drives net %s\n",
acc_fetch_fullname(gate_handle),
acc_fetch_name(net_handle) );
acc_close();

```

Figure 103-Using acc_handle_conn()

\subsection*{23.42 acc_handle_datapath()}
\begin{tabular}{||l|l|l||}
\hline \multicolumn{3}{|c|}{ acc_handle_datapath() } \\
\hline Synopsis: & Get a handle to a data path for an edge-sensitive module path. \\
\hline Syntax: & acc_handle_datapath (modpath_handle) \\
\hline \multicolumn{4}{|c|}{ Type } & Description \\
\cline { 2 - 3 } Returns: & handle & Handle of a data path \\
\hline & Type & Name \\
\cline { 2 - 3 } Arguments: & handle & modpath_handle \\
\hline
\end{tabular}

The ACC routine acc_next_datapath() shall return a handle to the data path associated with an edge-sensitive module path. If there is no data path, null shall be returned.

A data path is part of the Verilog HDL description for edge-sensitive module paths, as illustrated below:


The example shown in Figure 104 uses acc_handle_datapath() to find the data path corresponding to the specified module path and displays the source and destination port names for the data path.
```

PLI_INT32 display_datapath_terms(modpath)
handle modpath;
{
handle datapath = acc_handle_datapath(modpath);
handle pathin = acc_\overline{next_input(datapath, null);}
handle pathout = acc_next_output(datapath, null);
/* there is only one input and output to a datapath */
io_printf("DATAPATH INPUT: %s\n", acc_fetch_fullname(pathin));
io_printf("DATAPATH OUTPUT: %s\n", acc_fetch_fullname(pathout));
acc_release_object(pathin);
acc_release_object(pathout);
}

```

Figure 104—Using acc_handle_datapath()

\subsection*{23.43 acc_handle_hiconn()}


The ACC routine acc_handle_hiconn() shall return the hierarchically higher net connection for a scalar port or a bit-select of one of the following:

Vector port
Part-select of a port
Concatenation of scalar ports, vector ports, part-selects of ports, or other concatenations

The hiconn is the net connected one level above the hierarchical scope of a module port, as illustrated below:


The example shown in Figure 105 uses acc_handle_hiconn() and acc_handle_loconn() to display the higher and lower connections of a module port.
```

PLI_INT32 display_port_info(mod, index)
handle mod;
PLI_INT32 index;
{
handle port = acc_handle_port (mod, index);
handle hiconn, loconn, port_bit;
if (acc_fetch_size(port) == 1) {
hiconn = acc_handle_hiconn (port);
loconn = acc- handle-loconn (port);
io_printf ("' hi: %s lo: %s\n",
acc_fetch_fullname(hiconn), acc_fetch_fullname(loconn));
}
else {
port_bit = null;
whil\overline{e}(port_bit = acc_next_bit (port, port_bit))
{
hiconn = acc_handle_hiconn (port_bit);
loconn = acc_handle_loconn (port_bit);
io_printf (" hi: %s lo: %s\n",
acc_fetch_fullname(hiconn), acc_fetch_fullname(loconn));
}
}
}

```

Figure 105—Using acc_handle_hiconn() and acc_handle_loconn()

\subsection*{23.44 acc_handle_interactive_scope()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_handle_interactive_scope ()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get a handle to the current interactive scope of the software tool.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_handle_interactive_scope()} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type Description} & \\
\hline & handle & \multicolumn{2}{|l|}{Handle of a Verilog hierarchy scope} \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Descrip \\
\hline & None & & \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_fetch_type() or acc_fetch_fulltype() to determine the scope type returned Use acc_set_interactive_scope() to change the interactive scope} \\
\hline
\end{tabular}

The ACC routine acc_handle_interactive_scope() shall return a handle to the Verilog HDL design scope where the interactive mode of a software product is currently pointing.

A scope shall be
A top-level module
A module instance
A named begin-end block

A named fork-join block
A Verilog HDL task
A Verilog HDL function

\subsection*{23.45 acc_handle_loconn()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_handle_loconn()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Gets the hierarchically lower net connection to a scalar module port or a bit-select of a vector port.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_handle_loconn(port_ref_handle)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & handle & \multicolumn{2}{|l|}{Handle of a net} \\
\hline \multirow[b]{2}{*}{Arguments:} & \multicolumn{2}{|r|}{Type Name} & Description \\
\hline & handle & port_ref_handle & Handle to a scalar port or a bit-select of a vector port \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_next_loconn() to find all nets connected to a scalar port or bit-select of a port Use acc_handle_hiconn() to get the hierarchically higher net connection of a port} \\
\hline
\end{tabular}

The ACC routine acc_handle_loconn() shall return the hierarchically lower net connection for a scalar port or a bit-select of one of the following:

Vector port
Part-select of a port
Concatenation of scalar ports, vector ports, part-selects of ports, or other concatenations
The loconn is the net connected within the hierarchical scope of a module port, as illustrated below:


Refer to the usage example in 23.43 for an example of using acc_handle_loconn().

\subsection*{23.46 acc_handle_modpath()}


The ACC routine acc_handle_modpath() shall return a handle to a module path if one can be found. If a module path cannot be found the return value shall be null, the acc_error_flag will not be set. If any of the input args are improper a null shall be returned and the acc_error_flag will be set.

Table 162-How acc_handle_modpath() works
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ Setting of accEnableArgs } & \multicolumn{1}{c|}{ acc_handle_modpath(0 shall } \\
\hline "no_acc_handle_modpath" (the default setting) & \begin{tabular}{l} 
Use the name arguments and ignore both handle argu- \\
ments (the handle arguments can be dropped)
\end{tabular} \\
\hline \begin{tabular}{l} 
"acc_handle_modpath" and either source_name \\
or destination_name is null
\end{tabular} & \begin{tabular}{l} 
Use the handle argument of the null name argument; if \\
the name argument is not null, the name shall be used \\
and the associated handle argument ignored
\end{tabular} \\
\hline
\end{tabular}

A module path is the specify block path for delays in the Verilog HDL description. For example:


The example shown in Figure 106 uses acc_handle_modpath() to obtain handles for paths that connect the sources and destinations listed in the file pathconn. dat. The format of pathconn. dat is shown below.

```

nclude <stdio.h>
nclude "acc_user.h"
efine NAME_SIZE 256
I_INT32 get_paths()
FILE *infile;
PLI_BYTE8 mod_name[ NAME_SIZE], src_name[ NAME_SIZE], dest_name[ NAME_SIZE
handle path_handle, mod_handle;
/* initialize the environment for ACC routines */
acc_initialize();
/* set accPathDelimStr to " " */
acc_configure(accPathDelimStr, "_");
/* read delays from file - "r" means read only */
infile = fopen("pathconn.dat","r");
while (fscanf(infile, "%s %s %s",mod_name,src_name,dest_name) != EOF)
{
/* get handle for module mod_name */
mod_handle = acc_handle_object(mod_name);
path_handle = acc
if (path_handle)
io_printf("Path %s was found\n",
acc_fetch_fullname(path_handle) );
else
io_printf("Path %s_%s was not found\n", src_name, dest_name);
}
acc_close();

```

Figure 106—Using acc_handle_modpath()

\subsection*{23.47 acc_handle_notifier()}


The ACC routine acc_handle_notifier() shall return a handle to the notifier reg associated with a timing check.

The example shown in Figure 117 uses acc_handle_notifier() to display the name of a notifier associated with a timing check.

\subsection*{23.48 acc_handle_object()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|r|}{acc_handle_object} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get a handle for any named object.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_handle_object (object_name)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type Description} & \\
\hline & handle & Handle to an object & \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & \\
\hline & quoted string or PLI_BYTE8 * & object_name & Literal s ative hie \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_set_scope() to set the scope when using relative path names for an object} \\
\hline
\end{tabular}

The ACC routine acc_handle_object() shall return a handle to a named object. The object_name argument shall be a quoted string or pointer to a string. The object_name can include a Verilog hierarchy path. The routine shall search for the object using the rules given in Table 163.

Table 163-How acc_handle_object() works
\begin{tabular}{|c|l|}
\hline If object_name contains & \multicolumn{1}{c|}{ acc_handle_object() shall } \\
\hline \begin{tabular}{c} 
A full hierarchical path name \\
(a full hierarchical path begins with a top-level module)
\end{tabular} & Return a handle to the object; no search is performed \\
\hline No path name \\
or \\
a relative path name
\end{tabular} \begin{tabular}{l} 
Search for object starting in the current PLI scope, fol- \\
lowing search rules defined in Section 12.6
\end{tabular}

The ACC routine acc_handle_object() shall use the current PLI scope as a basis for searching for objects. The PLI scope shall default to the Verilog scope of the system task/function that called the C application of the user, and it can be changed from within the application using acc_set_scope().

Table 141 lists the objects in a Verilog HDL description for which acc_handle_object() shall return a handle.

Table 164-Named objects
\begin{tabular}{|c|c|}
\hline Modules & Named events \\
\hline Module ports & Parameters \\
\hline Data paths & Specparams \\
\hline Primitives & Named blocks \\
\hline Nets & Verilog HDL tasks \\
\hline Regs & Verilog HDL functions \\
\hline Integer, time and real variables & \\
\hline
\end{tabular}

The example shown in Figure 107 uses acc_handle_object() to retrieve handles for net names read from a file called primdelay.dat. The format of the file is shown below. Note that this example assumes that each net is driven by only one primitive.

```

include <stdio.h>
include "acc_user.h"
define NAME_SIZE 256
LI_INT32 write_prim_delays()
FILE *infile;
PLI_BYTE8 full_net_name[ NAME_SIZE] ;
double rise, fall;
handle net_handle, driver_handle, prim_handle;
/*initialize the environment for ACC routines*/
acc_initialize();
/*set accPathDelayCount parameter for rise and fall delays only*/
acc_configure(accPathDelayCount, "2");
/*read delays from file - "r" means read only*/
infile = fopen("primdelay.dat","r");
while (fscanf(infile,"%s %lf %lf",full_net_name,\&rise,\&fall) != EOF)
{
/*get handle for the net*/
net_handle = acc_handle_object(full_net_name);
/*get primitive connected to first net driver*/
driver_handle = acc_next_driver(net_handle, null);
prim_handle = acc_handle_parent(driver_handle);
/*replace delays with new values*/
acc_replace_delays(prim_handle, rise, fall);
}
acc_close();

```

Figure 107-Using acc_handle_object()

\subsection*{23.49 acc_handle_parent()}
\begin{tabular}{||l|l|l||}
\hline \multicolumn{3}{|c|}{ acc_handle_parent() } \\
\hline Synopsis: & Get a handle for the parent primitive instance or module instance of an object. \\
\hline Syntax: & acc_handle_parent (object_handle) \\
\hline \multicolumn{4}{|c|}{ Type } & Description \\
\cline { 2 - 4 } Returns: & handle & Handle of a primitive, port or module \\
\hline \multicolumn{4}{|c|}{ Type } & \multicolumn{2}{c|}{ Name } & Description \\
Arguments: & handle & object_handle \\
\hline
\end{tabular}

The ACC routine acc_handle_parent() shall return a handle to the parent of any object. A parent is an object that contains another object.

The parent of a terminal shall be the primitive that contains the terminal.
The parent of a port bit shall be the port that contains the bit.
The parent of any other object (except a top-level module) shall be the module instance that contains the object.
Top-level modules do not have parents. When a top-level module handle is passed to acc_handle_parent(), it shall return null.

The example shown in Figure 108 uses acc_handle_parent() to determine which terminals of a primitive drive a net.
```

include "acc_user.h"
'LI_INT32 get_primitives(net_handle)
.andle net_handle;
handle primitive_handle;
handle driver_handle;
/*get primitive that owns each terminal that drives the net*/
driver_handle = null;
while (driver_handle = acc_next_driver(net_handle, driver_handle) )
{
primitive_handle = acc_handle_parent(driver_handle);
io_printf("Primitive %s drives net %s\n",
acc_fetch_fullname(primitive_handle),
acc_fetch_fullname(net_handle) );
}

```

Figure 108—Using acc_handle_parent()
23.50 acc_handle_path()


The ACC routine acc_handle_path() shall return a handle to an intermodule path. An intermodule path shall be a net path that connects an output or inout port of one module to an input or inout port of another module.

The example shown in Figure 109 is a C code fragment that uses acc_handle_path() to fetch min:typ:max delays for the intermodule path referenced by intermod_path.
```

include "acc_user.h"
'LI_INT32 fetch_mintypmax_delays(port_output, port_input)
.andle port_output, port_input;

```
```

    handle intermod_path;
    ```
    handle intermod_path;
    double delay_array[ 9] ; 和_handle_path() returns a handle
    double delay_array[ 9] ; 和_handle_path() returns a handle
    . . . to a net path that represents the
    . . . to a net path that represents the
    acc_configure(accMinTypMaxDelays, "true"); connection from an output or inout
    acc_configure(accMinTypMaxDelays, "true"); connection from an output or inout
    .. port to an input (or inout) port
    .. port to an input (or inout) port
    intermod_path = acc_handle_path(port_output, port_input);
    intermod_path = acc_handle_path(port_output, port_input);
    acc_fetch_delays(intermod_path, delay_array);
```

    acc_fetch_delays(intermod_path, delay_array);
    ```

Figure 109—Using acc_handle_path()

\subsection*{23.51 acc_handle_pathin()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_handle_pathin()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get a handle for the first net connected to a module path source.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_handle_pathin(path_handle)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & handle & Handle to a net & \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Description \\
\hline & handle & path_handle & Handle of the module path \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_next_modpath() or acc_handle_modpath() to get path_handle} \\
\hline
\end{tabular}

The ACC routine acc_handle_pathin() shall return a handle to the net connected to the first source in a module path. If a module path has more than one input source, only the handle to the net connected to the first source shall be returned. For example:


The example shown in Figure 110 uses acc_handle_pathin() to find the net connected to the input of a path.
```

include "acc_user.h"
'LI_INT32 get_path_nets(path_handle)
an\overline{d}le path_handle;
handle pathin_handle, pathout_handle;
pathin_handle = acc_handle_pathin(path_handle);
pathout_handle = acc_handle_pathout(path_handle);
io_print̄f("Net conne\overline{c}ted to input is: %s\n",
acc_fetch_name(pathin_handle) );
io_printf("Net connected to output is: %s\n",
acc_fetch_name(pathout_handle) );

```

Figure 110-Using acc_handle_pathin()

\subsection*{23.52 acc_handle_pathout()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_handle_pathout()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get a handle for the first net connected to a module path destination.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_handle_pathout (path_handle)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & handle & \multicolumn{2}{|l|}{Handle to a net} \\
\hline \multicolumn{2}{|r|}{Type} & Name & Description \\
\hline Arguments: & handle & path_handle & Handle of the module path \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_next_modpath() or acc_handle_modpath() to get path_handle} \\
\hline
\end{tabular}

The ACC routine acc_handle_pathout() shall return a handle to the net connected to the first destination in a module path. If a module path has more than one output destination, only the handle to the net connected to the first destination shall be returned. For example:


The example shown in Figure 111 uses acc_handle_pathout() to find the net connected to the output of a path.
```

include "acc_user.h"
'LI_INT32 get_path_nets(path_handle)
andle path handle;
handle pathin_handle, pathout_handle;
pathin_handle = acc_handle_pathin(path_handle);
pathout_handle = acc_handle_pathout(path_handle);
io_printf("Net connected to input is: %s\n",
acc_fetch_name(pathin_handle) );
io_printf("Net connected to output is: %s\n",
acc_fetch_name(pathout_handle) );

```

Figure 111-Using acc_handle_pathout()

\subsection*{23.53 acc_handle_port()}
\begin{tabular}{||l|l|l||}
\hline \multicolumn{4}{|c|}{ acc_handle_port() } \\
\hline Synopsis: & Get a handle for a module port, based on the position of the port. \\
\hline Syntax: & acc_handle_port (module_handle, port_index) \\
\hline \multirow{3}{*}{ Returns: } & Type & Description \\
\cline { 2 - 4 } Arguments: & handle & Handle to a module port \\
\cline { 2 - 4 } & Type & Name \\
\hline
\end{tabular}

The ACC routine acc_handle_port() shall return a handle to a specific port of a module, based on the position of the port in the module declaration.

The index of a port shall be its position in a module definition in the source description. The indices shall be integers that start at \(\mathbf{0}\) and increase from left to right. Table 165 shows how port indices are derived.

Table 165—Deriving port indices
\begin{tabular}{|c|c|}
\hline For & Indices shall be \\
\hline Implicit ports: module \(\mathrm{A}(\mathrm{q}, \mathrm{a}, \mathrm{b})\); & \(\mathbf{0}\) for port \(\mathbf{q}\) 1 for port a 2 for port b \\
\hline ```
Explicit ports:
    module top;
        reg ra, rb;
        wire wq;
        explicit_port_mod epm1(.b(rb), .a(ra),.q(wq) );
    endmodule
    module explicit_port_mod(q, a, b);
        input a, b;
        output q;
        nand (q, a, b);
    endmodule
``` & \begin{tabular}{l}
0 for explicit port epm1.q \\
1 for explicit port epm1.a \\
2 for explicit port epm1.b
\end{tabular} \\
\hline
\end{tabular}

The example shown in Figure 112 uses acc_handle_port() to identify whether a particular module port is an output.
```

include "acc_user.h"
nt is_port_output(module_handle,port_index)
andle- module_handle;
'LI_INT32 port_index;
handle port_handle;
PLI_INT32 direction;
/*check port direction* /
port_handle = acc_handle_port(module_handle, port_index);
direction = acc_fetch_direction(port_handle);
if (direction == accOutput || direction == accInout)
return(true);
else
return(false);

```

Figure 112-Using acc_handle_port()

\subsection*{23.54 acc_handle_scope()}
\begin{tabular}{|c|c|c|c|c|}
\hline \multicolumn{5}{|c|}{acc_handle_scope ()} \\
\hline Synopsis: & \multicolumn{4}{|l|}{Get a handle to the scope that contains an object.} \\
\hline Syntax: & \multicolumn{4}{|l|}{acc_handle_scope (object_handle)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type Description} & & \\
\hline & handle & Handle of a scope & & \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & \multicolumn{2}{|r|}{Description} \\
\hline & handle & object_handle & Handle to an object & \\
\hline Related routines: & \multicolumn{4}{|l|}{Use acc_fetch_type() or acc_fetch_fulltype() to determine the scope type returned} \\
\hline
\end{tabular}

The ACC routine acc_handle_scope() shall return the handle to the scope of an object. A scope shall be
A top-level module
A module instance
A named begin-end block
A named fork-join block
A Verilog HDL task
A Verilog HDL function

The example shown in Figure 113 uses acc_handle_scope () to display the scope that contains an object.
```

PLI_INT32 get_scope(obj)
han\overline{dle obj;}
{
handle scope = acc_handle_scope(obj);
io_printf ("Scope %s contains object %s\n",
acc_fetch_fullname(scope), acc_fetch_name(obj);
}

```

Figure 113-Using acc_handle_scope()

\subsection*{23.55 acc_handle_simulated_net()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|r|}{acc_handle_simulated_net()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the simulated net associated with the collapsed net passed as an argument.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_handle_simulated_net(collapsed_net_handle)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type Description} & \\
\hline & handle & Handle of the simulated n & \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Desc \\
\hline & handle & collapsed_net_handle & Handle of a collapsed net \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_object_of_type() to determine if a net has been collapsed} \\
\hline
\end{tabular}

The ACC routine acc_handle_simulated_net() shall return a handle to the simulated net that is associated with a specified collapsed net. If a handle to a net that is not collapsed is passed into the routine, a handle to that same net shall be returned.

When a Verilog HDL source description connects modules together, a chain of nets with different scopes and names are connected, as is illustrated in the following simple diagram:


In this small circuit, nets out1, w5, and in1 are all tied together, effectively becoming the same net. Software products can collapse nets that are connected together within the data structure of the product. The resultant net after collapsing is referred to as the simulated net; the other nets are referred to as collapsed nets. The ACC routines can obtain a handle to any net, whether it is collapsed or not. The routine
acc_object_of_type() can be used to determine if a net has been collapsed, and the routine acc_handle_simulated_net() can be used to find the resultant net from the net collapsing process.

The example shown in Figure 114 uses acc_handle_simulated_net() to find all simulated nets within a particular scope. The application then displays each collapsed net, along with the simulated net. The ACC routine acc_object_of_type() is used with the property accCollapsedNet to determine whether a net has been collapsed onto another net.
```

include "acc_user.h"
'LI_INT32 display_simulated_nets()
handle mod handle;
handle simulated_net_handle;
handle net_handle;
/*reset environment for ACC routines*/
acc_initialize();
/*get scope-first argument passed to user-defined system task*/
/* associated with this routine*/
mod handle = acc_handle tfarg(1);
io_printf("In module %s:\n",acc_fetch_fullname(mod_handle) );
net_handle = null;
/*display name of each collapsed net and its net of origin*/
while(net_handle = acc_next_net(mod_handle,net_handle) )
{
if (acc_object_of_type(net_handle, accCollapsedNet) )
{
simulated_net_handle = acc_handle_simulated_net(net_handle);
io_printf(" - net %s was cōllapse\overline{d}
acc_fetch_name(net_handle),
acc_fetch_name(simulated_net_handle) );
}
}

```

Figure 114—Using acc_handle_simulated_net()

\subsection*{23.56 acc_handle_tchk()}


The ACC routine acc_handle_tchk() shall return a handle to a timing check based on arguments that describe the type of timing check, signals used, and edge qualifiers for the signals. The signals used to describe the timing check shall be passed as either signal names (passed as either a quoted string or a character string pointer) or signal handles. The number of signal arguments required by acc_handle_tchk() shall depend on the type of timing check.

Table 166 shows how the number of arguments for acc_handle_tchk() is determined.
Table 166-How acc_handle_tchk() works
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ If } & \multicolumn{1}{c|}{ acc_handle_tchk() shall } \\
\hline tchk_type is accWidth or accPeriod & \begin{tabular}{l} 
ignore arguments: second_arg_conn_name, \\
second_arg_edge_type, and optional \\
second_arg_conn_handle
\end{tabular} \\
\hline \begin{tabular}{l} 
tchk_type is accHold, accNochange, accRecovery, accSetup, or \\
accSkew
\end{tabular} & \begin{tabular}{l} 
use arguments: second_arg_conn_name, \\
second_arg_edge_type, and optional \\
second_arg_conn_handle
\end{tabular} \\
\hline \begin{tabular}{l} 
Default mode, or \\
acc_configure(accEnableArgs, no_acc_handle_tchk ) \\
has been called
\end{tabular} & \begin{tabular}{l} 
Use the name arguments and ignore both optional \\
handle arguments
\end{tabular} \\
\hline \begin{tabular}{l} 
The routine \\
acc_configure(accEnableArgs, acc_handle_tchk \()\) \\
has been called, and either first_arg_conn_name or \\
second_arg_conn_name is null
\end{tabular} & \begin{tabular}{l} 
Use the associated handle argument of the null \\
name argument if the name argument is not \\
null, the name shall be used and the associated \\
handle argument ignored
\end{tabular} \\
\hline
\end{tabular}

NOTE Unused arguments can be dropped if they do not precede any required arguments; otherwise, the unused arguments should be specified as null.

The routine acc_handle_tchk() shall use predefined edge group constants to represent groups of transitions among \(\mathbf{0}\), 1, and \(\mathbf{X}\) edge values, as described in Table 167. The routine shall treat transitions to or from a logic Z as transitions to or from a logic X .

Table 167-Edge group constants
\begin{tabular}{|c|c|}
\hline Edge group constant & Description of edge trigger \\
\hline accPosedge & Any positive transition: \\
accPosEdge & \(\mathbf{0}\) to \(\mathbf{1}\) \\
& \(\mathbf{0}\) to \(\mathbf{x}\) \\
& \(\mathbf{x}\) to \(\mathbf{1}\) \\
\hline accNegedge & Any negative transition: \\
accNegEdge & \(\mathbf{1}\) to \(\mathbf{0}\) \\
& \(\mathbf{1}\) to \(\mathbf{x}\) \\
& \(\mathbf{x}\) to \(\mathbf{0}\) \\
\hline accNoedge & Any transition: \\
accNoEdge & \(\mathbf{0}\) to \(\mathbf{1}\) \\
& \(\mathbf{1}\) to \(\mathbf{0}\) \\
& \(\mathbf{0}\) to \(\mathbf{x}\) \\
& \(\mathbf{x}\) to \(\mathbf{1}\) \\
& \(\mathbf{1}\) to \(\mathbf{x}\) \\
& \(\mathbf{x}\) to \(\mathbf{0}\) \\
\hline
\end{tabular}

The routine acc_handle_tchk() shall recognize predefined edge-specific constants that represent individual transitions among \(\mathbf{0}, \mathbf{1}\), and \(\mathbf{X}\) edge values that trigger timing checks, as described in Table 168.

Table 168—Edge specific constants
\begin{tabular}{|c|l|}
\hline Edge specific constant & \multicolumn{1}{|c|}{ Description of edge trigger } \\
\hline accEdge01 & Transition from \(\mathbf{0}\) to \(\mathbf{1}\) \\
\hline accEdge0x & Transition from \(\mathbf{0}\) to \(\mathbf{x}\) \\
\hline accEdgex \(\mathbf{~}\) & Transition from \(\mathbf{x}\) to \(\mathbf{1}\) \\
\hline accEdge10 & Transition from \(\mathbf{1}\) to \(\mathbf{0}\) \\
\hline accEdge1 \(\mathbf{~}\) & Transition from \(\mathbf{1}\) to \(\mathbf{x}\) \\
\hline accEdgex \(\mathbf{0}\) & Transition from \(\mathbf{x}\) to \(\mathbf{0}\) \\
\hline
\end{tabular}

The Verilog HDL allows multiple edges to be specified for timing checks. The routine acc_handle_tchk() shall recognize multiple edges using edge sums. Edge sums are lists of edge-specific constants connected by plus ( + ) signs. They represent the Verilog-HDL edge-control specifiers used by particular timing checks. Figure 115 shows a call to acc_handle_tchk() that accesses a \$width timing check containing edge-control specifiers.


Figure 115-Edge sums model edge-control specifiers
The example shown in Figure 116 uses acc_handle_tchk( \()\) to identify all cells in a module that contain either or both of the following timing checks:

A \$period timing check triggered by a positive edge on the clock signal clk
A \$setup timing check triggered on signal d by any transition and on signal clk by either of these clock edge transitions: \(\mathbf{1}\) to \(\mathbf{0}\) or \(\mathbf{X}\) to \(\mathbf{0}\)

Note that in this example:
a) Both calls to acc_handle_tchk() supply names for all relevant connections; therefore, the optional handle arguments are not supplied.
b) For \$period timing checks, acc_handle_tchk() ignores the second_arg_conn_name and second_arg_edge_type arguments; therefore, these arguments are not supplied.
```

include "acc_user.h"
'LI_INT32 get_ps_tchks()
handle module_handle, port_handle, net_handle, cell_handle;
/*initialize environment for ACC routines*/
acc_initialize();
/*get handle for module* /
module_handle = acc_handle_tfarg(1);
io_printf("Module is %s\n", acc_fetch_name(module_handle) );
/*scan all cells in module for: */
/* period timing checks triggered by a positive clock edge */
/* setup timing checks triggered by 1->0 and x->0 clock edges */
cell handle = null;
while(cell_handle = acc_next_cell(module_handle, cell_handle) )
{
if(acc_handle_tchk(cell_handle,accPeriod,"clk",accPosedge) )
io_printf("positive clock edge triggers period check in cell %s\n",
acc_fetch_fullname(cell_handle) );
if(acc_handle_t\overline{chk (cel̄l_handle,accSētup,"d",accNoedge,}
"clk",accEdge10+accEdgex0) )
io_printf("10 and x0 edges trigger setup check in cell %s\n",
acc_fetch_fullname(cell_handle) );
}
acc_close();

```

Figure 116-Using acc_handle_tchk()

\subsection*{23.57 acc_handle_tchkarg1()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_handle_tchkarg1()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get a handle for the timing check terminal connected to the first argument of a timing check.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_handle_tchkargl(tchk_handle)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type Description} & \\
\hline & handle & \multicolumn{2}{|l|}{Handle of a timing check terminal} \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Description \\
\hline & handle & tchk_handle & Handle of a timing check \\
\hline Related routines: & acc_han & nn() to get the net con & to a timing check terminal \\
\hline
\end{tabular}

The ACC routine acc_handle_tchkarg1() shall return a handle to the timing check terminal associated with the first argument of a timing check.

In order to trace a timing check terminal in the Verilog HDL description, or to display the name of the terminal, it is first necessary to obtain a handle to the net connected to the terminal. The routine acc_handle_conn() with the timing check terminal handle as the argument can be used to obtain the net handle.

The example shown in Figure 117 uses acc_handle_tchkarg1() and acc_handle_tchkarg2() to obtain the nets connected to the first and second arguments of each setup timing check in each cell under a module.
```

nclude "acc user.h"
I_INT32 show_check_nets()
{
handle module_handle, cell_handle;
handle tchk_handle, tchkarg1_handle, tchkarg2_handle, notifier_handle
PLI_INT32 tchk_type, counter;
/* initialize environment for ACC routines */
acc_initialize();
/* get handle for module* /
module_handle = acc_handle_tfarg(1);
io_prin}tf("module is s %s\n", acc_fetch_fullname(module_handle) );
/* scan all cells in module for timing checks */
cell_handle = null;
whil\overline{e (cell_handle = acc_next_cell(module_handle, cell_handle) )}
{
io_printf("cell is: %s\n", acc_fetch_fullname(cell_handle) );
counter = 0;
while (tchk_handle = acc_next_tchk(cell_handle, tchk_handle) )
{
/* get nets connected to timing check arguments */
tchk_type = acc_fetch_type(tchk_handle);
if (tchk_type == accSētup)
{
counter++;
io_printf(" for setup check \#%d:\n", counter);
tchkarg1_handle = acc_handle_tchkarg1(tchk_handle);
io_printf(" data net is %s\n",
acc fetch name(acc handle conn(tchkarg1 handle) );
tchkarg2_handle\overline{e}= ac\overline{c}_handle_\overline{tchkarg}\overline{2}(tchk_handle);
io_printf(" reference net is %s\n",
acc_fetch_name(acc_handle_conn(tchkarg2 handle) );
notifier_handle = acc_handle_notifier(tchk_handle);
if (notifier_handle != null)
io_printf(" notifier reg is %s\n",
acc_fetch_name(acc_handle_conn(notifier_handle) ) );
else
io_printf(" no notifier reg\n");
}
}
}
acc_close();

```

Figure 117-Using acc_handle_tchkarg1(), acc_handle_tchkarg2() and acc_handle_notifier()

\subsection*{23.58 acc_handle_tchkarg2()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_handle_tchkarg2()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get a handle for the timing check terminal connected to the second argument of a timing check.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_handle_tchkarg2(tchk_handle)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type Description} & \\
\hline & handle & Handle to a timing c & minal \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Description \\
\hline & handle & tchk_handle & Handle of a timing check \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_handle_conn() to get the net connected to a timing check terminal} \\
\hline
\end{tabular}

The ACC routine acc_handle_tchkarg2() shall return a handle to the timing check terminal associated with the second argument of a timing check.

In order to trace a timing check terminal in the Verilog HDL description, or to display the name of the terminal, it is first necessary to obtain a handle to the net connected to the terminal. The routine acc_handle_conn() with the timing check terminal handle as the argument can be used to obtain the net handle.

Refer to Figure 117 for an example of using acc_handle_tchkarg2().

\subsection*{23.59 acc_handle_terminal()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_handle_terminal()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get a handle for a primitive terminal based on the position of the primitive terminal.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_handle_terminal (primitive_handle, terminal_index)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & handle & \multicolumn{2}{|l|}{Handle of a primitive terminal} \\
\hline \multirow{3}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & handle & primitive_handle & Handle of a primitive \\
\hline & PLI_INT32 & terminal_index & Integer index of the desired terminal \\
\hline Related routines & \multicolumn{3}{|l|}{Use acc_handle_conn() to get the net connected to a primitive terminal} \\
\hline
\end{tabular}

The ACC routine acc_handle_terminal() shall return a handle of a primitive terminal based on the position of the terminal in the Verilog HDL source description.

The index of a terminal shall be its position in a gate, switch, or UDP declaration. The indices shall be integers that start at zero and increase from left to right. Table 169 shows how terminal indices are derived.

Table 169-Deriving terminal indices
\begin{tabular}{|c|c|}
\hline For & Indices shall be \\
\hline nand g1 (out, in1, in2); & \begin{tabular}{l}
\(\mathbf{0}\) for terminal out \\
\(\mathbf{1}\) for terminal in1 \\
\(\mathbf{2}\) for terminal in2
\end{tabular} \\
\hline
\end{tabular}

The example shown in Figure 118 uses acc_handle_terminal() to identify the name of a net connected to a primitive terminal.
```

include "acc_user.h"
'LI_INT32 print_terminal_net(gate_handle, term_index)
andle gate_handle;
'LI_INT32 term_index;
handle term_handle;
term_handle = -acc_handle_terminal (gate_handle, term_index);
io_p\overline{rintf("%s terminal net \#%d is %s\n",}
acc_fetch_name(gate_handle), term_index,
acc_fetch_name(acc_handle_conn(term_handle) ) );

```

Figure 118-Using acc_handle_terminal()
23.60 acc_handle_tfarg(), acc_handle_itfarg()


The ACC routine acc_handle_tfarg() shall return a handle to an argument in the current instance of a userdefined system task/function. The ACC routine acc_handle_itfarg() shall return a handle to an argument in a specific instance of a user-defined system task/function.

Argument numbers shall start at 1 and increase from left to right in the order that they appear in the system task or function call.

The system task/function argument can be:
A module instance
A primitive instance
A net, reg, integer variable, time variable, or real variable
A legal bit select of a net, reg, integer variable or time variable

Table 170-How acc_handle_tfarg() operates
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ When } & \multicolumn{1}{c|}{ acc_handle_tfarg() shall } \\
\hline \begin{tabular}{l} 
The system task or function argument is an \\
unquoted Verilog HDL identifier
\end{tabular} & Return a handle to the object \\
\hline \begin{tabular}{l} 
The system task or function argument is a quoted \\
string name of any object
\end{tabular} & \begin{tabular}{l} 
Function similar to acc_handle_object() by searching for an \\
object matching the string and, if found, returning a handle to \\
the object. \\
The object shall be searched for in the following order: \\
a) The current PLI scope [as set by acc_set_scope 0\(]\) \\
b) The scope of the system task/function
\end{tabular} \\
\hline
\end{tabular}

The example shown in Figure 119 uses acc_handle_tfarg() in a C language application that has the following characteristics:
a) It changes the rise and fall delays of a gate.
b) It takes three arguments the first is a Verilog HDL gate and the others are double-precision floating-point constants representing rise and fall delay values.
c) It associates through the PLI interface mechanism with a Verilog HDL system task called \$timing_task.

To invoke the application, the system task \$timing_task is called from the Verilog HDL source description, as in the following sample call:
```

\$timing_task(top.g12, 8.4, 9.2);

```

When Verilog encounters this call, it executes new_timing. A handle to the first argument, the gate top.g12, is retrieved using acc_handle_tfarg(), while the other two arguments the delay values are retrieved using acc_fetch_tfarg() .
```

\#include "acc_user.h"
PLI_INT32 new_timing()
{
handle gate_handle;
double new_rise, new_fall;
/*initialize and configure ACC routines*/
acc_initialize();
acc_configure(accToHiZDelay, "max");
/*get handle to gate*/
gate_handle = acc_handle_tfarg( 1 ); top.g12
/* get new delay values * /
new_rise = acc_fetch_tfarg( 2 ); 8.4
new_fall = acc_fetch_tfarg( 3 ); 9, 9.2
/*place new delays on the gate*/
acc_replace_delays(gate_handle,new_rise,new_fall);
/* report action */
io_printf("Primitive %s has new delays %d %d\n",
acc_fetch_fullname(gate_handle),
new_rise, new_fall);
acc_close();
}

```

Figure 119-Using acc_handle_tfarg()

\subsection*{23.61 acc_handle_tfinst()}


The ACC routine acc_handle_tfinst() is used to obtain a handle of the user-defined system task/function call that invoked the current PLI application.

\subsection*{23.62 acc_initialize()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_initialize()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Initializes the environment for ACC routines.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_initialize()} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{1 if successful; 0 if an error is encountered} \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Descri \\
\hline & None & & \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_configure() to set configuration parameter after calling acc_initialize() Use acc_close() at the end of a routine that called acc_initialize()} \\
\hline
\end{tabular}

The ACC routine acc_initialize() shall perform the following functions:
Initialize all configuration parameters to their default values
Allocate memory for string handling and other internal uses
The routine acc_initialize() should be called in a C language application before invoking any other ACC routines. Potentially, multiple PLI applications running in the same simulation session can interfere with each other because they share the same set of configuration parameters. To guard against application interference, both acc_initialize() and acc_close() reset any configuration parameters that have changed from their default values.

The example shown in Figure 120 uses acc_initialize() to initialize the environment for ACC routines.
```

:include "acc_user.h"
'LI_INT32 append_mintypmax_delays()
handle prim;

```

Figure 120—Using acc_initialize()

\subsection*{23.63 acc_next()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_next()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get handles to objects of each type specified in an array within the reference scope.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_next (object_type_array, reference_handle, object_handle)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & handle & \multicolumn{2}{|l|}{Handle of the object found} \\
\hline \multicolumn{2}{|r|}{Type} & Name & Description \\
\hline \multirow[t]{3}{*}{Arguments:} & \[
\begin{gathered}
\text { static } \\
\text { PLI_INT32 } \\
\text { array }
\end{gathered}
\] & object_type_array & Static integer array containing one or more predefined integer constants that represent the types of objects desired; the last element has to be \(\mathbf{0}\) \\
\hline & handle & reference_handle & Handle of a scope \\
\hline & handle & object_handle & Handle of the previous object found; initially null \\
\hline
\end{tabular}

The ACC routine acc_next() shall scan for and return handles to one or more types of objects within a scope. This routine performs a more general function than the object-specific acc_next_ routines, such as acc_next_net() and acc_next_primitive(), which scan only one type of object within a scope.

The objects for which acc_next() is to scan shall be listed as an array of object types or fulltypes in a static integer array. The array shall contain any number and combination of the predefined integer constants listed in Table 171. The array list shall be terminated by a \(\mathbf{0}\). The routine acc_next() can return objects in an arbitrary order.

The following \(C\) language statement is an example of declaring an array of object types called net_reg_list:
```

    static PLI_INT32 net_reg_list[ 3] = { accNet,accRegister,0} ;
    ```

When this array is passed to acc_next(), the ACC routine shall return handles to nets and regs within the reference object.

Note that a Verilog HDL function contains an object with the same name, size, and type as the function. If the function is scanned for objects of the type of the function, a handle to this object shall be returned.

The objects for which acc_next() shall obtain handles are listed in Table 171.

Table 171-Type and fulltype constants supported by acc_next()
\begin{tabular}{|c|c|c|}
\hline \multicolumn{2}{|c|}{Description} & Predefined integer constant \\
\hline \multirow[t]{9}{*}{General object types} & Integer variable & accIntegerVar \\
\hline & Module & accModule \\
\hline & Named event & accNamedEvent \\
\hline & Net & accNet \\
\hline & Primitive & accPrimitive \\
\hline & Real variable & accRealVar \\
\hline & Reg & accRegister \\
\hline & Time variable & accTimeVar \\
\hline & Parameter & accParameter \\
\hline \multirow[t]{3}{*}{Module fulltypes} & Top-level module & accTopModule \\
\hline & Module instance & accModuleInstance \\
\hline & Cell module instance & accCellInstance \\
\hline \multirow[t]{6}{*}{Net fulltypes} & Wire nets & accWire accTri \\
\hline & Wired-AND nets & accWand accTriand \\
\hline & Wired-OR nets & accWor accTrior \\
\hline & Pulldown, pullup nets & accTri0 accTri1 \\
\hline & Supply nets & accSupply0 accSupply1 \\
\hline & Storage nets & accTrireg \\
\hline \multirow[t]{3}{*}{Parameter fulltypes} & Integer parameters & accIntegerParam \\
\hline & Real parameters & accRealParam \\
\hline & String parameters & accStringParam \\
\hline
\end{tabular}

Table 171-Type and fulltype constants supported by acc_next() (continued)
\begin{tabular}{|c|c|c|}
\hline \multicolumn{2}{|c|}{Description} & Predefined integer constant \\
\hline \multirow[t]{9}{*}{Primitive fulltypes} & N -input, 1-output gates & accAndGate accNandGate accNorGate accOrGate accXnorGate accXorGate \\
\hline & 1-input, N -output gates & accBufGate accNotGate \\
\hline & Tri-state gates & \begin{tabular}{l}
accBufif0 \\
accBufif1 \\
accNotif0 \\
accNotif1
\end{tabular} \\
\hline & MOS gates & accNmosGate accPmosGate accRnmosGate accRpmosGate \\
\hline & CMOS gates & accCmosGate accRemosGate \\
\hline & Bidirectional pass gates & accRtranGate accRtranif0Gate accRtranif1Gate accTranGate accTranif0Gate accTranif1Gate \\
\hline & Pulldown, pullup gates & accPulldownGate accPullUpGate \\
\hline & Combinational UDP & accCombPrim \\
\hline & Sequential UDP & accSeqPrim \\
\hline
\end{tabular}

The example shown in Figure 121 uses acc_next() to find all nets and regs in a module. The application then displays the names of these nets and reg.
```

"include "acc_user.h"
'LI_INT32 display_nets_and_registers()
static PLI_INT32 net_reg_list[3] = {accNet,accRegister,0};
handle
mod_handle, obj_handle;
/*reset environment for ACC routines*/
acc_initialize();
/*get handle for module-first argument passed to*/
/* user-defined system task associated with this routine*/
mod_handle = acc_handle_tfarg(1);
io_printf("Module %s contains these nets and registers:\n",
acc_fetch_fullname(mod_handle) );
/*display names of all nets and registers in the module*/
obj_handle = null;
while (obj_handle = acc_next(net_reg_list,mod_handle,obj_handle) )
io_printf(" %s\n",_ acc_fetch_name(obj_handle) );
acc_close();

```

Figure 121—Using acc_next()

\subsection*{23.64 acc_next_bit()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_next_bit()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get handles to bits in a port or expanded vector.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_next_bit(reference_handle, bit_handle)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type Description} & \\
\hline & handle & \multicolumn{2}{|l|}{Handle of a port bit, vector bit or path terminal bit} \\
\hline \multirow{3}{*}{Arguments:} & Type & Name & Description \\
\hline & handle & reference_handle & Handle of a p \\
\hline & handle & bit_handle & Handle of the \\
\hline Related routines: &  & ) to return the next port ort() to return the handle type() to determine if a & module module port \(r\) is expanded \\
\hline
\end{tabular}

The ACC routine acc_next_bit() shall obtain handles to the bits of a vector port, an expanded vector, or a path terminal.

An expanded vector is a vector for which a software product shall permit access to the discrete bits of the vector. The routine acc_object_of_type() can be used to determine if a vector reference handle is expanded before calling acc_next_bit() with the vector handle. For example:
```

if (acc object of type(vector handle, accExpandedVector) )
while (bit_handle = acc_next_bit(vector_handle, bit_handle) )

```

When the reference_handle object is a vector, the first call to acc_next_bit() shall return the handle to the msb (leftmost bit) of the object. Subsequent calls shall return the handles to the remaining bits down to the lsb (rightmost bit). The call after the return of the handle to the lsb returns null. When the reference_handle is scalar, acc_next_bit() shall treat the object as a 1-bit vector.

The example shown in Figure 122 uses acc_next_bit() to display the lower connection of each bit of a port.
```

include "acc_user.h"
LI_INT32 display_port_bits(module_handle, port_number)
and\overline{le module_handle;}
LI_INT32 port_number;
handle port_handle, bit_handle;
/* get handle for port * /
port_handle = acc_handle_port(module_handle, port_number);
/* display port number and module instance name */
io_printf("Port %d of module %s contains the following bits: \n",
port_number, acc_fetch_fullname(module_handle) );
/* display lowēr hierarchi\overline{cal con̄nection of each bit */}
bit_handle = null;
while (bit_handle = acc_next_bit(port_handle, bit_handle) )
io_printf(" %s\n",acc_fetch_fullname(bit_handle) );

```

Figure 122-Using acc_next_bit() with module ports

The example shown in Figure 123 uses acc_next_bit() to assign a VCL monitor flag to each bit of a vector net.
```

-nclude "acc_user.h"
\lrcornerI_INT32 monitor_bits()
handle bit_handle, net_handle, mod_handle;
/* reset environment for ACC routines */
acc_initialize();
/* get handle for system task argument associated with this routine */
mod_handle = acc_handle_tfarg(1);
/* get handles to all nets in the module */
net handle = null;
while (net_handle = acc_next_net(mod_handle, net_handle) )
{
/* add VCL monitor each bit of expanded vector nets */
if (acc_object_of_type(net_handle, accExpandedVector) )
{
bit handle = null;
while (bit_handle = acc_next_bit(net_handle, bit_handle) )
acc_vcl_add(bit_handle, net_consumer, null, vcl_verilog_logic)
}
}

```

Figure 123-Using acc_next_bit() with a vector net
23.65 acc_next_cell()
\begin{tabular}{||l|l|l||}
\hline \multicolumn{4}{|c|}{ acc_next_cell() } \\
\hline Synopsis: & Get handles to cell instances within a region that includes the entire hierarchy below a module. \\
\hline Syntax: & acc_next_cell_(reference_handle, cell_handle) \\
\hline \multirow{4}{*}{ Returns: } & \multicolumn{2}{|c|}{ Type } \\
\cline { 2 - 4 } Arguments: & handle & Handle of a cell module \\
\hline & Type & Name \\
\cline { 2 - 4 } & handle & reference_handle \\
\hline
\end{tabular}

The ACC routine acc_next_cell() shall return handles to the cell module instances in the reference scope and all module instance scopes below the reference scope. The routine shall not find cells that are instantiated inside other cells.

A cell instance shall be a module instance that has either of these characteristics:

The module definition appears between the compiler directives' celldefine and 'endcelldefine.
The module definition is in a model library, where a library is a collection of module definitions in a file or directory that are read by library invocation options.

The example shown in Figure 124 uses acc_next_cell() to list all cell instances at or below a given hierarchy scope.
```

!include "acc_user.h"
'LI_INT32 list_cells()
handle module_handle, cell_handle;
/*initialize environment for ACC routines*/
acc_initialize();
/* get handle for module*/
module_handle = acc_handle_tfarg(1);
io_printf("%s contains the following cells:\n",
acc_fetch_fullname(module_handle) );
/*display names of all cells in the module*/
cell handle = null;
while\overline{e}(cell_handle = acc_next_cell(module_handle,cell_handle) )
io_printf(" %s\n",acc_fetch_fullname(cell_handle) );
acc_close();

```

Figure 124—Using acc_next_cell()
23.66 acc_next_cell_load()
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_next_cell_load()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get handles for cell loads on a net.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_next_cell_load (reference_handle, load_handle)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & handle & \multicolumn{2}{|l|}{Handle of a primitive input terminal} \\
\hline \multicolumn{2}{|r|}{Type} & Name & Description \\
\hline \multirow[t]{2}{*}{Arguments:} & handle & reference_handle & Handle of a scalar net or bit select of a vector net \\
\hline & handle & load_handle & Handle of the previous load found; initially null \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_next_load() to get a handle to all primitive input terminal loads} \\
\hline
\end{tabular}

The ACC routine acc_next_cell_load() shall return handles to the cell module instances that are driven by a net. The handle for a cell load shall be a primitive input terminal connected to an input or inout port of the cell load instance.

The routines acc_next_load() and acc_next_cell_load() have different functionalities. The routine acc_next_load() shall return every primitive input terminal driven by a net, whether it is inside a cell or a module instance. The routine acc_next_cell_load() shall return only one primitive input terminal per cell input or inout port driven by a net. Figure 125 illustrates the difference, using a circuit in which net1 drives primitive gates in cell1, cell2, and module1. For this circuit, acc_next_load() returns four primitive input terminals as loads on net1, while acc_next_cell_load() returns two primitive input terminals as loads on net1.


Figure \(\mathbf{1 2 5}\)-The difference between acc_next_load() and acc_next_cell_load()

The example shown in Figure 126 uses acc_next_cell_load() to find all cell loads on a net.
```

!include "acc_user.h"
'LI_INT32 get_cell_loads()
handle net_handle;
handle loa\overline{d_handle, load_net_handle;}
/*initialize environment for ACC routines*/
acc_initialize();
/*get handle for net*/
net_handle = acc_handle_tfarg(1);
/*display names of all cell loads on the net*/
load_handle = null;
while(load_handle = acc_next_cell_load(net_handle,load_handle) )
{
load_net handle = acc handle_conn(load_handle);
io_print\overline{f("Cell load is conneccted to: %}%\n",
acc_fetch_fullname(load_net_handle) );
}
acc_close();

```

Figure 126-Using acc_next_cell_load()

\subsection*{23.67 acc_next_child()}
\begin{tabular}{||l|l|l||}
\hline \multicolumn{4}{|c|}{ acc_next_child() } \\
\hline Synopsis: & Get handles for children of a module. \\
\hline Syntax: & acc_next_child_reference_handle, child_handle) \\
\hline \multirow{4}{*}{ Returns: } & Type & Description \\
\cline { 2 - 4 } Arguments: & handle & Handle of a module instance \\
\cline { 2 - 4 } & Type & \multicolumn{2}{c|}{ Name } \\
\hline & handle & reference_handle \\
handle & child_handle & Handle of a module \\
\hline
\end{tabular}

The ACC routine acc_next_child() shall return handles to the module instances (children) within the reference module. The routine shall also return handles to top-level modules, as shown in Table 172.

Table 172-How acc_next_child() works
\begin{tabular}{|c|l|}
\hline When & \multicolumn{1}{c|}{ acc_next_child() shall } \\
\hline The reference_handle is not null & \begin{tabular}{l} 
Scan for modules instantiated inside the module associated \\
with reference_handle
\end{tabular} \\
\hline The reference_handle is null & Scan for top-level modules (same as acc_next_topmod \((\) ) ) \\
\hline
\end{tabular}

The ACC routine acc_next_topmod() does not work with acc_collect() or acc_count(), but acc_next_child() with a null reference handle argument can be used in place of acc_next_topmod(). For example:
```

    acc_count(acc_next_child, null); /* counts top-level modules */
    acc_collect(acc_next_child, null, &count); /* collect top-level
    modules */

```

Figure 127 shows the use of acc_next_child() to display the names of all modules instantiated within a module.
```

!include "acc_user.h"
'LI_INT32 print_children(module_handle)
randle module_handle;
handle child_handle;
io_printf("Module %s contains the following module instances:\n",
acc_fetch_fullname(module_handle) );
child_handle = null;
while(child_handle = acc_next_child(module_handle, child_handle) )
io_print\overline{f(" %s\n",\overline{acc_fetch_name(chi\overline{l}d_handle) );}}\mathbf{~}=\overline{l}

```

Figure 127—Using acc_next_child()

\subsection*{23.68 acc_next_driver()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_next_driver()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get handles to primitive terminals that drive a net.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_next_driver(reference_handle, driver_handle)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & handle & \multicolumn{2}{|l|}{Handle of a primitive terminal} \\
\hline \multicolumn{2}{|r|}{Type} & Name & Description \\
\hline \multirow[t]{2}{*}{Arguments:} & handle & reference_handle & Handle of a scalar net or bit select of a vector net \\
\hline & handle & driver_handle & Handle of the previous driver found; initially null \\
\hline
\end{tabular}

The ACC routine acc_next_driver() shall return handles to the primitive output or inout terminals that drive a net.

The example shown in Figure 128 uses acc_next_driver() to determine which terminals of a primitive drive a net.
```

include "acc_user.h"
'LI_INT32 print_drivers(net_handle)
andle net_handle;
handle primitive_handle;
handle driver_handle;
io_printf("Net %s is driven by the following primitives:\n",
acc_fetch_fullname(net_handle) );
/*get primitive that owns each terminal that drives the net*/
driver_handle = null;
while (driver_handle = acc_next_driver(net_handle, driver_handle) )
{
primitive_handle = acc_handle_parent(driver_handle);
io_printf(" %s\n",
acc_fetch_fullname(primitive_handle) );
}

```

Figure 128-Using acc_next_driver()

\subsection*{23.69 acc_next_hiconn()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_next_hiconn()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get handles for hierarchically higher net connections to a module port.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_next_hiconn(reference_handle, net_handle)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & handle & \multicolumn{2}{|l|}{Handle of a net} \\
\hline \multirow{3}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & handle & reference_handle & Handle of a port \\
\hline & handle & net_handle & Handle of the previous net found; initially null \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_handle_hiconn() to get a handle to hierarchically higher connection of a specific port bit Use acc_next_loconn() to get handles to the hierarchically lower connection} \\
\hline
\end{tabular}

The ACC routine acc_next_hiconn() shall return handles to the hierarchically higher net connections to a module port. A hierarchically higher connection shall be the part of the net that appears outside the module, as shown in the following diagram:


When the reference handle passed to acc_next_hiconn() is a vector port, the routine shall return the hiconn nets bit-by-bit, starting with the msb (leftmost bit) and ending with the lsb (rightmost bit).

The example shown in Figure 129 uses acc_next_hiconn() and acc_next_loconn() to find and display all net connections made externally (hiconn) and internally (loconn) to a module port.
```

include "acc_user.h"
'LI_INT32 display_connections(module_handle, port_handle)
.andle module_handle, port_handle;
handle hiconn_net, loconn_net;
/*get and display low connections* /
io_printf("For module %s, port \#%d internal connections are:\n",
acc_fetch_fullname(module_handle),
acc-fetch_index(port handle) );
loconn net = nūll;
while (loconn_net = acc_next_loconn(port_handle, loconn_net) )
io_printf(" %s\n", acc_fetch_fullname(loconn_net) );
/*get and display high connections*/
io_printf("For module %s, port \#%d external connections are:\n",
acc fetch fullname(module handle),
acc_fetch_index(port_handle) );
hiconn_net = null;
while (hiconn_net = acc_next_hiconn(port_handle, hiconn_net) )
io_printf(" %s\n", acc_fetch_fullname(hiconn_net) );

```

Figure 129—Using acc_next_hiconn() and acc_next_loconn()

\subsection*{23.70 acc_next_input()}


The ACC routine acc_next_input() shall return handles to the input path terminals of a module path, the source terminals of a data path or the timing check terminals of a timing check. The routine acc_handle_conn() can be passed the input path terminal handle to derive the net connected to the terminal.

A module path is the specify block path for delays in the Verilog HDL description. A data path is part of the Verilog HDL description for edge-sensitive module paths, as shown in the following diagram:


The example shown in Figure 130 uses acc_next_input(). It accepts a handle to a scalar net or a net bitselect, and a module path. The application returns true if the net is connected to the input of the path.
```

int is net on path input(net, path)
handle net; /* scalar net or bit-select of vector net */
handle path;
{
handle pterm_in, pterm_conn, bit;
/* scan path input terminals */
pterm_in = null;
while (pterm_in = acc_next_input(path, pterm_in) )
{
/* retrieve net connected to path terminal */
pterm_conn = acc_handle_conn (pterm_in);
bit = null;
if (acc_object_of_type (pterm_conn, accExpandedVector) )
{
bit = null;
while (bit = acc_next_bit (pterm_conn, bit) )
if (acc_compare_handles (bit, net) )
return (true);
}
else
if (acc_compare_handles(bit, net) )
return (true);
}
return (false);
}

```

Figure 130—Using acc_next_input()

\subsection*{23.71 acc_next_load()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_next_load()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get handles to primitive terminals driven by a net.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_next_load (reference_handle, load_handle)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & handle & \multicolumn{2}{|l|}{Handle of a primitive terminal} \\
\hline \multicolumn{2}{|r|}{Type} & Name & Description \\
\hline \multirow[t]{2}{*}{Arguments:} & handle & reference_handle & Handle of a scalar net or bit select of a vector net \\
\hline & handle & load_handle & Handle of the previous load found; initially null \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_next_cell_load() to get cell module loads} \\
\hline
\end{tabular}

The ACC routine acc_next_load() shall return handles to the primitive loads that are being driven by a net. The handle for a load shall be a primitive input terminal.

The routines acc_next_load() and acc_next_cell_load() have different functionalities. The routine acc_next_load() shall return every primitive input terminal driven by a net, whether it is inside a cell or a module instance. The routine acc_next_cell_load() shall return only one primitive input terminal per cell input or inout port driven by a net. Figure 131 illustrates the difference, using a circuit in which net 1 drives primitive gates in cell1, cell2, and module1. For this circuit, acc_next_load() returns four primitive input terminals as loads on net1, while acc_next_cell_load() returns two primitive input terminals as loads on net1.


Figure 131 - The difference between acc_next_load() and acc_next_cell_load()

The example shown in Figure 132 uses acc_next_load() to find all terminals driven by a net.
```

include "acc_user.h"
'LI_INT32 get_loads()
handle net_handle, load_handle, load_net_handle;
/*initialize the environment for ACC routines*/
acc_initialize();
/*get handle for net* /
net_handle = acc_handle_tfarg(1);
io_printf("Net %s is driven by:\n",acc_fetch_fullname(net_handle) );
/*get primitive that owns each terminal driven by the net*/
load_handle = null;
whilè (load_handle = acc_next_load(net_handle, load_handle) )
{
load_net_handle = acc_handle_conn(load_handle);
io_printf(" %s ",
acc_fetch_fullname(load_net_handle) );
}
acc_close();

```

Figure 132-Using acc_next_load()

\subsection*{23.72 acc_next_loconn()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_next_loconn()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get handles to hierarchically lower net connections to a port of a module.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_next_loconn(reference_handle, net_handle)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & handle & \multicolumn{2}{|l|}{Handle of a net} \\
\hline \multicolumn{2}{|r|}{Type} & Name & Description \\
\hline \multirow[t]{2}{*}{Arguments:} & handle & reference_handle & Handle of a port \\
\hline & handle & net_handle & Handle of the previous net found; initially null \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_handle_loconn() to get a handle to hierarchically lower connection of a specific port bit Use acc_next_hiconn() to get handles to the hierarchically higher connection} \\
\hline
\end{tabular}

The ACC routine acc_next_loconn() shall return handles to the hierarchically lower net connections to a module port. A hierarchically lower connection shall be the part of the net that appears inside the module, as shown in the following diagram:


When the reference handle passed to acc_next_loconn() is a vector port, the routine shall return the loconn nets bit-by-bit, starting with the msb (leftmost bit) and ending with the lsb (rightmost bit).

Refer to Figure 129 for an example of using acc_next_loconn().

\subsection*{23.73 acc_next_modpath()}
\begin{tabular}{||l|l|l||}
\hline \hline \multicolumn{4}{|c|}{ acc_next_modpath() } \\
\hline Synopsis: & Get handles to module paths of a module. \\
\hline Syntax: & acc_next_modpath (reference_handle, path_handle) \\
\hline \multirow{4}{*}{ Returns: } & Type & Description \\
\cline { 2 - 4 } Arguments: & handle & Handle of a module path \\
\cline { 2 - 4 } & Type & Name \\
\hline
\end{tabular}

The ACC routine acc_next_modpath() shall return handles to the module paths in a module. A module path is the specify block path for delays in the Verilog HDL description. For example:


The example in Figure 133 uses acc_next_modpath() to list the nets connected to all module paths in a module.
```

nclude "acc_user.h"
,I_INT32 get_path_nets(module_handle)
.ndle module_handle;
randle path_handle, pathin_handle, pathout_handle;
*scan all paths in the module */
o_printf("For module %s:\n",acc_fetch_fullname(module_handle) );
ath_handle = null;
rhile (path_handle = acc_next_modpath(module_handle, path_handle) )
io_printf(" path %s connections are:\n",acc_fetch_name(path_handle) );
pathin_handle = acc_handle_pathin(path_handle);
pathout_handle = acc_handle_pathout(path_handle);
io_printf("net %s con̄nected to input\n",\overline{c}cc_fetch_name(pathin_handle) );
io_printf("net %s connected to output\n",acc_fetch_name(pathout_handle)

```

Figure 133-Using acc_next_modpath()
23.74 acc_next_net()
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_next_net()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get handles to nets in a module.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_next_net(reference_handle, net_handle)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & handle & \multicolumn{2}{|l|}{Handle of a net} \\
\hline \multicolumn{2}{|r|}{Type} & Name & Description \\
\hline \multirow[t]{2}{*}{Arguments:} & handle & reference_handle & Handle of a module \\
\hline & handle & net_handle & Handle of the previous net found; initially null \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_object_of_type() to determine if a net is scalar or vector, expanded or unexpanded Use acc_next_bit() to get handles to all bits of an expanded vector net} \\
\hline
\end{tabular}

The ACC routine acc_next_net() shall return handles to the nets within a module scope. The routine shall return a handle to a vector net as a whole; it does not return a handle to each individual bit of a vector net. The routine acc_object_of_type() can be used to determine if a net is vector or scalar and if it is expanded or unexpanded. The routine acc_next_bit() can be used to retrieve a handle for each bit of an expanded vector net.

The example shown in Figure 134 uses acc_next_net() to display the names of all nets in a module.
```

"include "acc_user.h"
'LI_INT32 display_net_names()
handle mod_handle, net_handle;
/*initialize environment for ACC routines*/
acc_initialize();
/* get handle for module* /
mod_handle = acc_handle_tfarg(1);
io_printf("Module %s contains the following nets:\n",
acc_fetch_fullname(mod_handle) );
/*display names of all nets in the module*/
net handle = null;
while (net_handle = acc_next_net(mod_handle, net_handle) )
io_printf(" %s\n", acc_fetch_name(net_handle) );
acc_close();

```

Figure 134—Using acc_next_net()

\subsection*{23.75 acc_next_output()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_next_output()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get handles to output path terminals of a module path or data path.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_next_output(reference_handle, terminal_handle)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & handle & \multicolumn{2}{|l|}{Handle to a module path terminal or data path terminal} \\
\hline \multicolumn{2}{|r|}{Type} & Name & Description \\
\hline \multirow[t]{2}{*}{Arguments:} & handle & reference_handle & Handle to a module path or data path \\
\hline & handle & terminal_handle & Handle of the previous terminal found; initially null \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_handle_conn() to get the net attached to the path terminal Use acc_release_object() to free memory allocated by acc_next_output()} \\
\hline
\end{tabular}

The ACC routine acc_next_output() shall return handles to the output path terminals of a module path or a data path. The routine acc_handle_conn() can be passed the output path terminal handle to derive the net connected to the terminal.

A module path is the specify block path for delays in the Verilog HDL description. A data path is part of the Verilog HDL description for edge-sensitive module paths, as shown in the following illustration:


The example shown in Figure 135 uses acc_next_output(). It accepts a handle to a scalar net or a net bitselect, and a module path. The application returns true if the net is connected to the output of the path.
```

int is_net_on_path output(net, path)
handle net; /* scalar net or bit-select of vector net */
handle path;
{
handle pterm_out, pterm_conn, bit;
/* scan path output terminals */
pterm_out = null;
while (pterm_out = acc_next_output(path, pterm_out) )
{
/* retrieve net connected to path terminal */
pterm_conn = acc_handle_conn (pterm_out);
if (acc_object_of_type (pterm_conn, accExpandedVector) )
{
bit = null;
while (bit = acc_next_bit (pterm_conn, bit) )
if (acc_compare_handles (bit, net) )
return (true);
}
else
if (acc_compare_handles (pterm_conn, net) )
return (true);
}
return (false);
}

```

Figure 135—Using acc_next_output()

\subsection*{23.76 acc_next_parameter()}


The ACC routine acc_next_parameter() shall return handles to the parameters in a scope. This handle can be passed to acc_fetch_paramtype() and acc_fetch_paramval() to retrieve the data type and value of the parameter. A scope is a module, task, function, or named block.

The example shown in Figure 136 uses acc_next_parameter() to scan for all parameters in a module.
```

nclude "acc_user.h"
I_INT32 print_parameter_values(module_handle)
ndle module_handle;
handle param_handle;
/*scan all parameters in the module and display values according to type
param_handle = null;
while (param_handle = acc_next_parameter(module_handle,param_handle) )
{
io_printf("Parameter %s = ",acc_fetch_fullname(param_handle) );
switch (acc_fetch_paramtype(param_han\overline{dle) )}
{
case accRealParam:
io_printf("%lf\n", acc_fetch_paramval(param_handle) );
break;
case accIntegerParam:
io_printf("%d\n", (PLI_INT32)acc_fetch_paramval(param_handle) );
break;
case accStringParam:
io_printf("%s\n",
(char*)(int)acc_fetch_paramval(param_handle) );
}
}

```

Figure 136—Using acc_next_parameter()

\subsection*{23.77 acc_next_port()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_next_port()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Gets handles to the ports of a module, or to ports which are connected to a given net or reg.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_next_port(reference, port_handle)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & handle & \multicolumn{2}{|l|}{Handle of a module port} \\
\hline \multirow{3}{*}{Arguments:} & Type & Name & Description \\
\hline & handle & reference_handle & Handle of a modul \\
\hline & handle & object_handle & Handle of the prev \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_fetch_direction() to determine the direction of a port Use acc_next_portout() to get handles to just output and inout ports} \\
\hline
\end{tabular}

The ACC routine acc_next_port() shall return handles to the input, output, and inout ports of a module. The handles shall be returned in the order specified by the port list in the module declaration, working from left to right.

The routine acc_next_port() shall be used two ways, as shown in Table 173.

Table 173—How acc_next_port() works
\begin{tabular}{|c|l|}
\hline If the reference handle is & \multicolumn{1}{c|}{ acc_next_port() shall return } \\
\hline A handle to a module & All ports of the module \\
\hline A handle to a net, reg or variable & \begin{tabular}{l} 
All ports connected to the net, reg or vari- \\
able within the scope of the net, reg or \\
variable
\end{tabular} \\
\hline
\end{tabular}

The example shown in Figure 137 uses acc_next_port() to find and display the input ports of a module.
```

include "acc_user.h"
'LI_INT32 display_inputs(module_handle)
.andle module_handle;
handle port_handle;
PLI_INT32 direction;
/*get handle for each module port*/
port_handle = null;
while}\mathrm{ (port_handle = acc_next_port(module_handle, port_handle) )
{
/*give the index of each input port*/
if (acc_fetch_direction(port_handle) == accInput)
io_printf("Port \#%d of %s is an input\n",
acc_fetch_index(port_handle),
acc_fetch_fullname(module_handle) );
}

```

Figure 137-Using acc_next_port() with a module handle
The example shown in Figure 138 uses acc_next_port() to find the port that is connected to a net, and then to display information about other nets connected to each bit of the same port.
```

'LI_INT32 display_port_connections()
handle net = acc_handle_tfarg(1);
handle port, bit;
port = bit = null;
while (port = acc_next_port(net, port) )
if (acc_object_of_type(port, accVectorPort) )
while (bit = acc_next_bit(port, bit) )
io_printf("PORTBIT: %s LOCONN: %s HICONN: %s/n",
acc_fetch_fullname(bit),
acc_fetch_fullname(acc_handle_loconn(bit) ),
acc_fetch_fullname(acc_handle_hiconn(bit) ) );

```

Figure 138—Using acc_next_port() with a net handle

\subsection*{23.78 acc_next_portout()}


The ACC routine acc_next_portout() shall return handles to the output and inout ports of a module. The handles shall be returned in the order specified by the port list in the module declaration, working from left to right.

The example shown in Figure 139 uses acc_next_portout() to find the output and inout ports of a module.
```

include "acc_user.h"
'LI_INT32 display_outputs(module_handle)
.an\overline{dle module_han\overline{dle;}}\mathbf{}/2
handle port_handle;
/* get handle for each module port* /
port_handle = null;
whil\overline{e (port_handle = acc_next_portout(module_handle, port_handle) )}
{
/*give the index of each output or inout port*/
io_printf("Port \#%d of %s is an output or inout\n",
acc_fetch_index(port_handle),
acc_fetch_fullname(mōdule_handle) );
}

```

Figure 139—Using acc_next_portout()

\subsection*{23.79 acc_next_primitive()}
\begin{tabular}{||l|l|l||}
\hline \multicolumn{4}{|c|}{ acc_next_primitive() } \\
\hline Synopsis: & Get handles to gates, switches, or user-defined primitives (UDPs) within a module. \\
\hline Syntax: & acc_next_primitive_(reference_handle, primitive_handle) \\
\hline \multirow{4}{*}{ Returns: } & Type & Description \\
\cline { 2 - 4 } Arguments: & handle & Handle of a primitive \\
\cline { 2 - 4 } & Type & \multicolumn{2}{c|}{ Name } & Description \\
\cline { 2 - 4 } & handle & reference_handle \\
handle & primitive_handle & Handle of the previous primitive found; initially null \\
\hline
\end{tabular}

The ACC routine acc_next_primitive() shall return handles to the built-in and user-defined primitives within a module.

The example shown in Figure 140 uses acc_next_primitive() to display the definition names of all primitives in a module.
```

include "acc_user.h"
'LI_INT32 get_primitive_definitions()
handle module_handle, prim_handle;
/*initialize environment for ACC routines*/
acc_initialize();
/*get handle for module* /
module_handle = acc_handle_tfarg(1);
io_printf("Module %s contains the following types of primitives:\n",
acc_fetch_fullname(module_handle) );
/*get and display defining names of all primitives in the module*/
prim_handle = null;
while\overline{e}}\mathrm{ (prim_handle = acc_next_primitive (module_handle,prim_handle) )
io_printf(" %s\n",
acc_fetch_defname(prim_handle) );
acc_close();

```

Figure 140—Using acc_next_primitive()

The example shown in Figure 141 uses acc_next_specparam() to scan for all specparams in a module.
```

aclude "acc_user.h"
I_INT32 print_specparam_values(module_handle)
.dle module_handle;
handle sparam_handle;
/*scan all parameters in the module and display values according to type
sparam_handle = null;
while (sparam_handle = acc_next_specparam(module_handle,sparam_handle)
{
io_printf("Specparam %s = ", acc_fetch_fullname(sparam_handle) );
switch (acc_fetch_paramtype(sparam_handle) )
{
case accRealParam:
io_printf("%lf\n", acc_fetch_paramval(sparam_handle) );
break;
case accIntegerParam:
io_printf("%d\n",
(int)acc_fetch_paramval(sparam_handle) ); break;
case accStringP
io_printf("%s\n",
(char*)(int)acc_fetch_paramval(sparam_handle));
}
}

```

Figure 141-Using acc_next_specparam()

\subsection*{23.82 acc_next_tchk()}


The ACC routine acc_next_tchk() shall return handles to the timing checks within a module. The handles can be passed to other ACC routines to get the nets or notifier in the timing check, and to read or modify timing check values.

The example shown in Figure 142 uses acc_next_tchk() to display information about setup timing checks.
```

nclude "acc_user.h"
.I_INT32 show_setup_check_nets()
handle mod_handle, cell_handle;
handle tchk_handle, tchk
PLI_INT32 tchk_type, counter;
/*initialize environment for ACC routines*/
acc_initialize();
/*get handle for module*/
mod_handle = acc_handle_tfarg(1);
/*scan all cells in module for timing checks*/
cell_handle = null;
while (cell_handle = acc_next_cell(mod_handle, cell_handle) )
{
io_printf("cell is: %s\n", acc_fetch_name(cell_handle) );
counter = 0;
tchk_handle = null;
while (tchk_handle = acc_next_tchk(cell_handle, tchk_handle) )
{
/*get nets connected to timing check arguments*/
tchk_type = acc_fetch_fulltype(tchk_handle);
if (tchk_type == accSetup)
{
counter++;
io_printf(" for setup check \#%d:\n",counter);
tchkarg1_handle = acc_handle_tchkarg1(tchk_handle);
tchkarg2_handle = acc_handle_tchkarg2(tchk_handle);
io_print\overline{f(" 1st`net is}%s\n 2\overline{nd net is %s\n",}
acc_fetch_name(acc_handle_conn(tchkarg1_handle) ),
acc_fetch_name(acc_handle_conn(tchkarg2_handle) ) );
}
}
}
acc_close();

```

Figure 142-Using acc_next_tchk()

\subsection*{23.83 acc_next_terminal()}
\begin{tabular}{||l|l|l||}
\hline \multicolumn{4}{|c|}{ acc_next_terminal() } \\
\hline Synopsis: & Get handles to terminals of a gate, switch, or user-defined primitive (UDP). \\
\hline Syntax: & acc_next_terminal (reference_handle, terminal_handle) \\
\hline \multirow{3}{*}{ Returns: } & Type & Description \\
\cline { 2 - 4 } Arguments: & handle & Handle of a primitive terminal \\
\cline { 2 - 4 } & Type & \multicolumn{2}{c|}{ Name } \\
\cline { 2 - 4 } & handle & reference_handle \\
handle & terminal_handle & Handle of a gate, switch or UDP \\
\hline
\end{tabular}

The ACC routine acc_next_terminal() shall return handles to the terminals on a primitive. The handles shall be returned in the order of the primitive instance statement, starting at terminal 0 (the leftmost terminal).

The example shown in Figure 143 uses acc_next_terminal() together with acc_handle_conn() to retrieve all nets connected to a primitive.
```

include "acc_user.h"
LI_INT32 display_terminals()
handle prim_handle,term_handle;
/*initialize environment for ACC routines*/
acc_initialize();
/*get handle for primitive*/
prim_handle = acc_handle_tfarg(1);
io_printf("Connections to primitive %s:\n",
acc_fetch_fullname(prim_handle) );
/*scan all terminals of the primitive
/* and display their nets*/
term handle = null;
while\overline{e}}\mathrm{ (term_handle = acc_next_terminal (prim_handle,term_handle) )
io_printf(" %s\n",
acc_fetch_name(acc_handle_conn(term_handle) ) );
acc_close();

```

Figure 143-Using acc_next_terminal()
23.84 acc_next_topmod()


The ACC routine acc_next_topmod() shall return handles to the top-level modules in a design.
The ACC routine acc_next_topmod() does not work with acc_collect() or acc_count(), but acc_next_child() with a null reference handle argument can be used in place of acc_next_topmod(). For example:
```

    acc_count(acc_next_child, null); /* counts top-level modules */
    acc_collect(acc_next_child, null, &count); /* collect top-level
    modules */

```

The example shown in Figure 144 uses acc_next_topmod() to display the names of all top-level modules.
```

include "acc_user.h"
'LI_INT32 show_top_modules()
handle module_handle;
/*initialize environment for ACC routines*/
acc_initialize();
/*scan all top-level modules*/
io_printf("The top-level modules are:\n");
module handle = null;
while (module_handle = acc_next_topmod (module_handle) )
/*display the instance name of each module*/
io_printf(" %s\n", acc_fetch_name(module_handle) );
acc_close();

```

Figure 144—Using acc_next_topmod()

\subsection*{23.85 acc_object_in_typelist()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_object_in_typelist()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Determine whether an object fits a type or fulltype, or special property, as specified in an input array.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_object_in_typelist(object_handle, object_type_array)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{true if the type, fulltype, or property of an object matches one specified in the array; false if there is no match} \\
\hline \multicolumn{2}{|r|}{Type} & Name & Description \\
\hline \multirow[t]{2}{*}{Arguments:} & handle & object_handle & Handle of an object \\
\hline & static integer array & object_type_array & Static integer array containing one or more predefined integer constants that represent the types and properties of objects desired; the last element shall be \(\mathbf{0}\) \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_object_of_type() to check for a match to a single predefined constant} \\
\hline
\end{tabular}

The ACC routine acc_object_in_typelist() shall determine whether an object fits one of a list of types, fulltypes, or special properties. The properties for which acc_object_in_typelist() is to check shall be listed as an array of constants in a static integer array. The array can contain any number and combination of the predefined integer constants, and it shall be terminated by a 0 .

The following C language statement shows an example of how to declare an array of object types called wired_nets:
```

static PLI_INT32
wired_nets[ 5] ={ accWand,accWor,accTriand,accTrior,0} ;

```

When this array is passed to acc_object_in_typelist(), the ACC routine shall return true if its object_handle argument is a wired net.

All type and fulltype constants shall be supported by acc_object_in_typelist(). These constants are listed in Table 113.

The special property constants supported by acc_object_in_typelist() are listed in Table 174.
The example shown in Figure 145 uses acc_object_in_typelist() to determine if a net is a wired net. The application then displays the name of each wired net found.
```

nclude "acc_user.h"
II_INT32 display_wired_nets()
static PLI_INT32 wired_nets[5]={accWand,accWor,accTriand,accTrior,0};
handle net_handle;
/*reset environment for ACC routines*/
acc_initialize();
/*get handle for net* /
net_handle = acc_handle_tfarg(1);
/*if a wired logic net, display its name*/
if (acc_object_in_typelist(net_handle,wired_nets) )
io_printf("Net-%s is a wire\overline{d net\n",acc_\overline{fetch_name(net_handle) );}}\mathbf{N}\mathrm{ ;}
else
io_printf("Net %s is not a wired net\n",acc_fetch_name(net_handle) )
acc_close();

```

Figure 145—Using acc_object_in_typelist()

\subsection*{23.86 acc_object_of_type()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_object_of_type()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Determine whether an object fits a specified type or fulltype, or special property.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_object_of_type(object_handle, object_type)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{true if the type, fulltype, or property of an object matches the object_type argument false if there is no match} \\
\hline \multirow{3}{*}{Arguments:} & & Type Name & Description \\
\hline & handle & object_handle & Handle of an object \\
\hline & PLI_INT32 & object_type & An integer constant that represents a type, fulltype, or special property \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_object_in_typelist() to check for a match to any of several predefined constants} \\
\hline
\end{tabular}

The ACC routine acc_object_of_type() shall determine whether an object fits a specified type, fulltype, or special property. The type, fulltype, or property is an integer constant, defined in acc_user.h. All type and fulltype constants shall be supported by acc_object_of_type(). These constants are listed in Table 113. The special property constants supported by acc_object_of_type() are listed in Table 174.

Table 174-Special object properties
\begin{tabular}{|c|c|}
\hline Property of object & Predefined integer constant \\
\hline Scalar & accScalar \\
\hline Vector & accVector \\
\hline Collapsed net & accCollapsedNet \\
\hline Expanded vector & accExpandedVector \\
\hline Unexpanded vector & accUnExpandedVector \\
\hline Hierarchy scope & accScope \\
\hline Module path with ifnone condition & accModPathHasIfnone \\
\hline
\end{tabular}

Simulated nets and collapsed nets are defined as follows. When a Verilog HDL source description connects modules together, a chain of nets with different scopes and names are connected, as is illustrated in the following simple diagram:


In this small circuit, nets out1, w5, and in1 are all tied together, effectively becoming the same net. Software products can collapse nets that are connected together within the data structure of the product. The resultant net after collapsing is referred to as a simulated net; the other nets are referred to as collapsed nets. The ACC routines can obtain a handle to any net, whether it is collapsed or not. The routine acc_object_of_type() can be used to determine if a net has been collapsed. The routine acc_handle_simulated_net() can be used to find the resultant net from the net collapsing process.

Expanded and unexpanded vectors determine if ACC routines can access a vector as a whole or access the bits within a vector. If a vector has the property accExpandedVector, then access to the discrete bits of the vector shall be permitted. This property has to be true in order for certain ACC routines, such as acc_next_bit(), to access each bit of a vector. If a vector has the property accUnExpandedVector, then access to the vector as a whole shall be permitted. This property has to be true in order for certain ACC routines to access the complete vector. A vector object can have just one of these properties true, or both can be true.
acc_object_of_type() with an accScope type constant will return true if the reference object is a Verilog scope. A scope is a module, task, function or named block.
acc_object_of_type() with an accModPathHasIfnone type constant will return true if the reference object is a Verilog module path, and there is an ifnone condition specified for the path.

The example shown in Figure 146 uses acc_object_of_type() to determine whether nets are collapsed nets. The application then displays each collapsed net, along with the simulated net.
```

include "acc_user.h"
'LI_INT32 display_collapsed_nets()
handle mod_handle;
handle net_handle;
handle simulated_net_handle;
/*reset environment for ACC routines*/
acc_initialize();
/*get scope-first argument passed to user-defined system task*/
/* associated with this routine*/
mod_handle = acc_handle_tfarg(1);
io_printf("In mo\overline{dule %s\overline{S\n",acc_fetch_fullname(mod_handle) );}}\mathbf{|}=\mp@code{l}
net_handle = null;
/*display name of each collapsed net and its net of origin*/
while (net_handle = acc_next_net(mod_handle,net_handle) )
{
if (acc_object_of_type(net_handle,accCollapsedNet) )
{
simulated_net_handle = acc_handle_simulated_net(net_handle);
io_printf(" net %s was collapsed onto net %s\n",
acc_fetch_name(net_handle),
acc_fetch_name(simulated_net_handle) );
}
}

```

Figure 146-Using acc_object_of_type()

\subsection*{23.87 acc_product_type()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_product_type ()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the software product type that is calling the PLI application.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_product_type()} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{A predefined integer constant representing the software product type} \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Description \\
\hline & None & & \\
\hline
\end{tabular}

The ACC routine acc_product_type() shall return a predefined integer constant that identifies the class of software product that is calling the PLI application. This information can be useful when a PLI application needs to customize the routine to specific types of software implementations. For example, a delay calculator might use typical delays for logic simulation and min:typ:max delays for timing analysis.

The integer constant values returned by acc_product_type() are listed in Table 175.

Table 175—Product types returned by acc_product_type()
\begin{tabular}{|c|c|}
\hline If the product is & acc_product_type() returns \\
\hline A logic simulator & accSimulator \\
\hline A timing analyzer & accTimingAnalyzer \\
\hline A fault simulator & accFaultSimulator \\
\hline Some other product & accOther \\
\hline
\end{tabular}

NOTE Software product vendors can define additional integer constants specific to their products.

The example shown in Figure 147 uses acc_product_type() to identify and display the product type being used.
```

nclude "acc_user.h"
I_INT32 show_application()

* reset environment for ACC routines */
cc_initialize();
* show application type and ACC routine version */
witch (acc_product_type() )
case accSimulator:
io_printf("Running logic simulation with PLI version %s\n",acc_version()
break;
case accTimingAnalyzer:
io_printf("Running timing analysis with PLI version %s\n",acc_version()
break;
case accFaultSimulator:
io_printf("Running fault simulation with PLI version %s\n",acc_version()
break;
default:
io_printf("Running other product with PLI version %s\n",acc_version());
}
cc_close();

```

Figure 147-Using acc_product_type()

\subsection*{23.88 acc_product_version()}


The ACC routine acc_product_version() shall return a pointer to a character string that indicates the version of the software product that called the PLI application. The return value for this routine is placed in the ACC internal string buffer. See 22.9 for explanation of strings in ACC routines.

The character string shall be in the following format:
```

<product_name> Version <version_number>

```

For example:
```

"Verilog Simulator Version OVIsim 1.0"

```

The string returned by acc_product_version() shall be defined by the software tool vendor.
The example shown in Figure 148 uses acc_product_version() to identify the version of the software product that is linked to ACC routines.
```

include "acc_user.h"
UI_INT32 show_versions()
/*initialize environment for ACC routines*/
acc_initialize();
/*show version of ACC routines*/
/* and version of Verilog that is linked to ACC routines*/
io_printf("Running %s with %s\n",acc_version(),acc_product_version() );
acc_close();

```

Figure 148—Using acc_product_version()

\subsection*{23.89 acc_release_object()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_release_object()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Deallocate memory allocated by calls to acc_next_input() and acc_next_output().} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_release_object (object_handle)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{0 if successful; 1 if an error is encountered} \\
\hline \multirow[b]{2}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & handle & object_handle & Handle to an input or output tern \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_next_input() to get handles to module path inputs and data path inputs Use acc_next_output() to get handles to module path outputs and data path outputs} \\
\hline
\end{tabular}

The ACC routine acc_release_object() shall deallocate memory that was allocated by a call to acc_next_input() or acc_next_output(). The routine should be called after using these ACC routines under the following circumstances:

Not all inputs or outputs were scanned.
The input or output path had only one terminal.
An error was returned.

The example shown in Figure 149 finds the data path corresponding to an input module path, and it displays the source and destination port names for the data path. The example calls acc_next_input() and acc_next_output() to get the first input and output, respectively, for a given path. Since these routines are only called once, acc_release_object() is called to free the memory allocated for the input and output handles.
```

PLI_INT32 display_datapath_terms(modpath)
.andle modpath;
handle datapath = acc_handle_datapath(modpath);
handle pathin = acc_next_input(datapath, null);
handle pathout = acc_next_output(datapath, null);
/* there is only one input and output to a data path */
io_printf("DATAPATH INPUT: %s\n", acc_fetch_fullname(pathin) );
io_printf("DATAPATH OUTPUT: %s\n", acc_fetch_fullname(pathout) );
acc_release_object(pathin);
acc_release_object(pathout) ;

```

Figure 149—Using acc_release_object()

\subsection*{23.90 acc_replace_delays()}

\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|r|}{acc_replace_delays() for min:typ:max delays (accMinTypMaxDelays set to true )} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Replace min:typ:max delay values for primitives, module paths, timing checks, module input ports, or intermodule paths; the delay values are contained in an array.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_append_delays (object_handle, array_ptr)} \\
\hline \multicolumn{3}{|c|}{Type} & Description \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{1 if successful; 0 if an error is encountered} \\
\hline \multicolumn{2}{|r|}{Type} & Name & Description \\
\hline \multirow[t]{2}{*}{Arguments:} & handle & object_handle & Handle of a primitive, module path, timing check, module input port, bit of a module input port, or intermodule path \\
\hline & double address & array_ptr & Pointer to array of min:typ:max delay values; the size of the array depends on the type of object and the setting of accPathDelayCount (see Section 22.8) \\
\hline
\end{tabular}

The ACC routine acc_replace_delays() shall work differently depending on how the configuration parameter accMinTypMaxDelays is set. When this parameter is set to false, a single delay per transition shall be assumed, and delays shall be passed as individual arguments. For this single delay mode, the first syntax table in this section shall apply.

When accMinTypMaxDelays is set to true, acc_replace_delays() shall pass one or more sets of minimum:typical:maximum delays contained in an array, rather than single delays passed as individual arguments. For this min:typ:max delay mode, the second syntax table in this section shall apply.

The number of delay values replaced by acc_replace_delays() shall be determined by the type of object and the setting of configuration parameters. Refer to 22.8 for a description of how the number of delay values are determined.

The routine acc_replace_delays() shall write delays in the timescale of the module that contains the object_handle.

When altering the delay via acc_replace_delays(), the value of the reject/error region will not be affected unless the limits exceed the value of the delay. If the reject/error limits exceed the delay they will be truncated down to the new delay limit.

The example shown in Figure 150 uses acc_replace_delays() to replace the current delays on a path with new delay values read from a file called pathdelay.dat. The format of the file is shown in the following diagram:

```

nclude <stdio.h>
nclude "acc_user.h"
efine NAME SIZE 256
I_INT32 write_path_delays()
FILE *infile;
PLI_BYTE8 full_module_name[ NAME_SIZE];
PLI_BYTE8 pathin_name[ NAME_SIZE], pathout_name[ NAME_SIZE];
double rise, fall;
handle mod_handle, path_handle;
/*initialize the environment for ACC routines*/
acc_initialize();
/* set accPathDelayCount parameter to return rise and fall delays only*/
acc_configure(accPathDelayCount, "2");
/*read delays from file - "r" means read only*/
infile = fopen("pathdelay.dat","r");
fscanf(infile, "%s %s %s %lf %lf",
full_module_name,pathin_name,pathout_name,\&rise,\&fall);
/*get handle for the module and the path*/
mod_handle = acc_handle_object(full_module_name);
path_handle = acc_handle_modpath(mod_handle,pathin_name,pathout_name);
/*replace delays with new values*/
acc_replace_delays(path_handle, rise, fall);
acc_close();

```

Figure 150-Using acc_replace_delays() in single delay mode
The example shown in Figure 151 uses acc_replace_delays() to scale the min:typ:max delays on all primitive delays inside cells within a given scope. The application fetches the existing delays for an object, multiplies the delays by a scale factor, and replaces the delays with the new, scaled values. This example assumes that the user application is associated through the PLI interface mechanism with a user-defined system task called \$scaleprimdelays. The scope and scale factors are passed as arguments as follows:

```

include "acc_user.h"
include "verīuser.h"
_I_INT32 scale_prim_delays()
array has to hold three sets
*) of min:typ:max values for
randle top, cell, prim; 年 rise, fall, and turn-off delays
int i;
louble da[ 9] ;
louble min_scale_factor, typ_scale_factor, max_scale_factor;
彐Cc initialize();
\existscc_configure(accMinTypMaxDelays,"true");
zop = acc_handle_tfarg(1); «_ argument \#1: Scope
nin_scale_factor = acc_fetch_tfarg(2); «_ argument \#2: Scale factor for minimum delay
-yp_scale_factor = acc_fetch_tfarg(3); «_ argument \#3: Scale factor for typical delay
nax_scale_factor = acc_fetch_tfarg(4); «_ argument \#4: Scale factor for maximum delay
Lo_printf("Scale min:typ:max delays for primitives in cells below %s\n",
acc fetch fullname(top) ) ;
Lo_printf("Scaling fāctors-min:typ:max-%4.2f:%4.2f:%4.2f\n",
min_scale_factor, typ_scale_factor, max_scale_factor);
vell = null;
vhile (cell = acc_next_cell(top, cell) )
{
prim = null;
while (prim = acc_next_primitive(cell, prim) )
{
acc_fetch_delays(prim,da);
for (i=0; i<9; i+=3)
da[i] = da[ i]*min_scale_factor; }\quad\mathrm{ da[6] typical
for (i=1; i<9; i+=3)
da[ i] = da[i]*typ_scale_factor;
for (i=2; i<9; i+=3)
da[ i] = da[ i]*max_scale_factor;
acc_replace_delays(prim,da);
}
}
\existscc_close();

```

Figure 151-Using acc_replace_delays() in min:typ:max delays mode

\subsection*{23.91 acc_replace_pulsere()}


The ACC routine acc_replace_pulsere() shall replace existing pulse handling reject_limit and e_limit values for a module path, intermodule path and module input port. The reject_limit and e_limit values are used to control how pulses are propagated through paths.

A pulse is defined as two transitions that occur in a shorter period of time than the delay. Pulse control values determine whether a pulse should be rejected, propagated through to the output, or considered an error. The pulse control values consist of a reject_limit and an e_limit pair of values, where:

The reject_limit shall set a threshold for determining when to reject a pulse any pulse less than the reject_limit shall not propagate to the output
The e_limit shall set a threshold for determining when a pulse is considered to be an error any pulse less than the e_limit and greater than or equal to the reject_limit shall propagate a logic x A pulse that is greater than or equal to the e_limit shall propagate
The example in Table 176 illustrates the relationship between the reject_limit and the e_limit.
Table 176-Pulse control example
\begin{tabular}{|l|l|}
\hline \multicolumn{1}{|c|}{ When } & \multicolumn{1}{c|}{ The pulse shall be } \\
\hline \begin{tabular}{l} 
reject_limit \(=10.5\) \\
e_limit \(=22.6\)
\end{tabular} & Rejected if \(<10.5\) \\
& An error if \(>=10.5\) and \(<22.6\) \\
& Passed if \(>=22.6\) \\
\hline
\end{tabular}

The following rules shall apply when specifying pulse handling values:
a) The value of reject_limit shall be less than or equal to the value of e_limit.
b) The reject_limit and e_limit shall not be greater than the delay.

If any of the limits do not meet the above rules, they will be truncated.

The number of pulse control values that acc_replace_pulsere() sets shall be controlled using the ACC routine acc_configure() to set the delay count configuration parameter accPathDelayCount, as shown in Table 177.

The minimum number of pairs of reject_limit and e_limit arguments to pass to acc_replace_pulsere() shall equal the value of accPathDelayCount. Any unused reject_limit and e_limit argument pairs shall be ignored by acc_replace_pulsere() and can be dropped from the argument list.

If accPathDelayCount is not set explicitly, it shall default to 6

Pulse control values shall be replaced using the timescale of the module that contains the object handle.
The example shown in Figure 152 uses acc_replace_pulsere() to replace rise and fall pulse handling values of paths listed in a file path. dat.
```

nclude <stdio.h>
nclude "acc user.h"
efine NAME_SIZE 256
I_INT32 replace_halfpulsevals()
'ILE *infile;
LI_BYTE8 mod_name[ NAME_SIZE] ;
LI_BYTE8 pathin_name[ NAME_SIZE], pathout_name[ NAME_SIZE] ;
andle mod, path;
ouble rise_reject_limit=0.0, rise_e_limit=0.0,
fall_reject_limit=0.0, fall__e_limit=0.0;
/*initialize environment for ACC routines*/
acc_initialize();
/*set accPathDelayCount to return two pairs of pulse handling values;*/
/* one each for rise and fall transitions*/
acc_configure(accPathDelayCount, "2");
/*read all module path specifications from file "path.dat"*/
infile = fopen("path.dat", "r");
while(fscanf(infile,"%s %s %s",mod_name,pathin_name,pathout_name)!=EOF)
{
mod=acc_handle_object(mod_name);
path=acc_handle_modpath(mod,pathin_name,pathout_name);
rise reject limít = .05;
if(a\overline{cc_replāce_pulsere(path, \&rise_reject_limit, \&rise_e_limit,}
\&fall_reject_limit, \&fall_e_limit) )
{
io_printf("rise reject limit = %lf, rise e limit = %lf\n",
rise_reject_limit, rise_e_limit);
io_printf("fall reject limit = %lf, fall e limit = %lf\n",
fall_reject_limit, fall_e_limit);
}
}
acc_close();

```

Figure 152-Using acc_replace_pulsere()

\subsection*{23.92 acc_reset_buffer()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_reset_buffer ()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Reset the string buffer to the beginning.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_reset_buffer()} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & \multicolumn{2}{|l|}{void} & \\
\hline \multirow[b]{2}{*}{Arguments} & Type & Name & Description \\
\hline & None & & \\
\hline Related routines: & \multicolumn{3}{|l|}{All ACC routines that return a pointer to a character string} \\
\hline
\end{tabular}

The ACC routine acc_reset_buffer() shall reset the string buffer to its beginning. The string buffer shall be used as temporary storage by other ACC routines that return a pointer to a character string. Refer to 22.9 for more information on the character string buffer.

\subsection*{23.93 acc_set_interactive_scope()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_set_interactive_scope ()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Set the interactive scope of a software tool.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_set_interactive_scope(scope, callback_flag)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & handle & \multicolumn{2}{|l|}{Handle of a Verilog hierarchy scope} \\
\hline \multirow{3}{*}{Arguments:} & Type & Name & Description \\
\hline & handle & scope & Handle to the scope which will be the new interactive scope \\
\hline & PLI_INT32 & callback_flag & If set to TRUE, then the misctf routines shall be called with reason reason_scope immediately. If set to FALSE, then the misctf routines are not called \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_handle_interactive_scope() to get a handle for the current interactive scope} \\
\hline
\end{tabular}

The ACC routine acc_set_interactive_scope() shall set the Verilog HDL design scope where the interactive mode of the software product is operating.

A scope shall be
A top-level module
A module instance
A named begin-end block
A named fork-join block
A Verilog HDL task
A Verilog HDL function

The following rules shall apply when specifying pulse handling values:
a) The reject_percentage and e_percentage shall be greater than or equal to 0.0 and less than or equal to 1.0.
b) The value of reject_percentage shall be less than or equal to the value of e_percentage.

The example shown in Figure 153 uses acc_set_pulsere() to set pulse control values for each path in a module such that all pulses between 0 and the path delay generate an \(X\) at the path output.
```

include "acc_user.h"
LI_INT32 set_pulse_control_e(module)
andle module;
handle path;
/*set pulse control values for all paths in the module*/
path = null;
while (path = acc_next_modpath(module, path) )
acc_set_pulsere(path, 0.0, 1.0);

```

Figure 153-Using acc_set_pulsere()

\subsection*{23.95 acc_set_scope()}


The ACC routine acc_set_scope() shall set the scope and search rules for the routine acc_handle_object(). The way that acc_set_scope() functions shall be dependent on the setting of configuration parameters as shown in Table 179.

\section*{Table 179-How acc_set_scope() works}
\begin{tabular}{|c|c|}
\hline If & acc_set_scope() shall \\
\hline Default mode, or acc_configure(accEnableArgs, no_acc_set_scope ) is called, and module_handle is a valid handle & Set the scope to the level of module_handle in the design hierarchy and ignore the optional module_name argument \\
\hline ```
Default mode, or
    acc_configure(accEnableArgs, no_acc_set_scope )
is called, and
    module_handle is null
``` & Set the scope to the top-level module that appears first in the source description \\
\hline The routine acc_configure(accEnableArgs, acc_set_scope ) has been called, and module_handle is a null & Set scope to the level of module_name in the design hierarchy \\
\hline The routine acc_configure(accEnableArgs, acc_set_scope ) has been called, and module_handle is a valid handle & Set scope to the level of module_handle in the design hierarchy and ignore the optional module_name argument \\
\hline The routine acc_configure(accEnableArgs, acc_set_scope ) has been called, and module_handle and module_name are both null & Set scope to the top-level module that appears first in the source description \\
\hline
\end{tabular}

To use the optional module_name argument, the configuration parameter accEnableArgs first has to be set by calling acc_configure() as follows:
acc_configure(accEnableArgs, "acc_set_scope");

If accEnableArgs is not set for acc_set_scope(), the routine shall ignore its optional argument. When the optional argument is not required for a call to acc_set_scope(), the argument can be dropped.

The example shown in Figure 154 uses acc_set_scope() to set a scope for the ACC routine acc_handle_object() to determine if a net is in a module.
```

include "acc_user.h"
'LI_INT32 is_net_in_module(module_handle,net_name)
.an\overline{dle module_handle;}
'LI_BYTE8 *net_name;
handle net_handle;
/* set scope to module* /
acc_set_scope(module_handle);
/*get handle for net*/
net_handle = acc_handle_object(net_name);
if (net_handle)
io_printf("Net %s found in module %s\n",
net_name,
acc_fetch_fullname(module_handle) );
else
io_printf("Net %s not found in module %s\n",
net_name,
acc_fetch_fullname(module_handle) );

```

Figure 154-Using acc_set_scope()

\subsection*{23.96 acc_set_value()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_set_value()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Set and propagate a value on a reg, variable, user-defined system function or a sequential UDP; procedurally assign a reg or variable; force a reg, variable, or net.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_set_value (object_handle, value_p, delay_p)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_INT32 & PLI_IN132 Zero if no errors; nonzero if an error occurred & an error occurred \\
\hline \multirow{4}{*}{Arguments:} & Type & Name & Description \\
\hline & handle & object_handle & Handle to a reg, variable, net, user-defined system function, or sequential UDP \\
\hline & p_setval_value & value_p & Pointer to a structure containing value to be set \\
\hline & p_setval_delay & delay_p & Pointer to a structure containing delay before value is set \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_fetch_value() to retrieve a logic value Use acc_fetch_size() to get the number of bits in a vector} \\
\hline
\end{tabular}

The ACC routine acc_set_value() shall set and propagate a value onto a reg, integer variable, time variable, real variable, or a sequential UDP. The routine shall also perform procedural assign/deassign or procedural force/release functions.

The acc_set_value() routine shall also return the value of a system function by passing a handle to the userdefined system function as the object handle. This should only occur during execution of the calltf routine for the system function. Attempts to use acc_set_value() with a handle to the system function when the calltf routine is not active shall be ignored. Should the calltf routine for a user defined system function fail to put a value during its execution, the default value of 0 shall be applied.

The logic value and propagation delay information shall be placed in separate structures. To use acc_set_value() to propagate a value, follow these basic steps:
a) Allocate memory for the structures s_setval_value, s_setval_delay, and if using vectors, s_acc_vecval.
b) Set the appropriate fields in each structure to the desired values.
c) Call acc_set_value() with an object handle and pointers to the s_setval_value and s_setval_delay structures.

The structure s_setval_value shall contain the value to be written. A value can be entered into this structure as a string, scalar, integer, real, or as an aval/bval pair. The s_setval_value structure is defined in acc_user.h and listed in Figure 155 (note that this structure is also used with the acc_fetch_value() routine).

The format field in the s_setval_value structure shall indicate the value type. The format shall be a predefined integer constant, listed in Table 180.

The value union in the s_setval_value structure shall be the value to be written. The value is placed in the appropriate field within the union for the format selected.
```

typedef struct t_setval_value
{
PLI_INT32 format;
union
{
PLI_BYTE8 *str;
PLI_INT32 scalar;
PLI_INT32 integer;
dou\overline{ble real;}
p_acc_vecval vector;
} value;
} s_setval_value, *p_setval_value, s_acc_value, *p_acc_value;

```

Figure 155 -The s_setval_value structure used by acc_set_value()

Table 180-Predefined constants for the format field of s_setval_value
\begin{tabular}{|c|l|}
\hline Value format & \multicolumn{1}{|c|}{ Definition } \\
\hline accScalarVal & One of: acc0, acc1, accZ, accX \\
\hline accVectorVal & aval and bval bit groups, with each group being an integer quantity \\
\hline accIntVal & An integer quantity \\
\hline
\end{tabular}

Table 180—Predefined constants for the format field of s_setval_value (continued)
\begin{tabular}{|c|l|}
\hline Value format & \multicolumn{1}{c|}{ Definition } \\
\hline accRealVal & A real-valued quantity \\
\hline accStringVal & \begin{tabular}{l} 
For integers and appropriately sized regs, any ASCII string; \\
for real-valued objects, any string that represents a real number
\end{tabular} \\
\hline accBinStrVal & A base 2 representation as a string \\
\hline accOctStrVal & A base 8 representation as a string \\
\hline accDecStrVal & A base 10 representation as a string \\
\hline accHexStrVal & A base 16 representation as a string \\
\hline
\end{tabular}

When the format field of the s_acc_vecval structure is set to accVectorVal, the value union field used shall be vector. The vector field is set to a pointer or an array of s_acc_vecval structures that contain aval/ bval pairs for each bit of the vector. The s_acc_vecval structure is listed in Figure 156.
```

typedef struct t_acc_vecval
{
PLI_INT32 aval;
PLI_INT32 bval;
} s_acc_vecval, *p_acc_vecval;

```

Figure 156-s_acc_vecval structure
The array of s_acc_vecval structures shall contain a record for every 32 bits of the vector, plus a record for any remaining bits. Memory has to be allocated by the user for the array of s_acc_vecval structures. If a vector has N bits, the size of the array shall be \(((\mathbb{N}-1) / 32)+1\) s_acc_vecval records. The routine acc_fetch_size() can be used to determine the value of N .

The lsb of the vector shall be represented by the lsb of the first record of s_acc_vecval array. The 33rd bit of the vector shall be represented by the lsb of the second record of the array, and so on. Each bit of the vector shall be encoded as an aval/bval pair. The encoding for each bit is shown in Table 181.

Table 181-Encoding of bits in the s_acc_vecval structure
\begin{tabular}{|c|c|c|}
\hline aval & bval & Value \\
\hline 0 & 0 & 0 \\
\hline 1 & 0 & 1 \\
\hline 0 & 1 & Z \\
\hline 1 & 1 & X \\
\hline
\end{tabular}

The structure s_setval_delay shall control how values are to be propagated into the Verilog HDL data structure. The structure is defined in acc_user. h and is listed in Figure 157.

The time field in the s setval delay structure shall indicate the delay that shall take place before a reg value assignment. The time field shall be of type s_acc_time structure, as shown in Figure 158.

The model field in the s_setval_delay structure shall determine how the delay shall be applied, and how other simulation events scheduled for the same object shall be affected. The delay model shall be specified using predefined integer constants, listed in Table 182 and Table 184.
```

typedef struct t_setval_delay
{
s acc time time;
PLII_INTT32 model;
} s_setval_delay, *p_setval_delay;

```

Figure 157-The s_setval_delay structure for acc_set_value()

Table 182—Predefined delay constants for the model field of s_setval_delay
\begin{tabular}{|c|c|l|}
\hline Integer constant & Delay model & \multicolumn{1}{c|}{ Description } \\
\hline accNoDelay & No delay & \begin{tabular}{l} 
Sets a reg, variable or sequential UDP to the \\
indicated value with no delay; other events \\
scheduled for the object are not affected
\end{tabular} \\
\hline accInertialDelay & Inertial delay & \begin{tabular}{l} 
Sets a reg or variable to the indicated value \\
after the specified delay; all scheduled events \\
on the object are removed before this event is \\
scheduled
\end{tabular} \\
\hline accTransportDelay & Modified transport delay & \begin{tabular}{l} 
Sets a reg or variable to the indicated value \\
after the specified delay; all scheduled events \\
on the object for times later than this event are \\
removed
\end{tabular} \\
\hline accPureTransportDelay & Pure transport delay & \begin{tabular}{l} 
Sets a reg or variable to the indicated value \\
after the specified delay; no scheduled events \\
on the object are removed
\end{tabular} \\
\hline
\end{tabular}

When setting the value of a sequential UDP, the model field shall be accNoDelay, and the new value shall be assigned with no delay even if the UDP instance has a delay.

The s_acc_time structure shall hold the delay value that shall be used by acc_set_value(). The s_acc_time structure is defined in acc_user. h and is listed in Figure 158.

The type field in the s_acc_time structure shall indicate the data type of the delay that shall be stored in the structure. The type shall be specified using predefined integer constants, listed in Table 183.

The low field shall be an integer that represents the lower 32 bits of a 64-bit delay value.
The high field shall be an integer that represents the upper 32 bits of a 64-bit delay value.
The real field shall be a double that represents the delay as a real number value.
```

typedef struct t_acc_time
{
PLI INT32 type;
PLI_INT32 low,
high;
double real;
} s_acc_time, *p_acc_time;

```

Figure 158-The s_acc_time structure for acc_set_value()

Table 183-Predefined time constants for the type field of s_acc_time
\begin{tabular}{|c|l|}
\hline Integer constant & \multicolumn{1}{c|}{ Description } \\
\hline accTime & \begin{tabular}{l} 
Delay is a 64-bit integer; time shall be scaled to the timescale in effect \\
for the module containing the object.
\end{tabular} \\
\hline accSimTime & \begin{tabular}{l} 
Delay is a 64-bit integer; time shall be scaled to the time units being \\
used by the simulator
\end{tabular} \\
\hline accRealTime & \begin{tabular}{l} 
Delay is a real number; time shall be scaled to the timescale in effect for \\
the module containing the object.
\end{tabular} \\
\hline
\end{tabular}

The routine acc_set_value() shall be used to perform a procedural continuous assignment of a value to a reg or variable or to deassign the reg or variable. This shall be the same functionality as the procedural assign and deassign keywords in the Verilog HDL.

The routine acc_set_value() shall also be used to perform a procedural force of a value onto a reg, variable or net, or to release the reg, variable or net. This shall be the same functionality as the procedural force and release keywords in the Verilog HDL.

When an object is deassigned or released using acc_set_value(), the current value of the object shall be returned to the s_setval_value structure.

To assign, deassign, force, or release an object using acc_set_value(), the s_setval_value and s_setval_delay structures shall be allocated and the fields shall be set to the appropriate values. For the model field of the s_setval_delay structure, one of the predefined constants listed in Table 184 shall be used.

Table 184—Predefined assign/force constants for the model field of s_setval_delay
\begin{tabular}{|c|l|}
\hline Integer constant & \multicolumn{1}{c|}{ Description } \\
\hline accAssignFlag & \begin{tabular}{l} 
Assigns a reg or variable to the indicated value with no delay; other \\
events scheduled for the object are overridden. \\
Same functionality as the Verilog HDL procedural assign keyword.
\end{tabular} \\
\hline accDeassignFlag & \begin{tabular}{l} 
Deassigns an assigned reg or variable; other events scheduled for the \\
object are no longer overridden. Same functionality as the Verilog HDL \\
procedural deassign keyword.
\end{tabular} \\
\hline accForceFlag & \begin{tabular}{l} 
Forces a value onto a reg, variable or net; other events scheduled for the \\
object are overridden. Same functionality as the Verilog HDL proce- \\
dural force keyword.
\end{tabular} \\
\hline accReleaseFlag & \begin{tabular}{l} 
Releases a forced reg, variable or net; other events scheduled for the \\
object are no longer overridden, and nets immediately return to the cur- \\
rent driven value. Same functionality as the Verilog HDL procedural \\
release keyword.
\end{tabular} \\
\hline
\end{tabular}

The example shown in Figure 159 uses acc_set_value() to set and propagate a value onto a reg. This example assumes the application is linked to a user-defined system task (using the PLI interface mechanism) called \$my_set_value (), which has the following usage for a four bit reg, \(r 1\) :
\$my_set_value(r1, "x011", 2.4);
```

?LI_INT32 my_set_value()
static s_setval_delay delay_s = {{accRealTime},accInertialDelay};
static s_setval_value value_s = { accBinStrVal} ;
handle reg = acc_handle_tfarg(1);
value_s.value.str = acc_fetch_tfarg_str(2);
delay s.time.real= acc fetch tfarg(3);
acc_set_value(reg, \&value_s, \&delay_s);

```

Figure 159-Using acc_set_value()

\subsection*{23.97 acc_vcl_add()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_vcl_add()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Set a callback to a consumer routine with value change information whenever an object changes value.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_vcl_add (object_handle, consumer_routine, user_data, vcl_flag)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & void & & \\
\hline \multirow{5}{*}{Arguments:} & Type & Name & Description \\
\hline & handle & object_handle & Handle to an object to be monitored (such as a reg or net) \\
\hline & C routine pointer & consumer_routine & Unquoted name of the C routine to be called when the object changes value \\
\hline & PLI_BYTE8 * & user_data & User-defined data that is passed back to the consumer routine when the object changes value \\
\hline & PLI_INT32 & vcl_flag & Predefined integer constant that selects the type of change information reported to the consumer routine \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_vcl_delete() to remove a VCL callback monitor} \\
\hline
\end{tabular}

The ACC routine acc_vcl_add() shall set up a callback monitor on an object that shall call a user-defined consumer routine when the object changes value. The consumer routine shall be passed logic value information or logic value and strength information about the object.

The acc_vcl_add() routine requires four arguments, as described in the following paragraphs.
The object_handle argument is a handle to the object to be monitored by an application. The VCL shall monitor value changes for the following objects:

Scalar regs and bit-selects of vector regs
Scalar nets, unexpanded vector nets, and bit-selects of expanded vector nets
Integer, real and time variables
Module ports
Primitive output or inout terminals
Named events

NOTE Adding a value change link to a module port is equivalent to adding a value change link to the loconn of the port. The vc_reason returned shall be based on the loconn of the port.

The object_handle passed to acc_vcl_add() is not returned when the consumer routine is called. However, the handle can be passed using the user_data argument.

The consumer_routine argument is a pointer to a C application. This application shall be called whenever the object changes value. When a value change callback occurs, the consumer_routine shall be passed the user_data argument and a pointer to a vc_record structure, which shall contain information about the change.

Refer to 22.10 for a full description of consumer routines and the vc_record structure.
The user_data argument is user-defined data, such as the object name, the object handle, the object value, or a pointer to a data structure. The value of the user_data argument shall be passed to the consumer routine each time a callback occurs. Note that the user_data argument is defined as character string pointer, and therefore any other type should be cast to a PLI_BYTE8*.

The \(v c l \_f l a g\) argument shall set the type of information the callback mechanism shall report. There are two types of flags, as shown in Table 185.

Table 185-vcl_flag constants used in acc_vcl_add()
\begin{tabular}{|c|l|}
\hline vcl_flag & \multicolumn{1}{c|}{ What it does } \\
\hline vcl_verilog_logic & \begin{tabular}{l} 
Indicates the VCL callback mechanism shall report \\
information on logic value changes
\end{tabular} \\
\hline vcl_verilog_strength & \begin{tabular}{l} 
Indicates the VCL callback mechanism shall report \\
information on logic value and strength changes
\end{tabular} \\
\hline
\end{tabular}

If an application calls acc_vel_add() with the same arguments more than once, the VCL callback mechanism shall only call the consumer routine once when the object changes value. If any of the VCL arguments, including the user_data, are different, the VCL callback mechanism shall call the consumer routine multiple times, once for each unique acc_vcl_add().

NOTE It is not recommended that multiple VCL flags be added with the same object, consumer and user_data. If multiple flags with the same values are added, then each call to acc_vcl_delete() with those values shall delete one flag; the order of deletion is indeterminate.

If multiple PLI applications monitor the same object at the same time, each application shall receive a separate call whenever that object changes value. Typically, multiple applications have distinct consumer routines and user_data pointers. These different consumer routines allow the value change information to be processed in different ways.

Refer to 22.10 for an example of using acc_vcl_add().

\subsection*{23.98 acc_vcl_delete()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_vcl_delete()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Removes a VCL callback monitor.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_vcl_delete(object_handle, consumer_routine, user_data, vcl_flag)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & void & & \\
\hline \multirow{5}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & handle & object_handle & Handle to the object to be monitored specified in the call to acc_vcl_add() \\
\hline & C routine pointer & consumer_routine & Unquoted name of the C routine specified in the call to acc_vcl_add() \\
\hline & PLI_BYTE8 * & user_data & User-defined data specified in the call to acc_vcl_add() \\
\hline & PLI_INT32 & vcl_flag & Predefined integer constant; vcl_verilog \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_vcl_add() to place a VCL callback monitor on an object} \\
\hline
\end{tabular}

The ACC routine acc_vcl_delete() shall remove a VCL callback monitor previously requested with a call to acc_vcl_add(). The acc_vcl_delete() routine requires four arguments, as described in the following paragraphs. When multiple PLI applications are monitoring the same object, acc_vcl_delete() shall stop monitoring the object only for the application associated with a specific acc_vel_add() call.

The object_handle argument is a handle to the object for which the VCL callback monitor is to be removed. This has to be a handle to the same object that was used when acc_vcl_add() was called.

The consumer_routine argument is the unquoted name of the C application called by the VCL callback monitor. This has to be the same C application that was specified when acc_vcl_add() was called.

The user_data argument is user-defined data that is passed to the consumer routine each time the object changes value. This has to be the same value that was specified when acc_vcl_add() was called.

The \(v c l\) flag argument is a predefined integer constant and has to be vcl_verilog. This constant shall be used in place of the vcl_flag values used with acc_vel_add().

Refer to 22.10 for an example of using acc_vcl_delete().

\subsection*{23.99 acc_version()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{acc_version()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get a pointer to a character string that indicates version number of the ACC routine software.} \\
\hline Syntax: & \multicolumn{3}{|l|}{acc_version()} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_BYTE8 * & \multicolumn{2}{|l|}{Character string pointer} \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Description \\
\hline & None & & \\
\hline Related routines: & \multicolumn{3}{|l|}{Use acc_product_version() to get the version of the software product in use Use acc_product_type() to get the type of software product in use} \\
\hline
\end{tabular}

The ACC routine acc_version() shall return a pointer to a character string that indicates the version of the ACC routines used in the software product that called the PLI application. The return value for this routine is placed in the ACC internal string buffer. See 22.9 for explanation of strings in ACC routines.

The character string shall be in the following format:
```

Access routines Version <version number>

```

For example, if the software product is using the IEEE Std 1364 PLI version of ACC routines, acc_version() might return a pointer to the following string:
```

"Access routines Version IEEE 1364 PLI"

```

NOTE The string returned by acc_version() shall be defined by the software product vendor.

The example shown in Figure 160 uses acc_version() to identify the version of ACC routines linked to the application.
```

include "acc_user.h"
II_INT32 show_versions()
/*initialize environment for ACC routines*/
acc_initialize();
/*show version of ACC routines*/
/* and version of Verilog that is linked to ACC routines*/
io_printf("Running %s with %s\n",acc_version(),acc_product_version() );
acc_close();

```

Figure 160—Using acc_version()

\section*{24. Using TF routines}

This clause provides an overview of the types of operations that are done with the PLI task/function (TF) routines. Detailed descriptions of the routines are provided in the next section.

\subsection*{24.1 TF routine definition}

The PLI TF routines, sometimes referred to as utility routines, provide a mechanism to manipulate the arguments of user-defined system tasks and functions and to synchronize interaction between a task and the simulator. Appropriate applications include stimulus generation, error checking, and interfaces to C models.

\subsection*{24.2 TF routine system task/function arguments}

The number of arguments passed to a system task shall be returned by \(\mathbf{t f}\) _nump(). A type for each argument shall be returned by tf_typep() and is primarily used to determine if an argument is writable.

An argument shall be considered read-only if, in the Verilog HDL source description, the argument cannot be used on the left-hand side of a procedural assignment statement. Signals declared as one of the net data types or the event data type, or bit-selects, part-selects, or concatenations of net data types, shall be readonly. A module instance name or a primitive instance name shall also be read-only.

Arguments shall be considered writable from the PLI if the arguments can be used on the left-hand side of procedural assignment in the Verilog HDL source description. Signals declared as reg, integer, time, or real shall be writable, as well as bit-selects, part-selects, and concatenations of these data types.

\section*{24.3 eading and writing system task/function argument values}

User-defined system task and function argument values can be determined and altered in a number of ways with the TF routines, depending on factors such as value type, data size, and desired format.

\subsection*{24.3.1 eading and writing 2-state parameter argument values}

To access the 2-state (logic 0 and 1) value of a system task/function argument of size less than or equal to 32 bits, the routine \(\mathbf{t f} \mathbf{g e t p}()\) can be used. To set the 2 -state value of an argument of size less than or equal to 32 bits, \(\mathbf{t f}\) _putp() can be used. If the argument is 33-64 bitstf_getlongp() and tf_putlongp() can be used. For arguments of type real, tf_getrealp() and tf_putrealp() can be used. Logic X and Z bits in the argument value shall be interpreted as 0 .

\subsection*{24.3.2 eading and writing 4-state values}

If 4-states (logic \(0,1, X\), and \(Z\) ) are required and a string representation of the value is appropriate, tf_strgetp() can be used to access the value. The routines tf_strdelputp(), tf_strlongdelputp(), and tf_strrealdelputp() can be used to write 4-state values to writable arguments. For applications with a high frequency of PLI calls, the overhead of these string-based routines can be excessive. The following paragraph describes an alternative.

4 -state values can also be accessed with the routine tf_exprinfo(). This routine shall create a persistent structure that contains the 4 -state value of an argument encoded in an s_vecval structure. After tf_exprinfo() has been called once for an argument, the pointer to the s_vecval structure can be saved. The argument value can be changed using that structure along with routines tf_propagatep() to send the value in the structure into a simulation and tf_evaluatep() to update the value in the structure to the current simulation value.

\subsection*{24.3.3 eading and writing strength values}

Strength values on scalar net arguments can be accessed with the routine tf_nodeinfo().

\subsection*{24.3.4 eading and writing to memories}

Memory array values can be accessed with the routine tf_nodeinfo(). This routine returns a pointer to a memval structure that represents the array in the Verilog HDL software product. Setting a value in the memval structure shall make it available for the software tool access, but this does not automatically cause the value to be propagated to any right-hand-side memory references.

\subsection*{24.3.5 eading and writing string values}

The routine tf_getestringp() shall return the string representation of a string constant or a vector argument. There is no direct method to write string values using TF routines, but it can be accomplished by writing 8bit ASCII character values to 8-bit reg elements in a vector reg using the tf_exprinfo() value structure.

\subsection*{24.3.6 Writing return values of user-defined functions}

2-state values can be set as the return value of a user-defined function using tf_putp(), tf_putlongp() and tf_putrealp() with an argument value of \(\mathbf{0}\). It is illegal to schedule the return value of a system function at a future simulation time. The routines \(\mathbf{t f}\) _strdelputp(), \(\mathbf{t f}\) _strlongdelputp(), and \(\mathbf{t f}\) _strrealdelputp() cannot be used to return the value of a system function. Should the calltf routine for a user defined system function fail to put a value during its execution, the default value of 0 shall be applied.

NOTE calling put routines to TF argument 0 (return of a function) shall only return a value in a calltf application, when the call to the function is active. The action of the put routine shall be ignored when the function is not active.

\subsection*{24.3.7 Writing the correct \(C\) data types}

It is important to ensure that the data type of the argument to any of the tf_put routines is consistent with the data type required by the routine and specified argument.

The following examples illustrate what cautions should be taken.
If the second argument of a system task/function instance is of type tf_readwritereal, meaning the argument is declared as a real variable in the Verilog HDL source description, the following tf_put routines shall produce valid results:
```

PLI INT32 i = 5;
tf_\overline{putp(2, i); /* write an integer value to 2ndargument */}

```

This example sets the second task/function argument to 5.0 assigning an integer value to a real variable is legal in the Verilog HDL.
```

double d = 5.7;
tf_putrealp(2, d); /* write a real value to 2ndargument */

```

This example sets the second task/function argument to 5.7.
The following routines, however, shall produce invalid results for the following reasons:
```

PLI_INT32 i = 5;
tf_\overline{putrealp(2, i); /* invalid result * /}

```

The statement PLI_INT32 i = 5 passes a 32-bit integer to tf_putrealp(), which expects a 64-bit double value type. Since there is no data type checking, tf_putrealp() shall read 32 bits of undefined data and try to use it as if it were valid data. The result is unpredictable.
```

float f = 5;
tf_putrealp(2, f); /* invalid result * /

```

The float statement passes a 32 -bit float to \(\mathbf{t f}\) _putrealp(), which is expecting a 64 -bit double value type. The result is unpredictable.
```

double d = 5.7;
tf_putp(2, d); /* invalid result * /

```

The tf_putp() routine shall take only the lower 32 bits of the 64 -bit double passed to it by the statement double \(d=5.7\).

\subsection*{24.4 Value change detection}

Value changes on system task/function arguments can be detected by enabling asynchronous callbacks with tf_asynchon(). The callbacks can be disabled with tf_asynchoff(). When argument change callbacks are enabled with tf_asynchon(), whenever an argument changes value, the misctf application associated with the user-defined system task/function shall be called back with three integer arguments: data, reason, and paramvc. Argument reason shall be reason_paramvc. The value change can be examined immediately, or a second callback can be requested later in the same time step (as described in 24.6). By setting a second callback at the end of the time step, an application can process all argument value changes within in a time step at once. The routines \(\mathbf{t f}\) copypvc_flag(), tf_movepvc_flag(), tf_testpvc_flag(), and tf_getpchange() can be used to determine all the arguments that changed in a time step.

\subsection*{24.5 Simulation time}

TF routines are provided to read simulation time and to scale delays to simulation time scales.
The routines tf_gettime() and tf_getlongtime() shall return the current simulation time in unsigned format. These times shall be scaled to the timescale of the module where the system task or function is invoked. The routine tf_strgettime() shall return unscaled simulation time in a string format.

PLI TF routines that involve time shall automatically scale delay values to the timescale of the module containing the instance of the user-defined task or function.

The routines tf_gettimeunit() and tf_gettimeprecision() can be used to obtain the timescale unit and precision of a module. These routines can also be used to obtain the internal simulation time unit, which is the smallest precision of all modules within a simulation. The routines tf_scale_longdelay(), tf_scale_realdelay(), tf_unscale_longdelay(), and tf_unscale_realdelay() can be used to convert between scaled delays and internal simulation time.

\subsection*{24.6 Simulation synchronization}

There are TF routines that allow synchronized calling of the misctf application associated with a userdefined system task or function. The misctf application can be called at the end of the current time step or at some future time step.

The routines tf_synchronize() and tf_rosynchronize() shall cause the misctf application associated with a user-defined system task to be called back in the current simulation time step.

The tf_synchronize() routine shall place a callback at the end of the inactive event queue for the current time step. The misctf application shall be called with reason_synch. It is possible for subsequent events to be added to the current time step after the tf_synchronize() callback (for this reason, when the callback occurs, the next scheduled time step cannot be determined). The misctf application can propagate new values in reason_synch mode.

The tf_rosynchronize() callback shall occur after all active, inactive, and nonblocking assign events for a time step have been processed. The misctf application shall be called with reason_rosynch. With reason_rosynch, it is possible to determine the time of the next scheduled time step using tf_getnextlongtime(). Values cannot be written to system task/function arguments during a reason_rosynch callback (the 'ro'indicates read-only). Placing a callback for tf_rosynchronize() during a callback for reason reason_rosynch will result in another reason_rosynch callback occurring during the same time slice.

The routine tf_setdelay() and its variations shall schedule the misctf application to be called back at a specified time with reason argument reason_reactivate. The routine tf_clearalldelays() shall remove any previously scheduled callbacks of this type.

\subsection*{24.7 Instances of user-defined tasks or functions}

The routine tf_getinstance() shall return a unique identifier for each instance of a user-defined system task or function in the Verilog HDL source description. This value can then be used as the instance_p argument to all the \(t f_{-} i^{*}\) routines so that the arguments of one instance can be manipulated from another task or function instance.

\subsection*{24.8 Module and scope instance names}

The full hierarchical path name of the module that contains an instance shall be returned by the routine tf_mipname(). The full name of the containing scope, which can be a Verilog HDL task or function, a named block, or a module instance, shall be returned by tf_spname().

\subsection*{24.9 Saving information from one system TF call to the next}

The TF routines tf_setworkarea() and tf_getworkarea() provide a special storage work area that can be used for:

Saving data during one call to a PLI application that can be retrieved in a subsequent call to the application.
Passing data from one type of PLI application to another, such as from a checktf application to a calltf application.

\subsection*{24.10 Displaying output messages}

The routine io_printf() can be used in place of the C printf() statement. This routine has essentially the same syntax and semantics as printf(), but it displays the output message to both the output channel of the software product which invoked the PLI application and to the log file of the software product.

The routine io_mcdprintf() is also similar to the C printf(), but permits writing information to files that were opened within the Verilog HDL source description using the \$fopen() built-in system function.

The routines \(\mathbf{t f}\) _warning(), \(\mathbf{t f}\) _error(), \(\mathbf{t f}\) _message(), and \(\mathbf{t f}\) _text() can be used to display warning and error messages that are automatically formatted to a similar format as the warning and error messages for the software product. The routines \(\mathbf{t f}\) error() and tf_message() shall also provide control for aborting the software product execution when an error is detected.

\subsection*{24.11 Stopping and finishing}

The routines \(\mathbf{t f}\) _dostop() and \(\mathbf{t f}\) _dofinish() are the PLI equivalents to the built-in system tasks \$stop and \(\$\) finish.

\section*{25. TF routine definitions}

This clause defines the PLI TF routines, explaining their function, syntax, and usage. The routines are listed in alphabetical order. See Clause 23 for conventions that are used in the definitions of the PLI routines.

\section*{25.1 io_mcdprintf()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{io_mcdprintf()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Write a formatted message to one or more files.} \\
\hline Syntax: & \multicolumn{3}{|l|}{io_mcdprintf(mcd, format, arg1,...arg12)} \\
\hline \multicolumn{3}{|c|}{Type} & Description \\
\hline \multicolumn{2}{|l|}{Returns: void} & & \\
\hline \multirow[b]{2}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & PLI_INT32 & mcd & An integer multi-channel descriptor value representing one or more open files \\
\hline & quoted string or PLI_BYTE8 * & format & A quoted character string or pointer to a character string that controls the message to be written \\
\hline (optional) & & \(\arg 1 \ldots \arg 12\) & 1 to 12 optional arguments of the format control string; the type of each argument should be consistent with how it is used in the format string \\
\hline Related routines: & \multicolumn{3}{|l|}{Use io_printf() to write messages to the output channel of the software product which invoked the PLI application and to the Verilog product log file} \\
\hline
\end{tabular}

The TF routine io_mcdprintf() shall write a formatted message to one or more open files, as described by the multi-channel descriptor mcd. This routine uses the descriptors created by the \$fopen system task or the VPI routine vpi_mcd_open(). See 17.2 .1 for the functional description of \$fopen, and 27.25 for the description of vpi_mcd_open().

The format strings shall use the same format as the C routine \(\operatorname{fprintf}()\).
The maximum number of arguments that can be used in the format control string is 12 .

\section*{25.2 io_printf()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{io_printf()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Print a formatted message to the output channel of the software product which invoked the PLI application and to the log file of the product.} \\
\hline Syntax: & \multicolumn{3}{|l|}{io_printf(format, arg1,...arg12)} \\
\hline \multicolumn{3}{|c|}{Type} & Description \\
\hline \multicolumn{2}{|l|}{Returns: void} & & \\
\hline & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline Arguments: & quoted string or PLI_BYTE8 * & format & A quoted character string or pointer to a character string that controls the message to be written \\
\hline (optional) & & \(\arg 1 \ldots \arg 12\) & 1 to 12 optional arguments of the format control string; the type of each argument should be consistent with how it is used in the format string \\
\hline Related routines: & \multicolumn{3}{|l|}{Use io_mcdprintf() to write a formatted message to one or more open files Use tf_message(), tf_error(), or tf_warning() to write error or warning messages} \\
\hline
\end{tabular}

The TF routine io_printf() shall write a formatted message as text output. The functionality is similar to the C printf() function. However, io_printf() differs from printf() because it ensures the message is written to both the output channel of the software product which invoked the PLI application and the output log file of the product.

The format control string uses the same formatting controls as the C printf() function (for example, \(\% \mathrm{~d}\) ).
The maximum number of arguments that can be used in the format control string is 12 .

\section*{25.3 mc_scan_plusargs()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{mc_scan_plusargs()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Scan software product invocation command line for plus ( + ) options.} \\
\hline Syntax: & \multicolumn{3}{|l|}{mc_scan_plusargs(startarg)} \\
\hline & \multicolumn{2}{|l|}{Type} & Description \\
\hline Returns: & PLI_BYTE8 * & \multicolumn{2}{|l|}{Pointer to a string with the result of the search} \\
\hline & Type & Name & Description \\
\hline Arguments: & quoted string or PLI_BYTE8 * & startarg & A quoted string or pointer to a character string with the first part of the invocation option to search for \\
\hline
\end{tabular}

The TF routine mc_scan_plusargs() shall scan all software product invocation command options and match a given string to a plus argument. The match is case sensitive.

The routine mc_scan_plusargs() shall

Return null if startarg is not found
Return the remaining part of the command argument if startarg is found (e.g., if the invocation option string is "+siz64", and startarg is "siz", then "64" is returned)

Return a pointer to a C string with a null terminator if there is no remaining part of a found plus argument

\section*{25.4 tf_add_long()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|r|}{tf_add_long()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Add two 64-bit integers.} \\
\hline Syntax: & \multicolumn{3}{|l|}{tf_add_long(aof_low1, aof_high1, low2, high2)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type} & Description \\
\hline & PLI_INT32 & \multicolumn{2}{|l|}{Always returns 0} \\
\hline \multirow{5}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_INT32 * & aof_low1 & Pointer to \\
\hline & PLI_INT32 * & aof_high1 & Pointer to \\
\hline & PLI_INT32 & low2 & Least sig \\
\hline & PLI_INT32 & high2 & Most sig \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_subtract_long() to subtract two 64-bit integers Use tf_multiply_long() to multiply two 64-bit integers Use tf_divide_long() to divide two 64-bit integers Use tf_compare_long() to compare two 64-bit integers} \\
\hline
\end{tabular}

The TF routine \(\mathbf{t f}\) _add_long() shall add two 64-bit values. After calling tf_add_long(), the variables used to pass the first operand shall contain the results of the addition. Figure 161 shows the high and low 32 bits of two 64-bit integers and how tf_add_long() shall add them.


Figure 161-Adding with tf_add_long()

\section*{25.5 tf_asynchoff(), tf_iasynchoff()}


The TF routines tf_asynchoff() and tf_iasynchoff() shall disable further calling of the misctf application for reason_paramve for the current instance or a specific instance of a user-defined system task or function.

Asynchronous calling is first enabled by the routines \(\mathbf{t f}\) _asynchon() or tf_iasynchon().

\section*{25.6 tf_asynchon(), tf_iasynchon()}


The TF routines tf_asynchon() and tf_iasynchon() shall enable a misctf user application to be called asynchronously whenever a system task/function argument value changes in the current instance or in a specific instance of a user-defined system task or function. After enabling, the routine specified by misctf in the PLI interface mechanism shall be called with a reason of reason_paramve each time any task/function argument changes value or strength. The index number of the argument that changed is passed to the misctf application as a third C argument, paramve.

The value change can be examined immediately, or a second callback can be requested later in the same time step (as described in Section 24.6). By setting a second callback at the end of the time step, an application can process all argument value changes within a time step at once. The routines tf_copypvc_flag(), tf_movepvc_flag(), tf_testpvc_flag(), and tf_getpchange() can be used to determine all the arguments that changed in a time step.

Task/function argument index numbering shall proceed from left to right, and the left-most argument shall be number 1 .

\section*{25.7 tf_clearalldelays(), tf_iclearalldelays()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|r|}{tf_clearalldelays(), tf_iclearalldelays()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Clear all scheduled reactivations by tf_setdelay() or tf_isetdelay().} \\
\hline Syntax: & \multicolumn{3}{|l|}{```
tf_clearalldelays()
tf_iclearalldelays(instance_p)
```} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type} & Description \\
\hline & PLI_INT32 & \multicolumn{2}{|l|}{Always returns 1} \\
\hline \multirow[b]{2}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a speci or function \\
\hline Related routines: & \multicolumn{3}{|l|}{\begin{tabular}{l}
Use tf_setdelay() or tf_isetdelay() to schedule a reactivation \\
Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function
\end{tabular}} \\
\hline
\end{tabular}

The TF routines tf_clearalldelays() and tf_iclearalldelays() shall clear all reactivation delays, which shall remove the effect of all previous tf_setdelay() or tf_isetdelay() calls for the current instance or specific instance of a user-defined system task or function.
25.8 tf_compare_long()
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_compare_long()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Compare two 64-bit integer values.} \\
\hline Syntax: & \multicolumn{3}{|l|}{tf_compare_long(low1, high1, low2, high2)} \\
\hline \multirow[b]{2}{*}{Returns:} & Type & \multicolumn{2}{|r|}{Description} \\
\hline & PLI_INT32 & \multicolumn{2}{|l|}{An integer flag indicating the result of the comparison} \\
\hline \multirow{5}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_UINT32 & low1 & Least significant 32 bits of first operand \\
\hline & PLI_UINT32 & high1 & Most significant 32 bits of first operand \\
\hline & PLI_UINT32 & low2 & Least significant 32 bits of second operand \\
\hline & PLI_UINT32 & high2 & Most significant 32 bits of second operand \\
\hline
\end{tabular}
\begin{tabular}{||l|l||}
\hline \multicolumn{1}{|l|}{} & \multicolumn{1}{c|}{ tf_compare_long() } \\
\hline \begin{tabular}{l} 
Related \\
routines:
\end{tabular} & \begin{tabular}{l} 
Use tf_add_long() to add two 64-bit integers \\
Use tf_subtract_long() to subtract two 64-bit integers \\
Use tf_multiply_long) to multiply two 64-bit integers \\
Use tf_divide_long() to divide two 64-bit integers
\end{tabular} \\
\hline
\end{tabular}

The TF routine tf_compare_long() shall compare two 64-bit integers and return one of the values given in Table 186.

Table 186—Return values for tf_compare_long()
\begin{tabular}{|c|c|}
\hline When & \begin{tabular}{c} 
tf_compare_long() shall \\
return
\end{tabular} \\
\hline operand1 < operand2 & \(\mathbf{- 1}\) \\
\hline operand 1 = operand2 & \(\mathbf{0}\) \\
\hline operand \(>\) operand 2 & \(\mathbf{1}\) \\
\hline
\end{tabular}

\section*{25.9 tf_copypvc_flag(), tf_icopypvc_flag()}


The TF routines tf_copypvc_flag() and tf_icopypve_flag() shall copy the current pve flag to the saved pvc flag and return the value of the flag that was copied. The argument narg is the index number of an argument in the current instance or a specific instance of a user-defined system task or function. Task/function argument index numbering shall proceed from left to right, with the left-most argument being number 1. If narg is \(\mathbf{- 1}\), then all argument pve flags shall be copied and the logical OR of all saved flags returned.

Argument Value Change (pvc) flags shall be used to indicate whether a particular user-defined system task or function argument has changed value. Each argument shall have two pve flags: a current pve flag, which shall be set by a software product when the change occurs, and a saved pvc flag, which shall be controlled by the user.

NOTE PVC flags shall not be set by the software product until tf_asynchon() or tf_iasynchon() has been called.

\subsection*{25.10 tf_divide_long()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|r|}{tf_divide_long()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Divide two 64-bit integers.} \\
\hline Syntax: & \multicolumn{3}{|l|}{tf_divide_long(aof_low1, aof_high1, low2, high2)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type} & Description \\
\hline & void & & \\
\hline \multirow{5}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_INT32 * & aof_low1 & Pointer to \\
\hline & PLI_INT32 * & aof_high1 & Pointer to \\
\hline & PLI_INT32 & low2 & Least sig \\
\hline & PLI_INT32 & high2 & Most sig \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_add_long() to add two 64-bit integers Use tf_subtract_long() to subtract two 64-bit integers Use tf_multiply_long() to multiply two 64-bit integers Use tf_compare_long() to compare two 64-bit integers} \\
\hline
\end{tabular}

The TF routine tf_divide_long() shall divide two 64-bit values. After calling tf_divide_long(), the variables used to pass the first operand shall contain the result of the division.

The operands shall be assumed to be in two s complement form. Figure 162 shows the high and low 32 bits of two 64-bit integers and how tf_divide_long() shall divide them.


Figure 162—Dividing with tf_divide_long()

\subsection*{25.11 tf_dofinish()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_dofinish()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Exit software product execution.} \\
\hline Syntax: & \multicolumn{3}{|l|}{tf_dofinish()} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type} & Description \\
\hline & PLI_INT32 & \multicolumn{2}{|l|}{Always returns 0} \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Description \\
\hline & None & & \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_dostop() to cause a product to enter interactive mode} \\
\hline
\end{tabular}

The TF routine \(\mathbf{t f}\) _dofinish() shall finish the software product execution the same as if a \(\$\) finish() built-in system task had been executed in the Verilog HDL source description.

\subsection*{25.12 tf_dostop()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_dostop()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Cause software product to enter interactive mode.} \\
\hline Syntax: & \multicolumn{3}{|l|}{tf_dostop()} \\
\hline & \multicolumn{2}{|l|}{Type} & Description \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{Always returns 0} \\
\hline & Type & \multirow[t]{2}{*}{Name} & Description \\
\hline Arguments: & None & & \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_dofinish() exit software product execution} \\
\hline
\end{tabular}

The TF routine tf_dostop() shall cause a software product to enter into its interactive mode as if a \$stop() built-in system task had been executed in the Verilog HDL source description.

\subsection*{25.13 tf_error()}


The TF routine tf_error() shall provide an error reporting mechanism compatible with error messages generated by the software product.

The format control string uses the same formatting controls as the C printf() function (for example, \(\% \mathrm{~d}\) ).

The maximum number of arguments that can be used in the format control string is five.
The location information (file name and line number) of the current instance of the user-defined system task or function is appended to the message using a format compatible with error messages generated by the software product.

The message is written to both the output channel of the software product which invoked the PLI application and the output log file of the product.

If tf_error() is called by the checktf application associated with the user-defined system task or function, the following rules shall apply:

If the checktf application is called when the Verilog HDL source code was being parsed or compiled, parsing or compilation shall be aborted after the error is reported.

If the checktf application is called when the user-defined task or function was invoked on the interactive command line, the interactive command shall be aborted.

\subsection*{25.14 tf_evaluatep(), tf_ievaluatep()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_evaluatep(), tf_ievaluatep()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Evaluate a system task/function argument expression.} \\
\hline Syntax: & \multicolumn{3}{|l|}{\begin{tabular}{l}
tf_evaluatep (narg) \\
tf_ievaluatep (narg, instance_p)
\end{tabular}} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type} & Description \\
\hline & PLI_INT32 & \multicolumn{2}{|l|}{0 if successful; 1 if an error occurred} \\
\hline \multirow{3}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & PLI_INT32 & narg & Index number of the user-defined system task or function argument \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_exprinfo() or tf_iexprinfo() to get a pointer to the s_tfexprinfo structure Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routines tf_evaluatep() and tf_ievaluatep() shall evaluate the current value of the specified argument in the current instance or a specific instance of a user-defined system task or function. The current value shall be returned to the value cell in the \(t f\) _exprinfo structure returned from a previous call to the routine tf_exprinfo() or tf_iexprinfo(). This can be a more efficient way to obtain the current value of an expression than to call tf_exprinfo() or tf_iexprinfo() repeatedly.

The argument narg shall be the index number of an argument in a user-defined system task or function. Task/function argument index numbering shall proceed from left to right, with the left-most argument being number 1.

\subsection*{25.15 tf_exprinfo(), tf_iexprinfo()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_exprinfo(), tf_iexprinfo()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get system task/function argument expression information.} \\
\hline Syntax: & \multicolumn{3}{|l|}{```
tf_exprinfo(narg, exprinfo_p)
tf_iexprinfo(narg, exprinfo_p, instance_p)
```} \\
\hline & \multicolumn{2}{|l|}{Type} & Description \\
\hline Returns: & struct t_tfexprinfo * & \multicolumn{2}{|l|}{Pointer to a structure containing the value of the second argument if successful; 0 if an error occurred} \\
\hline \multirow{4}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_INT32 & narg & Index number of the user-defined system task or function argument \\
\hline & struct t_tfexprinfo * & exprinfo_p & Pointer to a variable declared as a t_tfexprinfo structure type \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_nodeinfo() or tf_inodeinfo() for additional information on writable arguments Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routines \(\mathbf{t f}\) _exprinfo() and tf_iexprinfo() shall return a pointer to a structure containing general information about the specified argument in the current instance or a specific instance of a user-defined system task or function. The information shall be stored in the \(C\) structure s_tfexprinfo.

Memory space shall first be allocated to hold the information before calling tf_exprinfo() or tf_iexprinfo(). For example:
```

    {
    s_tfexprinfo info; /* declare a variable of the structure type
    * /
tf_exprinfo(n, \&info); /* pass tf_exprinfo a pointer to the variable
* /
}

```

This routine shall return the second argument, which is the pointer to the information structure. If narg is out of range, or if some other error is found, then \(\mathbf{0}\) shall be returned. The argument narg shall be the index number of an argument in a user-defined system task or function. Task/function argument index numbering shall proceed from left to right, with the left-most argument being number 1.

The s_tfexprinfo structure is defined in veriuser. h and is listed in Figure 163.
```

typedef struct t_tfexprinfo
{
PLI INT16 expr type;
PLI_INT16 padding;
struct t_vecval *expr_value_p;
double real_value;
PLI_BYTE8 *expr_string;
PLI_INT32 expr_ngroups;
PLI_INT32 expr_vec_size;
PLI_INT32 expr_sign;
PLI_INT32 expr_lhs_select;
PLI_INT32 expr_rhs_select;
} s_tfexprinfo, *p_tfexprinfo;

```

Figure 163-The s_tfexprinfo structure definition

The expr_type of the s_tfexprinfo structure shall indicate the Verilog HDL data type of the argument, and it shall be one of the predefined constants as given in Table 187 and defined in veriuser. h .

Table 187-Predefined constants used with tf_exprinfo()
\begin{tabular}{|c|c|}
\hline Predefined constant & Description \\
\hline tf_nullparam & For null or non-existent arguments \\
\hline tf_string & For string arguments \\
\hline tf_readonly & \begin{tabular}{c} 
For net, net bit, net part select and \\
constant integer arguments
\end{tabular} \\
\hline tf_readonlyreal & For constant real number arguments \\
\hline tf_readwrite & \begin{tabular}{c} 
For reg, integer and time variable \\
arguments
\end{tabular} \\
\hline
\end{tabular}

Table 187-Predefined constants used with tf_exprinfo() (continued)
\begin{tabular}{|c|c|}
\hline Predefined constant & Description \\
\hline tf_readwritereal & For real variable arguments \\
\hline tf_rwbitselect & \begin{tabular}{c} 
For bit-select of reg, integer and \\
time variable arguments
\end{tabular} \\
\hline tf_rwpartselect & \begin{tabular}{c} 
For part-select of reg, integer and \\
time variable arguments
\end{tabular} \\
\hline tf_rwmemselect & For memory word arguments \\
\hline
\end{tabular}

If the expression type is tf_readonly, tf_readwrite, \(\mathbf{t f}\) _rwbitselect, tf_rwpartselect, or tf_rwmemselect, the expr_value \(\quad p\) of the s_tfexprinfo structure shall be a pointer to an array of s_vecval structures that shall contain the resultant value of the expression. The s_vecval structure for representing vector values is defined in veriuser. h and is listed in Figure 164.
```

typedef struct t_vecval
{
PLI INT32 avalbits;
PLI_INT32 bvalbits;
s_vecval, *p_vecval;

```

Figure 164-The s_vecval structure definition
If the number of bits in the vector (defined by the expr_vec_size field of the s_tfexprinfo structure) is less than or equal to 32 , then there shall only be one s_vecval group in the expr_value p array. For 33 bits to 64 bits, there shall be two groups in the array, and so on. The number of groups shall also be given by the value of the expr_ngroups field of the s_tfexprinfo structure. The components avalbits and bvalbits of the s_vecval structure shall hold the bit patterns making up the value of the argument. The lsb in the value shall be represented by the lsb sin the avalbits and bvalbits components, and so on. The bit coding shall be as given in Table 188.

Table 188-avalbits/bvalbits encoding
\begin{tabular}{|c|c|}
\hline aval / bval & Logic value \\
\hline 00 & 0 \\
\hline 10 & 1 \\
\hline 01 & High impedance \\
\hline 11 & Unknown \\
\hline
\end{tabular}

If the expression type is tf_readonlyreal or tf_readwritereal, the real_value field of the s_tfexprinfo structure shall contain the value.

If the expression is of type \(\mathbf{t f}\) _string, the expr_string field of the s_tfexprinfo structure shall point to the string.

If the expression type is tf_readonly, tf_readwrite, \(\mathbf{t f}\) _rwbitselect, \(\mathbf{t f}\) _rwpartselect, or tf_rwmemselect, the expr_ngroups of the s_tfexprinfo structure shall indicate the number of groups for the argument expression value and determine the array size of the expr_value_p value structure pointer. If the expression type is tf_readonlyreal or tf_readwritereal, expr_ngroups shall be 0 .

If the expression type is tf_readonly, tf_readwrite, tf_rwbitselect, tf_rwpartselect, or tf_rwmemselect, the expr_vec_size field of the s_tfexprinfo structure shall indicate the total number of bits in the array of expr_value_p value structures. If the expression type is tf_readonlyreal or tf_readwritereal, expr_vec_size shall be 0 .

The expr_sign field of the s_tfexprinfo structure shall indicate the sign type of the expression. It shall be 0 for unsigned or nonzero for signed.

The expr_lhs_select and expr_rhs_select fields shall contain the select information about the object if it is a reg bit-select, net bit-select, part-select, variable array word-select, or memory word-select.

\subsection*{25.16 tf_getcstringp(), tf_igetcstringp()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|r|}{tf_getcstringp(), tf_igetcstringp()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get system task/function argument value as a string.} \\
\hline Syntax: & \multicolumn{3}{|l|}{```
tf_getcstringp(narg)
tf_igetcstringp(narg, instance_p)
```} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type} & Description \\
\hline & PLI_BYTE8 * & \multicolumn{2}{|l|}{Pointer to a character string} \\
\hline \multirow{3}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_INT32 & narg & Index number of the user-defi argument \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance or function \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf _getp() or \(\operatorname{tf}\) _igetp() to get an argument value as a 32 -bit integer Use tf_getlongp() or tf_igetlongp() to get an argument value as a 64-bit integer Use tf _getrealp() or tf _igetrealp() to get an argument value as a double Use tf_strgetp() or tf_istrgetp() to get an argument value as a formatted string Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routines tf_getcstringp() and tf_igetcstringp() shall return a character string representing the value of the specified argument in the current instance or a specific instance of a user-defined system task or function. If the argument identified by narg is a literal string, reg, integer variable, time variable, or an expression, then \(\mathbf{t f}\) _getcstringp() or tf_igetestringp() shall convert its value to a C language ASCII string by
a) Eliminating leading zeros
b) Converting each group of 8 bits to an ASCII character
c) Adding a \(\backslash 0\) string termination character to the end

If the argument identified by narg is null or if narg is out of range, then a null shall be returned. If the argument identified by narg is a real variable or an expression that evaluates to a real value, then tf_getestringp() and tf_igetcstringp() shall return NULL.

The argument narg shall be the index number of an argument in a user-defined system task or function. Task/function argument index numbering shall proceed from left to right, with the left-most argument being number 1.

\subsection*{25.17 tf_getinstance()}
\begin{tabular}{||l|l|l||}
\hline \multicolumn{4}{|c|}{ tf_getinstance() } \\
\hline Synopsis: & Get a pointer to the current instance of a user-defined system task or function. \\
\hline Syntax: & tf_getinstance () \\
\hline \multicolumn{4}{|c|}{ Type } & Description \\
\cline { 2 - 4 } Returns: & PLI_BYTE8 * & Pointer to a system task or function instance \\
\hline \multicolumn{4}{|c|}{} \\
Arguments: & Type & Name \\
\hline
\end{tabular}

The TF routine tf_getinstance() shall return a pointer that identifies the current instance of the user-defined task or function in the Verilog HDL source code. The pointer returned by tf_getinstance() can be used later in other TF routine calls to refer to this instance of the task or function. Many of the TF routines are in two forms. One deals with the current task or function instance. The other deals with some other instance of the task or function, where the instance pointer for the other instance was previously obtained using tf_getinstance() during a call to a user routine initiated by that instance.

\subsection*{25.18 tf_getlongp(), tf_igetlongp()}


The TF routines tf_getlongp() and tf_igetlongp() shall return a 64-bit integer value for the argument specified by narg in the current instance or a specific instance of a user-defined system task or function. If narg is out of range or the argument is null, then 0 shall be returned. Logic X and Z bits in the argument value shall be interpreted as 0 .

The argument narg shall be the index number of an argument in a user-defined system task or function. Task/function argument index numbering shall proceed from left to right, with the left-most argument being number 1.

\subsection*{25.19 tf_getlongtime(), tf_igetlongtime()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|r|}{tf_getlongtime(), tf_igetlongtime()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get current simulation time as a 64-bit integer.} \\
\hline Syntax: & \multicolumn{3}{|l|}{\begin{tabular}{l}
tf_getlongtime (aof_hightime) \\
tf_igetlongtime (aof_hightime, instance_p)
\end{tabular}} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type} & Descrip \\
\hline & PLI_INT32 & \multicolumn{2}{|l|}{Least significant (right-most) 32 bits of simulation time} \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & \\
\hline & PLI_INT32 * & aof_hightime & Pointer to most time \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a spec or function \\
\hline Related routines: & \multicolumn{3}{|l|}{\begin{tabular}{l}
Use tf_gettime() to get the simulation time as a 32-bit integer \\
Use tf_strgettime() to get the simulation time as a character string \\
Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function
\end{tabular}} \\
\hline
\end{tabular}

The TF routines tf_getlongtime() and tf_igetlongtime() shall return the simulation time as a 64-bit integer. The high 32 bits of simulation time shall be assigned to the aof_hightime argument, and the low 32 bits of time shall be returned.

Time shall be expressed in the timescale unit of the module containing the current instance or a specific instance of the user-defined system task or function.

\subsection*{25.20 tf_getnextlongtime()}


The TF routine tf_getnextlongtime() shall assign the 64-bit time of the next simulation event to aof_lowtime and aof_hightime, and it shall return an integer value that indicates the meaning of the time assigned. The time shall be expressed in the timescale units of the module containing the current userdefined system task or function instance.

The tf_getnextlongtime() routine shall only return the time for the next simulation event when it is called in a read-only synchronize mode. A read-only synchronize mode occurs when the misctf user application has been called with reason_rosynch. If tf_getnextlongtime() is not called in read-only synchronize mode, then the current simulation time shall be assigned.

Table 189 summarizes the functions of tf_getnextlongtime().

Table 189—Return values for tf_getnextlongtime()
\begin{tabular}{|c|c|c|}
\hline When & tf_getnextlongtime() shall return & tf_getnextlongtime() shall assign to aof_lowtime and aof_hightime \\
\hline tf_getnextlongtime() was called from a misctf application that was called with reason_rosynch & 0 & The next simulation time for which an event is scheduled \\
\hline There are no more future events scheduled & 1 & 0 \\
\hline tf_getnextlongtime() was not called from a misctf application that was called with reason_rosynch & 2 & The current simulation time \\
\hline
\end{tabular}

NOTE Case 2 shall take precedence over case 1 .
25.21 tf_getp(), tf_igetp()
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_getp(), tf_igetp()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get a system task/function argument value as an integer or character string pointer.} \\
\hline Syntax: & \multicolumn{3}{|l|}{```
tf_getp(narg)
tf_igetp(narg, instance_p)
```} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type} & Description \\
\hline & PLI_INT32 & \multicolumn{2}{|l|}{Integer value of an argument or character string pointer of argument string value} \\
\hline \multirow{3}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_INT32 & narg & Index number of the user-defined system task or function argument \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_getlongp() or tf_igetlongp() to Get an argument value as a 64-bit integer Use tf_getrealp() or tf_igetrealp() to get an argument value as a double Use tf_getcstringp() or tf_igetcstringp() to get an argument value as a string Use tf _strgetp() or tf _istrgetp() to get an argument value as a formatted string Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routines \(\mathbf{t f} \mathbf{g e t p}()\) and \(\mathbf{t f} \mathbf{i g e t p}()\) shall return a value of the argument specified by narg in the current instance or a specific instance of a user-defined system task or function. If the value of the argument is an integer or a real number, the routines shall return an integer value. If the argument is a literal string, then the routines shall return a pointer to a C type string (a string terminated by a \(\backslash 0\) character). If narg is out of range or the argument is null, then 0 shall be returned. Logic X and Z bits in the argument value shall be interpreted as 0 .

The argument narg shall be the index number of an argument in a user-defined system task or function. Task/function argument index numbering shall proceed from left to right, with the left-most argument being number 1.

The routines \(\mathbf{t f} \mathbf{g e t p}()\) and \(\mathbf{t f} \mathbf{g e t r e a l p}()\) differ in the value returned, as shown by the following example.
If the fourth argument in the user-defined system task or function has a value of 9.6 (a real value), then
\[
\text { PLI_INT32 ivalue }=\text { tf_getp(4) }
\]
would set ivalue to 10 , whereas
```

double dvalue = tf_getrealp(4)

```
would set dvalue to 9.6.

In the first example, note that the PLI_INT32 conversion rounds off the value of 9.6 to 10 (rather than truncating it to 9). In the second example, note that the real value has to be declared as a double (not as a float ). Rounding is performed following the Verilog HDL rules.

\subsection*{25.22 tf_getpchange(), tf_igetpchange()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_getpchange(), tf_igetpchange()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the index number of the next system task/function argument that changed value.} \\
\hline Syntax: & \multicolumn{3}{|l|}{\begin{tabular}{l}
tf_getpchange (narg) \\
tf_igetpchange(narg, instance_p)
\end{tabular}} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type} & Description \\
\hline & PLI_INT32 & \multicolumn{2}{|l|}{Index number of the argument that changed} \\
\hline \multirow{3}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_INT32 & narg & Index number of the user-defined system task or function argument \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{\begin{tabular}{l}
Use tf_asynchon() or tf_iasynchon() to enable pvc flags \\
Use tf_imovepvc_flag(-1) to save pvc flags before calling tf_getpchange() \\
Use tf_copypvc_flag() or tf_icopypvc_flag() to copy pve flags \\
Use tf_testpvc_flag() or tf_itestpvc_flag() to get the value of a saved pve flag \\
Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function
\end{tabular}} \\
\hline
\end{tabular}

The TF routines tf_getpchange() and tf_igetpchange() shall return the number of the next argument with a number greater than narg that changed value for the current instance or for a specific instance of a userdefined system task or function. The narg argument shall be 0 the first time this routine is called within a
given user routine invocation. The routines shall return the argument number if there is a change in an argument with a number greater than narg, and they shall return 0 if there are no changes in arguments greater than narg or if an error is detected. The routine shall use the saved pve flags, so it is necessary to execute tf_movepve_flag(-1) prior to calling the routine.

The argument narg shall be the index number of an argument in a user-defined system task or function. Task/function argument index numbering shall proceed from left to right, with the left-most argument being number 1.

PVC flags shall indicate whether a particular user-defined system task or function argument has changed value. Each argument shall have two pvc flags: a current pvc flag, which shall be set by a software product when the change occurs, and a saved pve flag, which shall be controlled by the user.

NOTE PVC flags shall not be set by the software product until tf_asynchon() or tf_iasynchon() has been called.

\subsection*{25.23 tf_getrealp(), tf_igetrealp()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_getrealp(), tf_igetrealp()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get a system task/function argument value as a double-precision value.} \\
\hline Syntax: & \multicolumn{3}{|l|}{```
tf_getrealp(narg)
tf_igetrealp(narg, instance_p)
```} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type} & Description \\
\hline & double & \multicolumn{2}{|l|}{Double-precision value of an argument} \\
\hline \multirow{3}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_INT32 & narg & Index number of the user-defined system task or function argument \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf _getp() or \(\operatorname{tf}\) _igetp() to get an argument value as a 32 -bit integer Use tf_getlongp() or tf_igetlongp() to get an argument value as a 64-bit integer Use tf_getcstringp() or tf_igetcstringp() to get an argument value as a string Use tf _strgetp() or tf _istrgetp() to get an argument value as a formatted string Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routines \(\mathbf{t f}\) _getrealp() and \(\mathbf{t f}\) _igetrealp() shall return a double-precision value of the argument specified by narg in the current instance or a specific instance of a user-defined system task or function. If narg is out of range or the argument is null, then 0 shall be returned. Logic X and Z bits in the argument value shall be interpreted as 0 .

The routines \(\mathbf{t f}\) _getrealp( ) and \(\mathbf{t f}\) _igetrealp() shall return 0.0 if the value being read is a literal string. Therefore, before calling these routines, tf_typep() or tf_itypep() should be called to check the type of the argument.

The argument narg shall be the index number of an argument in a user-defined system task or function. Task/function argument index numbering shall proceed from left to right, with the left-most argument being number 1.

\subsection*{25.24 tf_getrealtime(), tf_igetrealtime()}


The TF routines tf_getrealtime() and tf_igetrealtime() shall return the simulation time as a real number in double-precision format.

Time shall be expressed in the timescale unit of the module containing the current instance or a specific instance of a user-defined system task or function.

\subsection*{25.25 tf_gettime(), tf_igettime()}


The TF routines tf_gettime() and tf_igettime() shall return the lower 32 bits of simulation time as an integer.

Time shall be expressed in the timescale unit of the module containing the current instance or a specific instance of a user-defined system task or function.

\subsection*{25.26 tf_gettimeprecision(), tf_igettimeprecision()}


The TF routines \(\mathbf{t f}\) _gettimeprecision() and \(\mathbf{t f}\) _igettimeprecision() shall return the timescale precision for the module that contains the current instance or a specific instance of a user-defined system task or function. The time precision is set by the 'timescale Verilog HDL compiler directive in effect when the module was compiled. The routines shall return an integer code representing the time precision, as shown in Table 190.

\section*{Table 190-Code returned by tf_gettimeprecision() and tf_igettimeprecision()}
\begin{tabular}{|c|c|}
\hline Integer code returned & Simulation time precision \\
\hline 2 & 100 s \\
\hline 1 & 10 s \\
\hline 0 & 1 s \\
\hline-1 & 100 ms \\
\hline-2 & 10 ms \\
\hline-3 & 1 ms \\
\hline-4 & 100 s \\
\hline-5 & 10 s \\
\hline-6 & 1 s \\
\hline-7 & 100 ns \\
\hline-8 & 10 ns \\
\hline-9 & 1 ns \\
\hline-10 & 100 ps \\
\hline-11 & 10 ps \\
\hline-12 & 1 ps \\
\hline-13 & 100 fs \\
\hline-14 & 10 fs \\
\hline-15 & 1 fs \\
\hline
\end{tabular}

When tf_igettimeprecision() is called with a null instance pointer, the routine shall return the simulation time unit, which is the smallest time precision used by all modules in a design.

\subsection*{25.27 tf_gettimeunit(), tf_igettimeunit()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|r|}{tf_gettimeunit(), tf_igettimeunit()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the timescale unit of a module or a simulation.} \\
\hline Syntax: & \multicolumn{3}{|l|}{```
tf_gettimeunit()
tf_igettimeunit(instance_p)
```} \\
\hline \multirow[b]{2}{*}{Returns:} & Type & \multicolumn{2}{|r|}{Description} \\
\hline & PLI_INT32 & An integer value th & nts a time unit \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a spec or function or \(n\) \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_gettimeprecision() or tf_igettimeprecision() to get the timescale time precision Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routines \(\mathbf{t f}\) _gettimeunit() and \(\mathbf{t f}\) _igettimeunit() shall return the timescale time units for the module that contains the current instance or a specific instance of a user-defined system task or function. The time unit for a module is set by the 'timescale Verilog HDL compiler directive in effect when the module was compiled. The routines shall return an integer code representing the time unit, as shown in Table 191.

Table 191-Code returned by tf_gettimeunit() and tf_igettimeunit()
\begin{tabular}{|c|c|}
\hline Integer code returned & Simulation time unit \\
\hline 2 & 100 s \\
\hline 1 & 10 s \\
\hline 0 & 1 s \\
\hline-1 & 100 ms \\
\hline-2 & 10 ms \\
\hline-3 & 1 ms \\
\hline-4 & 100 s \\
\hline-5 & 10 s \\
\hline-6 & 1 s \\
\hline-7 & 100 ns \\
\hline-8 & 10 ns \\
\hline-9 & 1 ns \\
\hline-10 & 100 ps \\
\hline-11 & 10 ps \\
\hline-12 & 1 ps \\
\hline-13 & 100 fs \\
\hline-14 & 10 fs \\
\hline-15 & 1 fs \\
\hline
\end{tabular}

When tf_igettimeunit() is called with a null instance pointer, the routines shall return the simulation time unit, which is the smallest time precision used by all modules in a design.

\subsection*{25.28 tf_getworkarea(), tf_igetworkarea()}


The TF routines tf_getworkarea() and tf_igetworkarea() shall return the work area pointer value of the current instance or a specific instance of a user-defined system task or function. The value of the work area pointer shall be placed there by a previous call to the routine \(\mathbf{t f}\) _setworkarea() or tf_isetworkarea(). These routines can be used as a means for two user applications to share information. For example, a checktf user application might open a file and then place the file pointer into the workarea using tf_setworkarea(). Later, the calltf user application can retrieve the file pointer using tf_getworkarea().
25.29 tf_long_to_real()


The TF routine tf_long_to_real() shall convert a 64-bit integer to a real (double-precision floating-point) number. The variable pointed to by aof_real shall contain the converted number upon return from this routine.

\subsection*{25.30 tf_longtime_tostr()}


The TF routine tf_longtime_tostr() shall convert a 64-bit integer time value to a character string. The time value shall be unsigned.

\subsection*{25.31 tf_message()}


The TF routine tf_message() shall display warning or error message information using the warning and error message format for a software product. The location information (file name and line number) of the current instance of the user-defined system task or function shall be appended to the message using a format compatible with warning and error messages generated by the software product, and the message shall be written to both the output channel of the software product which invoked the PLI application and the output \(\log\) file of the product.

The level field shall indicate the severity level of the error, specified as a predefined constant. There shall be five levels: \(E R R \_E R R O R, E R R \_S Y S T E M, E R R \_I N T E R N A L, E R R \_M E S S A G E\), and \(E R R \_W A R N I N G\). If tf_message() is called by the checktf application associated with the user-defined system task or function, the following rules shall apply:

If the checktf application is called when the Verilog HDL source code was being parsed or compiled, and the level is \(E R R \_E R R O R, E R R \_S Y S T E M\), or \(E R R \_I N T E R N A L\), then parsing or compilation shall be aborted after an error message is reported.
If the checktf application is called when the Verilog HDL source code was being parsed or compiled, and the level is \(E R R \_W A R N I N G\) or \(E R R \_M E S S A G E\), then parsing or compilation shall continue after a warning message is reported.
If the checktf application is called when the user-defined task or function was invoked on the interactive command line, the interactive command shall be aborted after a warning message or error message is reported.

The facility and code fields shall be string arguments that can be used in the Verilog software product message syntax. These strings shall be less than 10 characters in length.

The message argument shall be a user-defined control string containing the message to be displayed. The control string shall use the same formatting controls as the C printf () function (for example, \%d). The message shall use up to a maximum of five variable arguments. There shall be no limit to the length of a variable argument. Formatting characters, such as \(\backslash n, \backslash t, \backslash b, \backslash f\), or \(\backslash r\), do not need to be included in the message the software product shall automatically format each message.

An example of a tf_message() call and the output generated are shown below. Note that the format of the output shall be defined by the software product.

Calling tf_message() with the arguments:
```

tf_message(ERR_ERROR, "User", "TFARG",
"Argument number %d is illegal in task %s", argnum,
taskname);

```

Might produce the output:
```

ERROR! Argument number 2 is illegal in task [ User-TFARG]
\$usertask

```

The routine tf_message() provides more control over the format and severity of error or warning messages than the routines \(\mathbf{t f}\) error() and \(\mathbf{t f}\) _warning() can provide. In addition, the routine \(\mathbf{t f}\) _message () can be used in conjunction with \(\mathbf{t f}\) _text () , which shall allow an error or warning message to be stored while a PLI application executes additional code before the message is printed and parsing or compilation of Verilog HDL source possibly aborted.

\subsection*{25.32 tf_mipname(), tf_imipname()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_mipname(), tf_imipname()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the hierarchical module instance path name as a string.} \\
\hline Syntax: & \multicolumn{3}{|l|}{```
tf_mipname()
tf_imipname(instance_p)
```} \\
\hline \multicolumn{3}{|c|}{Type} & Description \\
\hline Returns: & PLI_BYTE8 * & \multicolumn{2}{|l|}{Pointer to a string containing the hierarchical path name} \\
\hline \multirow[b]{2}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{\begin{tabular}{l}
Use tf_spname() or tf_ispname() to get the scope path name \\
Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function
\end{tabular}} \\
\hline
\end{tabular}

The TF routine tf_mipname() shall return the Verilog HDL hierarchical path name to the module instance containing the call to the current instance or a specific instance of a user-defined system task or function.

The string obtained shall be stored in a temporary buffer. If the string is needed across multiple calls to the PLI application, the string should be preserved.

\subsection*{25.33 tf_movepvc_flag(), tf_imovepvc_flag()}


The TF routines tf_movepvc_flag() and tf_imovepvc_flag() shall move the current pve flag to the saved pve flag and clear the current flag for the current instance or a specific instance of a user-defined system task or function. The routine shall return the value of the flag that was moved.

The argument narg shall be the index number of an argument in a specific instance of a user-defined system task or function. Task/function argument index numbering shall proceed from left to right, with the left-most argument being number 1. If narg is \(\mathbf{- 1}\), then all argument pve flags shall be moved and the logical OR of all saved flags returned.

PVC flags shall be used to indicate whether a particular user-defined system task or function argument has changed value. Each argument shall have two pve flags: a current pve flag, which shall be set by a software product when the change occurs, and a saved pve flag, which shall be controlled by the user.

NOTE PVC flags shall not be set by the software product until tf_asynchon() or tf_iasynchon() has been called.

\subsection*{25.34 tf_multiply_long()}


The TF routine tf_multiply_long() shall multiply two 64-bit values. After calling tf_multiply_long(), the variables used to pass the first operand shall contain the results of the multiplication. Figure 165 shows the high and low 32 bits of two 64-bit integers and how tf_multiply_long() shall multiply them.


Figure 165-Multiplying with tf_multiply_long()

\subsection*{25.35 tf_nodeinfo(), tf_inodeinfo()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_nodeinfo(), tf_inodeinfo()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get system task/function argument node information.} \\
\hline Syntax: & \multicolumn{3}{|l|}{\begin{tabular}{l}
tf_nodeinfo(narg, nodeinfo_p) \\
tf_inodeinfo(narg, nodeinfo_p, instance_p)
\end{tabular}} \\
\hline & Type & \multicolumn{2}{|r|}{Description} \\
\hline Returns: & struct t_tfnodeinfo * & \multicolumn{2}{|l|}{The value of the second argument if successful; 0 if an error occurred} \\
\hline \multicolumn{2}{|r|}{Type} & Name & Description \\
\hline \multirow[t]{3}{*}{Arguments:} & PLI_INT32 & narg & Index number of the user-defined system task or function argument \\
\hline & struct t_tfnodeinfo * & nodeinfo_p & Pointer to a variable declared as the t_tfnodeinfo structure type \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_exprinfo() or tf_iexprinfo() for general information on arguments Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routines \(\mathbf{t f}\) _nodeinfo() and \(\mathbf{t f}\) _inodeinfo() shall obtain information about the specified argument in the current instance or a specific instance of a user-defined system task or function.

The information shall be stored in the \(C\) structure \(s\) _tfnodeinfo as defined in the file veriuser. \(h\). The routine shall only be called for arguments that are of the types described in Table 192. Memory space shall first be allocated to hold the information before calling \(\mathbf{t f}\) _nodeinfo() or tf_inodeinfo(). For example:
```

    {
    s tfnodeinfo info; /* declare a variable of the structure type */
    t\overline{f_nodeinfo(n, &info); /* pass tf_nodeinfo a pointer to the variable}
    * /
}

```

The routines shall return the second argument, which is the pointer to the information structure. If narg is out of range, or if some other error is found, then \(\mathbf{0}\) shall be returned.

The argument narg shall be the index number of an argument in a user-defined system task or function. Task/function argument index numbering shall proceed from left to right, with the left-most argument being number 1.

The tf_nodeinfo() and tf_inodeinfo() routines shall support at least the following Verilog data types as a system task or system function argument:
scalar and vector regs
scalar and vector nets
integer, time and real variables
word select of a one-dimensional reg, integer or time array
null argument
The s_tfnodeinfo structure is defined in veriuser. h and is listed in Figure 166.
```

typedef struct t_tfnodeinfo
{
PLI_INT16 node_type;
PLI_INT16 padding;
union
{
struct t_vecval *vecval_p;
struct t_strengthval *strengthval_p;
PLI_BYTE\overline{8}}\mathrm{ *memoryval_p;
double *real_val_p;
} node_value;
PLI_BYTE8 * node_symbol;
PLI_INT32 node_ngroups;
PLI_INT32 node_vec_size;
PLI_INT32 node_sign;
PLI_INT32 node_ms_index;
PLI_INT32 node_ls_index;
PLI_INT32 node_mem_size;
PLI_INT32 node_lhs_element;
PLI_INT32 node_rhs_element;
PLI INT32 * node handle;
s_tfnodeinfo, *\overline{p}_tfnodeinfo;

```

Figure 166-The s_tfnodeinfo structure definition
The following paragraphs define the fields of the s_tfnodeinfo structure.
The node_type field of the s tfnodeinfo structure shall indicate the Verilog HDL data type of the argument, and is one of the predefined constants as given in Table 192 and defined in veriuser. h .

Table 192-Predefined constants for node_type
\begin{tabular}{|c|l|}
\hline Predefined constant & \multicolumn{1}{|c|}{ Description } \\
\hline tf_null_node & Not a writable argument \\
\hline tf_reg_node & Argument references a reg variable \\
\hline tf_integer_node & Argument references an integer variable \\
\hline tf_real_node & Argument references a real variable \\
\hline tf_time_node & Argument references a time variable \\
\hline tf_netvector_node & Argument references a vector net \\
\hline tf_netscalar_node & Argument references a scalar net \\
\hline tf_memory_node & Argument references a memory \\
\hline
\end{tabular}

The node_value field of the s_tfnodeinfo structure shall be a union of pointers to value structures defining the current value on the node referenced by the argument. The union member accessed shall depend on the node_type. The union members are given in Table 193.

Table 193-How the node_value union is used
\begin{tabular}{|c|c|}
\hline When the node_type is & The union member used is \\
\hline \begin{tabular}{c} 
tf_reg_node, tf_integer_node, tf_time_node, \\
or tf_netvector_node
\end{tabular} & vecval_p \\
\hline tf_real_node & real_val_p \\
\hline tf_netscalar_node & strengthval_p \\
\hline tf_memory_node & memoryval_p \\
\hline
\end{tabular}

If the node_type is \(\mathbf{t f}\) _reg_node, \(\mathbf{t f}\) _integer_node, \(\mathbf{t f}\) _time_node, or \(\mathbf{t f}\) _netvector_node, then node_value shall be a pointer to an array of s_vecval structures that gives the resultant value of the node. The s_vecval structure for representing vector values is defined in veriuser. h and is listed in Figure 167.
```

typedef struct t vecval
{
PLI INT32 avalbits;
PLI_INT32 bvalbits;
} s_vecval, *p_vecval;

```

Figure 167-The s_vecval structure definition
If the number of bits in the vector (defined by the node_vec_size field of the s_tfnodeinfo structure) is less than or equal to 32 , then there shall only be one s_vecval group in the node_value.vecval_p array. For 33 bits to 64 bits, two groups shall be in the array, and so on. The number of groups shall also be given by the value of node_ngroups. The fields for avalbits and bvalbits of the s_vecval structure shall hold the bit patterns making up the value of the argument. The lsb in the value shall be represented by the lsb s in the avalbits and bvalbits components, and so on. The bit coding shall be as given in Table 194.

Table 194-avalbits/bvalbits encoding
\begin{tabular}{|c|c|}
\hline aval / bval & Logic value \\
\hline 00 & 0 \\
\hline 10 & 1 \\
\hline 01 & High impedance \\
\hline 11 & Unknown \\
\hline
\end{tabular}

If the node_type field of the s_tfnodeinfo structure is tf_netscalar_node, then the node_value.strengthval_p field of the s_tfnodeinfo structure shall point to an s_strengthval structure of the form given in Figure 168.
```

typedef struct t_strengthval
{
PLI INT32 strength0;
PLI_INT32 strength1;
s_strengthval, *p_strengthval;

```

Figure 168-The s_strengthval structure definition

In the s strengthval structure, strength0 shall give the 0 -strength bit pattern for the value, and strength1 shall give the 1 -strength bit pattern. Refer to 7.10 for details about these bit patterns.

If the node_type field of the s_tfnodeinfo structure is tf_memory_node, then node_value.memoryval_p shall point to a memval structure giving the total contents of the memory. The structure is organized as shown in Figure 169.
```

struct
{
PLI_BYTE8 avalbits[ node ngroups] ;
PLI_BYTE8 bvalbits[ node_ngroups] ;
memval[ node_mem_size] ;

```

Figure 169-The memval structure definition
Note that a pointer to the memval structure data structure cannot be represented in C , so the node_value.memoryval_p field of the s_tfnodeinfo structure is declared as a pointer to a PLI_BYTE8 type. The memory element with the lowest number address in the Verilog array declaration shall be located in the first group of bytes, which is the byte group represented by memval[ 0\(]\).

The node_symbol field of the s_tfnodeinfo structure shall be a string pointer to the identifier of the argument.

If the node_type field of the \(s\) tfnodeinfo structure is \(\mathbf{t f}\) _reg_node, \(\mathbf{t f}\) _integer_node, \(\mathbf{t f}\) _time_node, or tf_netvector_node, then the \(\overline{n o d e \_n g r o u p s ~ f i e l d ~ o f ~ t h e ~ s ~ t f n o d e i n f o ~ s t r u c t u r e ~ s h a l l ~ i n d i c a t e ~ t h e ~ n u m-~}\) ber of groups for the argument nodevalue and shall determine the array size of the node_value.vecval_p value structure. If the node_type is tf_real_node, then node_ngroups shall be 0 .

If the node_type field of the s tfnodeinfo structure is tf_reg_node, tf_integer_node, tf_time_node, or tf_netvector_node, then the \(\overline{n o d e} \_v e c \_\)size field of the s_tfnodeinfo structure shall indicate the total number of bits in the array of the node_value.vecval_p structure. If node_type is tf_real_node, then node_vec_size shall be 0 .

The node_sign field of the \(s \_t f n o d e i n f o\) structure shall indicate the sign type of the node as follows: 0 for unsigned, nonzero for signed.

If the node_type is tf_memory_node, then node_mem_size shall indicate the number of elements in the node_value.memoryval_p structure.

If the node_type field of the s tfnodeinfo structure is tf_reg_node or tf_netvector_node, then the node_value.node_ms_element and node_value.node_ls_element fields shall contain the msb and lsb of the given vector.

If the node_type field of the s tfnodeinfo structure is tf_reg_node or tf_netvector_node, and the argument is a part-select, then the node_value.node_rhs_index and node_value.node_lhs_index fields shall contain the msb and lsb of the given part-select.

The field node_handle is not used.

\subsection*{25.36 tf_nump(), tf_inump()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_nump(), tf_inump()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get number of task or function arguments.} \\
\hline Syntax: & \multicolumn{3}{|l|}{```
tf_nump()
tf_inump(instance_p)
```} \\
\hline \multicolumn{3}{|c|}{Type} & Description \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{The number of arguments} \\
\hline \multirow[b]{2}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routines \(\mathbf{t f}\) _nump() and \(\mathbf{t f}\) _inump() shall return the number of task/function arguments specified in the current instance or a specific instance of a user-defined task or function statement in the Verilog source description. The number returned shall be greater than or equal to zero.

Note: null arguments are counted. Therefore, \(\$\) foo() returns a count of 1 and \(\$\) foo(,) returns a count of 2 . The routine tf_typep() returns a type of tf_nullparam for a null argument.

\subsection*{25.37 tf_propagatep(), tf_ipropagatep()}


The TF routines tf_propagatep() and tf_ipropagatep() shall write a value to an argument node of the current instance or a specific instance of a user-defined system task or function, and then propagate the value to any loads that read the value of the node.

In order to write values back into a Verilog software product data structure using tf_propagatep() and tf_ipropagatep(), the value shall first be placed into the value structure pointed to by the component expr_value_p as allocated by calling tf_exprinfo() or tf_iexprinfo(). The structure for tf_exprinfo() and \(\mathbf{t f}\) _iexprinfo () shall be used for all argument types except memories.

\subsection*{25.38 tf_putlongp(), tf_iputlongp()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_putlongp(), tf_iputlongp()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Write a 64-bit integer value to a system task/function argument or function return.} \\
\hline Syntax: & \multicolumn{3}{|l|}{\begin{tabular}{l}
tf_putlongp(narg, lowvalue, highvalue) \\
tf_iputlongp(narg, lowvalue, highvalue, instance_p)
\end{tabular}} \\
\hline \multicolumn{3}{|c|}{Type} & Description \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{0 if successful; 1 if an error occurred} \\
\hline \multicolumn{2}{|r|}{Type} & Name & Description \\
\hline \multirow[t]{4}{*}{Arguments:} & PLI_INT32 & narg & Index number of the user-defined system task or function argument or 0 to return a function value \\
\hline & PLI_INT32 & lowvalue & Least significant (right-most) 32 bits of value \\
\hline & PLI_INT32 & highvalue & Most significant (left-most) 32 bits of value \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf _putp() or tf _iputp() to put an argument value as a 32 -bit integer Use tf_putrealp() or tf_iputrealp() to put an argument value as a double Use tf _strdelputp() to put a value as a formatted string with delay Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routines \(\mathbf{t f}\) _putlongp() and tf_iputlongp() shall write a 64 -bit integer value to the argument specified by narg of the current instance or a specific instance of a user-defined system task or function. If narg is 0 , \(\mathbf{t} \mathbf{f} \mathbf{p u t l o n g p}()\) and \(\mathbf{t f}\) iputlongp() shall write the value as the return of a user-defined system function. If narg is out of range or the argument cannot be written to, then the routines shall do nothing. Should the calltf routine for a user defined system function fail to put a value during its execution, the default value of 0 shall be applied.

The argument narg shall be the index number of an argument in a user-defined system task or function. Task/function argument index numbering shall proceed from left to right, with the left-most argument being number 1.

The data type of the values to be written should be consistent with the type of put routine and the type of the argument to which the value shall be written. Refer to 24.3 for more details on proper data type selection with put routines.

NOTE calling put routines to TF argument 0 (return of a function) shall only return a value in a calltf application, when the call to the function is active. The action of the put routine shall be ignored when the function is not active.

\subsection*{25.39 tf_putp(), tf_iputp()}


The TF routine \(\mathbf{t f} \_\mathbf{p u t p}()\) and \(\mathbf{t f}\) _iputp() shall write an integer value to the argument specified by narg of the current instance or a specific instance of a user-defined system task or function. If narg is 0 , \(\mathbf{t f}\) _putp() or tf_iputp() shall write the value as the return of a user-defined system function. If narg is out of range or the argument cannot be written to, then the routines shall do nothing. Should the calltf routine for a user defined system function fail to put a value during its execution, the default value of 0 shall be applied.

The argument narg shall be the index number of an argument in a user-defined system task or function. Task/function argument index numbering shall proceed from left to right, with the left-most argument being number 1.

The data type of the value to be written should be consistent with the type of put routine and the type of the argument to which the value shall be written. Refer to Section 24.3 for more details on proper data type selection with put routines.

NOTE Calling put routines to TF argument 0 (return of a function) shall only return a value in a calltf application, when the call to the function is active. The action of the put routine shall be ignored when the function is not active.

\subsection*{25.40 tf_putrealp(), tf_iputrealp()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_putrealp(), tf_iputrealp()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Write a real value to a system task/function argument or function return.} \\
\hline Syntax: & \multicolumn{3}{|l|}{```
tf_putrealp(narg, value)
tf_iputrealp(narg, value, instance_p)
```} \\
\hline \multicolumn{3}{|c|}{Type} & Description \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{0 if successful; 1 if an error occurred} \\
\hline \multirow{4}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & PLI_INT32 & narg & Index number of the user-defined system task or function argument or 0 to return a function value \\
\hline & double & value & A double-precision value \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_putp() or tf_iputp() to put an argument value as a 32-bit integer Use tf_putlongp() or tf_iputlongp() to put an argument value as a 64-bit integer Use tf_strdelputp() to put a value as a formatted string with delay Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routines tf_putrealp() and tf_iputrealp() shall write a double-precision real value to the argument specified by narg of the current instance or a specific instance of a user-defined system task or function. If narg is 0 , \(\mathbf{t f} \_\)putrealp() and \(\mathbf{t f}\) _iputrealp() shall write the value as the return of a user-defined system function. If narg is out of range or the argument cannot be written to, then the routines shall do nothing. Should the calltf routine for a user defined system function fail to put a value during its execution, the default value of 0.0 shall be applied.

The argument narg shall be the index number of an argument in a user-defined system task or function. Task/function argument index numbering shall proceed from left to right, with the left-most argument being number 1.

The data type of the value to be written should be consistent with the type of put routine and the type of the argument to which the value shall be written. Refer to 24.3 for more details on proper data type selection with put routines.

NOTE calling put routines to TF argument 0 (return of a function) shall only return a value in a calltf application, when the call to the function is active. The action of the put routine shall be ignored when the function is not active.

\subsection*{25.41 tf_read_restart()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_read_restart()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get a block of data from a previously written save file.} \\
\hline Syntax: & \multicolumn{3}{|l|}{tf_read_restart(blockptr, blocklen)} \\
\hline & \multicolumn{2}{|l|}{Type} & Description \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{Nonzero if successful; zero if an error occurred} \\
\hline \multirow{3}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_BYTE8 * & blockptr & Pointer to block of saved data \\
\hline & PLI_INT32 & blocklen & Length of block \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_write_save() to save a block of data} \\
\hline
\end{tabular}

The TF routine tf_read_restart() shall read back a block of memory that was saved with tf_write_save(). This routine shall only be called from the misctf application when the misctf routine is invoked with reason_restart.

The argument blockptr shall be a pointer to an allocated block of memory to which the saved data shall be restored.

The argument blocklen shall be the length in bytes of the allocated block of memory. Exactly as many bytes have to be restored as were written with tf_write_save().

If any user task instance pointers have been saved (for use with tf_i* calls), tf_getinstance() has to be used to get new instance pointer values after the restart. If pointers to user data were saved, the application of the user has to implement a scheme to reconnect them properly.

\subsection*{25.42 tf_real_to_long()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_real_to_long()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Convert a real number to a 64-bit integer.} \\
\hline Syntax: & \multicolumn{3}{|l|}{tf_real_to_long(realvalue, aof_low, aof_high)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type} & Description \\
\hline & void & & \\
\hline \multirow{4}{*}{Arguments:} & Type & Name & Description \\
\hline & double & realvalue & Value to be converted \\
\hline & PLI_INT32 * & aof_low & Pointer to an integer variable for storing the least significant (right-most) 32 bits of the converted value \\
\hline & PLI_INT32 * & aof_high & Pointer to an integer variable for storing the most significant (left-most) 32 bits of the converted value \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_long_to_real() to convert a 64-bit integer to a real number} \\
\hline
\end{tabular}

The TF routine tf_real_to_long() shall convert a double-precision floating-point number to a 64-bit integer. The converted value shall be returned in the variables pointed to by aof_low and aof_high.

\subsection*{25.43 tf_rosynchronize(), tf_irosynchronize()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|r|}{tf_rosynchronize(), tf_irosynchronize()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Synchronize to end of simulation time step.} \\
\hline Syntax: & \multicolumn{3}{|l|}{```
tf_rosynchronize()
tf_irosynchronize(instance_p)
```} \\
\hline \multirow[b]{2}{*}{Returns:} & Type & & Description \\
\hline & PLI_INT32 & cessful; 1 if & ccurred \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific or function \\
\hline Related routines: & \multicolumn{3}{|l|}{\begin{tabular}{l}
Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function Use tf_synchronize() to synchronize to end of simulation time step \\
Use tf_getnextlongtime() to get next time at which a simulation event is scheduled
\end{tabular}} \\
\hline
\end{tabular}

The TF routines tf_rosynchronize() and tf_irosynchronize() shall schedule a callback to the misctf application associated with the current instance or a specific instance of a user-defined system task or function. The misctf application shall be called with a reason of reason_rosynch at the end of the current simulation time step.

The routines tf_synchronize() and tf_rosynchronize() have different functionality. The routine tf_synchronize() shall call the associated misctf application at the end of the current simulation time step with reason_synch, and the misctf application shall be allowed to schedule additional simulation events using routines such as tf_strdelputp().

The routine tf_rosynchronize() shall call the associated misctf application at the end of the current simulation time step with reason_rosynch, and the PLI shall not be allowed to schedule any new events. This guarantees that all simulation events for the current time are completed. Calls to routines such as tf_strdelputp() and tf_setdelay() are illegal during processing of the misctf application with reason reason_rosynch.

The routine tf_getnextlongtime() shall only return the next simulation time for which an event is scheduled when used in conjunction with the routines \(\mathbf{t f}\) _rosynchronize() and tf_irosynchronize().

\subsection*{25.44 tf_scale_longdelay()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_scale_longdelay()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Convert a 64-bit integer delay to the timescale of the module instance.} \\
\hline Syntax: & \multicolumn{3}{|l|}{tf_scale_longdelay(instance_p, delay_lo, delay_hi, aof_delay_lo, aof_delay_hi)} \\
\hline \multicolumn{3}{|c|}{Type} & Description \\
\hline Returns: & void & & \\
\hline \multirow{6}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline & PLI_INT32 & delay_lo & Least significant (right-most) 32 bits of the delay to be converted \\
\hline & PLI_INT32 & delay_hi & Most significant (left-most) 32 bits of the delay to be converted \\
\hline & PLI_INT32 * & aof_delay_lo & Pointer to a variable to store the least significant (rightmost) 32 bits of the conversion result \\
\hline & PLI_INT32 * & aof_delay_hi & Pointer to a variable to store the most significant (leftmost) 32 bits of the conversion result \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_scale_realdelay() to scale real number delays Use tf_unscale_longdelay() to convert a delay to the time unit of a module Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routine tf_scale_longdelay() shall convert a 64-bit integer delay into the timescale of the module containing the instance of the user-defined system task or function pointed to by instance_p. The arguments \(a o f \_d e l a y \_l o\) and aof_delay_hi shall contain the address of the converted delay returned by the routine.

\subsection*{25.45 tf_scale_realdelay()}


The TF routine tf_scale_realdelay() shall convert a double-precision floating-point delay into the timescale of the module containing the instance of the user-defined system task or function pointed to by instance_p. The argument aof_realdelay shall contain the address of the converted delay returned by the routine.

\subsection*{25.46 tf_setdelay(), tf_isetdelay()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_setdelay(), tf_isetdelay()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Activate the misctf application at a particular simulation time.} \\
\hline Syntax: & \multicolumn{3}{|l|}{```
tf_setdelay(delay)
tf_isetdelay(delay, instance_p)
```} \\
\hline & \multicolumn{2}{|l|}{Type} & Description \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{1 if successful; 0 if an error occurred} \\
\hline \multirow{3}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_INT32 & delay & 32-bit integer delay time \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_setlongdelay() or tf_isetlongdelay() for 64-bit integer reactivation delays Use tf_setrealdelay() or tf_isetrealdelay() for real number reactivation delays Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routines \(\mathbf{t f}\) _setdelay() and tf_isetdelay() shall schedule a callback to the misctf application associated with the current instance or a specific instance of a user-defined system task or function. The misctf application shall be called at a future reactivation time. The reactivation time shall be the current simulation time plus the specified delay. The misctf application shall be called at the reactivation time with a reason of reason_reactivate. The tf_setdelay() and tf_isetdelay() routines can be called several times with different delays, and several reactivations shall be scheduled. Multiple calls to tf_setdelay() and tf_isetdelay() for the same time step are permitted and shall result in multiple calls to the misctf application for that time step.

The delay argument shall be a 32-bit integer and shall be greater than or equal to 0 . The delay shall assume the timescale units specified for the module containing the specific system task call.

\subsection*{25.47 tf_setlongdelay(), tf_isetlongdelay()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_setlongdelay(), tf_isetlongdelay()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Activate the misctf application at a particular simulation time.} \\
\hline Syntax: & \multicolumn{3}{|l|}{```
tf_setlongdelay(lowdelay, highdelay)
tf_isetlongdelay(lowdelay, highdelay, instance_p)
```} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type} & Description \\
\hline & PLI_INT32 & \multicolumn{2}{|l|}{1 if successful; 0 if an error occurred} \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_INT32 & lowdelay & Least significant (right-most) 32 bits of the delay time to reactivation \\
\hline & PLI_INT32 & highdelay & Most significant (left-most) 32 bits of the delay time to reactivation \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{\begin{tabular}{l}
Use tf_setdelay() or tf_isetdelay() for 32-bit integer reactivation delays \\
Use tf_setrealdelay() or tf_isetrealdelay() for real number reactivation delays \\
Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function
\end{tabular}} \\
\hline
\end{tabular}

The TF routines tf_setlongdelay() and tf_isetlongdelay() shall schedule a callback to the misctf application associated with the current instance or a specific instance of a user-defined system task or function. The misctf application shall be called at a future reactivation time. The reactivation time shall be the current simulation time plus the specified delay. The misctf routine shall be called at the reactivation time with a reason of reason_reactivate. The tf_setlongdelay() and tf_isetlongdelay() routines can be called several times with different delays, and several reactivations shall be scheduled. Multiple calls to tf_setlongdelay() and tf_isetlongdelay() for the same time step are permitted and shall result in multiple calls to the misctf application for that time step.

The delay argument shall be a 64-bit integer and shall be greater than or equal to 0 . The delay shall assume the timescale units specified for the module containing the specific system task call.

\subsection*{25.48 tf_setrealdelay(), tf_isetrealdelay()}


The TF routines \(\mathbf{t f}\) _setrealdelay() and tf_isetrealdelay() shall schedule a callback to the misctf application associated with the current instance or a specific instance of a user-defined system task or function. The misctf application shall be called at a future reactivation time. The reactivation time shall be the current simulation time plus the specified delay. The misctf application shall be called at the reactivation time with a reason of reason_reactivate. The tf_setrealdelay() and tf_isetrealdelay() routines can be called several times with different delays, and several reactivations shall be scheduled. Multiple calls to tf_setrealdelay() and tf_isetrealdelay() for the same time step are permitted and shall result in multiple calls to the misctf application for that time step.

The delay argument shall be a double-precision value and shall be greater than or equal to 0.0 . The delay shall assume the timescale units specified for the module containing the specific system task call.

\subsection*{25.49 tf_setworkarea(), tf_isetworkarea()}

\section*{tf_setworkarea(), tf_isetworkarea()}

Synopsis: \(\quad\) Store user data pointer in work area.
```

Syntax: tf_setworkarea(workarea)
tf_isetworkarea(workarea, instance_p)

```

\section*{Type}

\section*{Description}

Returns: PLI_INT32 Always returns 0

The TF routines tf_setworkarea() and tf_isetworkarea() shall store a pointer to user data in the work area of the current instance or a specific instance of a user-defined system task or function. The pointer that is stored can be retrieved by calling tf_getworkarea() or tf_igetworkarea().

The work area can be used for
Saving information during one call to a PLI routine, which can be retrieved upon a subsequent invocation of the routine
Passing information from one type of PLI application to another, such as from a checktf application to a calltf application

Note that the workarea pointer is a PLI_BYTE8 * type. If the memory allocated for the user data is of some other type, it should be cast to PLI_BYTE8 *.
25.50 tf_sizep(), tf_isizep()
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_sizep(), tf_isizep()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the bit length of a system task/function argument.} \\
\hline Syntax: & \multicolumn{3}{|l|}{```
tf_sizep(narg)
tf_isizep(narg, instance_p)
```} \\
\hline & \multicolumn{2}{|l|}{Type} & Description \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{The number of bits of the system task/function argument} \\
\hline \multirow{3}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_INT32 & narg & Index number of the user-defined system task or function argument \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routines \(\mathbf{t f}\) _sizep() and \(\mathbf{t f}\) _isizep() shall return the value size in bits of the specified argument in the current instance or a specific instance of a user-defined system task or function.

If the specified argument is a literal string, \(\mathbf{t f}\) _sizep() and \(\mathbf{t f}\) _isizep() shall return the string length.
If the specified argument is real or if an error is detected, \(\mathbf{t f}\) _sizep() and \(\mathbf{t f}\) _isizep() shall return 0 .
The argument narg shall be the index number of an ARGUMENT in a user-defined system task or function. Task/function argument index numbering shall proceed from left to right, with the left-most argument being number 1.

\subsection*{25.51 tf_spname(), tf_ispname()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_spname(), tf_ispname()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get scope hierarchical path name as a string.} \\
\hline Syntax: & \multicolumn{3}{|l|}{```
tf_spname()
tf_ispname(instance_p)
```} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type} & Description \\
\hline & PLI_BYTE8 * & \multicolumn{2}{|l|}{Pointer to a character string with the hierarchical path name} \\
\hline \multirow[b]{2}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routines \(\mathbf{t f}\) _spname() and tf_ispname() shall return a pointer to the Verilog HDL hierarchical path name to the scope containing the call of a specific instance of a user-defined system task or function.

A scope shall be
A top-level module
A module instance
A named begin-end block
A named fork-join block
A Verilog HDL task
A Verilog HDL function

The string obtained shall be stored in a temporary buffer. If the string is needed across multiple calls to the PLI application, the string should be preserved.
25.52 tf_strdelputp(), tf_istrdelputp()


The TF routines tf_strdelputp() and \(\mathbf{t f}\) _istrdelputp() shall write a string value to the specified argument of the current instance or a specific instance of a user-defined system task or function. The actual change to the argument shall be scheduled as an event on the argument in the Verilog model at a future simulation time. An argument value of 0 (system function return) shall be illegal.

The bitlength argument shall define the value size in bits.
The format shall define the format of the value specified by value_p and shall be one of the characters given in Table 195.

\section*{Table 195—Format characters}
\begin{tabular}{cc} 
Format character & Description \\
\(\prime \mathrm{b}^{\prime}\) or \({ }^{\prime} \mathrm{B}^{\prime}\) & Value is in binary \\
\(\prime \mathrm{o}^{\prime}\) or \({ }^{\prime} \mathrm{O}^{\prime}\) & Value is in octal \\
\(\prime \mathrm{d}^{\prime}\) or \({ }^{\prime} \mathrm{D}^{\prime}\) & Value is in decimal \\
\(\prime \mathrm{h}^{\prime}\) or \({ }^{\prime} \mathrm{H}^{\prime}\) & Value is in hexadecimal
\end{tabular}

The delay argument shall represent the amount of time before the value shall be applied to the argument, and it shall be greater than or equal to 0 . The delay shall assume the timescale units of the module containing the instance of the user-defined system task or function.

The delaytype argument shall determine how the value shall be scheduled in relation to other simulation events on the same reg or variable. The delaytype shall be one of integer values shown in Table 196.

\subsection*{25.53 tf_strgetp(), tf_istrgetp()}


The TF routines \(\mathbf{t f}\) _strgetp() and \(\mathbf{t f}\) _istrgetp() shall return a pointer to a string that contains the value of the argument expression of the current instance or a specific instance of a user-defined system task or function.

The string format is specified by format, and shall be one of the following characters shown in Table 197.
Table 197-Format characters
\begin{tabular}{|c|c|}
\hline Format character & Description \\
\hline\({ }^{\prime} \mathrm{b}^{\prime}\) or \({ }^{\prime} \mathrm{B}^{\prime}\) & Value is in binary \\
\hline\(' \mathrm{o}^{\prime}\) or \(\mathrm{O}^{\prime}\) & Value is in octal \\
\hline\(' \mathrm{~d}^{\prime}\) or \(\mathrm{D}^{\prime}\) & Value is in decimal \\
\hline\({ }^{\prime} \mathrm{h}^{\prime}\) or \(\mathrm{H}^{\prime}\) & Value is in hexadecimal \\
\hline
\end{tabular}

The string value returned shall have the same form as output from the formatted built-in system task \$display() in terms of value lengths and value characters used. The length shall be of arbitrary size (not limited to 32 bits as with the \(\mathbf{t f} \mathbf{g e t p}()\) routine), and unknown and high-impedance values shall be obtained.

The referenced argument can be a string, in which case a pointer to the string shall be returned (the format shall be ignored in this case). The string obtained shall be stored in a temporary buffer. If the string is needed across multiple calls to the PLI application, the string should be preserved.

A null pointer shall be returned for errors.

\subsection*{25.54 tf_strgettime()}
\begin{tabular}{|c|c|c|c|c|}
\hline \multicolumn{5}{|c|}{tf_strgettime()} \\
\hline Synopsis: & \multicolumn{4}{|l|}{Get the current simulation time as a string.} \\
\hline Syntax: & \multicolumn{4}{|l|}{tf_strgettime()} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type} & \multicolumn{2}{|l|}{Description} \\
\hline & PLI_BYTE8* & \multicolumn{3}{|l|}{Pointer to a character string with the simulation time} \\
\hline \multirow[b]{2}{*}{Arguments:} & \multirow[t]{2}{*}{Type} & Name & & Description \\
\hline & & & No arguments & \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_gettime() to get simulation time as a 32 -bit integer value Use tf_getlongtime() to get simulation time as a 64-bit integer value Use tf_getrealtime() to get simulation time as a real value} & \\
\hline
\end{tabular}

The TF routine tf_strgettime() shall return a pointer to a string, which shall be the ASCII representation of the current simulation time. The string obtained shall be stored in a temporary buffer. If the string is needed across multiple calls to the PLI application, the string should be preserved.

Time shall be expressed in simulation time units, which is the smallest time precision used by all modules in a design.

\subsection*{25.55 tf_strlongdelputp(), tf_istrlongdelputp()}


The TF routines \(\mathbf{t f}\) _strlongdelputp() and \(\mathbf{t f}\) _istrlongdelputp() shall write a string value to the specified argument of the current instance or a specific instance of a user-defined system task or function. The actual change to the argument shall be scheduled as an event on the argument in the Verilog model at a future simulation time. An argument value of 0 (system function return) shall be illegal.

The bitlength argument shall define the value size in bits.
The format shall define the format of the value specified by value_p and shall be one of the characters shown in Table 198.

Table 198-Format characters
\begin{tabular}{|c|c|}
\hline Format character & Description \\
\hline\(' \mathrm{~b}^{\prime}\) or \({ }^{\prime} \mathrm{B}^{\prime}\) & Value is in binary \\
\hline\(' \mathrm{o}^{\prime}\) or \(\mathrm{O}^{\prime}\) & Value is in octal \\
\hline\(' \mathrm{~d}^{\prime}\) or \({ }^{\prime} \mathrm{D}^{\prime}\) & Value is in decimal \\
\hline\(' \mathrm{~h}^{\prime}\) or \(\mathrm{H}^{\prime}\) & Value is in hexadecimal \\
\hline
\end{tabular}

The delay argument shall represent the amount of time before the value shall be applied to the argument, and it shall be greater than or equal to 0 . The delay shall assume the timescale units of the module containing the instance of the user-defined system task or function.

The delaytype argument shall determine how the value shall be scheduled in relation to other simulation events on the same reg or variable. The delaytype shall be one of integer values shown in Table 199.

Table 199-delaytype codes
\begin{tabular}{|c|c|l|}
\hline delaytype code & Definition & \multicolumn{1}{c|}{ Description } \\
\hline \(\mathbf{0}\) & Inertial delay & \begin{tabular}{l} 
All scheduled events on the output argument in the Ver- \\
ilog model are removed before scheduling a new event
\end{tabular} \\
\hline \(\mathbf{1}\) & Pure transport delay & \begin{tabular}{l} 
No scheduled events on the output argument in the Ver- \\
ilog model are removed before scheduling a new \\
event the last event to be scheduled is not necessarily \\
the last one to occur
\end{tabular} \\
\hline \(\mathbf{2}\) & \begin{tabular}{l} 
All events that are scheduled for times later than the new \\
event on the output argument in the Verilog model are \\
removed before scheduling a new event
\end{tabular} \\
\hline
\end{tabular}

\subsection*{25.56 tf_strrealdelputp(), tf_istrrealdelputp()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_strrealdelputp(), tf_istrrealdelputp()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Write a value to a system task/function argument from string value specification, using a real number delay.} \\
\hline Syntax: & \multicolumn{3}{|l|}{```
tf_strrealdelputp(narg, bitlength, format, value_p, realdelay,
    delaytype)
tf_istrrealdelputp(narg, bitlength, format, value_p, realdelay,
    delaytype, instance_p)
```} \\
\hline \multicolumn{3}{|c|}{Type} & Description \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{1 if successful; 0 if an error is detected} \\
\hline \multirow{8}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & PLI_INT32 & narg & Index number of the user-defined system task or function argument \\
\hline & PLI_INT32 & bitlength & Number of bits the value represents \\
\hline & PLI_INT32 & format & A character in single quotes representing the radix (base) of the value \\
\hline & quoted string or PLI_BYTE8 * & value_p & Quoted character string or pointer to a character string with the value to be written \\
\hline & double & realdelay & Double-precision value representing the time delay before the value shall be written to the argument \\
\hline & PLI_INT32 & delaytype & Integer code representing the delay mode for applying the value \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_strdelputp() or tf_istrdelputp() for 32-bit integer delays Use tf_strlongdelputp() or tf_istrlongdelputp() for 64-bit integer delays Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routines tf_strrealdelputp() and tf_istrrealdelputp() shall write a string value to the specified argument of the current instance or a specific instance of a user-defined system task or function. The actual change to the argument shall be scheduled as an event on the argument in the Verilog model at a future simulation time. An argument value of 0 (system function return) shall be illegal.

The bitlength argument shall define the value size in bits.
The format shall define the format of the value specified by value_p and shall be one of the characters given in Table 200.

Table 200—Format characters
\begin{tabular}{|c|c|}
\hline Format character & Description \\
\hline\({ }^{\prime} \mathrm{b}^{\prime}\) or \({ }^{\prime} \mathrm{B}^{\prime}\) & Value is in binary \\
\hline\(' \mathrm{o}^{\prime}\) or \(\mathrm{O}^{\prime}\) & Value is in octal \\
\hline\(' \mathrm{~d}^{\prime}\) or \(\mathrm{D}^{\prime}\) & Value is in decimal \\
\hline \multicolumn{1}{c}{\(\mathrm{h}^{\prime}\) or \(\mathrm{H}^{\prime}\)} & Value is in hexadecimal \\
\hline
\end{tabular}

The delay argument shall represent the amount of time before the value shall be applied to the argument, and it shall be greater than or equal to 0 . The delay shall assume the timescale units of the module containing the instance of the user-defined system task or function.

The delaytype argument shall determine how the value shall be scheduled in relation to other simulation events on the same reg or variable. The delaytype shall be one of integer values shown in Table 201.

Table 201-delaytype codes
\begin{tabular}{|c|c|l|}
\hline delaytype code & Definition & \multicolumn{1}{c|}{ Description } \\
\hline \(\mathbf{0}\) & Inertial delay & \begin{tabular}{l} 
All scheduled events on the output argument in the Ver- \\
ilog model are removed before scheduling a new event
\end{tabular} \\
\hline \(\mathbf{1}\) & Modified transport delay & \begin{tabular}{l} 
All events that are scheduled for times later than the new \\
event on the output argument in the Verilog model are \\
removed before scheduling a new event
\end{tabular} \\
\hline \(\mathbf{2}\) & \begin{tabular}{l} 
No scheduled events on the output argument in the Ver- \\
ilog model are removed before scheduling a new \\
event the last event to be scheduled is not necessarily \\
the last one to occur
\end{tabular} \\
\hline
\end{tabular}

\subsection*{25.57 tf_subtract_long()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|r|}{tf_subtract_long()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Subtract two 64-bit integers.} \\
\hline Syntax: & \multicolumn{3}{|l|}{tf_subtract_long(aof_low1, aof_high1, low2, high2)} \\
\hline & \multicolumn{2}{|l|}{Type} & Description \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{Always returns 0} \\
\hline \multirow{5}{*}{Arguments:} & Type & Name & \\
\hline & PLI_INT32 * & aof_low1 & Pointer \\
\hline & PLI_INT32 * & aof_high1 & Pointer \\
\hline & PLI_INT32 & low2 & Least si \\
\hline & PLI_INT32 & high2 & Most sig \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_add_long() to add two 64-bit integers Use tf_multiply_long() to multiply two 64-bit integers Use tf_divide_long() to divide two 64-bit integers Use tf_compare_long() to compare two 64-bit integers} \\
\hline
\end{tabular}

The TF routine \(\mathbf{t f}\) _subtract_long() shall subtract two 64-bit values. After calling tf_subtract_long(), the variables used to pass the first operand shall contain the results of the subtraction. The operands shall be assumed to be in two s complement form. Figure 170 shows the high and low 32 bits of two 64 -bit integers and how tf_subtract_long() shall subtract them.


Figure 170—Subtracting with tf_subtract_long()
The example program fragment shown in Figure 171 uses tf_subtract_long() to calculate the relative time from the current time to the next event time (this example assumes that the code is executed during a misctf application call with reason of reason_rosynch).

The text message generated by this example is split between the two io_printf() calls. If done in a single io_printf(), the second call to tf_longtime_tostr() would overwrite the string from the first call, since the string is placed in a temporary buffer.
```

PLI INT32 currlow, currhigh;
PLI_INT32 relalow, relahigh;
currlow = tf_getlongtime(\&currhigh);
io_printf("At time %s: ", tf_longtime_tostr(currlow, currhigh));
if(tf_getnextlongtime(\&relalow, \&relahigh) == 0)
{
tf_subtract_long(\&relalow, \&relahigh, currlow, currhigh);
io_printf ("relative time to next event is %s",
tf_longtime_tostr(relalow, relahigh));
}
else
printf("there are no future events");

```

Figure 171—Using tf_subtract_long()

\subsection*{25.58 tf_synchronize(), tf_isynchronize()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_synchronize(), tf_isynchronize()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Synchronize to end of simulation time step.} \\
\hline Syntax: & \multicolumn{3}{|l|}{\begin{tabular}{l}
tf_synchronize() \\
tf_isynchronize (instance_p)
\end{tabular}} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type} & Description \\
\hline & PLI_INT32 & \multicolumn{2}{|l|}{0 if successful; 1 if an error occurred} \\
\hline \multirow[b]{2}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_rosynchronize() for read-only synchronization Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routines tf_synchronize() and tf_isynchronize() shall schedule a callback to the misctf application associated with the current instance or a specific instance of a user-defined system task or function. The misctf application shall be called with a reason of reason_synch at the end of the current simulation time step.

The routines tf_synchronize() and tf_rosynchronize() have different functionality. The routine tf_synchronize() shall call the associated misctf application at the end of the current simulation time step with reason_synch, and the misctf application shall be allowed to schedule additional simulation events using routines such as tf_strdelputp().

The routine tf_rosynchronize() shall call the associated misctf application at the end of the current simulation time step with reason_rosynch, and the PLI shall not be allowed to schedule any new events. This guarantees that all simulation events for the current time are completed. Calls to routines such as tf_strdelputp() and tf_setdelay() are illegal during processing of the misctf application with reason reason_rosynch.

The routine tf_getnextlongtime() shall only return the next simulation time for which an event is scheduled when used in conjunction with the routines \(\mathbf{t f}\) _rosynchronize() and tf_irosynchronize().

\subsection*{25.59 tf_testpvc_flag(), tf_itestpvc_flag()}
\begin{tabular}{||l|l|l||}
\hline \multicolumn{4}{|c|}{ tf_testpvc_flag(), tf_itestpvc_flag() } \\
\hline Synopsis: & Test system task/function argument value change flags. \\
\hline Syntax: & \begin{tabular}{l} 
tf_testpvc_flag (narg) \\
tf_itestpvc_flag (narg,
\end{tabular} \\
\hline \multirow{3}{|c|}{ instance_p) }
\end{tabular}

The TF routines tf_testpvc_flag() and tf_itestpve_flag() shall return value of the saved pve flag. The argument narg shall be the index number of an argument in a specific instance of a user-defined system task or function. Task/function argument index numbering shall proceed from left to right, with the left-most argument being number 1. If narg is \(\mathbf{- 1}\), then all argument pve flags shall be tested and the logical OR of all saved flags returned.

PVC flags shall be used to indicate whether a particular user-defined system task or function argument has changed value. Each argument shall have two pve flags: a current pve flag, which shall be set by a software product when the change occurs, and a saved pve flag, which shall be controlled by the user.

NOTE PVC flags shall not be set by the software product until tf_asynchon() or tf_iasynchon() has been called.

\subsection*{25.60 tf_text()}


The TF routine tf_text() shall store text messages about an error in a buffer, which will be printed when the routine tf_message() is called. The routine shall provide a method for a PLI application to store information about one or more errors before it calls the tf_message() TF routine. This allows an application to process all of a routine, such as syntax checking, before calling tf_message(), which can be set to abort processing after printing messages. An application shall be able to call tf_text() any number of times before it calls tf_message().

When the application calls tf_message(), the information stored by tf_text() shall be displayed before the information in the call to tf_message(). Each call to tf_message() shall clear the buffer where tf_text() stores its information.

The message argument is a user-defined control string containing the message to be displayed. The control string uses the same formatting controls as the C printf () function (for example, \%d). The message shall use up to a maximum of five variable arguments. There shall be no limit to the length of a variable argument. Formatting characters, such as \(\backslash n, \backslash t, \backslash b, \backslash f\), or \(\backslash r\), do not need to be included in the message the software product shall automatically format each message.

An example of using \(\mathbf{t f}\) _text() and tf_message() calls and the output generated follow. Note that the format of the output shall be defined by the software product.

Calling tf_text() and tf_message() with the arguments:
```

tf_text ("Argument number %d", argnum);
tf_message(ERR_ERROR, "User", "TFARG",
" is illegal in task %s", taskname);

```

Might produce the output:
```

ERROR! Argument number 2 is illegal in task [User-TFARG]
\$usertask

```
25.61 tf_typep(), tf_itypep()
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_typep(). tf_itypep()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get a system task/function argument type.} \\
\hline Syntax: & \multicolumn{3}{|l|}{```
tf_typep (narg)
tf_itypep(narg, instance_p)
```} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type} & Description \\
\hline & PLI_INT32 & \multicolumn{2}{|l|}{A predefined integer constant representing the Verilog HDL data type for the argument} \\
\hline \multirow{3}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_INT32 & narg & Index number of the user-defined system task or function argument \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routines tf_typep() and tf_itypep() shall return an integer constant indicating the type of an argument for the current instance or a specific instance of a user-defined system task or function. The integer constants shall be as shown in Table 202.

Table 202—Predefined tf_typep() constants
\begin{tabular}{|c|l|}
\hline Predefined constant & \multicolumn{1}{c|}{ Description } \\
\hline tf_nullparam & \begin{tabular}{l} 
The argument is a null expression (where no text has been given as the argu- \\
ment), or narg is out of range
\end{tabular} \\
\hline tf_string & The argument is a literal string \\
\hline tf_readonly & The argument is a expression with a value that can be read but not written \\
\hline tf_readwrite & The argument is a expression with a value that can be read and written \\
\hline tf_readonlyreal & The argument is a real number expression with a value that can be read but not written \\
\hline tf_readwritereal & The argument is a real number expression with a value that can be read and written \\
\hline
\end{tabular}

A read only expression shall be any expression that would be illegal as a left-hand-side construct in a Verilog HDL procedural assignment (e.g., an expression using net data types or event data types)
A read/write expression shall be any expression that would be legal as a left-hand-side construct in a Verilog HDL procedural assignments (e.g., an expression using reg, integer, time, or real data types)

The argument narg shall be the index number of an argument in a user-defined system task or function. Task/function argument index numbering shall proceed from left to right, with the left-most argument being number 1.

\subsection*{25.62 tf_unscale_longdelay()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_unscale_longdelay()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Convert a delay from internal simulation time units to the timescale of a particular module.} \\
\hline Syntax: & \multicolumn{3}{|l|}{tf_unscale_longdelay(instance_p, delay_lo, delay_hi, aof_delay_lo, aof_delay_hi)} \\
\hline \multicolumn{3}{|c|}{Type} & Description \\
\hline Returns: & void & & \\
\hline \multirow{6}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline & PLI_INT32 & delay_lo & Least significant (right-most) 32 bits of the delay to be converted \\
\hline & PLI_INT32 & delay_hi & Most significant (left-most) 32 bits of the delay to be converted \\
\hline & PLI_INT32 * & aof_delay_lo & Pointer to a variable to store the least significant (rightmost) 32 bits of the conversion result \\
\hline & PLI_INT32 * & aof_delay_hi & Pointer to a variable to store the most significant (leftmost) 32 bits of the conversion result \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_unscale_realdelay() to unscale real number delays Use tf_scale_longdelay() to convert a delay to the timescale of the module instance Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routine tf_unscale_longdelay() shall convert a 64-bit integer delay expressed in internal simulation time into the time units of the module containing the user-defined system task or function referenced by the instance_p pointer. The argument aof_delay_lo and aof_delay_hi shall contain the address of the converted delay returned by the routine.

\subsection*{25.63 tf_unscale_realdelay()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_unscale_realdelay()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Convert a delay expressed in internal simulation time units to the timescale of a particular module.} \\
\hline Syntax: & \multicolumn{3}{|l|}{tf_unscale_realdelay(instance_p, realdelay, aof_realdelay)} \\
\hline \multicolumn{3}{|c|}{Type} & Description \\
\hline Returns: & void & & \\
\hline \multirow{4}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & PLI_BYTE8 * & instance_p & Pointer to a specific instance of a user-defined system task or function \\
\hline & double & delay & Value of the delay to be converted \\
\hline & double * & aof_realdelay & Pointer to a variable to store the conversion result \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_unscale_longdelay() to unscale 64-bit integer delays Use tf_scale_realdelay() to convert a delay to the timescale of the module instance Use tf_getinstance() to get a pointer to an instance of a user-defined system task or function} \\
\hline
\end{tabular}

The TF routine tf_unscale_realdelay() shall convert a double-precision delay expressed in internal simulation time into the time units of the module containing the user-defined system task or function referenced by the instance_p pointer. The argument aof_realdelay shall contain the address of the converted delay returned by the routine.

\subsection*{25.64 tf_warning()}


The TF routine \(\mathbf{t f}\) _warning() shall provide a warning reporting mechanism compatible with warning messages generated by the software product.

The format control string shall use the same formatting controls as the C printf() function (for example, \%d).

The maximum number of arguments that shall be used in the format control string is 5 .
The location information (file name and line number) of the current instance of the user-defined system task or function shall be appended to the message using a format compatible with error messages generated by the software product.

The message shall be written to both the output channel of the software product which invoked the PLI application and the output log file of the product.

The \(\mathbf{t f}\) _warning() routine shall not abort parsing or compilation of Verilog HDL source code.

\subsection*{25.65 tf_write_save()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{tf_write_save()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Append a block of data to a save file.} \\
\hline Syntax: & \multicolumn{3}{|l|}{tf_write_save(blockptr, blocklen)} \\
\hline \multicolumn{3}{|c|}{Type} & Description \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{Nonzero value if successful, zero if an error is encountered} \\
\hline \multicolumn{2}{|r|}{Type} & Name & Description \\
\hline \multirow[t]{2}{*}{Arguments:} & PLI_BYTE8 * & blockptr & Pointer to the first byte of the block of data to be saved \\
\hline & PLI_INT32 & blocklen & Number of bytes are to be saved \\
\hline Related routines: & \multicolumn{3}{|l|}{Use tf_read_restart() to retrieve the data saved} \\
\hline
\end{tabular}

The TF routine tf_write_save() shall write user-defined data to the end of a save file being written by the \$save built-in system task. This routine shall be called from the misctf application when misctf is invoked with reason_save.

The argument blockptr shall be a pointer to an allocated block of memory containing the data to be saved.
The argument blocklen shall be the length in bytes of the allocated block of memory. Note that exactly as many bytes shall be restored using tf_read_restart() as were written with tf_write_save().

\section*{26. Using VPI routines}

Clause 26 and Clause 27 specify the Verilog Procedural Interface (VPI) for the Verilog HDL. This clause describes how the VPI routines are used, and Section 27 defines each of the routines in alphabetical order.

\subsection*{26.1 V I system tasks and functions}

User defined system tasks and functions are created using the routine vpi_register_systf() (see 27.34). The registration of system tasks must occur prior to elaboration or the resolution of references.

The intended use model would be to place a reference to a routine within the vlog_startup_routines[] array. This routine would register all user defined system tasks and functions when it is called.

VPI system tasks have compiletf, sizetf, and calltf routines which have the same use model as the corresponding checktf, sizetf and callff routines in the TF interface mechanism for user defined system tasks and functions (refer to Clause 21). The functionality provided in the TF interface mechanism for the misctf routine is supported via a set of callbacks, which can be registered using vpi_register_cb().

\subsection*{26.2 The V I interface}

The VPI interface provides routines that allow Verilog product users to access information contained in a Verilog design, and that allow facilities to interact dynamically with a software product. Applications of the VPI interface can include delay calculators and annotators, connecting a Verilog simulator with other simulation and CAE systems, and customized debugging tasks.

The functions of the VPI interface can be grouped into two main areas:
Dynamic software product interaction using VPI callbacks
Access to Verilog HDL objects and simulation specific objects

\subsection*{26.2.1 V I callbacks}

Dynamic software product interaction shall be accomplished with a registered callback mechanism. VPI callbacks shall allow a user to request that a Verilog HDL software product, such as a logic simulator, call a user-defined application when a specific activity occurs. For example, the user can request that the user application my_monitor() be called when a particular net changes value, or that my_cleanup () be called when the software product execution has completed.

The VPI callback facility shall provide the user with the means to interact dynamically with a software product, detecting the occurrence of value changes, advancement of time, end of simulation, etc. This feature allows applications such as integration with other simulation systems, specialized timing checks, complex debugging features, etc.

The reasons for which callbacks shall be provided can be separated into four categories:
Simulation event (e.g., a value change on a net or a behavioral statement execution)
Simulation time (e.g., the end of a time queue or after certain amount of time)
Simulator action/feature (e.g., the end of compile, end of simulation, restart, or enter interactive mode)
User-defined system task or function execution
VPI callbacks shall be registered by the user with the functions vpi_register_cb() and vpi_register_systf(). These routines indicate the specific reason for the callback, the application to be called, and what system and user data shall be passed to the callback application when the callback occurs. A facility is also provided to call the callback functions when a Verilog HDL product is first invoked. A primary use of this facility shall be for registration of user-defined system tasks and functions.

\subsection*{26.2.2 V I access to Verilog HDL objects and simulation objects}

Accessible Verilog HDL objects and simulation objects and their relationships and properties are described using data model diagrams. These diagrams are presented in 26.6. The data model diagrams indicate the routines and constants that are required to access and manipulate objects within an application environment. An associated set of routines to access these objects is defined in Clause 27.

The VPI interface also includes a set of utility routines for functions such as handle comparison, file handling, and redirected printing, which are described in Table 211.

VPI routines provide access to objects in an instantiated Verilog design. An instantiated design is one where each instance of an object is uniquely accessible. For instance, if a module \(m\) contains wire \(w\) and is instantiated twice as m 1 and m 2 , then \(\mathrm{m} 1 . \mathrm{w}\) and \(\mathrm{m} 2 . \mathrm{w}\) are two distinct objects, each with its own set of related objects and properties.

The VPI interface is designed as a simulation interface, with access to both Verilog HDL objects and specific simulation objects. This simulation interface is different from a hierarchical language interface, which would provide access to HDL information but would not provide information about simulation objects.

\subsection*{26.2.3 Error handling}

To determine if an error occurred, the routine vpi_chk_error() shall be provided. The vpi_chk_error() routine shall return a nonzero value if an error occurred in the previously called VPI routine. Callbacks can be set up for when an error occurs as well. The vpi_chk_error() routine can provide detailed information about the error.

\subsection*{26.2.4 Function availability}

Certain features of the VPI interface must occur early in the execution of a tool. In order to allow this process to occur in an orderly manner, some functionality must be restricted in these early stages. Specifically, when the routines within the vlog_startup_routines[] array are executed, there is very little functionality available. Only two routines can be called at this time:
```

vpi_register_systf()
vpi_register_cb()

```

In addition, the vpi_register_cb() routine can only be called for the following reasons:
```

cbEndOfCompile
cbStartOfSimulation
cbEndOfSimulation
cbUnresolvedSystf
cbError
cbPLIError

```

Refer to 27.34 for a further explanation of the use of the vlog_startup_routines[ ] array.
The next earliest phase is when the sizetf routines are called for the user defined system functions. At this phase, no additional access is permitted. After the sizetf routines are called, the routines registered for reason cbEndOfCompile are called. At this point, and continuing until the tool has finished execution, all functionality is available.

\subsection*{26.2.5 Traversing expressions}

The VPI routines provide access to any expression which can be written in the HDL. Dealing with these expressions can be complex, since very complex expressions can be written in the HDL. Expressions with
multiple operands will result in a handle of type vpiOperation. To determine how many operands, access the property vpiOpType. This operation will be evaluated after its subexpressions. Therefore, it has the least precedence in the expression.

An example of a routine which traverses an entire complex expression is listed below:
```

void traverseExpr(vpiHandle expr)
{
vpiHandle subExprI, subExprH;
switch (vpi_get(vpiExpr,expr))
{
case vpiOperation:
subExprI = vpi iterate(vpiOperand, expr);
if (subExprI)
while (subExprH = vpi_scan(subExprI))
traverseExpr(subExprH);
/* else it's of op type vpiNullop */
break;
default:
/* Do whatever to the leaf object. */
break;
}
}

```

\subsection*{26.3 V I object classifications}

VPI objects are classified using data model diagrams. These diagrams provide a graphical representation of those objects within a Verilog design to which the VPI routines shall provide access. The diagrams shall show the relationships between objects and the properties of each object. Objects with sufficient commonality are placed in groups. Group relationships and properties apply to all the objects in the group.

As an example, this simplified diagram shows that there is a one-to-many relationship from objects of type module to objects of type net, and a one-to-one relationship from objects of type net to objects of type module. Objects of type net have properties vpiName, vpiVector, and vpiSize, with data types string, boolean, and integer respectively.


The VPI data model diagrams are presented in 26.6.
For object relationships (unless a special tag is shown in the diagram), the type used for access is determined by adding vpi to the beginning of the word within the enclosure with each word's first letter being a capital. Using the above example, if an application has a handle to a net, and wants to go to the module instance where the net is defined, the call would be:
```

modH = vpi handle(vpiModule,netH);

```
where netH is a handle to the net. As another example, to access a named event object, use the type vpiNamedEvent.

\subsection*{26.3.1 Accessing object relationships and properties}

The VPI interface defines the C data type of vpiHandle. All objects are manipulated via a vpiHandle variable. Object handles can be accessed from a relationship with another object, or from a hierarchical name, as the following example demonstrates:
```

vpiHandle net;
net = vpi_handle_by_name("top.m1.w1", NULL);

```

This example call retrieves a handle to wire top.m1.w1 and assigns it to the vpiHandle variable net. The NULL second argument directs the routine to search for the name from the top level of the design.

The VPI interface provides generic functions for tasks, such as traversing relationships and determining property values. One-to-one relationships are traversed with routine vpi_handle(). In the following example, the module that contains net is derived from a handle to that net:
```

vpiHandle net, mod;
net = vpi_handle_by_name("top.m1.w1", NULL);
mod = vpi_handle(vpiModule, net);

```

The call to vpi_handle() in the above example shall return a handle to module top.m1.
Sometimes it is necessary to access a class of objects which do not have a name, or whose name is ambiguous with another class of objects which can be accessed from the reference handle. Tags are used in this situation.


In this example, the tags vpiLeftRange and vpiRightRange are used to access the expressions which make up the range of the part select. These tags are used instead of vpiExpr to get to the expressions. Without the tags, the VPI interface would not know which expression should be accessed. For example:
vpi_handle(vpiExpr, part_select_handle)
would be illegal when the reference handle (part_select_handle) is a handle to a part select, because the part select can refer to two expressions, a left-range and a right-range.

Properties of objects shall be derived with routines in the vpi_get family. The routine vpi_get() returns integer and boolean properties. Integer and boolean properties shall be defined to be of type PLI_INT32. For boolean properties, a value of 1 shall represent TRUE and a value of 0 shall represent FALSE. The routine vpi_get_str() accesses string properties. String properties shall be defined to be of type PLI_BYTE8 *. For example: to retrieve a pointer to the full hierarchical name of the object referenced by handie mod, the following call would be made:
```

PLI_BYTE8 *name = vpi_get_str(vpiFullName, mod);

```

In the above example, the pointer name shall now point to the string "top.m1".

One-to-many relationships are traversed with an iteration mechanism. The routine vpi_iterate() creates an object of type vpiIterator, which is then passed to the routine vpi_scan() to traverse the desired objects. In the following example, each net in module top.m1 is displayed:
```

vpiHandle itr;
itr = vpi_iterate(vpiNet,mod);
while (net = vpi_scan(itr) )
vpi_printf("\t%s\n", vpi_get_str(vpiFullName, net) );

```

As the above examples illustrate, the routine naming convention is a vpi prefix with _ word delimiters (with the exception of callback-related defined values, which use the \(c b\) prefix). Macro-defined types and properties have the vpi prefix, and they use capitalization for word delimiters.

The routines for traversing Verilog HDL structures and accessing objects are described in Clause 27.

\subsection*{26.3.2 bject type properties}

All objects have a vpiType property, which is not shown in the data model diagrams.
```

-> type
int: vpiType

```

Using vpi_get(vpiType, <object_handle>) returns an integer constant which represents the type of the object.

Using vpi_get_str(vpiType, <object_handle>) returns a pointer to a string containing the name of the type constant. The name of the type constant is derived from the name of the object as it is shown in the data model diagram (refer to 26.3 for a description of how type constant names are derived from object names).

Some objects have additional type properties which are shown in the data model diagrams; vpiDelayType, vpiNetType, vpiOpType, vpiPrimType, vpiResolvedNetType and vpiTchkType. Using vpi_get(<type_property>, <object_handle>) returns an integer constant which represents the additional type of the object. Refer to vpi_user.h in Annex G for the types which can be returned for these additional type properties. The constant names of the types returned for these additional type properties can be accessed using vpi_get_str().

\subsection*{26.3.3 bject file and line properties}

Most objects have two location properties, which are not shown in the data model diagrams:
```

-> location
int: vpiLineNo
str: vpiFile

```

The properties vpiLineNo and vpiFile can be affected by the line and file compiler directives. See Section 19 for more details on these compiler directives. These properties are applicable to every object that corresponds to some object within the HDL. The exceptions are objects of type:

\author{
vpiCallback \\ vpiDelayTerm \\ vpiDelayDevice \\ vpiInterModPath \\ vpiIterator \\ vpiTimeQueue
}

\subsection*{26.3.4 Delays and values}

Most properties are of type integer, boolean, or string. Delay and logic value properties, however, are more complex and require specialized routines and associated structures. The routines vpi_get_delays() and vpi_put_delays() use structure pointers, where the structure contains the pertinent information about delays. Similarly, simulation values are also handled with the routines vpi_get_value() and vpi_put_value(), along with an associated set of structures.

The routines, C structures, and some examples for handling delays and logic values are presented in Clause 27. See 27.14 for vpi_get_value(), 27.32 for vpi_put_value(), 27.9 for vpi_get_delays(), and 27.30 for vpi_put_delays().

Nets, primitives, module paths, timing checks, and continuous assignments can have delays specified within the HDL. Additional delays may exist, such as module input port delays or inter-module path delays, that do not appear within the HDL. To access the delay expressions that are specified within the HDL, use the method vpiDelay. These expressions shall be either an expression that evaluates to a constant if there is only one delay specified, or an operation if there are more than one delay specified. If multiple delays are specified, then the operation'supiOpType shall be vpiListOp. To access the actual delays being used by the tool, use the routine vpi_get_delays() on any of these objects.

\subsection*{26.4 List of V I routines by functional category}

The VPI routines can be divided into groups based on primary functionality.
VPI routines for simulation-related callbacks
VPI routines for system task/function callbacks
VPI routines for traversing Verilog HDL hierarchy
VPI routines for accessing properties of objects
VPI routines for accessing objects from properties
VPI routines for delay processing
VPI routines for logic and strength value processing
VPI routines for simulation time processing
VPI routines for miscellaneous utilities

Table 203 through Table 211 list the VPI routines by major category. Clause 27 defines each of the VPI routines, listed in alphabetical order.

Table 203-VPI routines for simulation related callbacks
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Register a simulation-related callback & vpi_register_cb() \\
\hline Remove a simulation-related callback & vpi_remove_cb() \\
\hline Get information about a simulation-related callback & vpi_get_cb_info() \\
\hline
\end{tabular}

Table 204-VPI routines for system task/function callbacks
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Register a system task/function callback & vpi_register_systf() \\
\hline Get information about a system task/function callback & vpi_get_systf_info() \\
\hline
\end{tabular}

Table 205—VPI routines for traversing Verilog HDL hierarchy
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Obtain a handle for an object with a one-to-one relationship & vpi_handle() \\
\hline Obtain handles for objects in a one-to-many relationship & \begin{tabular}{c} 
vpi_iterate() \\
vpi_scan()
\end{tabular} \\
\hline Obtain a handle for an object in a many-to-one relationship & vpi_handle_multi() \\
\hline
\end{tabular}

Table 206-VPI routines for accessing properties of objects
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Get the value of objects with types of int or bool & vpi_get() \\
\hline Get the value of objects with types of string & vpi_get_str() \\
\hline
\end{tabular}

Table 207-VPI routines for accessing objects from properties
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Obtain a handle for a named object & vpi_handle_by_name() \\
\hline Obtain a handle for an indexed object & vpi_handle_by_index() \\
\hline Obtain a handle to a word or bit in an array & vpi_handle_by_multi_index() \\
\hline
\end{tabular}

Table 208-VPI routines for delay processing
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Retrieve delays or timing limits of an object & vpi_get_delays() \\
\hline Write delays or timing limits to an object & vpi_put_delays() \\
\hline
\end{tabular}

Table 209—VPI routines for logic and strength value processing
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline Retrieve logic value or strength value of an object & vpi_get_value() \\
\hline Write logic value or strength value to an object & vpi_put_value() \\
\hline
\end{tabular}

Table 210-VPI routines for simulation time processing
\begin{tabular}{|c|c|}
\hline To & Use \\
\hline Find the current simulation time or the scheduled time of future events & vpi_get_time() \\
\hline
\end{tabular}

Table 211-VPI routines for miscellaneous utilities
\begin{tabular}{|l|c|}
\hline \multicolumn{1}{|c|}{ To } & Use \\
\hline \begin{tabular}{l} 
Write to the output channel of the software product which invoked the \\
PLI application and the current log file
\end{tabular} & vpi_printf() \\
\hline \begin{tabular}{l} 
Write to the output channel of the software product which invoked the \\
PLI application and the current log file using varargs
\end{tabular} & vpi_vprintf() \\
\hline Flush data from the current simulator output buffers & vpi_flush() \\
\hline Open a file for writing & vpi_mcd_open() \\
\hline Close one or more files & vpi_mcd_close() \\
\hline Write to one or more files & vpi_mcd_printf() \\
\hline Write to one or more open files using varargs & vpi_mcd_vprintf() \\
\hline Flush data from a given MCD output buffer & vpi_mcd_flush() \\
\hline Retrieve the name of an open file & vpi_mcd_name() \\
\hline Retrieve data about product invocation options & vpi_get_vlog_info() \\
\hline See if two handles refer to the same object & vpi_compare_objects() \\
\hline Obtain error status and error information about the previous call to a & vpi_chk_error() \\
\hline VPI routine & vpi_free_object() \\
\hline Free memory allocated by VPI routines & vpi_put_data() \\
\hline Add user-allocated storage to application saved data & vpi_get_data() \\
\hline Retrieve user-allocated storage from application saved data & vpi_put_userdata() \\
\hline Store user data in VPI work area & vpi_get_userdata() \\
\hline Retrieve user data from VPI work area & vpi_sim_control() \\
\hline Control simulation execution (stop, finish, etc.) & \\
\hline
\end{tabular}

\subsection*{26.5 Key to data model diagrams}

This subsection contains the keys to the symbols used in the data model diagrams. Keys are provided for objects and classes, traversing relationships, and accessing properties.

\subsection*{26.5.1 Diagram key for objects and classes}
Object Definition:
Bold letters in a solid enclosure indicate an object definition. The
properties of the object are defined in this location.

\subsection*{26.5.2 Diagram key for accessing properties}
\begin{tabular}{|c|c|}
\hline obj & Integer and boolean properties are accessed with the routine vpi_get(). These properties are of type PLI_INT32. \\
\hline \multirow[t]{2}{*}{\[
\begin{aligned}
& \text {-> vector } \\
& \quad \text { bool: vpiVector } \\
& \text {-> size } \\
& \quad \text { int: vpiSize }
\end{aligned}
\]} & Example: Given handle obj_h to an object of type vpiObj, test if the object is a vector, and get the size of the object. \\
\hline & ```
PLI_INT32 vect_flag = vpi_get(vpivector, obj_h);
PLI_INT32 size = vpi_get(vpiSize, obj_h);
``` \\
\hline obj & String properties are accessed with routine vpi_get_str(). String properties are of type PLI_BYTE8 *. \\
\hline \multirow[t]{2}{*}{\begin{tabular}{l}
-> name \\
str: vpiName \\
str: vpiFullName
\end{tabular}} & Example: \\
\hline & PLI_BYTE8 *name = vpi_get_str(vpiName, obj_h); \\
\hline object & Complex properties for time and logic value are accessed with the indicated routines. See the descriptions of the routines for usage. \\
\hline \multicolumn{2}{|l|}{-> complex} \\
\hline \begin{tabular}{l}
func1() \\
func2()
\end{tabular} & \\
\hline
\end{tabular}

\subsection*{26.5.3 Diagram key for traversing relationships}
A single arrow indicates a one-to-one relationship accessed with the routine vpi_handle().
Example: Given vpiHandle variable ref_h of type ref, access obj_h of type \(O b j\) :
obj_h = vpi_handle(Obj, ref_h);

A tagged one-to-one relationship is traversed similarly, using Tag instead of \(O b j\) :
Example:
obj_h = vpi_handle(Tag, ref_h);
A one-to-one relationship which originates from a circle is traversed using NULL for the ref_h:
Example:

A double arrow indicates a one-to-many relationship accessed with the routine vpi_scan().
Example: Given vpiHandle variable ref_h of type ref, scan objects of type \(O b j\) :
itr = vpi iterate(Obj, ref h);
itr = vpi iterate(Obj, ref h);
while (obj_h = vpi_scan(itr) )
while (obj_h = vpi_scan(itr) )
    /* process `obj_h' */
    /* process `obj_h' */
A tagged one-to-many relationship is traversed similarly, using Tag instead of \(O b j\) :
Example:
itr = vpi_iterate(Tag, ref_h);
itr = vpi_iterate(Tag, ref_h);
while (obj_h = vpi_scan(itr) )
while (obj_h = vpi_scan(itr) )
    /* proces̀s `obj_\overline{h' */}
    /* proces̀s `obj_\overline{h' */}
A one-to-many relationship which originates from a circle is traversed using NULL for the ref_h:
Example:
itr = vpi_iterate(Obj, NULL);
itr = vpi_iterate(Obj, NULL);
while (obj_h = vpi_scan(itr) )
while (obj_h = vpi_scan(itr) )
    /* process `obj_h' */
    /* process `obj_h' */

For relationships which do not have a tag, the type used for access is determined by adding vpi to the beginning of the word within the enclosure with each word's first letter being a capital. Refer to 26.3 for more details on VPI access constant names.

\section*{26.6 bject data model diagrams}

Subclauses 26.6.1 through 26.6 .43 contain the data model diagrams that define the accessible objects and groups of objects, along with their relationships and properties.

\subsection*{26.6.1 Module}


2 Passing a NULL handle to vpi_get() with types vpiTimePrecision or vpiTimeUnit shall return the smallest time precision of all modules in the instantiated design.

3 The properties vpiDefLineNo and vpiDefFile can be affected by the `line and `file compiler directives. See Clause 19 for more details on these compiler directives.

4 If a module is an element within a module array, the vpiIndex transition is used to access the index within the array. If a module is not part of a module array, this transition shall return NULL.

\subsection*{26.6.2 Instance arrays}


NOTE Traversing from the instance array to expr shall return a simple expression object of type vpiOperation with a vpiOpType of vpiListOp. This expression can be used to access the actual list of connections to the module or primitive instance array in the Verilog source code.

\subsection*{26.6.3 Scope}

26.6.4 I declaration

\subsection*{26.6.5 orts}


\section*{NOTES}

1 vpiHighConn shall indicate the hierarchically higher (closer to the top module) port connection.
2 vpiLowConn shall indicate the lower (further from the top module) port connection.
3 Properties scalar and vector shall indicate if the port is 1 bit or more than 1 bit. They shall not indicate anything about what is connected to the port.

4 Properties index and name shall not apply for port bits.
5 If a port is explicitly named, then the explicit name shall be returned. If not, and a name exists, then that name shall be returned. Otherwise, NULL shall be returned.

6 vpiPortIndex can be used to determine the port order. The first port has a port index of zero.
7 vpiHighConn and vpiLowConn shall return NULL if the port is not connected.
\(8 \quad\) vpiSize for a null port shall return 0 .

\subsection*{26.6.6 Nets and net arrays}


\section*{NOTES}

1 For vectors, net bits shall be available regardless of vector expansion.
(Notes continued on next page)

2 Continuous assignments and primitive terminals shall be accessed regardless of hierarchical boundaries.
3 Continuous assignments and primitive terminals shall only be accessed from scalar nets or bit selects.
4 For vpiPorts, if the reference handle is a bit then port bits shall be returned. If it is the entire vector, then a handle to the entire port shall be returned.

5 For vpiPortInst, if the reference handle is a bit or scalar, then port bits or scalar ports shall be returned, unless the highconn for the port is a complex expression where the bit index cannot be determined. If this is the case, then the entire port shall be returned. If the reference handle is a vector, then the entire port shall be returned.
6 For vpiPortInst, it is possible for the reference handle to be part of the highconn expression, but not connected to any of the bits of the port. This may occur if there is a size mismatch. In this situation, the port shall not qualify as a member for that iteration.

7 For implicit nets, vpiLineNo shall return \(\mathbf{0}\), and vpiFile shall return the file name where the implicit net is first referenced.
8 vpi_handle(vpiIndex, net_bit_handle) shall return the bit index for the net bit. vpi_iterate(vpiIndex, net_bit_handle) shall return the set of indices for a multidimensional net array bit select, starting with the index for the net bit and working outward.
9 Only active forces and assign statements shall be returned for vpiLoad.
10 Only active forces shall be returned for vpiDriver.
11 vpiDriver shall also return ports that are driven by objects other than nets and net bits.
12 vpiLocalLoad and vpiLocalDriver return only the loads or drivers that are local, i.e.: contained by the module instance which contains the net, including any ports connected to the net (output and inout ports are loads, input and inout ports are drivers).

13 For vpiLoad, vpiLocalLoad, vpiDriver and vpiLocalDriver iterators, if the object is vpiNet for a vector net, then all loads or drivers are returned exactly once as the loading or driving object. That is, if a part select loads or drives only some bits, the load or driver returned is the part select. If a driver is repeated, it is only returned once. To trace exact bit by bit connectivity pass a vpiNetBit object to vpi_iterate.

14 An iteration on loads or drivers for a variable bit-select shall return the set of loads or drivers for whatever bit that the bitselect is referring to at the beginning of the iteration.
15 vpiSimNet shall return a unique net if an implementation collapses nets across hierarchy (refer to Section 12.3.10 for the definition of simulated net and collapsed net).

16 The property vpiExpanded on an object of type vpiNetBit shall return the property's value for the parent.
17 The loads and drivers returned from vpi_iterate(vpiLoad, obj_handle) and vpi_iterate(vpiDriver, obj_handle) may not be the same in different implementations, due to allowable net collapsing (see Section 12.3.10). The loads and drivers returned from vpi_iterate(vpiLocalLoad, obj_handle) and vpi_iterate(vpiLocalDriver, obj_handle) shall be the same for all implementations.

18 The boolean property vpiConstantSelect returns TRUE if the expression that constitutes the index or indices evaluates to a constant, and FALSE otherwise.

19 vpi_get(vpiSize, net_handle) returns the number of bits in the net. vpi_get(vpiSize, net_array_handle) returns the total number of nets in the in the array.

20 vpi_iterate(vpiIndex, net_handle) shall return the set of indices for a net within an array, starting with the index for the net and working outward. If the net is not part of an array, a NULL shall be returned.

\subsection*{26.6.7 egs and reg arrays}

regs

vpilndex

\section*{NOTES}

1 Continuous assignments and primitive terminals shall be accessed regardless of hierarchical boundaries.
2 Continuous assignments and primitive terminals shall only be accessed from scalar regs and bit selects.

3 For vpiPorts, if the reference handle is a bit then port bits shall be returned. If it is the entire vector, then a handle to the entire port shall be returned.

4 For vpiPortInst, if the reference handle is a bit or scalar, then port bits or scalar ports shall be returned, unless the highconn for the port is a complex expression where the bit index cannot be determined. If this is the case, then the entire port shall be returned. If the reference handle is a vector, then the entire port shall be returned.

5 For vpiPortInst, it is possible for the reference handle to be part of the highconn expression, but not connected to any of the bits of the port. This may occur if there is a size mismatch. In this case, the port shall not qualify as a member for that iteration.

6 vpi_handle(vpiIndex, reg_bit_handle) shall return the bit index for the reg bit. vpi_iterate(vpiIndex, reg_bit_handle) shall return the set of indices for a multidimensional reg array bit select, starting with the index for the reg bit and working outward.

7 Only active forces and assign statements shall be returned for vpiLoad and vpiDriver.
8 For vpiLoad and vpiDriver iterators, if the object is vpiReg for a vectored reg, then all loads or drivers are returned exactly once as the loading or driving object. That is, if a part select loads or drives only some bits, the load or driver returned is the part select. If a driver is repeated, it is only returned once. To trace exact bit by bit connectivity, pass a vpiRegBit object to the iterator.

9 The loads and drivers returned from vpi_iterate(vpiLoad, obj_handle) and vpi_iterate(vpiDriver, obj_handle) may not be the same in different implementations due to allowable net collapsing (see Section 12.3.10).
10 An iteration on loads or drivers for a variable bit-select shall return the set of loads or drivers for whatever bit that the bitselect is referring to at the beginning of the iteration.

11 If the reg has a default initialization assignment, the expression can be accessed using vpi_handle(vpiExpr, reg_handle) or vpi_handle(vpiExpr, reg_bit_handle).
12 vpi_get(vpiSize, reg_handle) returns the number of bits in the reg. vpi_get(vpiSize, reg_array_handle) returns the total number of regs in the in the array.

13 vpi_iterate(vpiIndex, reg_handle) shall return the set of indices for a reg within an array, starting with the index for the reg and working outward. If the reg is not part of an array, a NULL shall be returned.

\subsection*{26.6.8 Variables}


\section*{NOTES}

1 A var select is a word selected from a variable array.
2 The VPI does not provide access to bits of variables. If a handle to bit select of a variable is obtained, the object shall be a vpiBitSelect in the simple expression class. The variable containing the bit can be accessed using vpiParent. Refer to Section 26.6.25.

3 The boolean property vpiArray shall be TRUE if the variable handle references an array of variables, and FALSE otherwise. If the variable is an array, iterate on vpiVarSelect to obtain handles to each variable in the array.

4 vpi_handle(vpiIndex, var_select_handle) shall return the index of a var select in a 1-dimensional array. vpi_iterate(vpiIndex, var_select_handle) shall return the set of indices for a var select in a multidimensional array, starting with the index for the var select and working outward.

5 vpiLeftRange and vpiRightRange shall apply to variables when vpiArray is TRUE, and represent the array range declaration. These relationships are only valid when vpiArray is TRUE.

6 vpiSize for a variable array shall return the number of variables in the array. For non-array variables, it shall return the size of the variable in bits.

7 vpiSize for a var select shall return the number of bits in the var select.
8 Variables whose boolean property vpiArray is TRUE do not have a value property.

\subsection*{26.6.9 Memory}


\section*{NOTES}

1 vpiSize for a memory shall return the number of words in the memory.
2 vpiSize for a memory word shall return the number of bits in the word.
3 A memory is a one-dimensional array of reg types. Since 1364-2000 supports multi-dimensional arrays of regs, access to arrays of regs has been generalized. Although the access provided in Section 26.6.9 is still allowed, the prefered method is to iterate using vpiRegArray. See Section 26.6.7.

\subsection*{26.6.10 bject range}


\subsection*{26.6.11 Named event}


NOTE vpi_iterate(vpiIndex, named_event_handle) shall return the set of indices for a named event within an array, starting with the index for the named event and working outward. If the named event is not part of an array, a NULL shall be returned.

\subsection*{26.6.12 arameter, specparam}


\section*{NOTES}

1 Obtaining the value from the object parameter shall return the final value of the parameter after all module instantiation overrides and defparams have been resolved.
2 vpiLhs from a param assign object shall return a handle to the overridden parameter.

\subsection*{26.6.13 rimitive, prim term}


\section*{NOTES}

1 vpiSize shall return the number of inputs.
2 For primitives, vpi_put_value() shall only be used with sequential UDP primitives.
3 vpiTermIndex can be used to determine the terminal order. The first terminal has a term index of zero.
4 If a primitive is an element within a primitive array, the vpiIndex transition is used to access the index within the array. If a primitive is not part of a primitive array, this transition shall return NULL.

\subsection*{26.6.14 UD}


\section*{NOTES}

1 Only string (decompilation) and vector (ASCII values) shall be obtained for table entry objects using vpi_get_value(). Refer to the definition of vpi_get_value() for additional details.

2 vpiPrimType returns vpiSeqPrim for sequential UDP's and vpiCombPrim for combinatorial UDP's.

\subsection*{26.6.15 Module path, path term}


\subsection*{26.6.16 Intermodule path}

\[
\begin{aligned}
& \text {-> delay } \\
& \quad \text { vpi_get_delays() } \\
& \text { vpi_put_delays() }
\end{aligned}
\]

NOTE To get to an intermodule path, vpi_handle_multi(vpiInterModPath, port1, port2) can be used.

\subsection*{26.6.17 Timing check}


\section*{NOTES}

1 The vpiTchkRefTerm is the first terminal for all tchks except \$setup, where vpiTchkDataTerm is the first terminal and vpiTchkRefTerm is the second terminal.

2 When iterating for the expressions in a tcheck the handle returned for what is known as the data, ref, and notifier terminal will have the type vpiTchkTerm. All other arguments will have types matching the expression.
26.6.18 Task, function declaration


NOTE A Verilog HDL function shall contain an object with the same name, size, and type as the function.

\subsection*{26.6.19 Task and function call}


\section*{NOTES}

1 The system task or function that invoked an application shall be accessed with vpi_handle(vpiSysTfCall, NULL)
2 vpi_get_value() shall return the current value of the system function.
3 If the vpiUserDefn property of a system task or function call is true, then the properties of the corresponding systf object shall be obtained via vpi_get_systf_info().

4 All user-defined system tasks or functions shall be retrieved using vpi_iterate(), with vpiUserSystf as the type argument, and a NULL reference argument.

5 Arguments to PLI tasks or functions are not evaluated until an application requests their value. Effectively, the value of any argument is not known until the application asks for it. When an argument is an HDL or system function call, the function cannot be evaluated until the application asks for its value. If the application never asks for the value of the function, it is never evaluated. If the application has a handle to an HDL or system function it may ask for its value at any time in the simulation. When this happens the function is called and evaluated at this time.

6 A null argument is an expression with a vpiType of vpiOperation and a vpiOpType of vpiNullOp.
7 The property vpiDecompile will return a string with a functionally equivalent system task or function call to what was in the original HDL. The arguments will be decompiled using the same manner as any expression is decompiled. See Section 26.6.26 for a description of expression decompilation.

\subsection*{26.6.20 Frames}


\section*{NOTES}

1 It shall be illegal to place value change callbacks on automatic variables.
2 It shall be illegal to put a value with a delay on automatic variables.
3 There is at most only one active frame at any time. To get a handle to the currently active frame, use vpi_handle(vpiFrame, NULL). The frame to stmt transition shall return the currently active statement within the frame.

4 Frame handles must be freed using vpi_free_object() once the application no longer needs the handle. If the handle is not freed it shall continue to exist, even after the frame has completed execution.

\subsection*{26.6.21 Delay terminals}


NOTES
1 The value of the input delay term shall change before the delay associated with the delay device.
2 The value of the output delay term shall not change until after the delay has occurred.

\subsection*{26.6.22 Net drivers and loads}


\subsection*{26.6.23 eg drivers and loads}


\subsection*{26.6.24 Continuous assignment}


\section*{NOTES}

1 The size of a cont assign bit is always scalar.
2 Callbacks for value changes can be placed onto cont assign or a cont assign bit.
3 vpiOffset shall return zero for the least significant bit.

\subsection*{26.6.25 Simple expressions}


\section*{NOTES}

1 For vectors, the vpiUse relationship shall access any use of the vector or part-selects or bit-selects thereof.
2 For bit-selects, the vpiUse relationship shall access any specific use of that bit, any use of the parent vector, and any partselect that contains that bit.

\subsection*{26.6.26 Expressions}


\section*{NOTES}

1 For an operator whose type is vpiMultiConcat, the first operand shall be the multiplier expression. The remaining operands shall be the expressions within the concatenation.

2 The property vpiDecompile will return a string with a functionally equivalent expression to the original expression within the HDL. Parenthesis shall be added only to preserve precedence. Each operand and operator shall be separated by a single space character. No additional white space shall be added due to parenthesis.
26.6.27 rocess, block, statement, event statement


\subsection*{26.6.28 Assignment}


\subsection*{26.6.29 Delay control}


NOTE For delay control associated with assignment, the statement shall always be NULL.

\subsection*{26.6.30 Event control}


NOTE For event control associated with assignment, the statement shall always be NULL.

\subsection*{26.6.31 epeat control}


NOTE For delay control and event control associated with assignment, the statement shall always be NULL.

\subsection*{26.6.32 While, repeat, wait}

26.6.33 For

\subsection*{26.6.34 Forever}

\subsection*{26.6.35 If, if-else}


\subsection*{26.6.36 Case}


NOTES
1 The case item shall group all case conditions that branch to the same statement.
2 vpi_iterate() shall return NULL for the default case item since there is no expression with the default case.

\subsection*{26.6.37 Assign statement, deassign, force, release}


\subsection*{26.6.38 Disable}


\subsection*{26.6.39 Callback}


NOTES
1 To get information about the callback object, the routine vpi_get_cb_info() can be used.
2 To get callback objects not related to the above objects, the second argument to vpi_iterate() shall be NULL.

\subsection*{26.6.40 Time queue}


\section*{NOTES}

1 The time queue objects shall be returned in increasing order of simulation time.
2 vpi_iterate() shall return NULL if there is nothing left in the simulation queue.
3 If any events after read only sync remain in the current queue, then it shall not be returned as part of the iteration.

\subsection*{26.6.41 Active time format}


NOTE If \$timeformat() has not been called, vpi_handle(vpiActiveTimeFormat,NULL) shall return a NULL.

\subsection*{26.6.42 Attributes}


\subsection*{26.6.43 Iterator}


\footnotetext{
1 vpi_handle(vpiUse, iterator_handle) shall return the reference handle used to create the iterator.
2 It is possible to have a \(N U L L\) reference handle, in which case vpi_handle(vpiUse, iterator_handle) shall return NULL.
}

\section*{27. VPI routine definitions}

This clause describes the Verilog Procedural Interface (VPI) routines, explaining their function, syntax, and usage. The routines are listed in alphabetical order. See Clause 23 for the conventions used in the definitions of the PLI routines.

\section*{27.1 vpi_chk_error()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{vpi_chk_error()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Retrieve information about VPI routine errors.} \\
\hline Syntax: & \multicolumn{3}{|l|}{vpi_chk_error (error_info_p)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type Description} & \\
\hline & PLI_INT32 & \multicolumn{2}{|l|}{returns the error severity level if the previous VPI routine call resulted in an error and \(\mathbf{0}\) (false) if no error occurred} \\
\hline \multirow[b]{2}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & p_vpi_error_info & error_info_p & Pointer to a structure containing error information \\
\hline
\end{tabular}

The VPI routine vpi_chk_error() shall return an integer constant representing an error severity level if the previous call to a VPI routine resulted in an error. The error constants are shown in Table 212. If the previous call to a VPI routine did not result in an error, then vpi_chk_error() shall return \(\mathbf{0}\) (false). The error status shall be reset by any VPI routine call except vpi_chk_error(). Calling vpi_chk_error() shall have no effect on the error status.

Table 212-Return error constants for vpi_chk_error()
\begin{tabular}{|c|c|}
\hline Error Constant & Severity Level \\
\cline { 1 - 1 } vpiNotice & \multirow{2}{*}{ lowest severity } \\
\cline { 1 - 1 } vpiWarning & \\
\cline { 1 - 1 } vpiError & \\
\cline { 1 - 1 } vpiSystem & \\
\cline { 1 - 1 } vpiInternal & \\
\hline
\end{tabular}

If an error occurred, the s_vpi_error_info structure shall contain information about the error. If the error information is not needed, a NULL can be passed to the routine. The s_vpi_error_info structure used by vpi_chk_error() is defined in vpi_user. h and is listed in Figure 172.
```

'pedef struct t_vpi_error_info
PLI_INT32 state; /* vpi[ Compile,PLI,Run] * /
PLI_INT32 level; /* vpi[ Notice,Warning,Error,System,Internal] *
PLI BYTE8 *message;
PLI_BYTE8 *product;
PLI BYTE8 * code;
PLI_BYTE8 * file;
PLI_INT32 line;
s_vpi_error_info, *p_vpi_error_info;

```

Figure 172-The s_vpi_error_info structure definition

\section*{27.2 vpi_compare_objects()}


The VPI routine vpi_compare_objects() shall return \(\mathbf{1}\) (true) if the two handles refer to the same object. Otherwise, \(\mathbf{0}\) (false) shall be returned. Handle equivalence cannot be determined with a \(\mathrm{C}==\) comparison.

\section*{27.3 vpi_control()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{vpi_control()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Pass information from user code to simulator.} \\
\hline Syntax: & \multicolumn{3}{|l|}{vpi_control (operation, varargs)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{1 (true) if successful; 0 (false) on a failure} \\
\hline \multirow{3}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_INT32 & operation & select type of operation \\
\hline & & varargs & variable number of operation specific arguments \\
\hline Related routines: & & & \\
\hline
\end{tabular}

The VPI routine vpi_control () shall pass information from a user PLI application to a Verilog software tool, such as a simulator. The following control constants are defined as part of the VPI standard:
\begin{tabular}{ll} 
vpiStop & \begin{tabular}{l} 
causes the \$stop built-in Verilog system task to be executed upon return of \\
the user function. This operation shall be passed one additional integer \\
argument, which is the same as the diagnostic message level argument
\end{tabular} \\
passed to \$stop (see 17.4.2). \\
vpiFinish \\
causes the \$finish built-in Verilog system task to be executed upon return \\
of the user function. This operation shall be passed one additional integer \\
argument, which is the same as the diagnostic message level argument \\
passed to \$finish (see 17.4.1).
\end{tabular}

\section*{27.4 vpi_flush()}

\section*{vpi_flush()}

Synopsis: Flushes the data from the simulator output channel and log file output buffers.
\begin{tabular}{|c|c|c|c|}
\hline \multirow[t]{2}{*}{Syntax:} & \multicolumn{3}{|l|}{vpi_flush()} \\
\hline & Type & Description & \\
\hline \multirow[t]{2}{*}{Returns:} & PLI_INT32 & 0 if successful, non-zero if unsuccessful & \\
\hline & Type & Name & Description \\
\hline Arguments: & \multicolumn{3}{|l|}{None} \\
\hline Related routines: & \begin{tabular}{l}
Use vpi_printf() \\
Use vpi_vprintf() \\
Use vpi_mcd_pr
\end{tabular} & write a finite number of arguments to the write a variable number of arguments to \(f()\) to write one or more opened files & utput channel and log file or output channel and log file \\
\hline
\end{tabular}

The routine vpi_flush() shall flush the output buffers for the simulator s output channel and current log file.

\section*{27.5 vpi_free_object()}
vpi_free_object()

\section*{Synopsis:}

The VPI routine vpi_free_object() shall free memory allocated for objects. It shall generally be used to free memory created for iterator objects. The iterator object shall automatically be freed when vpi_scan() returns NULL either because it has completed an object traversal or encountered an error condition. If neither of these conditions occur (which can happen if the code breaks out of an iteration loop before it has scanned every object), vpi_free_object() should be called to free any memory allocated for the iterator. This routine can also optionally be used for implementations that have to allocate memory for objects. The routine shall return \(\mathbf{1}\) (true) on success and \(\mathbf{0}\) (false) on failure.

\section*{27.6 vpi_get()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{vpi_get()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get the value of an integer or boolean property of an object.} \\
\hline Syntax: & \multicolumn{3}{|l|}{vpi_get(prop, obj)} \\
\hline \multicolumn{4}{|c|}{Type \(\quad\) Description} \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{Value of an integer or boolean property} \\
\hline \multirow{3}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_INT32 & prop & An integer constant representing the property of an object for which to obtain a value \\
\hline & vpiHandle & obj & Handle to an object \\
\hline Related routines: & \multicolumn{3}{|l|}{Use vpi_get_str() to get string properties} \\
\hline
\end{tabular}

The VPI routine vpi_get() shall return the value of integer and boolean object properties. These properties shall be of type PLI_INT32. Boolean properties shall have a value of \(\mathbf{1}\) for TRUE and \(\mathbf{0}\) for FALSE. For integer object properties such as vpiSize, any integer shall be returned. For integer object properties that return a defined value, refer to Annex G for the value that shall be returned. Note for object property vpiTimeUnit or vpiTimePrecision, if the object is NULL, then the simulation time unit shall be returned. Should an error occur, vpi_get() shall return vpiUndefined.

\section*{27.7 vpi_get_cb_info()}


The VPI routine vpi_get_cb_info() shall return information about a simulation-related callback in an s_cb_data structure. The memory for this structure shall be allocated by the user.

The s_cb_data structure used by vpi_get_cb_info() is defined in vpi_user. h and is listed in Figure 173.
```

!pedef struct t_cb_data
PLI_INT32 reason; /* callback reason * /
PLI_INT32 (* cb_rtn)(struct t_cb_data *); /* call routine * /
vpiHandle obj; /* trigger object */
p_vpi_time time; /* callback time */
p_vpi_value value; /* trigger object value */
PEI_INTT32 index; /* index of the memory word or var select
that changed * /
PLI_BYTE8 *user_data;
s_c\overline{b_data, *p_cb_\overline{data;}}\mathbf{}/2

```

Figure 173-The s_cb_data structure definition

\section*{27.8 vpi_get_data()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{vpi_get_data()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get data from an implementation's save/restart location} \\
\hline Syntax: & \multicolumn{3}{|l|}{vpi_get_data(id, dataLoc, numOfBytes)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{The number of bytes retrieved} \\
\hline \multicolumn{2}{|r|}{Type} & Name & Description \\
\hline \multirow[t]{3}{*}{Arguments:} & PLI_INT32 & id & A save/restart ID returned from vpi_get(vpiSaveRestartID, NULL) \\
\hline & PLI_BYTE8 * & dataLoc & Address of application allocated storage \\
\hline & PLI_INT32 & numOfBytes & Number of bytes to be retrieved from save/restart location \\
\hline Related routines: & \multicolumn{3}{|l|}{Use vpi_put_data() to write saved data} \\
\hline
\end{tabular}

The routine shall place numOfBytes of data into the memory location pointed to by dataLoc from a simulation s save/restart location. This memory location has to be properly allocated by the application. The first call for a given \(i d\) will retrieve the data starting at what was placed into the save/restart location with the first call to vpi_put_data() for a given id. The return value shall be the number of bytes retrieved. On a failure the return value shall be 0 . Each subsequent call shall start retrieving data where the last call left off. It shall be a warning for an application to retrieve more data than what was placed into the simulation save/restart location for a given id. In this case the dataLoc shall be filled with the data that is left for the given id and the remaining bytes shall be filled with ' \(\backslash 0\) '. The return value shall be the actual number of bytes retrieved. It shall be acceptable for an application to retrieve less data then what was stored for a given id with vpi_put_data(). This routine can only be called from a user application routine that has been called for reason cbStartOfRestart or cbEndOfRestart. The recommended way to get the id for vpi_get_data() is to
pass it as the value for user_data when registering for cbStartOfRestart or cbEndOfRestart from the cbStartOfSave or cbEndOfSave user application routine. An application can get the path to the simulations save/restart location by calling vpi_get_str(vpiSaveRestartLocation, NULL) from a user application routine that has been called for reason cbStartOfRestart or cbEndOfRestart.

The following example illustrates using vpi_get_data():
```

/* Uses the global pointer firstWrk * /
PLI_INT32 consumer_restart(p_cb_data data)
{
struct myStruct *wrk; /* myStruct is defined in vpi_put_data()
example */
PLI_INT32 status;
PLI-INT32 cnt, size;
PLI_INT32 id = (PLI_INT32)data->user_data;
/* Get the number of structures. * /
status = vpi_get_data(id,(PLI_BYTE8 *)\&cnt),sizeof(PLI_INT32));
assert(status > \overline{0); /* Check returned status. */}
size = cnt * sizeof(struct myStruct);
/* allocate memory for the structures * /
cnt * = sizeof(struct myStruct);
firstWrk = malloc(cnt);
/* retrieve the data structures * /
if (cnt != vpi_get_data(id, (PLI_BYTE8 *) firstWrk,cnt))
return(1); /齐 erर्ror. */
firstWrk = wrk;
/* Fix the next pointers in the link list. */
for (wrk = firstWrk; cnt > 0; cnt--)
{
wrk->next = wrk + 1;
wrk = wrk->next;
}
wrk->next = NULL;
return(SUCCESS);
}

```

\section*{27.9 vpi_get_delays()}
\begin{tabular}{|l|l|l||}
\hline \multicolumn{4}{|c|}{ vpi_get_delays() } \\
\hline Synopsis: & Retrieve the delays or pulse limits of an object. \\
\hline Syntax: & vpi_get_delays (obj, delay_p) \\
\hline \multirow{3}{*}{ Returns: } & Type & Description \\
\hline \multirow{3}{*}{ Arguments: } & void & Type \\
\cline { 2 - 4 } & vpiHandle & Name \\
\cline { 2 - 4 } & p_vpi_delay & obj \\
\hline \multicolumn{3}{|c|}{\begin{tabular}{l} 
Related \\
routines:
\end{tabular}} \\
\hline
\end{tabular}

The VPI routine vpi_get_delays() shall retrieve the delays or pulse limits of an object and place them in an s_vpi_delay structure that has been allocated by the user. The format of the delay information shall be controlled by the time_type flag in the s_vpi_delay structure. This routine shall ignore the value of the type flag in the s_vpi_time structure.

The s_vpi_delay and s_vpi_time structures used by both vpi_get_delays() and vpi_put_delays() are defined in vpi_user. \(\overline{\mathrm{h}}\) and \(\overline{\text { are }}\) listed in Figure 174 and Figure 175.
```

pedef struct t_vpi_delay
struct t_vpi_time *da; /* pointer to user allocated array of
delay values */
PLI_INT32 no_of_delays; /* number of delays */
PLI_INT32 time_type; /* [ vpiScaledRealTime, vpiSimTime,
or vpiSuppressTime] */
PLI_INT32 mtm_flag; /* true for mtm values */
PLI_INT32 appēnd_flag; /* true for append */
PLI_INT32 pulsere_flag; /* true for pulsere values */
s_vpi_delay, *p_vpi_delay;

```

Figure 174-The s_vpi_delay structure definition
```

pedef struct t_vpi_time
PLI_INT32 type; /* [ vpiScaledRealTime, vpiSimTime, vpiSuppressTime] *
PLI_UINT32 high, low; /* for vpiSimTime */
dou\overline{ble real; /* for vpiScaledRealTime */}
s_vpi_time, *p_vpi_time;

```

Figure 175-The s_vpi_time structure definition
The \(d a\) field of the s_vpi_delay structure shall be a user-allocated array of s_vpi_time structures.

This array shall store delay values returned by vpi_get_delays(). The number of elements in this array shall be determined by

The number of delays to be retrieved
The mtm_flag setting
The pulsere_flag setting
The number of delays to be retrieved shall be set in the no_of_delays field of the s_vpi_delay structure. Legal values for the number of delays shall be determined by the type of object.

For primitive objects, the no_of_delays value shall be 2 or 3 .
For path delay objects, the no_of_delays value shall be 1,2,3, 6, or 12 .
For timing check objects, the no_of_delays value shall match the number of limits existing in the timing check.
For intermodule path objects, the no_of_delays value shall be 2 or 3 .
The user allocated s_vpi delay array shall contain delays in the same order in which they occur in the Verilog HDL description. The number of elements for each delay shall be determined by the flags mtm_flag and pulsere_flag, as shown in Table 213.

Table 213-Size of the s_vpi_delay->da array
\begin{tabular}{|c|c|c|}
\hline Flag values & Number of s vpi time array elements required for \(\bar{s}-v p i \_d e l a y->d a\) & Order in which delay elements shall be filled \\
\hline \[
\begin{gathered}
\text { mtm_flag }=\text { FALSE } \\
\text { pulsere_flag = FALSE }
\end{gathered}
\] & no_of_delays & \begin{tabular}{l}
1st delay: da[ 0] -> 1st delay \\
2nd delay: da[ 1] -> 2nd delay . . .
\end{tabular} \\
\hline \[
\begin{aligned}
\text { mtm_flag } & =\text { TRUE } \\
\text { pulsere_flag } & =\text { FALSE }
\end{aligned}
\] & 3 * no_of_delays & \begin{tabular}{l}
1st delay: da[ 0] -> min delay \\
da[ 1] -> typ delay \\
da[ 2] -> max delay \\
2nd delay: ...
\end{tabular} \\
\hline \begin{tabular}{l}
\(\mathbf{m t m}\) _flag \(=\) FALSE \\
pulsere_flag = TRUE
\end{tabular} & 3 * no_of_delays & ```
1st delay: da[ 0] -> delay
    da[ 1] -> reject limit
    da[ 2] -> error limit
2nd delay element: ...
``` \\
\hline \[
\mathbf{m t m} \_\mathbf{f l a g}=T R U E
\]
pulsere_flag = TRUE & 9 * no_of_delays & 1st delay: \begin{tabular}{rl} 
da[ 0] & \(->\) min delay \\
\(d a[1]\) & \(->\) typ delay \\
\(d a[2]\) & \(->\) max delay \\
\(d a[3]\) & \(->\) min reject \\
\(d a[4]\) & \(->\) typ reject \\
\(d a[5]\) & \(->\) max reject \\
\(d a[6]\) & \(->\) min error \\
\(d a[7]\) & \(->\) typ error \\
da[ 8] & \(->\) max error
\end{tabular}
2nd delay: ... \\
\hline
\end{tabular}

The delay structure has to be allocated before passing a pointer to vpi_get_delays(). In the following example, a static structure, prim_da, is allocated for use by each call to the vpi_get_delays() function.
```

display_prim_delays(prim)
vpiHandle prim;
{
static s_vpi_time prim_da[ 3] ;
static s_vpi_delay delay_s = { NULL, 3, vpiScaledRealTime} ;
static p_vpi_delay delay_p = \&delay_s;
delay_s.da = prim_da;
vpi_get_delays(prim, delay_p);
vpi_pri\overline{ntf("Delays for primitive %s: %6.2f %6.2f %6.2f\n",}
vpi_get_str(vpiFullName, prim)
delay_p->da[ 0].real, delay_p->da[ 1] .real, delay_p-
>da[ 2].real);
}

```

\subsection*{27.10 vpi_get_str()}


The VPI routine vpi_get_str() shall return string property values. The string shall be placed in a temporary buffer that shall be used by every call to this routine. If the string is to be used after a subsequent call, the string should be copied to another location. Note that a different string buffer shall be used for string values returned through the s_vpi_value structure.

The following example illustrates the usage of vpi_get_str().
```

PLI BYTE8 *str;
vpi\overline{Handle mod = vpi_handle_by name("top.mod1",NULL);}
vpi_printf ("Module top.mod1 is an instance of %s\n",
vpi_get_str(vpiDefName, mod));

```

\subsection*{27.11 vpi_get_systf_info()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{vpi_get_systf_info()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Retrieve information about a user-defined system task/function-related callback.} \\
\hline Syntax: & \multicolumn{3}{|l|}{vpi_get_systf_info(obj, systf_data_p)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & void & & \\
\hline \multirow{3}{*}{Arguments:} & Type & Type Name & Description \\
\hline & vpiHandle & obj & Handle to a system task/function-related callback \\
\hline & p_vpi_systf_data & systf_data_p & Pointer to a structure containing callback information \\
\hline Related routines: & \multicolumn{3}{|l|}{Use vpi_get_cb_info() to retrieve information about a simulation-related callback} \\
\hline
\end{tabular}

The VPI routine vpi_get_systf_info() shall return information about a user-defined system task or function callback in an s_vpi_systf_data structure. The memory for this structure shall be allocated by the user.

The s_vpi_systf_data structure used by vpi_get_systf_info() is defined in vpi_user.h and is listed in Figure 176.
```

ypedef struct t_vpi_systf_data
PLI_INT32 type; /* vpiSysTask, vpiSysFunc * /
PLI_INT32 sysfunctype; /* vpiSysTask, vpi[ Int,Real,Time,Sized,
SizedSigned] Func */
PLI_BYTE8 *tfname; /* first character must be `\$' */
PLI_INT32 (*calltf)(PLI_BYTE8 *);
PLI-INT32 (* compiletf)(\overline{PLI_BYTE8 *);}
PLI_INT32 (*sizetf)(PLI_BYTE8 *); /* for sized function
callbacks only */
PLI_BYTE8 *user_data;
s_vpi_systf_data, *p_vpi_systf_data;

```

Figure 176-The s_vpi_systf_data structure definition

\subsection*{27.12 vpi_get_time()}
\begin{tabular}{||l|l|l||}
\hline \multicolumn{4}{|c|}{ vpi_get_time() } \\
\hline Synopsis: & Retrieve the current simulation time. \\
\hline Syntax: & vpi_get_time (obj, time_p) \\
\hline \multirow{3}{*}{ Returns: } & Type & Description \\
\cline { 2 - 4 } \multirow{3}{*}{ Arguments: } & void & Type \\
\cline { 2 - 4 } & vpiHandle & Name \\
\cline { 2 - 4 } & p_vpi_time & obj \\
\hline \multicolumn{4}{|c|}{\begin{tabular}{l} 
Related \\
routines:
\end{tabular}} & time_p & Pointer to a structure containing time information \\
\hline
\end{tabular}

The VPI routine vpi_get_time() shall retrieve the current simulation time, using the time scale of the object. If obj is NULL, the simulation time is retrieved using the simulation time unit. The time \(p\) - \(>\) type field shall be set to indicate if scaled real or simulation time is desired. The memory for the time \(\quad p\) structure shall be allocated by the user.

The s_vpi_time structure used by vpi_get_time() is defined in vpi_user.h and is listed in Figure 177 [this is the same time structure as used by vpi_put_value()].
```

pedef struct t_vpi_time
PLI_INT32 type; /* [ vpiScaledRealTime, vpiSimTime, vpiSuppressTime]
PLI_UINT32 high, low; /* for vpiSimTime */
double real; /* for vpiScaledRealTime */
s_vpi_time, *p_vpi_time;

```

Figure 177-The s_vpi_time structure definition

\subsection*{27.13 vpi_get_userdata()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{vpi_get_userdata()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get user-data value from an implementation's system task/function instance storage location} \\
\hline Syntax: & \multicolumn{3}{|l|}{vpi_get_userdata (obj)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & void * & \multicolumn{2}{|l|}{user-data value associated with a system task instance or system function instance} \\
\hline \multirow[b]{2}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & vpiHandle & obj & handle to a system task instance or system function instance \\
\hline Related routines: & \multicolumn{3}{|l|}{Use vpi_put_userdata() to write data into the user data storage area} \\
\hline
\end{tabular}

This routine shall return the value of the user-data associated with a previous call to vpi_put_userdata() for a user-defined system task or function call handle. If no user-data had been previously associated with the object, or if the routine fails, the return value shall be NULL.

\subsection*{27.14 vpi_get_value()}


The VPI routine vpi_get_value() shall retrieve the simulation value of VPI objects. The value shall be placed in an s_vpi_value structure, which has been allocated by the user. The format of the value shall be set by the format field of the structure.

When the format field is vpiObjTypeVal, the routine shall fill in the value and change the format field based on the object type, as follows:
```

For an integer, vpiIntVal
For a real, vpiRealVal
For a scalar, either vpiScalar or vpiStrength
For a time variable, vpiTimeVal with vpiSimTime
For a vector, vpiVectorVal

```

The buffer this routine uses for string values shall be different from the buffer that vpi_get_str() shall use. The string buffer used by vpi_get_value() is overwritten with each call. If the value is needed, it should be saved by the application.

The s_vpi_value, s_vpi_vecval and s_vpi_strengthval structures used by vpi_get_value() are defined in vpi_user. h and are listed in Figure 178, Figure 179, and Figure 180.
```

ypedef struct t_vpi_value
PLI_INT32 format; /* vpi[[ Bin,Oct,Dec,Hex] Str,Scalar,Int,Real,String,
Vector,Strength,Suppress,Time,ObjType] Val */
union
{
PLI_BYTE8 *str; /* string value */
PLI_INT32 scalar; /* vpi[ 0,1,X,Z] */
PLI_INT32 integer; /* integer value */
dou\overline{ble real; /* real value */}
struct t_vpi_time *time; /* time value */
struct t_vpi__vecval *vector; /* vector value */
struct t_vpi_strengthval *strength; /* strength value */
PLI_BYTE8 *misc; /* ...other */
} value;
s_vpi_value, *p_vpi_value;

```

Figure 178-The s_vpi_value structure definition
```

ypedef struct t_vpi_vecval
/* following fields are repeated enough times to contain vector */
PLI_INT32 aval, bval; /* bit encoding: ab: 00=0, 10=1, 11=X, 01=Z *,
s_vpi_vecval, *p_vpi_vecval;

```

Figure 179-The s_vpi_vecval structure definition
```

ypedef struct t_vpi_strengthval
PLI_INT32 logic; /* vpi[ 0,1,X,Z] */
PLI_INT32 s0, s1; /* refer to strength coding in the LRM *,
s_vpi_strengthval, *p_vpi_strengthval;

```

Figure 180-The s_vpi_strengthval structure definition
For vectors, the \(p_{-} v p i \_v e c v a l\) field shall point to an array of s_vpi_vecval structures. The size of this array shall be determined by the size of the vector, where array_size \(=((\) vector_size-1)/32 +1\()\). The lsb of the vector shall be represented by the lsb of the 0 -indexed element of s_vpi_vecval array. The 33rd bit of the vector shall be represented by the lsb of the 1 -indexed element of the array, and so on. The memory for the union members str, time, vector, strength, and misc of the value union in the s_vpi_value structure shall be provided by the routine vpi_get_value(). This memory shall only be valid until the next call to vpi_get_value(). [Note that the user must provide the memory for these members when calling vpi_put_value()]. When a value change callback occurs for a value type of vpiVectorVal, the system shall create the associated memory (an array of s_vpi_vecval structures) and free the memory upon the return of the callback.

Table 214-Return value field of the s_vpi_value structure union
\begin{tabular}{|c|c|c|}
\hline Format & Union member & Return description \\
\hline vpiBinStrVal & str & String of binary character(s) \([\mathbf{1 , 0 , \mathbf { x } , \mathbf { z } ]}\) \\
\hline vpiOctStrVal & str & ```
String of octal character(s) [0—7, \(\mathbf{x}, \mathbf{X}, \mathbf{z}, \mathbf{Z}\)
\(\mathbf{x}\) when all the bits are \(\mathbf{x}\)
\(\mathbf{X}\) when some of the bits are \(\mathbf{x}\)
\(\mathbf{z}\) when all the bits are \(\mathbf{z}\)
\(\mathbf{Z}\) when some of the bits are \(\mathbf{z}\)
``` \\
\hline vpiDecStrVal & str & String of decimal character(s) [0-9] \\
\hline vpiHexStrVal & str & \begin{tabular}{l}
String of hex character(s) [0—f, \(\mathbf{x}, \mathbf{X}, \mathbf{z}, \mathbf{Z}\) \(\mathbf{x}\) when all the bits are \(\mathbf{x}\) \\
\(\mathbf{X}\) when some of the bits are \(\mathbf{x}\) \(\mathbf{z}\) when all the bits are \(\mathbf{z}\) \\
\(\mathbf{Z}\) when some of the bits are \(\mathbf{z}\)
\end{tabular} \\
\hline vpiScalarVal & scalar & vpi1, vpi0, vpiX, vpiZ, vpiH, vpiL \\
\hline vpiIntVal & integer & Integer value of the handle. Any bits \(x\) or \(z\) in the value of the object are mapped to a 0 \\
\hline vpiRealVal & real & Value of the handle as a double \\
\hline vpiStringVal & str & A string where each 8-bit group of the value of the object is assumed to represent an ASCII character \\
\hline vpiTimeVal & time & Integer value of the handle using two integers \\
\hline vpiVectorVal & vector & \(a v a l / b v a l ~ r e p r e s e n t a t i o n ~ o f ~ t h e ~ v a l u e ~ o f ~ t h e ~ o b j e c t ~\) \\
\hline vpiStrengthVal & strength & Value plus strength information \\
\hline vpiObjTypeVal & & Return a value in the closest format of the object \\
\hline
\end{tabular}

If the format field in the s_vpi_value structure is set to vpiStrengthVal, the value.strength pointer must point to an array of s_vpi_strengthval structures. This array must have at least as many elements as there are bits in the vector. If the object is a reg or variable, the strength will always be returned as strong.

If the logic value retrieved by vpi_get_value() needs to be preserved for later use, the user must allocate storage and copy the value. The following example can be used to copy a value which was retrieved into an s_vpi_value structure into another structure with user-allocated storage.
```

/*
* Copy s_vpi_value structure - must first allocate pointed to
fields.
* nvalp must be previously allocated.
* Need to first determine size for vector value.
* /
void copy_vpi_value(s_vpi_value *nvalp, s_vpi_value *ovalp,
PLI_INT32 blen, PLI_INT32 nd_alloc)
{
int i;
PLI_INT32 numvals;
nvalp->format = ovalp->format;
switch (nvalp->format) {
/* all string values */
case vpiBinStrVal: case vpiOctStrVal: case vpiDecStrVal:

```
```

    case vpiHexStrVal: case vpiStringVal:
        if (nd_alloc) nvalp->value.str = malloc(strlen(ovalp-
    >value.str)+1);
strcpy(nvalp->value.str, ovalp->value.str);
break;
case vpiScalarVal:
nvalp->value.scalar = ovalp->value.scalar;
break;
case vpiIntVal:
nvalp->value.integer = ovalp->value.integer;
break;
case vpiRealVal:
nvalp->value.real = ovalp->value.real;
break;
case vpiVectorVal:
numvals = (blen + 31) >> 5;
if (nd alloc)
{
nvalp->value.vector = (p_vpi_vecval)
malloc(numvals*sizeof(s_vpi_vecval));
}
/* t_vpi_vecval is really array of the 2 integer a/b sections */
/* memcpy or bcopy better here */
for (i = 0; i <numvals; i++)
nvalp->value.vector[ i] = ovalp->value.vector[ i] ;
break;
case vpiStrengthVal:
if (nd_alloc)
{
nvalp->value.strength = (p vpi strengthval)
malloc(sizeof(s_vpi_strengthval));
}
/* assumes C compiler supports struct assign */
* (nvalp->value.strength) = *(ovalp->value.strength);
break;
case vpiTimeVal:
nvalp->value.time = (p_vpi_time) malloc(sizeof(s_vpi_time));
/* assumes C compiler supports struct assign */
*(nvalp->value.time) = * (ovalp->value.time);
break;
/* not sure what to do here? */
case vpiObjTypeVal: case vpiSuppressVal:
vpi_printf(
"**ERR: can not copy vpiObjTypeVal or vpiSuppressVal formats -
not for filled records.\n");
break;
}
}

```

To get the ASCII values of UDP table entries (see Table 40), the p_vpi_vecval field shall point to an array of s vpi vecval structures. The size of this array shall be determined by the size of the table entry (no. of symbols per table entry), where array_size \(=(\) (table_entry_size-1)/4 +1). Each symbol shall require two bytes; the ordering of the symbols within s_vpi_vecval shall be the most significant byte of abit first, then the least significant byte of abit, then the most significant byte of bbit and then the least significant byte of bbit. Each symbol can be either one or two characters; when it is a single character, the second byte of the pair shall be an ASCII \(\backslash 0\).

Real valued objects shall be converted to an integer using the rounding defined in 3.9 .2 before being returned in a format other than vpiRealVal and vpiStringVal. If the format specified is vpiStringVal then the value shall be returned as a string representation of a floating point number. The format of this string shall be in decimal notation with at most 16 digits of precision.

If a constant object's vpiConstType is vpiStringVal, the value shall be retrieved using either a format of vpiStringVal or vpiVectorVal.

The misc field in the s vpi value structure shall provide for alternative value types, which can be implementation specific. If this field is utilized, one or more corresponding format types shall also be provided.

In the following example, the binary value of each net that is contained in a particular module and whose name begins with a particular string is displayed. [This function makes use of the strcmp () facility normally declared in a string. h C library.]
```

void display certain net values(mod, target)
vpiHandle mod;
PLI_BYTE8 *target;
{
static s_vpi_value value s = { vpiBinStrVal};
static p_vpi_value value_p = \&value_s;
vpiHandle net, itr;
itr = vpi iterate(vpiNet, mod);
while (net = vpi_scan(itr))
{
PLI_BYTE8 *net_name = vpi_get_str(vpiName, net);
if (strcmp(targ}et, net name) == 0
{
vpi_get_value(net, value_p);
vpi_printf("Value of net %s: %s\n",
vpi_get_str(vpiFullName, net),value_p->value.str);
}
}
}

```

The following example illustrates the use of vpi_get_value() to access UDP table entries. Two sample outputs from this example are provided after the example.
```

/*
* hUDP must be a handle to a UDP definition
*/
static void dumpUDPTableEntries(vpiHandle hUDP)
{
vpiHandle hEntry, hEntryIter;
s vpi value value;
P\overline{LI} IN\T32 numb;
PLI_INT32 udpType;
PLI INT32 item;
PLI_INT32 entryVal;
PLI-INT32 *abItem;
PLI INT32 cnt, cnt2;
numb = vpi_get(vpiSize, hUDP);
udpType = vpi_get(vpiPrimType, hUDP);
if (udpType == vpiSeqPrim)
numb++; /* There is one more table entry for state */
numb++; /* There is a table entry for the output */
hEntryIter = vpi_iterate(vpiTableEntry, hUDP);

```
```

    if (!hEntryIter)
    return;
    value.format = vpiVectorVal;
    while(hEntry = vpi_scan(hEntryIter))
    {
                vpi_printf("\n");
                /* Show the entry as a string */
                value.format = vpiStringVal;
                vpi_get_value(hEntry, &value);
                vpi_printf("%s\n", value.value.str);
                /* Decode the vector value format */
                value.format = vpiVectorVal;
                vpi_get value(hEntry, &value);
                abItem = (PLI_INT32 *)value.value.vector;
                for(cnt=((numb}-1)/2+1);cnt>0;cnt--
                    {
                        entryVal = *abItem;
                        abItem++;
                        /* Rip out 4 characters */
                                for (cnt2=0;cnt2<4;cnt2++)
                            {
                        item = entryVal&0xff;
                        if (item)
                        vpi_printf("%c", item);
                            else
                            vpi_printf("_");
                            entry\overline{Val = entryVal>>8;}
                            }
                }
        }
    vpi_printf("\n");
    }

```

For a UDP table of:
\begin{tabular}{lll}
1 & 0 & \(: ?: 1 ;\) \\
0 & \((01)\) & \(: ?:-;\) \\
\((10)\) & 0 & \(: 0: 1 ;\)
\end{tabular}

The output from the preceding example would be:
```

10:1
0_1 __1
01:0
10

```
\(\qquad\)
```

00:1
0 0 1

```

For a UDP table entry of:
\begin{tabular}{lll}
1 & 0 & \(: ?: 1 ;\) \\
0 & \((01)\) & \(: ?:-;\) \\
\((10)\) & 0 & \(: 0: 1 ;\)
\end{tabular}

The output from the preceding example would be:
\[
\begin{aligned}
& \text { 10:?:1 } \\
& 01 \text { ? } \\
& \overline{0}(\overline{0} 1): \bar{?}:- \\
& 100 \text { - ? } \\
& (1 \overline{0}) \overline{0}: \overline{0}: 1 \\
& \text { _001_1_0 }
\end{aligned}
\]

\subsection*{27.15 vpi_get_vlog_info()}


The VPI routine vpi_get_vlog_info() shall obtain the following information about Verilog product execution:

The number of invocation options ( \(\arg c\) )
Invocation option values ( argv)
Product and version strings

The information shall be contained in an s_vpi_vlog_info structure. The routine shall return \(\mathbf{1}\) (true) on success and \(\mathbf{0}\) (false) on failure.

The s_vpi_vlog_info structure used by vpi_get_vlog_info() is defined in vpi_user.h and is listed in Figure \(18 \overline{1}\).
```

typedef struct t_vpi_vlog_info
{
PLI_INT32 argc;
PLI BYTE8 **argv;
PLI-BYTE8 *product;
PLI_BYTE8 *version;
} s_vpi_vlog_info, *p_vpi_vlog_info;

```

Figure 181-The s_vpi_vlog_info structure definition

The format of the \(\arg v\) array is that each pointer in the array shall point to a NULL terminated character array which contains the string located on the tool's invocation command line. There shall be argc entries in the argv array. The value in entry zero shall be the tool's name.

The argument following \(\mathrm{a}-\mathrm{f}\) argument shall contain a pointer to a NULL terminated array of pointers to characters. This new array shall contain the parsed contents of the file. The value in entry zero shall contain the name of the file. The remaining entries shall contain pointers to NULL terminated character arrays containing the different options in the file. The last entry in this array shall be a NULL. If one of the options is \(a-£\) then the next pointer shall behave the same as described above.

\subsection*{27.16 vpi_handle()}


The VPI routine vpi_handle() shall return the object of type type associated with object ref. The one-to-one relationships that are traversed with this routine are indicated as single arrows in the data model diagrams.

The following example application displays each primitive that an input net drives.
```

void display_driven_primitives(net)
vpiHandle net;
{
vpiHandle load, prim, itr;
vpi_printf("Net %s drives terminals of the primitives: \n",
vpi_get_str(vpiFullName, net));
itr = vpi_iterate(vpiLoad, net);
if (!itr)
return;
while (load = vpi_scan(itr))
{
switch(vpi_get(vpiType, load))
{
case vpiGate:
case vpiSwitch:
case vpiUdp:
prim = vpi_handle(vpiPrimitive, load);
vpi_printf("\t%s\n", vpi_get_str(vpiFullName,
prim));
}
}
}

```

\subsection*{27.17 vpi_handle_by_index()}
\begin{tabular}{||l|l|l||}
\hline \multicolumn{4}{|c|}{ vpi_handle_by_index() } \\
\hline Synopsis: & Get a handle to an object using its index number within a parent object. \\
\hline Syntax: & vpi_handle_by_index (obj, index) \\
\hline \multirow{4}{*}{ Returns: } & Type & Description \\
\cline { 2 - 4 } Arguments: & vpiHandle & Handle to an object \\
\cline { 2 - 4 } & vpiHandle & Name \\
\cline { 2 - 4 } & PLI_INT32 & obj \\
\hline
\end{tabular}

The VPI routine vpi_handle_by_index() shall return a handle to an object based on the index number of the object within a parent object. This function can be used to access all objects that can access an expression using vpiIndex. Argument obj shall represent the parent of the indexed object. For example, to access a netbit, obj would be the associated net, while for a memory word, obj would be the associated memory.

\subsection*{27.18 vpi_handle_by_multi_index()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{vpi_handle_by_multi_index()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Obtain a handle to a sub object using an array of indexes and a parent object.} \\
\hline Syntax: & \multicolumn{3}{|l|}{vpi_handle_by_multi_index(obj, num_index, index_array)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & vpiHandle & \multicolumn{2}{|l|}{Handle to an object of type vpiRegBit, vpiNetBit, vpiRegWord, or vpiNetWord} \\
\hline \multirow{4}{*}{Arguments:} & Type & Name & Description \\
\hline & vpiHandle & obj & handle to an object \\
\hline & PLI_INT32 & num_index & number of indexes in the index array \\
\hline & PLI_INT32 * & index_array & array of indexes. Left most index first \\
\hline Related routines: & \multicolumn{3}{|l|}{} \\
\hline
\end{tabular}

The VPI routine vpi_handle_by_multi_index() shall return a handle to an object based on the list of indexes and parent object passed in. The argument num_index will contain the number of indexes in the provided array index_array. The order of the indexes provided, shall be for the left most select first, progressing to the right most select last. This function can be used to access all objects whose property vpiMultiArray is TRUE. This routine shall only provide access to a bit or word of the parent object.

\subsection*{27.19 vpi_handle_by_name()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{vpi_handle_by_name()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Get a handle to an object with a specific name.} \\
\hline Syntax: & \multicolumn{3}{|l|}{vpi_handle_by_name (name, scope)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & vpiHandle & Handle to an obj & \\
\hline \multirow{3}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_BYTE8 * & name & A character string or pointer to a string containing the name of an object \\
\hline & vpiHandle & scope & Handle to a Verilog HDL scope \\
\hline
\end{tabular}

The VPI routine vpi_handle_by_name() shall return a handle to an object with a specific name. This function can be applied to all objects with a fullname property. The name can be hierarchical or simple. If scope is NULL, then name shall be searched for from the top level of hierarchy. If a scope object is provided, then search within that scope only.

\subsection*{27.20 vpi_handle_multi()}


The VPI routine vpi_handle_multi() can be used to return a handle to an object of type vpiInterModPath associated with a list of output port and input port reference objects. The ports shall be of the same size and can be at different levels of the hierarchy.

\subsection*{27.21 vpi_iterate()}


The VPI routine vpi_iterate() shall be used to traverse one-to-many relationships, which are indicated as double arrows in the data model diagrams. The vpi_iterate() routine shall return a handle to an iterator, whose type shall be vpiIterator, which can used by vpi_scan() to traverse all objects of type type associated with object ref. To get the reference object from the iterator object use vpi_handle(vpiUse, iterator_handle). If there are no objects of type type associated with the reference handle ref, then the vpi_iterate() routine shall return NULL.

The following example application uses vpi_iterate() and vpi_scan() to display each net (including the size for vectors) declared in the module. The example assumes it shall be passed a valid module handle.
```

void display_nets(mod)
vpiHandle mo\overline{d};
{
vpiHandle net;
vpiHandle itr;
vpi_printf("Nets declared in module %s\n",
vpi_get_str(vpiFullName, mod));
itr = vpi_iterate(vpiNet, mod);
while (ne\overline{t = vpi_scan(itr))}
{
vpi_printf("\t%s", vpi_get_str(vpiName, net));
if (vpi_get(vpiVector, - net))
{
vpi_printf(" of size %d\n", vpi__get(vpiSize, net));
}
else vpi_printf("\n");
}
}

```

\subsection*{27.22 vpi_mcd_close()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{vpi_mcd_close()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Close one or more files opened by vpi_mcd_open().} \\
\hline Syntax: & \multicolumn{3}{|l|}{vpi_mcd_close (mcd)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type Description} & \\
\hline & PLI_UINT32 & \multicolumn{2}{|l|}{0 if successful, the mcd of unclosed channels if unsuccessful} \\
\hline \multirow[b]{2}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_UINT32 & mcd & A multi-channel descriptor repre \\
\hline Related routines: & \multicolumn{3}{|l|}{\begin{tabular}{l}
Use vpi_mcd_open() to open a file \\
Use vpi_mcd_printf() to write to an opened file Use vpi_mcd_vprintf() to write a variable number of arguments to an opened file Use vpi_mcd_flush() to flush a file output buffer Use vpi_mcd_name() to get the name of a file represented by a channel descriptor
\end{tabular}} \\
\hline
\end{tabular}

The VPI routine vpi_mcd_close() shall close the file(s) specified by a multi-channel descriptor, mcd. Several channels can be closed simultaneously, since channels are represented by discrete bits in the integer \(m c d\). On success this routine shall return a 0 ; on error it shall return the \(m c d\) value of the unclosed channels. This routine can also be used to close file descriptors which were opened using the system function \$fopen(). See 17.2.1 for the functional description of \$fopen().

The following descriptors are predefined, and cannot be closed using vpi_mcd_close():
descriptor 1 is for the output channel of the software product which invoked the PLI application and the current log file

\subsection*{27.23 vpi_mcd_flush()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{vpi_mcd_flush()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Flushes the data from the given MCD output buffers.} \\
\hline Syntax: & \multicolumn{3}{|l|}{vpi_mcd_flush (mcd)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{0 if successful, non-zero if unsuccessful} \\
\hline \multirow[b]{2}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & PLI_UINT32 & mcd & A multi-channel descriptor repr to write \\
\hline Related routines: & \multicolumn{3}{|l|}{Use vpi_mcd_printf() to write a finite number of arguments to an opened file Use vpi_mcd_vprintf() to write a variable number of arguments to an opened file Use vpi_mcd_open() to open a file Use vpi_mcd_close() to close a file Use vpi_mcd_name() to get the name of a file represented by a channel descriptor} \\
\hline
\end{tabular}

The routine vpi_mcd_flush() shall flush the output buffers for the file(s) specified by the multi-channel descriptor, mcd.

\subsection*{27.24 vpi_mcd_name()}


The VPI routine vpi_mcd_name() shall return the name of a file represented by a single-channel descriptor, \(c d\). On error, the routine shall return NULL. This routine shall overwrite the returned value on subsequent calls. If the application needs to retain the string, it should copy it. This routine can be used to get the name of any file, opened using the system function \$fopen or the VPI routine vpi_mcd_open(). The channel descriptor cd could be an fd file descriptor returned from \$fopen (indicated by the most significant bit being set) or an mcd multi-channel descriptor returned by either the system function \$fopen or the VPI routine vpi_mcd_open(). See 17.2.1 for the functional description of \$fopen.

\subsection*{27.25 vpi_mcd_open()}


The VPI routine vpi_med_open() shall open a file for writing and shall return a corresponding multi-channel description number (mcd). The channel descriptor 1 (least significant bit) is reserved for representing the output channel of the software product which invoked the PLI application and the log file (if one is currently open). The channel descriptor 32 (most significant bit) is reserved to represent a file descriptor (fd) returned from the Verilog HDL \$fopen system function.

The mod descriptor returned by vpi_mcd_open() routine is compatible with the mcd descriptors returned from the \$fopen system function. The mcd descriptors returned from vpi_mcd_open() and from \$fopen may be shared between the HDL system tasks which use mod descriptors and the VPI routines which use mcd descriptors. Note that the \$fopen system function can also return fd file descriptors (indicated by the most significant bit being set). An fd is not compatible with the med descriptor returned by vpi_mcd_open(). See 17.2.1 for the functional description of \$fopen.

The vpi_mcd_open() routine shall return a 0 on error. If the file has already been opened either by a previous call to vpi_mcd_open() or using \$fopen in the Verilog source code, then vpi_mcd_open(), shall return the descriptor number.

\subsection*{27.26 vpi_mcd_printf()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{vpi_mcd_printf()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Write to one or more files opened with vpi_mcd_open() or \$fopen.} \\
\hline Syntax: & \multicolumn{3}{|l|}{vpi_mcd_printf(mcd, format, ...)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_INT32 & \multicolumn{2}{|l|}{The number of characters written} \\
\hline \multirow{3}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_UINT32 & mcd & A multi-channel descriptor repre to write \\
\hline & PLI_BYTE8 * & format & A format string using the C fprin \\
\hline Related routines: & \multicolumn{3}{|l|}{\begin{tabular}{l}
Use vpi_mcd_vprintf() to write a variable number of arguments to an opened file \\
Use vpi_mcd_open() to open a file \\
Use vpi_mcd_close() to close a file \\
Use vpi_mcd_flush() to flush a file output buffer \\
Use vpi_mcd_name() to get the name of a file represented by a channel descriptor
\end{tabular}} \\
\hline
\end{tabular}

The VPI routine vpi_mcd_printf() shall write to one or more channels (up to 31 ) determined by the mcd. An mcd of 1 (bit 0 set) corresponds to the channel 1 , an mcd of 2 (bit 1 set) corresponds to channel 2, an mcd of 4 (bit 2 set) corresponds to channel 3, and so on. Channel 1 is reserved for the output channel of the software product which invoked the PLI application and the current log file. The most significant bit of the descriptor is reserved by the tool to indicate that the descriptor is actually a file descriptor instead of an mcd. vpi_mcd_printf() shall also write to a file represented by an mcd which was returned from the Verilog HDL \$fopen system function. vpi_mcd_printf() shall not write to a file represented by an fd file descriptor returned from \$fopen (indicated by the most significant bit being set). See 17.2 .1 for the functional description of \$fopen

Several channels can be written to simultaneously, since channels are represented by discrete bits in the integer mcd.

The text written shall be controlled by one or more format strings. The format strings shall use the same format as the C fprintf() routine. The routine shall return the number of characters printed, or EOF if an error occurred.

\subsection*{27.27 vpi_mcd_vprintf()}


This routine performs the same function as vpi_mcd_printf(), except that varargs has already been started.

\subsection*{27.28 vpi_printf()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{vpi_printf()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Write to the output channel of the software product which invoked the PLI application and the current product \(\log\) file.} \\
\hline Syntax: & \multicolumn{3}{|l|}{vpi_printf(format, ...)} \\
\hline \multirow[b]{2}{*}{Returns:} & \multicolumn{2}{|l|}{Type Description} & \\
\hline & PLI_INT32 & \multicolumn{2}{|l|}{The number of characters written} \\
\hline \multirow[b]{2}{*}{Arguments:} & \multicolumn{2}{|l|}{Type Name} & Description \\
\hline & PLI_BYTE8 * & format & A format string using the C prin \\
\hline Related routines: & \multicolumn{3}{|l|}{Use vpi_vprintf() to write a variable number of arguments Use vpi_mcd_printf() to write to an opened file Use vpi_mcd_flush() to flush a file output buffer Use vpi_mcd_vprintf() to write a variable number of arguments to an opened file} \\
\hline
\end{tabular}

The VPI routine vpi_printf() shall write to both the output channel of the software product which invoked the PLI application and the current product log file. The format string shall use the same format as the C printf () routine. The routine shall return the number of characters printed, or EOF if an error occurred.

\subsection*{27.29 vpi_put_data()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{vpi_put_data()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Put data into an implementation's save/restart location} \\
\hline Syntax: & \multicolumn{3}{|l|}{vpi_put_data(id, dataLoc, numOfBytes)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & PLI_INT32 & PLI_INT32 The number of bytes written & \\
\hline \multirow{4}{*}{Arguments:} & Type & Name & Description \\
\hline & PLI_INT32 & id & A save/restart ID returned from vpi_get(vpiSaveRestartID, NULL) \\
\hline & PLI_BYTE8 * & dataLoc & Address of application allocated storage \\
\hline & PLI_INT32 & numOfBytes & Number of bytes to be added to save/restart location \\
\hline Related routines: & \multicolumn{3}{|l|}{Use vpi_get_data() to retrieve saved data} \\
\hline
\end{tabular}

This routine shall place numOfBytes, which must be greater than zero, of data located at dataLoc into an implementation's save/restart location. The return value shall be the number of bytes written. A zero shall be returned if an error is detected. There shall be no restrictions:
on how many times the routine can be called for a given id. on the order applications put data using the different ids.

The data from multiple calls to vpi_put_data() with the same \(\boldsymbol{i d}\) shall be stored by the simulator in such a way that the opposing routine vpi_get_data() can pull data out of the save/restart location using different size chunks. This routine can only be called from a user application routine that has been called for the reason cbStartOfSave or cbEndOfSave. A user can get the path to the implementation's save/restart location by calling vpi_get_str(vpiSaveRestartLocation, NULL) from a user application routine that has been called for reason cbStartOfSave or cbEndOfSave.

The following example illustrates using vpi_put_data():
```

/* example of how to place data into a save/restart location */
struct myStruct{
struct myStruct * next;
PLI_INT32 d1;
PLI_INT32 d2;
}
struct myStruct *firstWrk; /* This data structure created
elsewhere. */
PLI_INT32 consumer_save(p_cb_data data)
{
struct myStruct *wrk;
s_cb_data cbData;
v\overline{p}iHāndle cbHdl;
PLI_INT32 id = 0;
PLI_INT32 cnt = 0;

```
```

    /* Get the number of structures */
    wrk = firstWrk;
    while (wrk)
        {
            cnt++;
            wrk = wrk->next;
    }
    /* now save the data */
    wrk = firstWrk;
/* save the number of data structures */
id = vpi_get(vpiSaveRestartID, NULL);
/*
* save the different data structures. Please note that
* a pointer is being saved. While this is allowed an
* application must change it to something useful on a restart.
* /
while (wrk)
{
wrk = wrk->next;
}
/* register a call for restart */
/*
* We need the "id" so that the saved data can be retrieved.
* Using the user_data field of the callback structure is the
* easiest way to pass this information to retrieval
operation.
* /
cbData.user_data = (PLI_BYTE8 *)id;
cbData.reasōn = cbStartŌfRestart;
cbData.cb_rtn = consumer_restart; /*
* Please see
vpi_get_data()
* for a description of
this
* routine.
* /
cbData.value = NULL;
cbData.time = NULL;
cbHdl = vpi_register_cb(\&cbData);
vpi_free_object(cbHd\overline{l});
retürn(0);
}

```
27.30 vpi_put_delays()


The VPI routine vpi_put_delays() shall set the delays or timing limits of an object as indicated in the delay_p structure. The same ordering of delays shall be used as described in the vpi_get_delays() function. If only the delay changes, and not the pulse limits, the pulse limits shall retain the values they had before the delays where altered.

The s_vpi_delay and s_vpi_time structures used by both vpi_get_delays() and vpi_put_delays() are defined in vpi_user. \(\overline{\mathrm{h}}\) and \(\overline{\text { are }}\) listed in Figure 182 and Figure 183.
```

sedef struct t_vpi_delay
;truct t_vpi_time *da; /* pointer to user allocated array of delay values
'LI_INT32}\mathrm{ no_of_delays; /* number of delays */
'LI_INT32 time_type; /* [ vpiScaledRealTime,vpiSimTime,vpiSuppressTime]
'LI_INT32 mtm_\overline{flag; /* true for mtm values */}

```

```

'LI_INT32 pulsere_flag; /* true for pulsere values */
;_vpi_delay, *p_vpi_delay;

```

Figure 182-The s_vpi_delay structure definition
```

\jmathedef struct t_vpi_time
?LI_INT32 type; /* [ vpiScaledRealTime, vpiSimTime, vpiSuppressTime] *,
?LI_UINT32 high, low; /* for vpiSimTime */
double real; /* for vpiScaledRealTime */
s_vpi_time, *p_vpi_time;

```

Figure 183-The s_vpi_time structure definition
The \(d a\) field of the s_vpi_delay structure shall be a user-allocated array of s_vpi_time structures. This array stores the delay values to be written by vpi_put_delays(). The number of elements in this array is determined by:

The number of delays to be written
The mtm_flag setting
The pulsere_flag setting
The number of delays to be set shall be set in the no_of_delays field of the s_vpi_delay structure. Legal values for the number of delays shall be determined by the type of object.

For primitive objects, the no_of_delays value shall be 2 or 3 .
For path delay objects, the no_of_delays value shall be 1, 2, 3, 6, or 12 .
For timing check objects, the no_of_delays value shall match the number of limits existing in the timing check.
For intermodule path objects, the no_of_delays value shall be 2 or 3 .

The user allocated s_vpi_delay array shall contain delays in the same order in which they occur in the Verilog HDL description. The number of elements for each delay shall be determined by the flags mtm_flag and pulsere_flag, as shown in Table 215.

Table 215-Size of the s_vpi_delay->da array
\begin{tabular}{|c|c|c|}
\hline Flag values & Number of s vpi time array elements required for \(\bar{s}\) - \(v p i \_d e l a y->d a\) & Order in which delay elements shall be filled \\
\hline \[
\begin{gathered}
\text { mtm_flag }=\text { FALSE } \\
\text { pulsere_flag }=\text { FALSE }
\end{gathered}
\] & no_of_delays & \begin{tabular}{l}
1st delay: da[ 0] -> 1st delay \\
2nd delay: da[ 1] -> 2nd delay
\end{tabular} \\
\hline \[
\begin{gathered}
\text { mtm_flag }=\text { TRUE } \\
\text { pulsere_flag }=\text { FALSE }
\end{gathered}
\] & 3 * no_of_delays & \begin{tabular}{l}
1st delay: da[ 0] \(->\) min delay \\
da[ 1] -> typ delay \\
da[ 2] -> max delay \\
2nd delay: ...
\end{tabular} \\
\hline \(\mathbf{m t m}\) _flag \(=\) FALSE pulsere_flag = TRUE & 3 * no_of_delays & ```
1st delay: da[ 0] -> delay
    da[ 1] -> reject limit
    da[ 2] -> error limit
2nd delay element: ...
``` \\
\hline \[
\begin{gathered}
\text { mtm_flag }=\text { TRUE } \\
\text { pulsere_flag }=\text { TRUE }
\end{gathered}
\] & 9 * no_of_delays &  \\
\hline
\end{tabular}

The following example application accepts a module path handle, rise and fall delays, and replaces the delays of the indicated path.
```

void set_path_rise_fall_delays(path, rise, fall)
vpiHandle pat\overline{h}
double rise, fall;
{
static s_vpi_time path_da[ 2] ;
static s_vpi_delay delay_s = {NULL, 2, vpiScaledRealTime};
static p_vpi_delay delay_p = \&delay_s;
delay_s.da = path_da;
path_da[ 0].real = rise;
path_da[ 1].real = fall;
vpi_put_delays(path, delay_p);
}

```

\subsection*{27.31 vpi_put_userdata()}


This routine will associate the value of the input userdata with the specified user-defined system task or function call handle. The stored value can later be retrieved with the routine vpi_get_userdata(). The routine will return a value of 1 on success or a 0 if it fails.

\subsection*{27.32 vpi_put_value()}
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{vpi_put_value()} \\
\hline Synopsis: & \multicolumn{3}{|l|}{Set a value on an object.} \\
\hline Syntax: & \multicolumn{3}{|l|}{vpi_put_value (obj, value_p, time_p, flags)} \\
\hline \multicolumn{4}{|c|}{Type Description} \\
\hline Returns: & vpiHandle & \multicolumn{2}{|l|}{Handle to the scheduled event caused by vpi_put_value()} \\
\hline \multirow{5}{*}{Arguments:} & Type & Name & Description \\
\hline & vpiHandle & obj & Handle to an object \\
\hline & p_vpi_value & value_p & Pointer to a structure with value information \\
\hline & p_vpi_time & time_p & Pointer to a structure with delay information \\
\hline & PLI_INT32 & flags & Integer constants that set the delay mode \\
\hline Related routines: & \multicolumn{3}{|l|}{Use vpi_get_value() to retrieve the value of an expression} \\
\hline
\end{tabular}

The VPI routine vpi_put_value() shall set simulation logic values on an object. The value to be set shall be stored in an s_vpi_value structure that has been allocated. The legal values which may be specified for each value format are listed in Table 214. The delay time before the value is set shall be stored in an s_vpi_time structure that has been allocated. The routine can be applied to nets, regs, variables, variable selects, memory words, named events, system function calls, sequential UDPs, and scheduled events. The flags argument shall be used to direct the routine to use one of the following delay modes:
vpiInertialDelay
vpiTransportDelay
vpiPureTransportDelay
vpiNoDelay
vpiForceFlag
vpiReleaseFlag
vpiCancelEvent

All scheduled events on the object shall be removed before this event is scheduled.

All events on the object scheduled for times later than this event shall be removed (modified transport delay).

No events on the object shall be removed (transport delay).
The object shall be set to the passed value with no delay. Argument time_p shall be ignored and can be set to NULL.
The object shall be forced to the passed value with no delay (same as the Verilog HDL procedural force). Argument time \(\_p\) shall be ignored and can be set to NULL.

The object shall be released from a forced value (same as the Verilog HDL procedural release). Argument time_p shall be ignored and can be set to NULL. The value \(p\) shall be updated with the value of the object after its release.

A previously scheduled event shall be cancelled. The object passed to vpi_put_value() shall be a handle to an object of type vpiSchedEvent.

If the flags argument also has the bit mask vpiReturnEvent, vpi_put_value() shall return a handle of type vpiSchedEvent to the newly scheduled event, provided there is some form of a delay and an event is scheduled. If the bit mask is not used, or if no delay is used, or if an event is not scheduled, the return value shall be NULL.

The handle to the event can be cancelled by calling vpi_put_value() with the flag set to vpiCancelEvent. The value_ \(p\) and time_p arguments to vpi_put_value() are not needed for cancelling an event, and can be set to NULL. It shall not be an error to cancel an event that has already occurred. The scheduled event can be tested by calling vpi_get() with the flag vpiScheduled. If an event is cancelled, it shall simply be removed from the event queue. Any effects that were caused by scheduling the event shall remain in effect (e.g., events that where cancelled due to inertial delay).

Calling vpi_free_object() on the handle shall free the handle but shall not affect the event.
When vpi_put_value() is called for an object of type vpiNet or vpiNetBit, and with modes of vpiInertialDelay, vpiTransportDelay, vpiPureTransportDelay, or vpiNoDelay, the value supplied overrides the resolved value of the net. This value shall remain in effect until one of the drivers of the net changes value. When this occurs, the net shall be re-evaluated using the normal resolution algorithms.

It shall be illegal to specify the format of the value as vpiStringVal when putting a value to a real variable or a system function call of type vpiRealFunc. It shall be illegal to specify the format of the value as vpiStrengthVal when putting a value to a vector object.

When vpi_put_value() with a vpiForce flag is used, it shall perform a procedural force of a value onto the same types of objects as supported by a procedural force. A vpiRelease flag shall release the forced value. This shall be the same functionality as the procedural force and release keywords in the Verilog HDL (refer to 9.3.2)

Sequential UDPs shall be set to the indicated value with no delay regardless of any delay on the primitive instance. Putting values to UDP instances must be done using the vpiNoDelay flag. Attempting to use the other delay modes shall result in an error.

Calling vpi_put_value() on an object of type vpiNamedEvent shall cause the named event to toggle. Objects of type vpiNamedEvent shall not require an actual value and the value_p argument may be NULL.

The vpi_put_value() routine shall also return the value of a system function by passing a handle to the userdefined system function as the object handle. This should only occur during execution of the calltf routine for the system function. Attempts to use vpi_put_value() with a handle to the system function when the calltf routine is not active shall be ignored. Should the calltf routine for a user defined system function fail to put a value during its execution, the default value of 0 will be applied. Putting return values to system functions must be done using the vpiNoDelay flag.

The vpi_put_value() routine shall only return a system function value in a calltf application, when the call to the system function is active. The action of vpi_put_value() to a system function shall be ignored when the system function is not active. Putting values to system function must be done using the vpiNoDelay flag.

The s_vpi_value and s_vpi_time structures used by vpi_put_value() are defined in vpi_user.h and are listed in Figure 184 and Figure 185.
```

ypedef struct t_vpi_value
PLI_INT32 format; /* vpi[[ Bin,Oct,Dec,Hex] Str,Scalar,Int,Real,String,
Vector,Strength,Suppress,Time,ObjType] Val */
union
{
PLI_BYTE8 *str; /* string value */
PLI_INT32 scalar; /* vpi[ 0,1,X,Z] */
PLI_INT32 integer; /* integer value */
double real; /* real value */
struct t_vpi_time *time; /* time value */
struct t_vpi_vecval *vector; /* vector value */
struct t_vpi_strengthval *strength; /* strength value */
PLI_BYTE8 *misc; /* ...other */
} valūe;
s_vpi_value, *p_vpi_value;

```

Figure 184-The s_vpi_value structure definition
```

rpedef struct t_vpi_time
PLI_INT32 type; /* [ vpiScaledRealTime, vpiSimTime, vpiSuppressTime] *
PLI_UINT32 high, low; /* for vpiSimTime */
dou\overline{ble real; /* for vpiScaledRealTime */}
s_vpi_time, *p_vpi_time;

```

Figure 185-The s_vpi_time structure definition
```

rpedef struct t_vpi_vecval
/* following fields are repeated enough times to contain vector */
PLI_INT32 aval, bval; /* bit encoding: ab: 00=0, 10=1, 11=X, 01=Z ,
s_vpi_vecval, *p_vpi_vecval;

```

Figure 186-The s_vpi_vecval structure definition
```

ypedef struct t_vpi_strengthval
PLI_INT32 logic; /* vpi[ 0,1,X,Z] */
PLI_INT32 s0, s1; /* refer to strength coding below */
s_vpi_strengthval, *p_vpi_strengthval;

```

Figure 187-The s_vpi_strengthval structure definition
For vpiScaledRealTime, the indicated time shall be in the timescale associated with the object.

\subsection*{27.33 vpi_register_cb()}


The VPI routine vpi_register_cb() is used for registration of simulation-related callbacks to a user-provided application for a variety of reasons during a simulation. The reasons for which a callback can occur are divided into three categories:

Simulation event Simulation time Simulation action or feature

How callbacks are registered for each of these categories is explained in the following paragraphs.
The \(c b \_d a t a \_p\) argument shall point to a s_cb_data structure, which is defined in vpi_user. h and given in Figure 188.
```

'pedef struct t_cb_data
PLI_INT32 reason; /* callback reason * /
PLI_INT32 (*cb_rtn)(struct t_cb_data *); /* call routine */
vpiHandle obj; /* trigger object */
p_vpi_time time; /* callback time */
p_vpi_value value; /* trigger object value */
PLI_INTT32 index; /* index of the memory word or var select
that changed */
PLI_BYTE8 *user_data;
s_cb_data, *p_cb_data;

```

\section*{Figure 188-The s_cb_data structure definition}

For all callbacks, the reason field of the s cb data structure shall be set to a predefined constant, such as cbValueChange, cbAtStartOfSimTime, \(\overline{\mathbf{c} b E} \bar{n}\) dOfCompile, etc. The reason constant shall determine when the user application shall be called back. Refer to the vpi_user. h file listing in Annex \(G\) for a list of all callback reason constants.

The \(c b \_r t h\) field of the \(s \_c b \_d a t a\) structure shall be set to the application routine, which shall be invoked when the simulator executes the callback. The use of the remaining fields are detailed in the following subsections.

\subsection*{27.33.1 Simulation-event-related callbacks}

The vpi_register_cb() callback mechanism can be registered for callbacks to occur for simulation events, such as value changes on an expression or terminal, or the execution of a behavioral statement. When the cb_data_p->reason field is set to one of the following, the callback shall occur as described below:
\begin{tabular}{ll} 
cbValueChange & \begin{tabular}{l} 
After value change on an expression or terminal, or execution of an event \\
statement
\end{tabular} \\
cbStmt & Before execution of a behavioral statement \\
cbForce/cbRelease & After a force or release has occurred \\
cbAssign/cbDeassign & \begin{tabular}{l} 
After a procedural assign or deassign statement has been executed
\end{tabular} \\
cbDisable & \begin{tabular}{l} 
After a named block or task containing a system task or function has been \\
disabled
\end{tabular}
\end{tabular}

The following fields shall need to be initialized before passing the s_cb_data structure to vpi_register_cb():
\begin{tabular}{ll}
\(c b \_\)_data_p->obj & \begin{tabular}{l} 
This field shall be assigned a handle to an expression, terminal, or \\
statement for which the callback shall occur. For force and release \\
callbacks, if this is set to NULL, every force and release shall generate a
\end{tabular} \\
callback.
\end{tabular}

Table 216-Value format field of cb_data_p->value->format
\begin{tabular}{|c|l|}
\hline Format & \multicolumn{1}{|c|}{ Registers a callback to return } \\
\hline vpiBinStrVal & String of binary character(s) \([\mathbf{1 , 0 , \mathbf { x } , \mathbf { z } ]}\) \\
\hline vpiOctStrVal & String of octal character(s) \([\mathbf{0}-\mathbf{7}, \mathbf{x}, \mathbf{X}, \mathbf{z}, \mathbf{Z}\) \\
\hline vpiDecStrVal & String of decimal character(s) \([0-9]\) \\
\hline vpiHexStrVal & String of hex character(s) \([\mathbf{0}-\mathbf{f}, \mathbf{x}, \mathbf{x , ~ \mathbf { z } , \mathbf { 4 }}\) \\
\hline vpiScalarVal & vpi1, vpi0, vpiX, vpiZ, vpiH, vpiL \\
\hline vpiIntVal & Integer value of the handle \\
\hline vpiRealVal & Value of the handle as a double \\
\hline vpiStringVal & An ASCII string \\
\hline vpiTimeVal & Integer value of the handle using two integers \\
\hline vpiVectorVal & aval/bval representation of the value of the object \\
\hline vpiStrengthVal & Value plus strength information of a scalar object only \\
\hline vpiObjectVal & Return a value in the closest format of the object \\
\hline
\end{tabular}

When a simulation event callback occurs, the user application shall be passed a single argument, which is a pointer to an s cb data structure [this is not a pointer to the same structure that was passed to vpi_register_cb()]. The time and value information shall be set as directed by the time type and value format fields in the call to vpi_register_cb(). The user_data field shall be equivalent to the user_data field passed to vpi_register_cb(). The user application can use the information in the passed structure and information retrieved from other VPI interface routines to perform the desired callback processing.
cbValueChange callbacks can be placed onto event statements. When the event statement is executed, the callback routine will be called. Since event statements do not have a value, when the callback routine is called, the value field of the s_cb_data structure will be NULL.

For a cbValueChange callback, if the obj is a memory or a variable array, the value in the s_cb_data structure shall be the value of the memory word or variable select that changed value. The index field shall contain the index of the memory word or variable select that changed value. If a cbValueChange callback is registered and the format is set to vpiStrengthVal then the callback shall occur whenever the object changes strength, including changes that do not result in a value change.

For cbForce, cbRelease, cbAssign and cbDeassign callbacks, the object returned in the obj field shall be a handle to the force, release, assign or deassign statement. The value field shall contain the resultant value of the LHS expression. In the case of a release, the value field shall contain the value after the release has occurred.

For a cbDisable callback, obj shall be a handle to a system task call, system function call, named begin, named fork, task, or function.

It is illegal to attempt to place a callback for reason cbForce, cbRelease, or cbDisable on a variable bit select.

The following example shows an implementation of a simple monitor functionality for scalar nets, using a simulation-event-related callback.
```

setup_monitor(net)
vpiHandle net;
{
static s vpi time time s = { vpiSimTime} ;
static s_vpi_value value_s = {vpiBinStrVal};
static s_cb_data cb_data_s =
{ cbValue\overline{Change, my_monitor, NULL, \&time_s, \&value_s};}
PLI_BYTE8 *net_name = vpi_get_str(vpiFullNäme, net);
cb_data_s.obj = net;
cb_data_s.user_data = malloc(strlen(net_name)+1);
strcpy(cb_data_s.user_data, net_name);
vpi_registerer_c\overline{b}(\&cb_dāta_s);
}
my_monitor(cb_data_p)
p_\overline{cb_data cb_\overline{data_\overline{p}; {}}\mathbf{C}=\mp@code{l}
vpi_printf("%d %d: %s = %s\n",
cb_data_p->time->high, cb_data_p->time->low,
cb_data_p->user_data,
cb_data_p->value->value.str);
}

```

\subsection*{27.33.1.1 Callbacks on Individual Statements}

When cbStmt is used in the reason field of the s_cb_data structure, the other fields in the structure will be defined as follows:
\begin{tabular}{ll}
\(c b \_d a t a \_p->c b \_r t n\) & The function to call before the given statement executes. \\
\(c b \_\)data_p->obj & \begin{tabular}{l} 
A handle to the statement on which to place the callback (the allowable \\
objects are listed in Table 217).
\end{tabular} \\
\(c b \_\)data_p->time & \begin{tabular}{l} 
A pointer to an s_vpi_time structure, wherein only the type is used, to \\
indicate the type of time which will be returned when the callback is \\
made. This type can be vpiScaledRealTime, vpiSimTime, or \\
vpiSuppressTime if no time information is needed by the callback \\
routine.
\end{tabular} \\
\(c b \_d a t a \_p->\) value & Not used. \\
\(c b \_d a t a \_p->\) index & Not used. \\
\(c b \_d a t a \_p->u s e r \_d a t a\) & Data to be passed to the callback function.
\end{tabular}

Just before the indicated statement executes, the indicated function will be called with a pointer to a new s_cb_data structure, which will contain the following informations:
cb_data_p->reason cbStmt.
cb_data_p->cb_rtn The same value as that passed to vpi_register_cb().
cb_data_p->obj
cb_data_p->time A pointer to an s_vpi_time structure, which will contain the current simulation time, of the type (vpiScaledRealTime or vpiSimTime) indicated in the call to vpi_register_cb(). If the value in the call to vpi_register_cb() was vpiSuppressTime, then the time pointer in the s_cb_data structure will be set to NULL.
\[
\begin{array}{ll}
\text { cb_data_p->value } & \text { always NULL. } \\
\text { cb_data_p->index } & \text { always set to } 0 . \\
\text { cb_data_p->user_data } & \text { The value passed in as } u \text { user_data in the call to vpi_register_cb(). }
\end{array}
\]

Multiple calls to vpi_register_cb() with the same data shall result in multiple callbacks.
Placing callbacks on statements which reside in protected portions of the code shall not be allowed, and shall cause vpi_register_cb() to return a NULL, with an appropriate error message printed.

\subsection*{27.33.1.2 Behavior by Statement Type}

Every possible object within the stmt class qualifies for having a cbStmt callback placed on it. Each possible object is listed in Table 217, for further clarification.

\section*{Table 217-cbStmt callbacks}
\begin{tabular}{|l|l|}
\hline \begin{tabular}{l} 
vpiBegin \\
vpiNamedBegin \\
vpiFork \\
vpiNamedFork
\end{tabular} & \begin{tabular}{l} 
One callback will occur prior to any of the statements within the block execut- \\
ing. The handle returned in the obj field will be the handle to the block object.
\end{tabular} \\
\hline \begin{tabular}{l} 
vpiIf \\
vpiIfElse
\end{tabular} & \begin{tabular}{l} 
The callback will occur before the condition expression in the if statement is \\
evaluated.
\end{tabular} \\
\hline vpiWhile & \begin{tabular}{l} 
A callback will occur prior to the evaluation of the condition expression on \\
every iteration of the loop.
\end{tabular} \\
\hline vpiRepeat & \begin{tabular}{l} 
A callback will occur when the repeat statement is first encountered, and on \\
every subsequent iteration of the repeat loop.
\end{tabular} \\
\hline vpiFor & \begin{tabular}{l} 
A callback will occur prior to any of the control expressions being evaluated. \\
Then on every iteration of the loop, a callback will occur prior to the evaluation \\
of the incremental statement.
\end{tabular} \\
\hline vpiForever & \begin{tabular}{l} 
A callback will occur when the forever statement is first encountered, and on \\
every subsequent iteration of the forever loop.
\end{tabular} \\
\hline \begin{tabular}{l} 
vpiWait \\
vpiCase \\
vpiAssignment \\
vpiAssignStmt \\
vpiDeassign \\
vpiDisable \\
vpiForce \\
viRelease \\
vpiEventStmt
\end{tabular} & \begin{tabular}{l} 
The callback will occur before the statement executes.
\end{tabular} \\
\hline vpiDelayControl & \begin{tabular}{l} 
The callback will occur when the delay control is encountered, before the delay \\
occurs.
\end{tabular} \\
\hline vpiEventControl & \begin{tabular}{l} 
The callback will occur when the event control is encountered, before the event \\
has occurred.
\end{tabular} \\
\hline \begin{tabular}{l} 
vpiTaskCall \\
vpiSysTaskCall
\end{tabular} & The callback will occur before the given task is executed. \\
\hline
\end{tabular}

\subsection*{27.33.1.3 egistering Callbacks on a Module-wide Basis}
vpi_register_cb() allows a handle to a module instance in the objfield of the s_cb_data structure. When this is done, the effect will be to place a callback on every statement which can have a callback placed on it.

When using vpi_register_cb() on a module object, the call will return a handle to a single callback object which can be passed to vpi_remove_cb() to remove the callback on every statement in the module instance.

Statements which reside in protected portions of the code shall not have callbacks placed on them.

\subsection*{27.33.2 Simulation-time-related callbacks}

The vpi_register_cb() can register callbacks to occur for simulation time reasons, include callbacks at the beginning or end of the execution of a particular time queue. The following time-related callback reasons are defined:
\begin{tabular}{ll} 
cbAtStartOfSimTime & \begin{tabular}{l} 
Callback shall occur before execution of events in a specified time queue. \\
A callback can be set for any time, even if no event is present.
\end{tabular} \\
cbReadWriteSynch & Callback shall occur after execution of events for a specified time. \\
cbReadOnlySynch & \begin{tabular}{l} 
Same as cbReadWriteSynch, except that writing values or scheduling \\
events before the next scheduled event is not allowed.
\end{tabular} \\
cbNextSimTime & Callback shall occur before execution of events in the next event queue. \\
cbAfterDelay & \begin{tabular}{l} 
Callback shall occur after a specified amount of time, before execution of \\
events in a specified time queue. A callback can be set for anytime, even if \\
no event is present.
\end{tabular}
\end{tabular}

For reason cbNextSimTime, the time field in the time structure is ignored. The following fields shall need to be set before passing the s_cb_data structure to vpi_register_cb():
\[
\begin{array}{ll}
c b \_d a t a \_p->\text { time->type } & \begin{array}{l}
\text { This field shall be set to either vpiScaledRealTime or vpiSimTime, } \\
\text { depending on what time information the user application requires during }
\end{array} \\
\text { the callback. vpiSuppressTime (or NULL for the cb_data_p->time } \\
\text { field) will result in an error. }
\end{array}
\]

The following situations will generate an error and no callback will be created:
Attempting to place a cbAtStartOfSimTime callback with a delay of zero when simulation has progressed into a time slice, and the application is not currently within a cbAtStartOfSimTime callback.
Attempting to place a cbReadWriteSynch callback with a delay of zero at read-only synch time.
Placing a callback for cbAtStartOfSimTime and a delay of zero during a callback for reason cbAtStartOfSimTime will result in another cbAtStartOfSimTime callback occurring during the same time slice.

The value fields are ignored for all reasons with simulation-time-related callbacks.
When the \(c b \_d a t a \_p->t i m e->t y p e\) is set to vpiScaledRealTime, the \(c b \_d a t a \_p->o b j\) field shall be used as the object for determining the time scaling.

When a simulation-time-related callback occurs, the user callback application shall be passed a single argument, which is a pointer to an s_cb_data structure [this is not a pointer to the same structure that was passed to vpi_register_cb()]. The \(\overline{\text { time }} \overline{-}\) structure shall contain the current simulation time. The user_data field shall be equivalent to the user_data field passed to vpi_register_cb().

The callback application can use the information in the passed structure and information retrieved from other interface routines to perform the desired callback processing.

\subsection*{27.33.3 Simulator action and feature related callbacks}

The vpi_register_cb() routine can register callbacks to occur for simulator action reasons or simulator feature reasons. Simulator action reasons are callbacks such as the end of compilation or end of simulation. Simulator feature reasons are software-product-specific features, such as restarting from a saved simulation state or entering an interactive mode. Actions are differentiated from features in that actions shall occur in all VPI-compliant products, whereas features might not exist in all VPI-compliant products.

The following action-related callbacks shall be defined:
\begin{tabular}{ll} 
cbEndOfCompile & End of simulation data structure compilation or build \\
cbStartOfSimulation & Start of simulation (beginning of time 0 simulation cycle) \\
cbEndOfSimulation & End of simulation (e.g., \$finish system task executed) \\
cbError & Simulation run-time error occurred \\
cbPLIError & Simulation run-time error occurred in a PLI function call \\
cbTchkViolation & Timing check error occurred \\
cbSignal & A signal occurred
\end{tabular}

Examples of possible feature related callbacks are
\begin{tabular}{ll} 
cbStartOfSave & Simulation save state command invoked \\
cbEndOfSave & Simulation save state command completed \\
cbStartOfRestart & Simulation restart from saved state command invoked \\
cbEndOfRestart & Simulation restart command completed \\
cbEnterInteractive & \begin{tabular}{l} 
Simulation entering interactive debug mode (e.g., \$stop system task \\
executed)
\end{tabular} \\
cbExitInteractive & \begin{tabular}{l} 
Simulation exiting interactive mode \\
cbInteractiveScopeChange
\end{tabular} \\
Simulation command to change interactive scope executed \\
cbUnresolvedSystf & Unknown user-defined system task or function encountered
\end{tabular}

The only fields in the s_cb_data structure that shall need to be setup for simulation action/feature callbacks are the reason, cb_rtn, \(\overline{\text { and }}\) user_data (if desired) fields.
vpi_register_cb() can be used to set up a signal handler. To do this, set the reason field to cbSignal and set the index field to one of the legal signals specified by the operating system. When this signal occurs, the simulator will trap the signal, proceed to a safe point (if possible), then call the callback routine.

When a simulation action/feature callback occurs, the user routine shall be passed a pointer to an s_cb_data structure. The reason field shall contain the reason for the callback. For cbTchkViolation callbacks, the obj field shall be a handle to the timing check. For cbInteractiveScopeChange, obj shall be a handle to the new scope. For cbUnresolvedSystf, user_data shall point to the name of the unresolved task or function. On a cbError callback, the routine vpi_chk_error() can be called to retrieve error information.

When an implementation restarts the only VPI callbacks that shall exist are those for cbStartOfRestart and cbEndOfRestart. Note when a user registers for these two callbacks the user_data field should not be a pointer into memory. The reason for this is that the executable used to restart an implementation may not be the exact same one used to save the implementation state. A typical use of the user_data field, for these two callbacks would be to store the ID returned from a call to vpi_put_data().

With the exception of cbStartOfRestart and cbEndOfRestart callbacks, when a restart occurs all registered callbacks shall be removed.

The following example shows a callback application that reports cpu usage at the end of a simulation. If the user routine setup_report_cpu() is placed in the vlog_startup_routines list, it shall be called just after the simulator is invoked.
```

static PLI_INT32 initial_cputime_g;
void report_cpu()
{
PLI_INT32 total = get_current_cputime() - initial_cputime_g;
vpi_printf("Simulation complete. CPU time used: %d\n",
total);
}
void setup_report_cpu()
{
static s_cb_data cb_data_s = { cbEndOfSimulation,
report_cpu} ;
initial_cputime_g = get_current_cputime();
vpi_register_cb(\&cb_data_s);
}

```

\subsection*{27.34 vpi_register_systf()}


The VPI routine vpi_register_systf() shall register callbacks for user-defined system tasks or functions. Callbacks can be registered to occur when a user-defined system task or function is encountered during compilation or execution of Verilog HDL source code.

The systf_data_p argument shall point to a s_vpi_systf_data structure, which is defined in vpi_user.h and listed in Figure 189.
```

pedef struct t_vpi_systf_data
PLI_INT32 type; /* vpiSysTask, vpiSysFunc */
PLI_INT32 sysfunctype; /* vpiSysTask, vpi[ Int,Real,Time,Sized,
SizedSigned] Func */
PLI BYTE8 *tfname; /* first character must be `\$' */
PLI-INT32 (*calltf)(PLI BYTE8 *);
PLI_INT32 (* compiletf)(\overline{PLI_BYTE8 *);}
PLI_INT32 (*sizetf)(PLI_BYTE8 *); /* for sized function
callbacks only */
PLI_BYTE8 *user_data;
s_vpi_syst\overline{f_data, *p_vpi_systf_data;}

```

Figure 189-The s_vpi_systf_data structure definition

\subsection*{27.34.1 System task and function callbacks}

User-defined Verilog system tasks and functions that use VPI routines can be registered with vpi_register_systf(). The following system task/function-related callbacks are defined.

The type field of the s_vpi_systf_data structure shall register the user application to be a system task or a system function. The type field value shall be an integer constant of vpiSysTask or vpiSysFunc.

The sysfunctype field of the s_vpi_systf_data structure shall define the type of value that a system function shall return. The sysfiunctype field shall be an integer constant of vpiIntFunc, vpiRealFunc, vpiTimeFunc, vpiSizedFunc or vpiSizedSignedFunc. This field shall only be used when the type field is set to vpiSysFunc.
tfname is a character string containing the name of the system task or function as it will be used in Verilog source code. The name shall begin with a dollar sign ( \(\$\) ), and shall be followed by one or more ASCII characters which are legal in Verilog HDL simple identifiers. These are the characters A through Z, a through z, 0 through 9 , underscore ( \(\_\)), and the dollar sign (\$). The maximum name length shall be the same as for Verilog HDL identifiers.

The compiletf, calltf, and sizetf fields of the s_vpi_systf_data structure shall be pointers to the userprovided applications that are to be invoked by the system task/function callback mechanism. One or more of the compiletf, calltf, and sizetf fields can be set to NULL if they are not needed. Callbacks to the applications pointed to by the compiletf and sizetf fields shall occur when the simulation data structure is compiled or built (or for the first invocation if the system task or function is invoked from an interactive mode). Callbacks to the application pointed to by the callff routine shall occur each time the system task or function is invoked during simulation execution.

The sizetf application shall only be called if the PLI application type is vpiSysFunc and the sysfunctype is vpiSizedFunc or vpiSizedSignedFunc. If no sizetf is provided, a user-defined system function of type vpiSizedFunc or vpiSizedSignedFunc shall return 32-bits.

The contents of the user_data field of the s_vpi_systf_data structure shall be the only argument passed to the compiletf, sizetf, and calltf routines when they are called. This argument shall be of the type PLI_BYTE8 *.

The following two examples illustrate allocating and filling in the s_vpi_systf_data structure and calling the vpi_register_systf() function. These examples show two different C programming methods of filling in the structure fields. A third method is shown in 27.34.3.
```

/*
* VPI registration data for a $list_nets system task
    */
void listnets_register()
{
    s_vpi_systf_data tf_data;
    tf_data.type = vpiSysTask;
    tf_data.tfname = "$list_nets";
tf-data.calltf = ListCall`;
tf data.compiletf = ListCheck;
vpi_register_systf(\&tf_data);
}
/*
* VPI registration data for a $my_random system function
    * /
void my_random_init()
{
    s_vpi_systf_data func_data;
    p_vpi_systf_data func_data_p = &func_data;
    PLI_BYTE8 *my workarea;
    my w
    func_data_p->type = vpiSysFunc;
    func_data_p->sysfunctype = vpiSizedFunc;
    func_data_p->tfname = "$my_random";
func-data p->calltf = my rāndom;
func_data_p->compiletf = my_random_compiletf;
func_data_p->compiletf = my_random_sizetf;
func_data_p->user_data = my_workarea;
vpi_\overline{register_syst\overline{f}(func_data_\overline{p});}
}

```

\subsection*{27.34.2 Initializing V I system task/function callbacks}

A means of initializing system task/function callbacks and performing any other desired task just after the simulator is invoked shall be provided by placing routines in a NULL-terminated static array, vlog_startup_routines. A C function using the array definition shall be provided as follows:
```

void (*vlog_startup_routines[] ) ();

```

This C function shall be provided with a VPI-compliant product. Entries in the array shall be added by the user. The location of vlog_startup_routines and the procedure for linking vlog_startup_routines with a software product shall be defined by the product vendor. (Note that callbacks can also be registered or removed at any time during an application routine, not just at startup time).

This array of C functions shall be for registering system tasks and functions. User tasks and functions that appear in a compiled description shall generally be registered by a routine in this array.

The following example uses vlog_startup_routines to register the system task and system function which were defined in the examples in 27.34.1.

Note that a tool vendor shall supply a file which contains the vlog startup routines array. The names of the PLI application register functions are added to this vendor supplied file.
```

extern void listnets_register();
extern void my random_init();
void (*vlog_startup_routines[]) () =
{
listnets_register,
my_random_init,
O
}

```

\subsection*{27.34.3 egistering multiple system tasks and functions}

Multiple system tasks and functions can be registered at least two different ways:
Allocate and define separate s_vpi_systf_data structures for each system task or function, and call vpi_register_systf() once for each structure. This is the method which was used by the examples in 27.34.1 and 27.34.2.
Allocate a static array of s_vpi_systf_data structures, and call vpi_register_systf() once for each structure in the array. If the final element in the array is set to zero, then the calls to vpi_register_systf() can be placed in a loop which terminates when it reaches the 0 .

The following example uses a static structure to declare three system tasks and functions, and places vpi_register_systf() in a loop to register them.
```

/*In a vendor product file which contains vlog_startup_routines
...* /
extern void register my systfs();
extern void my_init();
void (*vlog_startup_routines[] )() =
{
setup_report_cpu, /* user routine example in 27.33.3 */
register_my_systfs, /* user routine listed below */
0 /* must be last entry in list */
}

```
}
```

```
/* In a user provided file... */
```

/* In a user provided file... */
void register my systfs()
void register my systfs()
{
{
static s_vpi_systf_data systfTestList[] = {
static s_vpi_systf_data systfTestList[] = {
{vpiSysTask, 0, "$my_task", my_task_calltf,
        {vpiSysTask, 0, "$my_task", my_task_calltf,
my_task_comptf,0,0},
my_task_comptf,0,0},
{vpiSysFunc, vpiIntFunc, "$my_int_func", my_int_func_calltf,
        {vpiSysFunc, vpiIntFunc, "$my_int_func", my_int_func_calltf,
my_int_func_comptf, 0,0},
my_int_func_comptf, 0,0},
{vpiSysFunc, vpiSizedFunc, "$my_sized_func",
        {vpiSysFunc, vpiSizedFunc, "$my_sized_func",
my_sized_func_calltf, my_sized_func_comptf,
my_sized_func_calltf, my_sized_func_comptf,
my_sized_func_sizetf,0},
my_sized_func_sizetf,0},
0} ;
0} ;
p_vpi_systf_data systf_data_p = \&(systfTestList[ 0] );
p_vpi_systf_data systf_data_p = \&(systfTestList[ 0] );
while (systf data_p->type)
while (systf data_p->type)
vpi_register_systf(systf_data_p++);

```
        vpi_register_systf(systf_data_p++);
```


### 27.35 vpi_remove_cb()

| vpi_remove_cb() |  |  |  |
| :---: | :---: | :---: | :---: |
| Synopsis: | Remove a simulation callback registered with vpi_register_cb(). |  |  |
| Syntax: | vpi_remove_cb (cb_obj) |  |  |
| Type Description |  |  |  |
| Returns: | PLI_INT32 | $\mathbf{1}$ (true) if successful; $\mathbf{0}$ (false) on a failure |  |
| Arguments: | Type | Name | Descri |
|  | vpiHandle | cb_obj | Handle to the callback object |
| Related routines: | Use vpi_register_cb() to register callbacks for simulation-related events |  |  |

The VPI routine vpi_remove_cb() shall remove callbacks that were registered with vpi_register_cb(). The argument to this routine shall be a handle to the callback object. The routine shall return a $\mathbf{1}$ (true) if successful, and a $\mathbf{0}$ (false) on a failure. After vpi_remove_cb() is called with a handle to the callback, the handle is no longer valid.

### 27.36 vpi_scan()

| vpi_scan() |  |  |  |
| :---: | :---: | :---: | :---: |
| Synopsis: | Scan the Verilog HDL hierarchy for objects with a one-to-many relationship. |  |  |
| Syntax: | vpi_scan(itr) |  |  |
| Returns: | Type Description |  |  |
|  | vpiHandle | Handle to an object |  |
| Arguments: | Type | Name | Desc |
|  | vpiHandle | itr | Handle to an iterator object |
| Related routines: | Use vpi_iterat Use vpi_handl Use vpi_handl | obtain an iterator to obtain handles to ulti() to obtain a | with a one-to-one relationship bject with a many-to-one re |

The VPI routine vpi_scan() shall traverse the instantiated Verilog HDL hierarchy and return handles to objects as directed by the iterator itr. The iterator handle shall be obtained by calling vpi_iterate() for a specific object type. Once vpi_scan() returns NULL, the iterator handle is no longer valid and cannot be used again.

The following example application uses vpi_iterate() and vpi_scan() to display each net (including the size for vectors) declared in the module. The example assumes it shall be passed a valid module handle.

```
void display nets(mod)
vpiHandle mod;
{
    vpiHandle net;
        vpiHandle itr;
        vpi_printf("Nets declared in module %s\n",
        vpi_get_str(vpiFullName, mod));
        itr = vpi_iterate(vpiNet, mod);
        while (net = vpi_scan(itr))
        {
            vpi_printf("\t%s", vpi_get_str(vpiName, net));
            if (vpi_get(vpiVector, net))
            {
                vpi_printf(" of size %d\n", vpi_get(vpiSize, net));
            }
        else vpi_printf("\n");
    }
}
```


### 27.37 vpi_vprintf()

| vpi_vprintf() |  |  |  |
| :---: | :---: | :---: | :---: |
| Synopsis: | Write to the output channel of the software product which invoked the PLI application and the current product $\log$ file using varargs which are already started. |  |  |
| Syntax: | vpi_vprintf(format, ap) |  |  |
| Type Description |  |  |  |
| Returns: | PLI_INT32 | The number of characters written |  |
| Arguments: | Type | Name | Description |
|  | PLI_BYTE8 * | format | A format string using the C prin |
|  | va_list | ap | An already started varargs list |
| Related routines: | Use vpi_printf() to write a finite number of arguments Use vpi_mcd_printf() to write to an opened file Use vpi_mcd_vprintf() to write a variable number of arguments to an opened file |  |  |

This routine performs the same function as vpi_printf(), except that varargs has already been started.

## Annex A

(normative)

## Formal syntax definition

The formal syntax of Verilog HDL is described using Backus-Naur Form (BNF).

## A. 1 Source text

## A.1.1 Library source text

library_text $::=\{$ library_descriptions $\}$
library_descriptions ::=
library_declaration
| include_statement
| config_declaration
library_declaration ::=
library library_identifier file_path_spec [ \{ , file_path_spec \}]
[ -incdir file_path_spec [ \{ , file_path_spec \}];
file_path_spec ::= file_path
include_statement $::=$ include $<$ file_path_spec> ;

## A.1.2 Configuration source text

```
config_declaration ::=
    config config_identifier ;
    design_statement
    {config_rule_statement}
    endconfig
design_statement ::= design { [library_identifier.]cell_identifier };
config_rule_statement ::=
    default_clause liblist_clause
    | inst_clause liblist_clause
    | inst_clause use_clause
    | cell_clause liblist_clause
    | cell_clause use_clause
default_clause ::= default
inst_clause ::= instance inst_name
inst_name ::= topmodule_identifier{.instance_identifier}
cell_clause ::= cell [ library_identifier.]cell_identifier
liblist_clause ::= liblist [{library_identifier}]
use_clause ::= use [library_identifier.]cell_identifier[:config]
```


## A.1.3 Module and primitive source text

```
source_text ::= { description }
description ::=
    module_declaration
    | udp_declaration
module_declaration ::=
    { attribute_instance } module_keyword module_identifier [ module_parameter_port_list ]
    [ list_of_ports ] ; { module_item }
    endmodule
    | { attribute_instance } module_keyword module_identifier [ module_parameter_port_list ]
        [ list_of_port_declarations ] ; { non_port_module_item }
        endmodule
module_keyword ::= module | macromodule
```


## A.1.4 Module parameters and ports

```
module_parameter_port_list ::= # ( parameter_declaration { , parameter_declaration } )
list_of_ports ::= ( port { , port } )
list_of_port_declarations ::=
    ( port_declaration {, port_declaration } )
    |()
port ::=
    [ port_expression ]
    | . port_identifier ([ port_expression ])
port_expression ::=
        port_reference
    | { port_reference {, port_reference } }
port_reference ::=
        port_identifier
    | port_identifier [ constant_expression ]
    | port_identifier [ range_expression ]
port_declaration ::=
        {attribute_instance} inout_declaration
    | {attribute_instance} input_declaration
    | {attribute_instance} output_declaration
```


## A.1.5 Module items

```
module_item ::=
    module_or_generate_item
    | port_declaration;
    | { attribute_instance } generated_instantiation
    | { attribute_instance } local_parameter_declaration
    | { attribute_instance } parameter_declaration
    | { attribute_instance } specify_block
    | { attribute_instance } specparam_declaration
```

```
module_or_generate_item ::=
    \{ attribute_instance \} module_or_generate_item_declaration
    | \{ attribute_instance \} parameter_override
    | \{ attribute_instance \} continuous_assign
    | \{ attribute_instance \} gate_instantiation
    | \{ attribute_instance \} udp_instantiation
    | \{ attribute_instance \} module_instantiation
    | \{ attribute_instance \} initial_construct
    | \{ attribute_instance \} always_construct
module_or_generate_item_declaration ::=
        net_declaration
    | reg_declaration
    | integer_declaration
    | real_declaration
    | time_declaration
    | realtime_declaration
    | event_declaration
    | genvar_declaration
    | task_declaration
    | function_declaration
non_port_module_item ::=
        \{ attribute_instance \} generated_instantiation
    | \{ attribute_instance \} local_parameter_declaration
    | \{ attribute_instance \} module_or_generate_item
    | \{ attribute_instance \} parameter_declaration
    | \{ attribute_instance \} specify_block
    | \{ attribute_instance \} specparam_declaration
parameter_override ::= defparam list_of_param_assignments ;
```


## A. 2 Declarations

## A.2.1 Declaration types

## A.2.1.1 Module parameter declarations

local_parameter_declaration ::=
localparam [ signed ] [ range ] list_of_param_assignments ;
| localparam integer list_of_param_assignments ;
| localparam real list_of_param_assignments ;
| localparam realtime list_of_param_assignments ;
| localparam time list_of_param_assignments ;
parameter_declaration ::=
parameter [ signed ] [ range ] list_of_param_assignments ;
| parameter integer list_of_param_assignments ;
| parameter real list_of_param_assignments ;
| parameter realtime list_of_param_assignments ;
| parameter time list_of_param_assignments ;
specparam_declaration ::= specparam [ range ] list_of_specparam_assignments ;

## A.2.1.2 ort declarations

```
inout_declaration ::= inout [ net_type ] [ signed ] [ range ]
    list_of_port_identifiers
input_declaration ::= input [ net_type ] [ signed ] [ range ]
        list_of_port_identifiers
output_declaration ::=
    output [ net_type ] [ signed ] [ range ]
        list_of_port_identifiers
    | output [ reg ] [ signed ] [ range ]
        list_of_port_identifiers
    | output reg [ signed ] [ range ]
        list_of_variable_port_identifiers
    | output [ output_variable_type ]
        list_of_port_identifiers
    | output output_variable_type
        list_of_variable_port_identifiers
```


## A.2.1.3 Type declarations

```
event_declaration ::= event list_of_event_identifiers ;
genvar_declaration ::= genvar list_of_genvar_identifiers ;
integer_declaration ::= integer list_of_variable_identifiers ;
net_declaration ::=
    net_type [ signed ]
        [ delay3 ] list_of_net_identifiers ;
    | net_type [ drive_strength ] [ signed ]
        [ delay3 ] list_of_net_decl_assignments ;
    | net_type [ vectored | scalared ] [ signed ]
        range [ delay3 ] list_of_net_identifiers ;
    | net_type [ drive_strength ] [ vectored | scalared ] [ signed ]
        range [ delay3 ] list_of_net_decl_assignments ;
    | trireg [ charge_strength ] [ signed ]
        [ delay3 ] list_of_net_identifiers ;
    | trireg [ drive_strength ] [ signed ]
        [ delay3 ] list_of_net_decl_assignments ;
    | trireg [ charge_strength ] [ vectored | scalared ] [ signed ]
        range [ delay3 ] list_of_net_identifiers ;
    | trireg [ drive_strength ] [ vectored | scalared ] [ signed ]
        range [ delay3 ] list_of_net_decl_assignments ;
real_declaration ::= real list_of_real_identifiers ;
realtime_declaration ::= realtime list_of_real_identifiers ;
reg_declaration ::= reg [ signed ] [ range ]
    list_of_variable_identifiers ;
time_declaration ::= time list_of_variable_identifiers ;
```


## A.2.2 Declaration data types

## A.2.2.1 Net and variable types

net_type ::=
supply0 | supply1
|tri $\quad$ triand | trior | tri0 | tri1
|wire $\quad \mid$ wand $\mid$ wor
output_variable_type ::= integer | time
real_type ::=
real_identifier [ = constant_expression ]
| real_identifier dimension \{ dimension \}
variable_type ::=
variable_identifier [ = constant_expression ]
| variable_identifier dimension \{ dimension \}

## A.2.2.2 Strengths

drive_strength ::=
( strength0, strength1)
| (strength1, strength0)
| ( strength0, highz1 )
| ( strength1, highzo )
| (highz0 , strength1 )
| (highz1, strength0)
strength0 $::=$ supply0 $\mid$ strong $0 \mid$ pull0 $\mid$ weak0
strength1 ::= supply1 | strong1 | pull1 | weak1
charge_strength ::=( small $) \mid($ medium $) \mid$ ( large $)$

## A.2.2.3 Delays

delay3 ::= \# delay_value | \# ( delay_value [, delay_value [, delay_value ] ])
delay 2 ::= \# delay_value $\mid$ \# ( delay_value [, delay_value ] )
delay_value ::=
unsigned_number
| parameter_identifier
| specparam_identifier
| mintypmax_expression

## A.2.3 Declaration lists

```
list_of_event_identifiers ::= event_identifier [ dimension { dimension }]
    {, event_identifier [ dimension { dimension }] }
list_of_genvar_identifiers ::= genvar_identifier {, genvar_identifier }
list_of_net_decl_assignments ::= net_decl_assignment {, net_decl_assignment }
list_of_net_identifiers ::= net_identifier [ dimension { dimension }]
    {, net_identifier [ dimension { dimension }] }
list_of_param_assignments ::= param_assignment {, param_assignment }
list_of_port_identifiers ::= port_identifier {, port_identifier }
list_of_real_identifiers ::= real_type { , real_type }
list_of_specparam_assignments ::= specparam_assignment { , specparam_assignment }
list_of_variable_identifiers ::= variable_type {, variable_type }
list_of_variable_port_identifiers ::= port_identifier [ = constant_expression ]
                                    {, port_identifier [ = constant_expression ] }
```


## A.2.4 Declaration assignments

```
net_decl_assignment ::= net_identifier = expression
param_assignment ::= parameter_identifier = constant_expression
specparam_assignment ::=
    specparam_identifier = constant_mintypmax_expression
    | pulse_control_specparam
pulse_control_specparam ::=
    PATHPULSE$ = (reject_limit_value [, error_limit_value ] );
    |PATHPULSE$specify_input_terminal_descriptor$specify_output_terminal_descriptor
    =( reject_limit_value [, error_limit_value ] );
error_limit_value ::= limit_value
reject_limit_value ::= limit_value
limit_value ::= constant_mintypmax_expression
```


## A.2.5 Declaration ranges

```
dimension ::= [ dimension_constant_expression : dimension_constant_expression ]
```

range $::=$ [ msb_constant_expression : 1sb_constant_expression ]

## A.2.6 Function declarations

function_declaration ::=
function [ automatic ] [ signed ] [ range_or_type ] function_identifier ;
function_item_declaration \{ function_item_declaration \}
function_statement
endfunction
| function [ automatic ] [ signed ] [ range_or_type ] function_identifier ( function_port_list ) ;
block_item_declaration \{ block_item_declaration \}
function_statement
endfunction
function_item_declaration ::=
block_item_declaration
| tf_input_declaration ;
function_port_list $::=\{$ attribute_instance $\}$ tf_input_declaration $\{,\{$ attribute_instance $\}$
tf_input_declaration \}
range_or_type $::=$ range $\mid$ integer $\mid$ real $\mid$ realtime $\mid$ time

## A.2.7 Task declarations

```
task_declaration ::=
    task [ automatic ] task_identifier ;
    { task_item_declaration }
    statement
    endtask
    task [ automatic ] task_identifier ( task_port_list ) ;
    {block_item_declaration }
    statement
    endtask
```

```
task_item_declaration ::=
            block_item_declaration
    | { attribute_instance } tf_input_declaration ;
    | { attribute_instance } tf_output_declaration;
    | { attribute_instance } tf_inout_declaration ;
task_port_list ::= task_port_item { , task_port_item }
task_port_item ::=
            { attribute_instance } tf_input_declaration
    | { attribute_instance } tf_output_declaration
    | { attribute_instance } tf_inout_declaration
tf_input_declaration ::=
            input [ reg ] [ signed ] [ range ] list_of_port_identifiers
            input [ task_port_type ] list_of_port_identifiers
tf_output_declaration ::=
            output [ reg ] [ signed ] [ range ] list_of_port_identifiers
            output [ task_port_type ] list_of_port_identifiers
tf_inout_declaration ::=
            inout [ reg ] [ signed ] [ range ] list_of_port_identifiers
            | inout [ task_port_type ] list_of_port_identifiers
task_port_type ::=
            time | real | realtime | integer
```


## A.2.8 Block item declarations

block_item_declaration ::=
\{ attribute_instance \} block_reg_declaration
| \{ attribute_instance \} event_declaration
| \{ attribute_instance \} integer_declaration
| \{ attribute_instance \} local_parameter_declaration
| \{ attribute_instance \} parameter_declaration
| \{ attribute_instance \} real_declaration
| \{ attribute_instance \} realtime_declaration
| \{ attribute_instance \} time_declaration
block_reg_declaration $::=$ reg [ signed ] [ range ]
list_of_block_variable_identifiers ;
list_of_block_variable_identifiers ::=
block_variable_type \{, block_variable_type \}
block_variable_type ::=
variable_identifier
| variable_identifier dimension \{ dimension \}

## A. 3 rimitive instances

## A.3.1 rimitive instantiation and instances

```
gate_instantiation ::=
    cmos_switchtype [delay3]
        cmos_switch_instance { , cmos_switch_instance } ;
    | enable_gatetype [drive_strength] [delay3]
        enable_gate_instance {, enable_gate_instance };
    | mos_switchtype [delay3]
        mos_switch_instance { , mos_switch_instance };
    | n_input_gatetype [drive_strength] [delay2]
        n_input_gate_instance {, n_input_gate_instance } ;
    | n_output_gatetype [drive_strength] [delay2]
        n_output_gate_instance {, n_output_gate_instance };
    | pass_en_switchtype [delay2]
        pass_enable_switch_instance {, pass_enable_switch_instance };
    | pass_switchtype
        pass_switch_instance { , pass_switch_instance };
    | pulldown [pulldown_strength]
        pull_gate_instance { , pull_gate_instance };
    | pullup [pullup_strength]
        pull_gate_instance { , pull_gate_instance };
cmos_switch_instance ::= [ name_of_gate_instance ] ( output_terminal , input_terminal,
        ncontrol_terminal , pcontrol_terminal )
enable_gate_instance ::= [ name_of_gate_instance ] ( output_terminal , input_terminal, enable_terminal )
mos_switch_instance ::= [ name_of_gate_instance ] ( output_terminal, input_terminal, enable_terminal )
n_input_gate_instance ::= [name_of_gate_instance ] (output_terminal, input_terminal {, input_terminal } )
n_output_gate_instance ::= [ name_of_gate_instance ] ( output_terminal { , output_terminal } ,
        input_terminal )
pass_switch_instance ::= [ name_of_gate_instance ] ( inout_terminal , inout_terminal )
pass_enable_switch_instance ::= [ name_of_gate_instance ] ( inout_terminal, inout_terminal,
        enable_terminal )
pull_gate_instance ::= [ name_of_gate_instance ] ( output_terminal )
name_of_gate_instance ::= gate_instance_identifier [range ]
```


## A.3.2 rimitive strengths

```
pulldown_strength ::=
    ( strength0, strength1 )
    | ( strength1, strength0 )
    | ( strength0 )
pullup_strength ::=
    ( strength0 , strength1 )
    |(strength1, strength0 )
    | ( strength1 )
```


## A.3.3 rimitive terminals

```
enable_terminal ::= expression
inout_terminal ::= net_lvalue
input_terminal ::= expression
ncontrol_terminal ::= expression
output_terminal ::= net_lvalue
pcontrol_terminal ::= expression
```


## A.3.4 rimitive gate and switch types

```
cmos_switchtype ::= cmos | remos
enable_gatetype ::= bufif0 | bufif1 | notif0 | notif1
mos_switchtype ::= nmos | pmos | rnmos | rpmos
n_input_gatetype::= and | nand | or | nor | xor | xnor
n_output_gatetype ::= buf | not
pass_en_switchtype ::= tranif0 | tranif1 | rtranif1 | rtranif0
pass_switchtype ::= tran | rtran
```


## A. 4 Module and generated instantiation

## A.4.1 Module instantiation

```
module_instantiation ::=
    module_identifier [ parameter_value_assignment ]
    module_instance {, module_instance };
parameter_value_assignment ::= # ( list_of_parameter_assignments )
list_of_parameter_assignments ::=
    ordered_parameter_assignment {, ordered_parameter_assignment } |
    named_parameter_assignment {, named_parameter_assignment }
ordered_parameter_assignment ::= expression
named_parameter_assignment ::= . parameter_identifier ([ expression ])
module_instance ::= name_of_instance ([ list_of_port_connections ])
name_of_instance ::= module_instance_identifier [ range ]
list_of_port_connections ::=
    ordered_port_connection {, ordered_port_connection }
    | named_port_connection {, named_port_connection }
ordered_port_connection ::= {attribute_instance } [ expression ]
named_port_connection ::= { attribute_instance } .port_identifier ([ expression ] )
```


## A.4.2 Generated instantiation

```
generated_instantiation ::= generate { generate_item } endgenerate
generate_item_or_null ::= generate_item |;
```

```
generate_item ::=
    generate_conditional_statement
    | generate_case_statement
    | generate_loop_statement
    | generate_block
    | module_or_generate_item
generate_conditional_statement ::=
    if ( constant_expression ) generate_item_or_null [ else generate_item_or_null ]
generate_case_statement ::= case ( constant_expression)
                genvar_case_item { genvar_case_item } endcase
genvar_case_item ::= constant_expression {, constant_expression } :
                    generate_item_or_null | default [ : ] generate_item_or_null
generate_loop_statement ::= for (genvar_assignment ; constant_expression ; genvar_assignment )
            begin : generate_block_identifier { generate_item } end
genvar_assignment ::= genvar_identifier = constant_expression
generate_block ::= begin [ : generate_block_identifier ] { generate_item } end
```


## A. 5 UD declaration and instantiation

## A.5.1 UD declaration

```
udp_declaration ::=
    { attribute_instance } primitive udp_identifier ( udp_port_list ) ;
    udp_port_declaration { udp_port_declaration }
    udp_body
    endprimitive
    | { attribute_instance } primitive udp_identifier (udp_declaration_port_list ) ;
    udp_body
    endprimitive
```


## A.5.2 UD ports

```
udp_port_list ::= output_port_identifier, input_port_identifier { , input_port_identifier }
udp_declaration_port_list ::=
    udp_output_declaration, udp_input_declaration { , udp_input_declaration }
udp_port_declaration ::=
    udp_output_declaration;
    | udp_input_declaration ;
    | udp_reg_declaration ;
udp_output_declaration ::=
    { attribute_instance } output port_identifier
    | { attribute_instance } output reg port_identifier [ = constant_expression ]
udp_input_declaration ::= { attribute_instance } input list_of_port_identifiers
udp_reg_declaration ::= { attribute_instance } reg variable_identifier
```


## A.5.3 UD body

udp_body ::= combinational_body $\mid$ sequential_body
combinational_body $::=$ table combinational_entry $\{$ combinational_entry \} endtable
combinational_entry ::= level_input_list : output_symbol ;
sequential_body $::=$ [ udp_initial_statement ] table sequential_entry \{ sequential_entry \} endtable
udp_initial_statement $::=$ initial output_port_identifier = init_val ;
init_val ::= 1'b0 $\left|\mathbf{1}^{\prime} \mathbf{b} \mathbf{1}\right| \mathbf{1}^{\prime} \mathbf{b x}\left|\mathbf{1}^{\mathbf{\prime}} \mathbf{b X}\right| \mathbf{1}^{\mathbf{\prime}} \mathbf{B 0}\left|\mathbf{1}^{\mathbf{\prime}} \mathbf{B 1}\right| \mathbf{1}^{\mathbf{\prime}} \mathbf{B x}\left|\mathbf{1}^{\prime} \mathbf{B X}\right| \mathbf{1} \mid \mathbf{0}$
sequential_entry ::= seq_input_list : current_state : next_state ;
seq_input_list $::=$ level_input_list $\mid$ edge_input_list
level_input_list $::=$ level_symbol \{ level_symbol \}
edge_input_list $::=\{$ level_symbol $\}$ edge_indicator $\{$ level_symbol $\}$
edge_indicator $::=($ level_symbol level_symbol $) \mid$ edge_symbol
current_state $::=$ level_symbol
next_state ::= output_symbol|-
output_symbol ::=0|1|x|X
level_symbol $::=\mathbf{0}|\mathbf{1}| \mathbf{x}|\mathbf{X}| \boldsymbol{?}|\mathbf{b}| \mathbf{B}$
edge_symbol $::=\mathbf{r}|\mathbf{R}| \mathbf{f}|\mathbf{F}| \mathbf{p}|\mathbf{P}| \mathbf{n}|\mathbf{N}|$ *

## A.5.4 UD instantiation

udp_instantiation ::= udp_identifier [ drive_strength ] [ delay2 ]
udp_instance $\{$, udp_instance $\}$;
udp_instance ::= [ name_of_udp_instance ] ( output_terminal , input_terminal
\{ , input_terminal $\}$ )
name_of_udp_instance $::=$ udp_instance_identifier [range ]

## A. 6 Behavioral statements

## A.6.1 Continuous assignment statements

```
continuous_assign ::= assign [ drive_strength ] [ delay3 ] list_of_net_assignments ;
list_of_net_assignments ::= net_assignment { , net_assignment }
net_assignment ::= net_lvalue = expression
```


## A.6.2 rocedural blocks and assignments

```
initial_construct ::= initial statement
always_construct ::= always statement
blocking_assignment ::= variable_lvalue = [ delay_or_event_control ] expression
nonblocking_assignment ::= variable_lvalue <= [ delay_or_event_control ] expression
procedural_continuous_assignments ::=
    assign variable_assignment
    deassign variable_lvalue
    | force variable_assignment
    force net_assignment
    release variable_lvalue
    release net_lvalue
```

```
function_blocking_assignment ::= variable_lvalue = expression
function_statement_or_null ::=
    function_statement
    | { attribute_instance };
```


## A.6.3 arallel and sequential blocks

```
function_seq_block ::= begin [ : block_identifier
    {block_item_declaration } ] { function_statement } end
variable_assignment ::= variable_lvalue = expression
par_block ::= fork [ : block_identifier
                            {block_item_declaration }] { statement } join
seq_block ::= begin [ : block_identifier
    { block_item_declaration } ] { statement } end
```


## A.6.4 Statements

statement ::=
\{ attribute_instance \} blocking_assignment ;
| \{ attribute_instance \} case_statement
| \{ attribute_instance \} conditional_statement
| \{ attribute_instance \} disable_statement
| \{ attribute_instance \} event_trigger
| \{ attribute_instance \} loop_statement
| \{ attribute_instance \} nonblocking_assignment ;
| \{ attribute_instance \} par_block
| \{ attribute_instance \} procedural_continuous_assignments ;
| \{ attribute_instance \} procedural_timing_control_statement
| \{ attribute_instance \} seq_block
| \{ attribute_instance \} system_task_enable
| \{ attribute_instance \} task_enable
| \{ attribute_instance \} wait_statement
statement_or_null ::=
statement
| \{ attribute_instance \};
function_statement $::=$
\{ attribute_instance \} function_blocking_assignment ;
| \{ attribute_instance \} function_case_statement
| \{ attribute_instance \} function_conditional_statement
| \{ attribute_instance \} function_loop_statement
| \{ attribute_instance \} function_seq_block
| \{ attribute_instance \} disable_statement
| \{ attribute_instance \} system_task_enable

## A.6.5 Timing control statements

delay_control ::=
\# delay_value
| \# ( mintypmax_expression )
delay_or_event_control ::=
delay_control
| event_control
$\mid$ repeat ( expression ) event_control
disable_statement $::=$
disable hierarchical_task_identifier ;
| disable hierarchical_block_identifier ;
event_control ::=
@ event_identifier

- ( event_expression)
- ©
- @ (*)
event_trigger ::=
-> hierarchical_event_identifier ;
event_expression ::=
expression
| hierarchical_identifier
| posedge expression
| negedge expression
| event_expression or event_expression
| event_expression, event_expression
procedural_timing_control_statement $::=$
delay_or_event_control statement_or_null
wait_statement ::=
wait ( expression ) statement_or_null


## A.6.6 Conditional statements

conditional_statement ::=
if ( expression )
statement_or_null [ else statement_or_null ]
| if_else_if_statement
if_else_if_statement ::=
if ( expression ) statement_or_null
\{ else if ( expression ) statement_or_null \}
[ else statement_or_null]
function_conditional_statement $::=$
if ( expression ) function_statement_or_null
[ else function_statement_or_null]
| function_if_else_if_statement
function_if_else_if_statement ::=
if ( expression ) function_statement_or_null
\{ else if ( expression ) function_statement_or_null \}
[ else function_statement_or_null]

## A.6.7 Case statements

```
case_statement ::=
    case ( expression)
        case_item { case_item } endcase
    | casez ( expression )
        case_item { case_item } endcase
    | casex ( expression)
        case_item { case_item } endcase
case_item ::=
    expression {, expression } : statement_or_null
    | default [:] statement_or_null
function_case_statement ::=
    case ( expression)
        function_case_item { function_case_item } endcase
    | casez ( expression)
        function_case_item { function_case_item } endcase
    | casex ( expression )
        function_case_item { function_case_item } endcase
function_case_item ::=
    expression {, expression }: function_statement_or_null
    | default [:] function_statement_or_null
```


## A.6.8 Looping statements

```
function_loop_statement ::=
    forever function_statement
    repeat ( expression ) function_statement
    | while ( expression ) function_statement
    | for (variable_assignment ; expression ; variable_assignment )
            function_statement
loop_statement ::=
    forever statement
    repeat ( expression ) statement
    while ( expression) statement
    |for ( variable_assignment ; expression ; variable_assignment )
        statement
```


## A.6.9 Task enable statements

system_task_enable $::=$ system_task_identifier [ ( expression \{ , expression \} ) ] ; task_enable ::= hierarchical_task_identifier [ ( expression \{ , expression \} )];

## A. 7 Specify section

## A.7.1 Specify block declaration

specify_block ::= specify \{ specify_item \} endspecify
specify_item ::=
specparam_declaration
| pulsestyle_declaration
| showcancelled_declaration
| path_declaration
| system_timing_check
pulsestyle_declaration ::=
pulsestyle_onevent list_of_path_outputs ;
| pulsestyle_ondetect list_of_path_outputs ;
showcancelled_declaration ::=
showcancelled list_of_path_outputs ;
| noshowcancelled list_of_path_outputs ;

## A.7.2 Specify path declarations

```
path_declaration ::=
        simple_path_declaration ;
    | edge_sensitive_path_declaration ;
    | state_dependent_path_declaration ;
simple_path_declaration ::=
    parallel_path_description = path_delay_value
    | full_path_description = path_delay_value
parallel_path_description ::=
    ( specify_input_terminal_descriptor [ polarity_operator ] => specify_output_terminal_descriptor )
full_path_description ::=
    ( list_of_path_inputs [ polarity_operator ] *> list_of_path_outputs )
list_of_path_inputs ::=
    specify_input_terminal_descriptor {, specify_input_terminal_descriptor }
list_of_path_outputs ::=
    specify_output_terminal_descriptor {, specify_output_terminal_descriptor }
```


## A.7.3 Specify block terminals

```
specify_input_terminal_descriptor ::=
    input_identifier
    | input_identifier [ constant_expression ]
    | input_identifier [ range_expression ]
specify_output_terminal_descriptor ::=
        output_identifier
    | output_identifier [ constant_expression ]
    | output_identifier [ range_expression ]
input_identifier ::= input_port_identifier | inout_port_identifier
output_identifier ::= output_port_identifier | inout_port_identifier
```


## A.7.4 Specify path delays

```
path_delay_value ::=
    list_of_path_delay_expressions
    | ( list_of_path_delay_expressions )
```

```
list_of_path_delay_expressions ::=
    t_path_delay_expression
    | trise_path_delay_expression,tfall_path_delay_expression
    | trise_path_delay_expression,tfall_path_delay_expression,tz_path_delay_expression
    | t01_path_delay_expression,t10_path_delay_expression, t0z_path_delay_expression,
    tz1_path_delay_expression,t1z_path_delay_expression,tz0_path_delay_expression
    | t01_path_delay_expression,t10_path_delay_expression,t0z_path_delay_expression,
        tz1_path_delay_expression,t1z_path_delay_expression,tz0_path_delay_expression
    t0x_path_delay_expression,tx1_path_delay_expression, t1x_path_delay_expression,
    tx0_path_delay_expression, txz_path_delay_expression, tzx_path_delay_expression
t_path_delay_expression ::= path_delay_expression
trise_path_delay_expression ::= path_delay_expression
tfall_path_delay_expression ::= path_delay_expression
tz_path_delay_expression ::= path_delay_expression
t01_path_delay_expression ::= path_delay_expression
t10_path_delay_expression ::= path_delay_expression
t0z_path_delay_expression ::= path_delay_expression
tz1_path_delay_expression ::= path_delay_expression
tlz_path_delay_expression ::= path_delay_expression
tz0_path_delay_expression ::= path_delay_expression
t0x_path_delay_expression ::= path_delay_expression
tx1_path_delay_expression ::= path_delay_expression
t1x_path_delay_expression ::= path_delay_expression
tx0_path_delay_expression ::= path_delay_expression
txz_path_delay_expression ::= path_delay_expression
tzx_path_delay_expression ::= path_delay_expression
path_delay_expression ::= constant_mintypmax_expression
edge_sensitive_path_declaration ::=
    parallel_edge_sensitive_path_description = path_delay_value
    | full_edge_sensitive_path_description = path_delay_value
parallel_edge_sensitive_path_description ::=
    ([ edge_identifier ] specify_input_terminal_descriptor =>
    specify_output_terminal_descriptor [ polarity_operator ] : data_source_expression )
full_edge_sensitive_path_description ::=
    ([ edge_identifier ] list_of_path_inputs *>
    list_of_path_outputs [ polarity_operator ] : data_source_expression )
data_source_expression ::= expression
edge_identifier ::= posedge | negedge
state_dependent_path_declaration ::=
        if ( module_path_expression ) simple_path_declaration
        | if ( module_path_expression ) edge_sensitive_path_declaration
        | ifnone simple_path_declaration
polarity_operator ::=+ | -
```


## A.7.5 System timing checks

## A.7.5.1 System timing check commands

system_timing_check ::=
\$setup_timing_check
| \$hold _timing_check
| \$setuphold_timing_check
| Srecovery_timing_check
| Sremoval_timing_check
| \$recrem_timing_check
| Sskew_timing_check
| \$timeskew_timing_check
| \$fullskew_timing_check
| \$period_timing_check
| \$width_timing_check
| \$nochange_timing_check
\$setup_timing_check ::=
\$setup ( data_event, reference_event, timing_check_limit [, [ notify_reg ] ]);
\$hold _timing_check ::=
\$hold (reference_event, data_event, timing_check_limit [, [ notify_reg ] ]) ;
\$setuphold_timing_check ::=
\$setuphold ( reference_event, data_event, timing_check_limit, timing_check_limit [, [ notify_reg][, [ stamptime_condition ][, [ checktime_condition ]
[,[delayed_reference][,[delayed_data ]]]]]]);
\$recovery_timing_check ::=
\$recovery (reference_event, data_event, timing_check_limit [, [ notify_reg ] ]) ;
\$removal_timing_check ::=
\$removal (reference_event, data_event, timing_check_limit [, [ notify_reg ] ]) ;
\$recrem_timing_check ::=
\$recrem ( reference_event, data_event, timing_check_limit, timing_check_limit [, [ notify_reg ] [, [ stamptime_condition ] [, [ checktime_condition ] [,[delayed_reference][,[delayed_data]]]]]);
\$skew_timing_check ::=
\$skew (reference_event, data_event, timing_check_limit [, [ notify_reg ] ]) ;
\$timeskew_timing_check ::=
\$timeskew ( reference_event, data_event, timing_check_limit
[, [ notify_reg ][,[ event_based_flag ][,[remain_active_flag ]]]]);
\$fullskew_timing_check ::=
\$fullskew (reference_event, data_event, timing_check_limit, timing_check_limit
[, [ notify_reg ][,[event_based_flag ][,[remain_active_flag ]]]]);
\$period_timing_check ::=
\$period (controlled_reference_event, timing_check_limit [, [ notify_reg ] ]) ;
\$width_timing_check ::=
\$width ( controlled_reference_event , timing_check_limit, threshold [, [ notify_reg] ]) ;
\$nochange_timing_check ::=
\$nochange (reference_event, data_event, start_edge_offset, end_edge_offset [, [ notify_reg ] ]) ;

## A.7.5.2 System timing check command arguments

checktime_condition ::= mintypmax_expression
controlled_reference_event ::= controlled_timing_check_event
data_event ::= timing_check_event
delayed_data ::=
terminal_identifier
| terminal_identifier [ constant_mintypmax_expression ]
delayed_reference ::=
terminal_identifier
| terminal_identifier [ constant_mintypmax_expression ]
end_edge_offset ::= mintypmax_expression
event_based_flag ::= constant_expression
notify_reg ::= variable_identifier
reference_event ::= timing_check_event
remain_active_flag ::= constant_mintypmax_expression
stamptime_condition ::= mintypmax_expression
start_edge_offset ::= mintypmax_expression
threshold ::=constant_expression
timing_check_limit ::= expression

## A.7.5.3 System timing check event definitions

timing_check_event ::=
[timing_check_event_control] specify_terminal_descriptor [ \&\&\& timing_check_condition ]
controlled_timing_check_event ::=
timing_check_event_control specify_terminal_descriptor [ \&\&\& timing_check_condition ]
timing_check_event_control ::=
posedge
| negedge
| edge_control_specifier
specify_terminal_descriptor ::=
specify_input_terminal_descriptor
| specify_output_terminal_descriptor
edge_control_specifier ::= edge [ edge_descriptor [, edge_descriptor ] ]
edge_descriptor ${ }^{1}$ ::=
01
10
| z_or_x zero_or_one
| zero_or_one z_or_x
zero_or_one ::= $\mathbf{0} \mid \mathbf{1}$
z_or_X ::= $\mathbf{x}|\mathbf{X}| \mathbf{z} \mid \mathbf{Z}$
timing_check_condition ::=
scalar_timing_check_condition
| ( scalar_timing_check_condition )

```
scalar_timing_check_condition ::=
    expression
    \(\mid \sim\) expression
    | expression == scalar_constant
    \(\mid\) expression \(===\) scalar_constant
    | expression != scalar_constant
    | expression !== scalar_constant
scalar_constant ::=
```

    1'b0 | 1'b1 | 1'B0 | 1'B1 |'b0 |'b1 |'B0 |'B1 | \(\mathbf{1} \mid \mathbf{0}\)
    
## A. 8 Expressions

## A.8.1 Concatenations

```
concatenation ::= { expression {, expression } }
constant_concatenation ::= { constant_expression { , constant_expression } }
constant_multiple_concatenation ::= { constant_expression constant_concatenation }
module_path_concatenation ::= { module_path_expression {, module_path_expression } }
module_path_multiple_concatenation ::= { constant_expression module_path_concatenation }
multiple_concatenation ::= { constant_expression concatenation }
net_concatenation ::= { net_concatenation_value { , net_concatenation_value } }
net_concatenation_value ::=
    hierarchical_net_identifier
    | hierarchical_net_identifier [ expression ] {[ expression ] }
    | hierarchical_net_identifier [ expression ] { [ expression ] } [range_expression ]
    | hierarchical_net_identifier [ range_expression ]
    | net_concatenation
variable_concatenation ::= { variable_concatenation_value { , variable_concatenation_value } }
variable_concatenation_value ::=
    hierarchical_variable_identifier
    | hierarchical_variable_identifier [ expression ] { [ expression ] }
    | hierarchical_variable_identifier [ expression ] { [ expression ] } [ range_expression ]
    | hierarchical_variable_identifier [ range_expression ]
    | variable_concatenation
```


## A.8.2 Function calls

```
constant_function_call ::= function_identifier { attribute_instance }
    ( constant_expression { , constant_expression } )
function_call ::= hierarchical_function_identifier{ attribute_instance }
    ( expression {, expression })
genvar_function_call ::= genvar_function_identifier { attribute_instance }
    ( constant_expression { , constant_expression } )
system_function_call ::= system_function_identifier
    [( expression {, expression } )]
```


## A.8.3 Expressions

base_expression ::= expression
conditional_expression $::=$ expression1 ? \{ attribute_instance \} expression2 : expression3
constant_base_expression ::= constant_expression
constant_expression ::=
constant_primary
| unary_operator \{ attribute_instance \} constant_primary
| constant_expression binary_operator \{ attribute_instance \} constant_expression
| constant_expression ? \{ attribute_instance \} constant_expression : constant_expression
| string
constant_mintypmax_expression ::= constant_expression
| constant_expression : constant_expression : constant_expression
constant_range_expression ::=
constant_expression
| msb_constant_expression : lsb_constant_expression
| constant_base_expression +: width_constant_expression
| constant_base_expression -: width_constant_expression
dimension_constant_expression ::= constant_expression
expression $1::=$ expression
expression $2::=$ expression
expression $3::=$ expression
expression ::=
primary
| unary_operator \{ attribute_instance \} primary
| expression binary_operator \{ attribute_instance \} expression
| conditional_expression
| string
lsb_constant_expression ::= constant_expression
mintypmax_expression ::=
expression
| expression : expression : expression
module_path_conditional_expression ::= module_path_expression ? \{ attribute_instance \}
module_path_expression : module_path_expression
module_path_expression $::=$
module_path_primary
| unary_module_path_operator \{ attribute_instance \} module_path_primary
| module_path_expression binary_module_path_operator \{ attribute_instance \} module_path_expression
| module_path_conditional_expression
module_path_mintypmax_expression ::=
module_path_expression
| module_path_expression : module_path_expression : module_path_expression
msb_constant_expression ::= constant_expression
range_expression ::=
expression
msb_constant_expression : lsb_constant_expression
base_expression +: width_constant_expression
| base_expression -: width_constant_expression
width_constant_expression ::= constant_expression

## A.8.4 rimaries

```
constant_primary ::=
    constant_concatenation
    | constant_function_call
    ( constant_mintypmax_expression )
    | constant_multiple_concatenation
    |genvar_identifier
    number
    parameter_identifier
    | specparam_identifier
module_path_primary ::=
    number
    | identifier
    | module_path_concatenation
    | module_path_multiple_concatenation
    | function_call
    | system_function_call
    | constant_function_call
    | ( module_path_mintypmax_expression )
primary ::=
    number
    | hierarchical_identifier
    | hierarchical_identifier [ expression ] { [ expression ] }
    | hierarchical_identifier [ expression ] { [ expression ] } [range_expression ]
    | hierarchical_identifier [ range_expression ]
    | concatenation
    | multiple_concatenation
    | function_call
    | system_function_call
    | constant_function_call
    | ( mintypmax_expression )
```


## A.8.5 Expression left-side values

```
net_lvalue ::=
        hierarchical_net_identifier
    | hierarchical_net_identifier [ constant_expression ] { [ constant_expression ] }
    | hierarchical_net_identifier [ constant_expression ] { [ constant_expression ] } [
        constant_range_expression ]
    | hierarchical_net_identifier [ constant_range_expression ]
    | net_concatenation
variable_lvalue ::=
    hierarchical_variable_identifier
    | hierarchical_variable_identifier [ expression ] { [ expression ] }
    | hierarchical_variable_identifier [ expression ] { [ expression ] } [range_expression ]
    | hierarchical_variable_identifier [ range_expression ]
    | variable_concatenation
```


## A.8.6 perators

unary_operator ::=
binary_operator ::=


unary_module_path_operator ::=
$!|\sim| \&|\sim \&||\sim| \wedge\left|\sim^{\wedge}\right| \wedge \sim$
binary_module_path_operator ::=
$==\left.|!=|\boldsymbol{\&}||| | \boldsymbol{\&}| |\right|^{\wedge}|\wedge \sim| \sim^{\wedge}$

## A.8.7 Numbers

number ::=
decimal_number
| octal_number
| binary_number
| hex_number
| real_number
real_number ${ }^{1}::=$
unsigned_number. unsigned_number
| unsigned_number [. unsigned_number ] exp [ sign ] unsigned_number
$\exp ::=\mathbf{e} \mid \mathbf{E}$
decimal_number ::=
unsigned_number
| [ size ] decimal_base unsigned_number
| [ size ] decimal_base x_digit \{ _ \}
| [ size ] decimal_base z_digit \{ _ \}
binary_number ::= [ size ] binary_base binary_value
octal_number ::= [ size ] octal_base octal_value
hex_number ::= [ size ] hex_base hex_value
sign ::=+|-
size ::= non_zero_unsigned_number
non_zero_unsigned_number ${ }^{1}::=$ non_zero_decimal_digit $\{$ _ decimal_digit \}
unsigned_number ${ }^{1}::=$ decimal_digit $\left\{\_\right.$decimal_digit \}
binary_value ${ }^{1}::=$ binary_digit $\{$ _ | binary_digit \}
octal_value $^{1}::=$ octal_digit $\{$ _ octal_digit \}
hex_value ${ }^{1}::=$ hex_digit $\{$ _ | hex_digit \}
decimal_base $\left.{ }^{1}::={ }^{\boldsymbol{r}} \mathbf{[ s |} \mathbf{S}\right] \mathbf{d} \mid \Upsilon[\mathbf{s} \mid \mathbf{S}] \mathbf{D}$
binary_base $\left.{ }^{1}::={ }^{\prime} \mathbf{[ s | S} \mid \mathbf{S}\right] \mathbf{b} \mid{ }^{9}[\mathbf{s} \mathbf{S} \mathbf{S}] \mathbf{B}$
octal_base ${ }^{1}::={ }^{\prime}[\mathbf{s} \mid \mathbf{S}] \mathbf{0} \mid{ }^{\prime}[\mathbf{s} \mid \mathbf{S}] \mathbf{O}$
hex_base $\left.{ }^{1}::={ }^{\prime}[\mathbf{s} \mid \mathbf{S}] \mathbf{h} \mid{ }^{\mathfrak{T}} \mathbf{[ s |} \mathbf{S}\right] \mathbf{H}$
non_zero_decimal_digit ::=1|2|3|4|5|6|7|8|9
decimal_digit ::= $0|\mathbf{1}| \mathbf{2} \mid \mathbf{3 | 4 | 5 | 6 | 7 | \mathbf { 7 } | \mathbf { 9 }}$
binary_digit ::= x_digit | z_digit | $\mathbf{0} \mid \mathbf{1}$
octal_digit ::= x_digit $\mid$ z_digit $|\mathbf{0}| \mathbf{1}|\mathbf{2}| \mathbf{3}|\mathbf{4}| \mathbf{5}|\mathbf{6}| \mathbf{7}$

```
hex_digit ::=
    x_digit z_digit \(^{\mathbf{0}} \mathbf{0}|\mathbf{1}| \mathbf{2}|\mathbf{3}| \mathbf{4}|\mathbf{5}| \mathbf{6}|\mathbf{7}| \mathbf{8} \mid \mathbf{9}\)
    \(|\mathbf{a}| \mathbf{b}|\mathbf{c}| \mathbf{d}|\mathbf{e}| \mathbf{f}|\mathbf{A}| \mathbf{B}|\mathbf{C}| \mathbf{D}|\mathbf{E}| \mathbf{F}\)
x_digit \(::=\mathbf{x} \mid \mathbf{X}\)
z_digit ::= \(\mathbf{z}|\mathbf{Z}|\) ?
```


## A.8.8 Strings

string ::= " \{ Any_ASCII_Characters_except_new_line \}"

## A. 9 General

## A.9.1 Attributes

attribute_instance $::=(*$ attr_spec $\{$, attr_spec $\} *$ )
attr_spec ::=
attr_name = constant_expression
| attr_name
attr_name $::=$ identifier

## A.9.2 Comments

comment ::=
one_line_comment
| block_comment
one_line_comment $::=/ /$ comment_text $\backslash n$
block_comment $::=/ *$ comment_text */
comment_text $::=\{$ Any_ASCII_character $\}$

## A.9.3 Identifiers

arrayed_identifier ::=
simple_arrayed_identifier
| escaped_arrayed_identifier
block_identifier ::= identifier
cell_identifier ::= identifier
config_identifier ::= identifier
escaped_arrayed_identifier ::= escaped_identifier [range ]
escaped_hierarchical_identifier ${ }^{4}::=$
escaped_hierarchical_branch
\{ .simple_hierarchical_branch | .escaped_hierarchical_branch \}
escaped_identifier $::=\backslash$ \{Any_ASCII_character_except_white_space\} white_space
event_identifier $::=$ identifier
function_identifier ::= identifier
gate_instance_identifier ::= arrayed_identifier
generate_block_identifier ::= identifier
genvar_function_identifier ::= identifier /* Hierarchy disallowed */
genvar_identifier ::= identifier
hierarchical_block_identifier ::= hierarchical_identifier
hierarchical_event_identifier $::=$ hierarchical_identifier
hierarchical_function_identifier ::= hierarchical_identifier
hierarchical_identifier ::=
simple_hierarchical_identifier
| escaped_hierarchical_identifier
hierarchical_net_identifier ::= hierarchical_identifier
hierarchical_variable_identifier ::= hierarchical_identifier
hierarchical_task_identifier ::= hierarchical_identifier
identifier ::=
simple_identifier
| escaped_identifier
inout_port_identifier ::= identifier
input_port_identifier ::= identifier
instance_identifier $::=$ identifier
library_identifier ::= identifier
memory_identifier $::=$ identifier
module_identifier ::= identifier
module_instance_identifier $::=$ arrayed_identifier
net_identifier ::= identifier
output_port_identifier ::= identifier
parameter_identifier ::= identifier
port_identifier ::= identifier
real_identifier ::= identifier
simple_arrayed_identifier ::= simple_identifier [range ]
simple_hierarchical_identifier ${ }^{3}::=$
simple_hierarchical_branch [ .escaped_identifier ]
simple_identifier ${ }^{2}::=\left[\mathbf{a - z A}-\mathbf{Z}_{-}\right]\left\{\left[\mathbf{a}-\mathbf{z A}-\mathbf{Z 0} \mathbf{- 9} \_\$\right]\right\}$
specparam_identifier $::=$ identifier
system_function_identifier $\left.{ }^{5}::=\mathbf{\$} \mathbf{a - z A} \mathbf{Z O} \mathbf{- 9} \_\$\right]\left\{\left[\mathbf{a - z A}-\mathbf{Z 0} \mathbf{- 9} \_\$\right]\right\}$
system_task_identifier ${ }^{5}::=\mathbf{\$}\left[\mathbf{a}-\mathbf{z A}-\mathbf{Z 0} \mathbf{- 9} \_\$\right]\left\{\left[\mathbf{a - z A}-\mathbf{Z 0}-\mathbf{9} \_\$\right]\right\}$
task_identifier ::= identifier
terminal_identifier $::=$ identifier
text_macro_identifier ::= simple_identifier
topmodule_identifier $::=$ identifier
udp_identifier ::= identifier
udp_instance_identifier $::=$ arrayed_identifier
variable_identifier ::= identifier

## A.9.4 Identifier branches

```
simple_hierarchical_branch \({ }^{3}::=\)
    simple_identifier [ [ unsigned_number ] ]
        [ \{ .simple_identifier [ [ unsigned_number ] ] \}]
escaped_hierarchical_branch \({ }^{4}::=\)
    escaped_identifier [ [ unsigned_number ] ]
        [ \{ .escaped_identifier [[unsigned_number ] ] \} ]
```


## A.9.5 White space

white_space ::= space | tab | newline | eof ${ }^{6}$

## NOTES

1) Embedded spaces are illegal.
2) A simple_identifier and arrayed_reference shall start with an alpha or underscore (_) character, shall have at least one character, and shall not have any spaces.
3) The period (.) in simple_hierarchical_identifier and simple_hierarchical_ branch shall not be preceded or followed by white_space.
4) The period in escaped_hierarchical_identifier and escaped_hierarchical_ branch shall be preceded by white_space, but shall not be followed by white_space.
5) The $\$$ character in a system_function_identifier or system_task_identifier shall not be followed by white_space. A system_function_identifier or system_task_identifier shall not be escaped.
6) End of file.

## Annex B

## (normative)

## List of keywords

Keywords are predefined nonescaped identifiers that define Verilog language constructs. An escaped identifier shall not be treated as a keyword.

| always | if | release |
| :---: | :---: | :---: |
| and | ifnone | repeat |
| assign | incdir | rnmos |
| automatic | include | rpmos |
| begin | initial | rtran |
| buf | inout | rtranif0 |
| bufif0 | input | rtranif1 |
| bufif1 | instance | scalared |
| case | integer | showcancelled |
| casex | join | signed |
| casez | large | small |
| cell | liblist | specify |
| cmos | library | specparam |
| config | localparam | strong0 |
| deassign | macromodule | strong1 |
| default | medium | supply0 |
| defparam | module | supply1 |
| design | nand | table |
| disable | negedge | task |
| edge | nmos | time |
| else | nor | tran |
| end | noshowcancelled | tranif0 |
| endcase | not | tranif1 |
| endconfig | notif0 | tri |
| endfunction | notif1 | tri0 |
| endgenerate | or | tri1 |
| endmodule | output | triand |
| endprimitive | parameter | trior |
| endspecify | pmos | trireg |
| endtable | posedge | unsigned |
| endtask | primitive | use |
| event | pullo | vectored |
| for | pullı | wait |
| force | pulldown | wand |
| forever | pullup | weak0 |
| fork | pulsestyle_onevent | weak1 |
| function | pulsestyle_ondetect | while |
| generate | remos | wire |
| genvar | real | wor |
| highz0 | realtime | xnor |
| highz1 | reg | xor |

## Annex C

## (informative)

## System tasks and functions

The system tasks and functions described in this annex are for informative purposes only and are not part of the IEEE standard Verilog HDL.

This annex describes system tasks and functions as companions to the system tasks and functions described in Clause 17. The system tasks and functions described in this annex may not be available in all implementations of the Verilog HDL. The following system tasks and functions are described in this annex:

| \$countdrivers | $[\mathrm{C} .1]$ |
| :--- | :--- |
| \$getpattern | $[\mathrm{C} .2]$ |
| \$incsave | $[\mathrm{C} .8]$ |
| \$input | $[\mathrm{C} .3]$ |
| \$key | $[\mathrm{C} .4]$ |
| \$list | $[\mathrm{C} .5]$ |
| \$log | $[\mathrm{C} .6]$ |
| \$nokey | $[\mathrm{C} .4]$ |
| \$nolog | $[\mathrm{C} .6]$ |
| \$reset | $[\mathrm{C} .7]$ |
| \$reset_count | $[\mathrm{C} .7]$ |


| \$reset_value | $[$ C. 7$]$ |
| :--- | :--- |
| \$restart | $[\mathrm{C} .8]$ |
| \$save | $[\mathrm{C} .8]$ |
| \$scale | $[\mathrm{C} .9]$ |
| \$scope | $[\mathrm{C} .10]$ |
| \$showscopes | $[\mathrm{C} .11]$ |
| \$showvars | $[\mathrm{C} .12]$ |
| \$sreadmemb | $[\mathrm{C} .13]$ |
| \$sreadmemh | [C.13] |

The word tool in this annex refers to an implementation of Verilog HDL, typically a logic simulator.

## C. 1 \$countdrivers

Syntax:
\$countdrivers (net, [ net_is_forced, number_of_01x_drivers, number_of_0_drivers, number_of_1_drivers, number_of_x_drivers ] );

The \$countdrivers system function is provided to count the number of drivers on a specified net so that bus contention can be identified.

This system function returns a 0 if there is no more than one driver on the net and returns a 1 otherwise (indicating contention). The specified net shall be a scalar or a bit-select of a vector net. The number of parameters to the system function may vary according to how much information is desired.

If additional parameters are supplied to the \$countdrivers function, each parameter returns the information described in Table C1.

Table C1 arameter return value for \$countdriver function

| Parameter | Return value |
| :--- | :--- |
| net_is_forced | 1 if net is forced <br> 0 otherwise |
| number_of_01x_drivers | An integer representing the number of <br> drivers on the net that are in 0, 1, or x <br> state. This represents the total number <br> of drivers that are not forced |
| number_of_0_drivers | An integer representing the number of <br> drivers on the net that are in 0 state |
| number_of_1_drivers | An integer representing the number of <br> drivers on the net that are in 1 state |
| number_of_x_drivers | An integer representing the number of <br> drivers on the net that are in x state |

## C. 2 \$getpattern

Syntax:
Sgetpattern ( mem_element );
The system function \$getpattern provides for fast processing of stimulus patterns that have to be propagated to a large number of scalar inputs. The function reads stimulus patterns that have been loaded into a memory using the \$readmemb or \$readmemh system tasks.

Use of this function is limited, however: it may only be used in a continuous assignment statement where the lefthand side is a concatenation of scalar nets, and the parameter to the system function is a memory element reference.

## Example:

The following example shows how stimuli stored in a file can be read into a memory using \$readmemb and applied to the circuit one pattern at a time using \$getpattern.

The memory in_mem is initialized with the stimulus patterns by the \$readmemb task. The integer variable index selects which pattern is being applied to the circuit. The for loop increments the integer variable index periodically to sequence the patterns.

```
module top;
parameter in_width=10,
        patterns=200,
        step=20;
reg [ 1:in_width] in_mem[ 1:patterns] ;
integer index;
// declare scalar inputs
wire il,i2,i3,i4,i5,i6,i7,i8,i9,i10;
// assign patterns to circuit scalar inputs (a new pattern
// is applied to the circuit each time index changes value)
assign {i1,i2,i3,i4,i5,i6,i7,i8,i9,i10} = $getpattern(in_mem[ index]);
initial begin
    // read stimulus patterns into memory
    $readmemb("patt.mem", in_mem);
    // step through patterns (note that each assignment
    // to index will drive a new pattern onto the circuit
    // inputs from the $getpattern system task specified above
    for (index = 1; index <= patterns; index = index + 1)
        #step;
end
    // instantiate the circuit module - e.g.
mod1 cct (o1,o2,o3,o4,o5, i1,i2,i3,i4,i5,i6,i7,i8,i9,i10);
endmodule
```


## C. 3 \$input

Syntax:

## \$input ("filename");

The \$input system task allows command input text to come from a named file instead of from the terminal. At the end of the command file, the input is switched back to the terminal.

## C. 4 \$key and \$nokey

Syntax:
\$key [("filename")];
\$nokey;
A key file is created whenever interactive mode is entered for the first time during simulation. The key file contains all of the text that has been typed in from the standard input. The file also contains information about asynchronous interrupts.

The \$nokey and \$key system tasks are used to disable and re-enable output to the key file. An optional file name parameter for \$key causes the old key file to be closed, a new file to be created, and output to be directed to the new file.

## C. 5 \$list

Syntax:

> \$list [ ( hierarchical_name ) ] ;

When invoked without a parameter, \$list produces a listing of the module, task, function, or named block that is defined as the current scope setting. If an optional parameter is supplied, it shall refer to a specific module, task, function or named block, in which case the specified object is listed.

## C. 6 \$log and \$nolog

Syntax:

```
$log[("filename")];
$nolog;
```

A log file contains a copy of all the text that is printed to the standard output. The log file may also contain, at the beginning of the file, the host command that was used to run the tool.

The $\mathbf{\$ n o l o g}$ and $\mathbf{\$ l o g}$ system tasks are used to disable and re-enable output to the log file. The $\mathbf{\$ n o l o g}$ task disables output to the log file, while the \$log task re-enables the output. An optional file name parameter for \$log causes the old file to be closed, a new $\log$ file to be created, and output to be directed to the new $\log$ file.

## C. 7 \$reset, \$reset_count, and \$reset_value

Syntax:

```
$reset [( stop_value [, reset_value, [ diagnostics_value ] ])];
$reset_count;
$reset_value;
```

The \$reset system task enables a tool to be reset to its Time 0 state so that processing (e.g., simulation) can begin again.

The \$reset_count system function keeps track of the number of times the tool is reset. The \$reset_value system function returns the value specified by the reset_value parameter argument to the \$reset system task. The \$reset_value system function is used to communicate information from before a reset of a tool to the time 0 state to after the reset.

The following are some of the simulation methods that can be employed with this system task and these system functions:

Determine the force statements a design needs to operate correctly, reset the simulation time to 0 , enter these force statements, and start to simulate again
Reset the simulation time to 0 and apply new stimuli
Determine that debug system tasks, such as \$monitor and \$strobe, are keeping track of the correct nets or regs, reset the simulation time to 0 , and begin simulation again

The \$reset system task tells a tool to return the processing of the design to its logical state at time 0 . When a tool executes the \$reset system task, it takes the following actions to stop the process:

1) Disables all concurrent activity, initiated in either initial or always procedural blocks in the source description or through interactive mode (disables, for example, all force and assign statements, the current \$monitor system task, and any other active tasks)
2) Cancels all scheduled simulation events

After a simulation tool executes the \$reset system task, the simulation is in the following state:
The simulation time is 0 .
All regs and nets contain their initial values.
The tool begins to execute the first procedural statements in all initial and always blocks.

The stop_value argument indicates whether interactive mode or processing is entered immediately after resetting of the tool. A value of 0 or no argument causes interactive mode to be entered after resetting the tool. A nonzero value passed to \$reset causes the tool to begin processing immediately.

The reset value argument is an integer, which specifies the value that shall be returned by the \$reset_value system function after the tool is reset. All declared integers return to their initial value after reset, but entering an integer as this argument allows access to what its value was before the reset with the \$reset_value system function. This argument provides a means of communicating information from before the reset of a tool to after the reset of the tool.

The diagnostic_value specifies the kind of diagnostic messages a tool displays before it resets the time to 0 . Increasing integer values results in increased information. A value of zero results in no diagnostic message

## C. 8 \$save, \$restart, and \$incsave

Three system tasks \$save, \$restart, and \$incsave work in conjunction with one another to save the complete state of simulation into a permanent file such that the simulation state can be reloaded at a later time and processing can continue where it left off.

Syntax:

$$
\begin{aligned}
& \text { \$save("file_name"); } \\
& \text { \$restart("file_name"); } \\
& \text { \$incsave("incremental_file_name"); }
\end{aligned}
$$

All three system tasks take a file name as a parameter. The file name has to be supplied as a string enclosed in quotation marks.

The \$save system task saves the complete state into the host operating system file specified as a parameter.
The \$incsave system task saves only what has changed since the last invocation of \$save. It is not possible to do an incremental save on any file other than the one produced by the last \$save.

The \$restart system task restores a previously saved state from a specified file.
Restarting from an incremental save is similar to restarting from a full save, except that the name of the incremental save file is specified in the restart command. The full save file that the incremental save file was based upon shall still be present, as it is required for a successful restart. If the full save file has been changed in any way since the incremental save was performed, errors will result.

The incremental restart is useful for going back in time. If a full save is performed near the beginning of processing, and an incremental save is done at regular intervals, then going back in time is as simple as restarting from the appropriate file.

## Example:

```
module checkpoint;
initial
    #500 $save("save.dat"); // full save
always begin // incremental save every 10000 units,
                            // files are recycled every 40000 units
    #100000 $incsave("inc1.dat");
    #100000 $incsave("inc2.dat");
    #100000 $incsave("inc3.dat");
    #100000 $incsave("inc4.dat");
end
endmodule
```


## C. 9 \$scale

Syntax:
\$scale ( hierarchical_name ) ;
The \$scale function takes a time value from a module with one time unit to be used in a module with a different time unit. The time value is converted from the time unit of one module to the time unit of the module that invokes \$scale.

## C. 10 \$scope

Syntax.
\$scope ( hierarchical_name ) ;

The \$scope system task allows a particular level of hierarchy to be specified as the scope for identifying objects. This task accepts a single parameter argument that shall be the complete hierarchical name of a module, task, function, or named block. The initial setting of the interactive scope is the first top-level module.

## C. 11 \$showscopes

Syntax.
\$showscopes [ ( n ) ];

The \$showscopes system task produces a complete list of modules, tasks, functions, and named blocks that are defined at the current scope level. An optional integer parameter can be given to \$showscopes. A nonzero parameter value causes all the modules, tasks, functions, and named blocks in or below the current hierarchical scope to be listed. No parameter or a zero value results in only objects at the current scope level to be listed.

## C. 12 \$showvars

Syntax.
\$showvars [ ( list_of_variables )] ;
The \$showvars system task produces status information for reg and net variables, both scalar and vector. When invoked without parameters, \$showvars displays the status of all variables in the current scope. When invoked with a list of variables, \$showvars shows only the status of the specified variables. If the list of variables includes a bit-select or part-select of a reg or net, then the status information for all the bits of that reg or net are displayed.

## C. 13 \$sreadmemb and \$sreadmemh

Syntax:

> \$sreadmemb ( mem_name, start_address, finish_address, string $\{$, string $\}) ;$
> \$sreadmemh $($ mem_name, start_address, finish_address, $\operatorname{string}\{$, string $\}) ;$

The system tasks \$sreadmemb and \$sreadmemh load data into memory mem_name from a character string.

The \$sreadmemh and \$sreadmemb system tasks take memory data values and addresses as string arguments. The start and finish addresses indicate the bounds for where the data from strings will be stored in the memory. These strings take the same format as the strings that appear in the input files passed as arguments to \$readmemb and \$readmemh.

## Annex D

## (informative)

## Compiler directives

The compiler directives described in this annex are for informative purposes only and are not part of the IEEE standard Verilog HDL.

This annex describes additional compiler directives as companions to the compiler directives described in Clause 19. The compiler directives described in this annex may not be available in all implementations of the Verilog HDL. The following compiler directives are described in this annex:

| 'default_decay_time | $[\mathrm{D.1]}$ | 'delay_mode_path | [D.4] |
| :--- | :--- | :--- | :--- |
| 'default_trireg_strength | $[\mathrm{D.2]}$ | 'delay_mode_unit | $[\mathrm{D.5]}$ |
| 'delay_mode_distributed | $[\mathrm{D.3]}$ | 'delay_mode_zero | [D.6] |

The word tool in this annex refers to an implementation of Verilog HDL, typically a logic simulator.

## D. 1 'default_decay_time

The 'default_decay_time compiler directive specifies the decay time for the trireg nets that do not have any decay time specified in the declaration. This compiler directive applies to all of the trireg nets in all the modules that follow it in the source description. An argument specifying the charge decay time shall be used with this compiler directive.

Syntax:
'default_decay_time integer_constant | real_constant | infinite

## Examples:

Example 1 The following example shows how the default decay time for all trireg nets can be set to 100 time units:
'default_decay_time 100
Example 2 The following example shows how to avoid charge decay on trireg nets:
'default_decay_time infinite
The keyword infinite specifies no charge decay for all the trireg nets that do not have decay time specification.

## D. 2 'default_trireg_strength

The 'default_trireg_strength compiler directive specifies the charge strength of trireg nets.

Syntax:
'default_trireg_strength integer_constant
The integer constant shall be between 0 and 250. It indicates the relative strength of the capacitance on the trireg net.

## D. 3 'delay_mode_distributed

The 'delay_mode_distributed compiler directive specifies the distributed delay mode for all modules that follow this directive in the source description.

Syntax:
'delay_mode_distributed
This compiler directive shall be used before the declaration of the module whose delay mode is being controlled.

## D. 4 'delay_mode_path

The 'delay_mode_path compiler directive specifies the path delay mode for all modules that follow this directive in the source description.

Syntax:
'delay_mode_path
This compiler directive shall be used before the declaration of the module whose delay mode is being controlled.

## D. 5 delay_mode_unit

The 'delay_mode_unit compiler directive specifies the unit delay mode for all modules that follow this directive in the source description.

Syntax:
'delay_mode_unit

This compiler directive shall be used before the declaration of the module whose delay mode is being controlled.

## D. 6 'delay_mode_zero

The 'delay_mode_zero compiler directive specifies the zero-delay mode for all modules that follow this directive in the source description.

Syntax:
'delay_mode_zero

This compiler directive shall be used before the declaration of the module whose delay mode is being controlled.

## Annex E

## (normative)

## acc_user.h

```
/*****************************************************************************
    * acc_user.h
    * IEEE 1364-2000 Verilog HDL Programming Language Interface (PLI).
    * This file contains the constant definitions, structure definitions, and
    * routine declarations for the Verilog Programming Language Interface ACC
    * access routines.
*
    ****************************************************************************/
#ifndef ACC USER H
#define ACC_USER_H
#ifdef "\overline{C"cplusplus}
extern "\overline{C"}
#endif
/* ---------------------------------------------------------------------------------------------
/*--------------------------- Portability Help --------------------------------------------
/*------------------------------------------------------------------------------------
/* Sized variables */
#ifndef PLI TYPES
#define PLI TYPES
typedef int-` PLI_INT32;
typedef unsigned int PLI_UINT32;
typedef short PLI_INT16;
typedef unsigned short PLI-UINT16;
typedef char PLI_BYTE8;
typedef unsigned char PLI_UBYTE8;
#endif
/* export a symbol * /
#if WIN32
#ifndef PLI DLLISPEC
#define PLI_DLLISPEC __declspec(dllimport)
#define ACC}\mp@subsup{}{}{-}\mathrm{ USER DEFINED DLLISPEC 1
#endif
#else
#ifndef PLI DLLISPEC
#define PLI-DLLISPEC
#endif
#endif
/* import a symbol */
#if WIN32
#ifndef PLI_DLLESPEC
#define PLI_DLLESPEC declspec(dllexport)
#define ACC_USER_DEFI\overline{NED_DLLESPEC 1}
#endif
#else
#ifndef PLI_DLLESPEC
#define PLI-DLLESPEC
#endif
#endif
/* mark a function as external */
#ifndef PLI EXTERN
#define PLI_EXTERN
#endif
```

```
/* mark a variable as external */
#ifndef PLI VEXTERN
#define PLI_VEXTERN extern
#endif
#ifndef PLI PROTOTYPES
#define PLI-PROTOTYPES
#define PROT̄O PARAMS (params) params
/* object is imported by the application */
#define XXTERN PLI EXTERN PLI DLLISPEC
/* object is exported by the application */
#define EETERN PLI_EXTERN PLI_DLLESPEC
#endif
/*
    * The following group of defines exists purely for backwards compatibility
    * /
#ifndef PLI EXTRAS
#define PLI EXTRAS
#define bool int
#define true
#define TRUE
#define
Lalse - < 
#define FALSE _ - < 
#define null OL
#endif
```



/*----------------------------- general defines ------------------------------------ /
typedef PLI INT32 *HANDLE;
\#ifndef VPI USER CDS H
typedef $\mathrm{PLI}^{-}$INT3 $\overline{2}{ }^{\text {}}$ handle;
\#endif

\#define accModule 20
\#define accScope 21
\#define accNet 25
\#define accReg 30
\#define accRegister accReg
\#define accPort 35
\#define accTerminal 45
\#define accInputTerminal 46
\#define accOutputTerminal 47
\#define accInoutTerminal 48
\#define accCombPrim 140
\#define accSeqPrim $\quad 142$
\#define accAndGate 144
\#define accNandGate 146
\#define accNorGate 148
\#define accOrGate 150
\#define accXorGate 152
\#define accXnorGate 154
\#define accBufGate 156
\#define accNotGate 158
\#define accBufif0Gate 160
\#define accBufif1Gate 162
\#define accNotif0Gate 164
\#define accNotif1Gate 166
\#define accNmosGate 168
\#define accPmosGate 170
\#define accCmosGate 172
\#define accRnmosGate 174
\#define accRpmosGate 176
\#define accRcmosGate 178
\#define accRtranGate 180
\#define accRtranif0Gate 182
\#define accRtranif1Gate 184

| \#define | accTranGate | 186 |
| :---: | :---: | :---: |
| \#define | accTranif0Gate | 188 |
| \#define | accTranif1Gate | 190 |
| \#define | accPullupGate | 192 |
| \#define | accPulldownGate | 194 |
| \#define | accIntegerParam | 200 |
| \#define | accIntParam | accIntegerParam |
| \#define | accRealParam | 202 |
| \#define | accStringParam | 204 |
| \#define | accPath | 206 |
| \#define | accTchk | 208 |
| \#define | accPrimitive | 210 |
| \#define | accBit | 212 |
| \#define | accPortBit | 214 |
| \#define | accNetBit | 216 |
| \#define | accRegBit | 218 |
| \#define | accParameter | 220 |
| \#define | accSpecparam | 222 |
| \#define | accTopModule | 224 |
| \#define | accModuleInstance | 226 |
| \#define | accCellInstance | 228 |
| \#define | accModPath | 230 |
| \#define | accWirePath | 234 |
| \#define | accInterModPath | 236 |
| \#define | accScalarPort | 250 |
| \#define | accBitSelectPort | 252 |
| \#define | accPartSelectPort | 254 |
| \#define | accVectorPort | 256 |
| \#define | accConcatPort | 258 |
| \#define | accWire | 260 |
| \#define | accWand | 261 |
| \#define | accWor | 262 |
| \#define | accTri | 263 |
| \#define | accTriand | 264 |
| \#define | accTrior | 265 |
| \#define | accTrio | 266 |
| \#define | accTril | 267 |
| \#define | accTrireg | 268 |
| \#define | accSupply0 | 269 |
| \#define | accSupply1 | 270 |
| \#define | accNamedEvent | 280 |
| \#define | acceventVar | accNamedEvent |
| \#define | accIntegerVar | 281 |
| \#define | accIntVar | 281 |
| \#define | accRealVar | 282 |
| \#define | accTimeVar | 283 |
| \#define | accScalar | 300 |
| \#define | accVector | 302 |
| \#define | accCollapsedNet | 304 |
| \#define | accExpandedVector | 306 |
| \#define | accUnExpandedVector | 307 |
| \#define | accProtected | 308 |
| \#define | accSetup | 366 |
| \#define | acchold | 367 |
| \#define | accWidth | 368 |
| \#define | accPeriod | 369 |
| \#define | accRecovery | 370 |
| \#define | accSkew | 371 |
| \#define | accNochange | 376 |
| \#define | accNoChange | accNochange |
| \#define | accSetuphold | 377 |
| \#define | accInput | 402 |
| \#define | accoutput | 404 |
| \#define | accInout | 406 |
| \#define | accMixedIo | 407 |
| \#define | accPositive | 408 |
| \#define | accNegative | 410 |
| \#define | accUnknown | 412 |
| \#define | accPathTerminal | 420 |
| \#define | accPathInput | 422 |
| \#define | accPathOutput | 424 |
| \#define | accDataPath | 426 |




```
typedef struct t_acc_time
{
    PLI_INT32 type;
    PLI_INT32 low,
        high;
    double real;
} s_acc_time, *p_acc_time;
/*----------------- data structure used with acc set value() ----------------*/
typedef struct t_setval_delay
{
    s_acc_time time;
    PLI INTT32 model;
} s_sētval_delay, *p_setval_delay;
/*--------------------- data structure of vector values ----------------------*/
typedef struct t_acc_vecval
{
    PLI_INT32 aval;
    PLI INT32 bval;
} s_a\overline{ccc_vecval, *p_acc_vecval;}
/*------ data structure used with acc_set_value() and acc_fetch_value() -----*/
typedef struct t setval value
{
    PLI_INT32 format;
    uniōn
            {
                PLI BYTE8 *str;
                PLI-INT32 scalar;
                    PLI-INT32 integer;
                dou\overline{le real;}
                p_acc_vecval vector;
            } value;
} s_setval_value, *p_setval_value, s_acc_value, *p_acc_value;
/*----------------------- structure for VCL strengths ---------------------------*/
typedef struct t_strengths
{
    PLI UBYTE8 logic value;
    PLI_UBYTE8 streng}th1
    PLI-UBYTE8 strength2;
} s_strengths, *p_strengths;
/*--------------- structure passed to callback routine for VCL --------------*/
typedef struct t_vc_record
{
    PLI_INT32 vC_reason;
    PLI-INT32 vC`hightime;
    PLI_INT32 vc_lowtime;
    PLI_BYTE8 *usēr_data;
    uniōn
            {
                PLI UBYTE8 logic_value;
                dou\overline{ble real_value;}
                handle vector handle;
                s_strengths strengths_s;
        out value
} s_vc_record, *p_vc_record;
/*------------- structure used with acc fetch location() routine ------------**/
typedef struct t_location
{
    PLI_INT32 line_no;
    PLI BYTE8 * fileñame;
} s_location, *p_location;
/*---------- structure used with acc_fetch_timescale_info() routine ----------*/
typedef struct t_timescale_info
```

\{
PLI_INT16 unit;
PLI-INT16 precision;
\} s_tïmescale_info, *p_timescale_info;

XXTERN PLI_INT32 acc_append_delays PROTO_PARAMS((handle object, ...));
XXTERN PLI-INT32 acc_append_pulsere PROTO_PARAMS((handle object, double vallr, double val̄̄x, ...));
XXTERN void acc_close PROTO_PARAMS((void));
XXTERN handle *acc-collect PROTO_PARAMS ( (handle (*p_next_routine) (), handle scope_object, PLI_INT32 *aof_count)) ;
XXTERN PLI_INT32 - acc_compare_handles PROTO_PARAMS ((handle h1, handle h2));
XXTERN PLI_INT32
*value));
XXTERN PLI INT32
object_hand̄le));
XXTERN PLI INT32
XXTERN PLI-BYTE8 *
XXTERN double
XXTERN PLI INT32
XXTERN PLI-BYTE8
XXTERN PLI BYTE8
XXTERN PLI-INT32
XXTERN PLI_INT32
XXTERN PLI-INT32
XXTERN PLI_INT32
XXTERN PLI-BYTE8
XXTERN PLI_INT32
XXTERN PLI-INT32
XXTERN double
XXTERN PLI_INT32
acc_configure PROTO_PARAMS̄((PLI_INT32 item, PLI_BYTE8
acc_count PROTO_PARAMS ((handle (*next_func) (), handle
acc_fetch_argc PROTO_PARAMS((void));
*acc-fetch_argv PROTO-PARAMS ((void));
acc_fetch_attribute $\overline{\text { PROTO_PARAMS ( }}$ (handle object, ...));
acc_fetch_attribute_int PR̄OTO_PARAMS((handle object, ...));
*acc_fetch_attribute_-str PROTO-PARAMS ((handle object, ....));
*acc-fetch_defname P $\bar{R} O T O$ PARAMS̄((handle object handle)); acc_fetch_delay_mode PRŌTO_PARAMS ((handle objēet_p)); acc-fetch delays PROTO PARĀMS ((handle object, ...-));
acc-fetch_direction PRŌTO_PARAMS((handle object_handle)); acc-fetch ${ }^{-}$edge PROTO PARAM̄ ((handle acc_obj));
*acc_fetch_fullname Pर्ROTO_PARAMS((handle-object_handle)); acc-fetch-fulltype PROTO-PARAMS ((handle object-h)); acc_fetch_index PROTO_PARAMS((handle object_hañle)); acc_fetch_itfarg PROTO PARAMS ((PLI INT32 n, handle tfinst));

tfinst));
XXTERN PLI_BYTE8
*acc_fetch_itfarg_str PROTO_PARAMS ((PLI_INT32 n, handle
tfinst));
XXTERN PLI_INT32 acc_fetch_location PROTO_PARAMS ((p_location location_p,
handle objēct));
XXTERN PLI BYTE8
XXTERN PLI-INT32
*acc_fetch_name PROTO_PARAMS((handle object_handle));
acc_fetch_paramtype $\overline{\text { PROTO }}$ _PARAMS ( (handle param_p));
acc_fetch_paramval PROTO_Р्РARAMS ((handle param));
acc_fetch_polarity PROTO-PARAMS((handle path));
acc_fetch_precision PROTO_PARAMS((void));
acc_fetch_pulsere PROTO_PÁRAMS ( (handle path_p, double *vallr,
XXTERN PLI INT32

PLI_INT32 ${ }^{\text {® }}$ lsb));
XXTĒRN PLI INT32 acc fetch size PROTO PARAMS((handle obj h));
XXTERN doū̄le acc-fetch_tfarg PROTO_PARAMS((PLI_INT32-n));
XXTERN PLI_INT32 acc_fetch_tfarg_int Pर्ROTO_PARAMS ( (PLI_INT32 n));
XXTERN PLI_BYTE8 *acc_fetch_tfarg_str PROTO_PARAMS ((PLI_INT32 n));
XXTERN voī̄ acc-fetch ${ }^{-}$times $\bar{c} a l e \quad i n f o ~ \overline{P R O T O}$ PARAMS $($ (handle obj,
p timescale info aof timescale info));

XXTERN PLI-BYTE8 *acc-fetch type str PR̄TO PARAMS ((PLI INT32 ${ }^{-}$type));
XXTERN PLI_BYTE8 *acc_fetch_value PROTO_PARAMS ( (handle ōbject_handle, PLI_BYTE8

XXTERN void - -acc_free $\overline{\mathrm{P} R O T O}$ _PARAMS ((handle *array_ptr));
XXTERN handle acc_handle_by_name PROTO_PARAMS((PLI_BYTE8 *inst_name, handle
scope_p));
XXTER $\bar{N}$ handle
acc_handle_condition PROTO_PARAMS ((handle obj));
acc_handle_conn PROTO_PARAM̄ ((handle term_p));
acc_-handle_datapath P $\bar{R} O T O$ PARAMS ( (handle $\bar{p}$ ath)) ;
acc-handle-hiconn PROTO PARAMS ((handle port ref));
acc_handle_interactive_-̄scope PROTO PARAMS((v̌oid));
acc_handle_itfarg PROTO_PARAMS ((PLİ_INT32 n, handle tfinst));
acc-handle-loconn PROTO-PARAMS((han̄̄le port_ref));
acc handle_modpath PROTŌ PARAMS ( (handle mod_p, PLI_BYTE8
andle
XXTERN handle
*pathin_name, PLI_BYTĒ̄ *pathōut_name, ...));

XXTERN handle
XXTERN handle
XXTERN handle XXTERN handle destination)); XXTERN handle
XXTERN handle XXTERN handle port num, ...)) XXTE $\bar{R} N$ handle XXTERN handle XXTERN handle tchk type, PLI BYTE XXTERN handle handle XXTERN handle XXTERN handle PLI INT32 terminal XXTERN handle XXTERN handle XXTERN PLI INT32 XXTERN han $\bar{d} l e$ handle h object));
XXTERN hāndle XXTERN handle XXTERN handle load));
XXTERN handle child));
XXTERN handle XXTERN handle XXTERN handle XXTERN handle XXTERN handle XXTERN handle XXTERN handle XXTERN handle XXTERN handle param));
XXTERN handle
XXTERN handle
XXTERN handle
prim));
XXTERN handle scope));
XXTERN handle sparam));
XXTERN handle
XXTERN handle term));
XXTERN handle
XXTERN PLI INT32
*type list));
XXTERN PLI_INT32
type));
XXTERN PLI INT32
XXTERN PLI BYTE8
XXTERN PLI INT32
XXTERN PLI ${ }^{-}$INT32
XXTERN PLI TNT32
vallr, double vallx
XXTERN void
acc reset
XXTERN PLI_INT32 acc_-set_interactive_scōpe PROTO_PARAMS (handle scope,
PLI INT32 callback flag));
XXTĒRN PLI_INT32 -acc_set_pulsere PROTO_PARAMS ( (handle path_p, double vallr, double val1e));
XXTERN PLI_BYTE8 *acc_set_scope PROTO_PARAMS ((handle object, ...));
XXTERN PLI-INT32 acc_set-value PROTO_PARAMS (handle obj, p_setval value
setval_p, $\left.\left.\bar{p} \_s e t v a l \_d e l a y ~ d e ̄ l a y \_p\right)\right) ; ~ ; ~$
XXTERN void -acc vcl add PROTO PARAMS ( (handle object p, PLI INT32

XXTERN void - - acc_vcl_delete PROTO_PARAMS ((hande object-p, PLI INT32
(* consumer) (p_vc_record), $\overline{\text { PLI_ }}$ _BYTE8 *user__data, PLI_INT32 vcl_flags));

```
XXTERN PLI_BYTE8 *acc_version PROTO_PARAMS((void));
/*---------------------------------------------------------------------------------*/
/*------------------------ global variable definitions ----------------------------*/
/* ---------------------------------------------------------------------------------------------
PLI_VEXTERN PLI_DLLISPEC PLI_INT32 acc_error_flag;
/* --------------------------------------------------------------------------------------* /
/*---------------------------- macro definitions -----------------------------------*/
/*----------------------------------------------------------------------------------------*/
#define acc_handle_calling_mod_m acc_handle_parent((handle)tf_getinstance())
#undef PLI_EXTERN
#undef PLI_VEXTERN
#ifdef ACC_USER_DEFINED_DLLISPEC
#undef ACC_USER_DEFINED_DLLISPEC
#undef PLI DLLISPEC
#endif
#ifdef ACC USER DEFINED DLLESPEC
#undef ACC_USER-DEFINED_-DLLESPEC
#undef PLI_DLLE\overline{S}PEC
#endif
#ifdef PLI PROTOTYPES
#undef PLI-PROTOTYPES
#undef PROTO PARAMS
#undef XXTER\overline{N}
#undef EETERN
#endif
#ifdef __cplusplus
}
#endif
#endif /* ACC_USER_H * /
```


## Annex F

## (normative)

## veriuser.h

```
/*****************************************************************************
    * veriuser.h
    * IEEE 1364-2000 Verilog HDL Programming Language Interface (PLI).
*
    * This file contains the constant definitions, structure definitions, and
    * routine declarations for the Verilog Programming Language Interface TF
    * task/function routines.
*
    ****************************************************************************/
#ifndef VERIUSER_H
#define VERIUSER_H
#ifdef __cplusplus
extern "\overline{C" {}
#endif
/*-------------------------------------------------------------------------------------
/*--------------------------- Portability Help ----------------------------------------
/* ------------------------------------------------------------------------------------
/* Sized variables */
#ifndef PLI_TYPES
#define PLI TYPES
typedef int }\mp@subsup{}{}{-}\mathrm{ PLI_INT32;
typedef unsigned int PLI_UINT32;
typedef short PLI INT16;
typedef unsigned short PLI_UINT16;
typedef char PLI_BYTE8;
typedef unsigned char PLI_UBYTE8;
#endif
/* export a symbol * /
#if WIN32
#ifndef PLI_DLLISPEC
#define PLI_DLLISPEC __declspec(dllimport)
#define VERIUUSER DEFINED DLLISPEC 1
#endif
#else
#ifndef PLI_DLLISPEC
#define PLI_DLLISPEC
#endif
#endif
/* import a symbol * /
#if WIN32
#ifndef PLI_DLLESPEC
#define PLI_DLLESPEC __declspec(dllexport)
#define VERİUSER DEFI\overline{NED DLLESPEC 1}
#endif
#else
#ifndef PLI DLLESPEC
#define PLI_DLLESPEC
#endif
#endif
```

```
/* mark a function as external */
#ifndef PLI EXTERN
#define PLI_EXTERN
#endif
/* mark a variable as external */
#ifndef PLI VEXTERN
#define PLI_VEXTERN extern
#endif
#ifndef PLI_PROTOTYPES
#define PLI PROTOTYPES
#define PROTO_PARAMS (params) params
/* object is defined imported by the application */
#define XXTERN PLI_EXTERN PLI_DLLISPEC
/* object is exported by the application * /
#define EETERN PLI EXTERN PLI DLLESPEC
#endif
/*
    * The following group of defines exists purely for backwards compatibility
* /
#ifndef PLI_EXTRAS
#define PLI EXTRAS
#define boo\overline{l int}
#define true 1
#define TRUE 1
#define false 0
#define FALSE 0
#define null OL
#endif
```



```
/*-------------- values for reason parameter to misctf routines --------------*/
#define reason_checktf 1
#define REASON_CHECKTF reason_checktf
#define reason`sizetf 2
#define REASON_SIZETF reason_sizetf
#define reason calltf 3
#define REASON CALLTF reason calltf
#define reason save 4
#define REASON_SAVE reason_save
#define reason_restart 5
#define REASON RESTART reason restart
#define reason_disable 6
#define REASON_DISABLE reason_disable
#define reason paramvc 7
#define REASON_PARAMVC reason_paramvc
#define reason synch 8
#define REASON_SYNCH reason_synch
```

```
#define reason_finish 9
#define REASON_FINISH reason_finish
#define reason reactivate 10
#define REASON_REACTIVATE reason_reactivate
#define reason rosynch 11
#define REASON_ROSYNCH reason_rosynch
#define reason_paramdrc 15
#define REASON PARAMDRC
#define reason_endofcompile
#define REASON ENDOFCOMPILE
#define reason_scope
#define RFASON SCOPE
#define reason_interactive
#define REASON_INTERACTIVE
#define reason reset
#define REASON }\mp@subsup{}{}{-}\mathrm{ RESET
#define reason endofreset
#define REASON ENDOFRESET
#define reason_force
#define REASON FORCE
#define reason_release
#define REASON_RELEASE reason_release
#define reason startofsave 27
#define reason_startofrestart 28
#define REASON_MAX 28
/*-- types used by tf_typep() and expr_type field in tf_exprinfo structure --*/
#define tf nullparam 0
#define TF_NULLPARAM tf_nullparam
#define tf string
    1
#define TF_STRING
tf_string
#define tf_readonly
#define TF READONLY
#define tf_readwrite
#define TF READWRITE
#define tf_rwbitselect
#define TF_RWBITSELECT
    12
f_rwbitselect
#define tf rwpartselect
#define TF_RWPARTSELECT
tf_rwpartselect
#define tf rwmemselect
1 4
#define TF_RWMEMSELECT
tf_rwmemselect
#define tf_readonlyreal
15
#define TF READONLYREAL
tf readonlyreal
#define tf_readwritereal
1 6
#define TF_READWRITEREAL
tf_readwritereal
/*---------- types used by node type field in tf_nodeinfo structure ---------**/
#define tf_null node 100
#define TF_NULL_NODE
    tf_null_node
#define tf reg_node
    101
#define TF_'REG_NODE
    tf_reg_node
#define tf_intēger_node
    102
#define TF INTEGER NODE
    tf_integer_node
#define tf_time no\overline{de}
103
#define TF_TIME_NODE
    tf_time_node
#define tf_netvēctor node
    104
#define TF_NETVECTOR_NODE
    tf_netvector_node
#define tf netscalar node
    105
#define TF_NETSCALAR_NODE
    tf_netscalar_node
#define tf_memory_node
    106
#define TF_MEMORY_NODE
tf memory node
#define tf_real_node
107
#define TF_REAL_NODE
tf_real_node
```

```
/* ----------------------------------------------------------------------------------*/
/*-------------------------- structure definitions ------------------------------------
/*---------------------------------------------------------------------------------------*/
/*----- structure used with tf_exprinfo() to get expression information -----*/
typedef struct t_tfexprinfo
{
    PLI_INT16 expr_type;
    PLI_INT16 padding;
    strūct t_vecval *expr_value_p;
    double real_value;
    PLI_BYTE8 *expr_string;
    PLI_INT32 expr_ngroups;
    PLI_INT32 expr_vec_size;
    PLI_INT32 expr_sign;
    PLI_INT32 expr_lhs_select;
    PLI-INT32 expr-rhs-select;
} s_t\overline{f}exprinfo, *p_tfexp
/*------- structure for use with tf_nodeinfo() to get node information ------*/
typedef struct t_tfnodeinfo
{
        PLI_INT16 node_type;
        PLI_INT16 padding;
        uniōn
            {
                struct t_vecval *vecval_p;
                struct t_strengthval *strengthval_p;
                PLI BYTE8 * *memoryval_p;
                dou\overline{ble *real_val_p;};
            } node_value;
    PLI BYTE}\overline{8}\mathrm{ * node symbol;
    PLI_INT32 node_ngroups;
    PLI_INT32 node_vec_size;
    PLI_INT32 node_sign}
    PLI_INT32 node_ms_index;
    PLI_INT32 node_ls_index;
    PLI_INT32 node_mem_size;
    PLI_INT32 node_lhs_element;
    PLI_INT32 node_rhs_element;
    PLI_INT32 * node_handle;
} s_tf
/*--------------------- data structure of vector values ------------------------*/
typedef struct t_vecval
{
    PLI INT32 avalbits;
    PLI_INT32 bvalbits;
} s_vecval, *p_vecval;
/*--------------- data structure of scalar net strength values ----------------*/
typedef struct t_strengthval
{
    PLI_INT32 strength0;
    PLI_INT32 strength1;
} s_strengthval, *p_strengthval;
```

```
/*------------------------------------------------------------------------------------*/
```

/*------------------------------------------------------------------------------------*/
/*---------------------------- routine definitions -------------------------------------
/*---------------------------- routine definitions -------------------------------------
/* ------------------------------------------------------------------------------------------------

```
/* ------------------------------------------------------------------------------------------------
```

```
XXTERN void io_mcdprintf PROTO_PARAMS((PLI_INT32 mcd, PLI_BYTE8
* format, ...));
XXTERN void io_printf PROTO PARAMS((PLI BYTE8 * format, ...));
XXTERN PLI_BYTE8 *mc_scan_plusargs PROTO_PARAMS((PLI_BYTE8 *plusarg));
XXTERN PLI INT32 tf add long PROTO PARAMS((PLI INT32 * aof lowtime1,
PLI_INT32 \staraof_hightime1, \overline{PLI_INT32 lowtim}e2, PLI_INT3}2 hightime\overline{2));
XXT\overline{ERN PLI_INT\overline{32}}\mathrm{ tf_asȳnchoff PROTO_PARAMS((void));}
XXTERN PLI_INT32 tf_asynchon PROTO_PARAMS((void));
XXTERN PLI_INT32 tf_clearalldelays PROTO_PARAMS((void));
XXTERN PLI_INT32 tf_compare_long PROTO_PARAMS((PLI_UINT32 low1, PLI_UINT32
high1, PLI_UINT32 low2, PLI_UINT32 high2));
XXTERN PLI_INT32 tf_COpypvc_flag PROTO_PARAMS((PLI_INT32 nparam));
XXTERN voi\overline{d tf_divide I}\⿴⿱冂一⿱一一厶儿
PLI_INT32 *aof_high1, PLI_INT32 low2, PLI_INT32 high2));
XXTERN PLI INT3}2 t\overline{f dofinish PROTO PARAMS((void));
XXTERN PLI_INT32 tf_dostop PROTO_PA\overline{RAMS ((void));}
XXTERN PLI_INT32 tf_error PROTO_PARAMS((PLI_BYTE8 * fmt, ...));
XXTERN PLI_INT32 tf_evaluatep PROTO PARAMS((PLI_INT32 pnum));
```



```
pinfo));
XXTERN PLI_BYTE8 *tf_getcstringp PROTO_PARAMS((PLI_INT32 nparam));
XXTERN PLI_BYTE8 *tf_getinstance PROTO_PARAMS((voi\overline{d}));
XXTERN PLI_INT32 tf_getlongp PROTO_PARAMS((PLI_INT32 *aof_highvalue,
PLI_INT32 pnum));
XXTERN PLI_INT32 tf_getlongtime PROTO_PARAMS((PLI_INT32 *aof_hightime));
XXTERN PLI_INT32 tf_getnextlongtime PRO
PLI_INT32 *aof_hightime));
XXTERN PLI INT}\overline{3}2 tf getp PROTO PARAMS((PLI INT32 pnum))
XXTERN PLI_INT32 tf_getpchange PROTO_PARAMS
XXTERN double tf_getrealp PROTO_PARAMS((PLI_INT32 pnum));
XXTERN double tf_getrealtime PROTTO_PARAMS((\overline{void));}
XXTERN PLI_INT32 tf_gettime PROTO_PARAMS((void));
XXTERN PLI_INT32 tf_gettimeprecision PROTO PARAMS((void));
XXTERN PLI_INT32 tf_gettimeunit PROTO_PARAMS((void));
XXTERN PLI BYTE8 *tf getworkarea PROTO PARAMS((void));
XXTERN PLI_INT32 tf_iasynchoff PROTO_PARAMS((PLI_BYTE8 *inst));
```



```
XXTERN PLI_INT32 tf_iclearalldelays PROTO PARAMS((PLI_BYTE8 * inst));
XXTERN PLI_INT32 tf_icopypvc_flag PROTO_P\overline{ARAMS((PLI_IN}T32 nparam,
PLI BYTE8 *inst));
XXTERN PLI_INT32 tf_ievaluatep PROTO_PARAMS((PLI_INT32 pnum, PLI_BYTE8
*inst));
XXTERN p_tfexprinfo tf_iexprinfo PROTO_PARAMS((PLI_INT32 pnum, p_tfexprinfo
pinfo, PLI_BYTE8 *inst));
XXTERN PLI_BYTE8 *tf_igetcstringp PROTO_PARAMS((PLI_INT32 nparam, PLI_BYTE8
*inst));
XXTERN PLI_INT32 tf_igetlongp PROTO_PARAMS((PLI_INT32 * aof_highvalue,
PLI_INT32 p
XXTERN PLI_INT32 - tf_igetlongtime PROTO_PARAMS((PLI_INT32 *aof_hightime,
PLI BYTE8 *inst));
XXTERN PLI_INT32 tf_igetp PROTO_PARAMS((PLI_INT32 pnum, PLI_BYTE8 *inst));
XXTERN PLI_INT32 tf_igetpchange PROTO_PARAMS((PLI_INT32 nparam, PLI_BYTE8
*inst));
XXTERN double tf_igetrealp PROTO_PARAMS((PLI_INT32 pnum, PLI_BYTE8
*inst));
XXTERN double tf_igetrealtime PROTO_PARAMS((PLI_BYTE8 *inst));
XXTERN PLI_INT32 tf_igettime PROTO_PARAMM((PLI_BYTE}8 * inst))
XXTERN PLI INT32 tf igettimeprecision PROTO PA\overline{RAMS ((PLI BYTE8 *inst));}
XXTERN PLI_INT32 tf_igettimeunit PROTO_PARAMS((PLI_BYTE}\overline{8}*inst))
XXTERN PLI BYTE8 *tf igetworkarea PROTO PARAMS((PLI BYTE8 *inst));
XXTERN PLI_BYTE8 *tf imipname PROTO PARAMMS((PLI BYTE8 * cell));
XXTERN PLI_INT32 tf_imovepvc_flag PROTO_PARAMS\overline{(}(PLI_INT32 nparam,
PLI_BYTE8 `
```

```
XXTERN p_tfnodeinfo tf_inodeinfo PROTO_PARAMS((PLI_INT32 pnum, p_tfnodeinfo
pinfo, PLI_BYTE8 *inst));
XXTERN PLI INT32 tf inump PROTO PARAMS((PLI BYTE8 * inst));
XXTERN PLI_INT32 tf_ipropagatep - PROTO_PARAMS ((PLI_INT32 pnum, PLI_BYTE8
*inst));
XXTERN PLI_INT32 tf_iputlongp PROTO_PARAMS((PLI_INT32 pnum, PLI_INT32
lowvalue, PLI_INT32 highvalue, PLI_BYTE8 *inst));
XXTERN PLI_INT32 tf_iputp PROTO_PARAMS((PLI_INT32 pnum, PLI_INT32 value,
PLI_BYTE8 *inst));
XXTE\overline{RN PLI_INT32 tf_iputrealp PROTO_PARAMS((PLI_INT32 pnum, double value,}
PLI_BYTE8 `inst));
XXTĒRN PLI_INT32 tf_irosynchronize PROTO_PARAMS((PLI_BYTE8 *inst));
XXTERN PLI_INT32 tf isetdelay PROTO PARAMS((PLI_INT3}\overline{2}\mathrm{ delay, PLI_BYTE8
*inst));
XXTERN PLI_INT32 tf_isetlongdelay PROTO_PARAMS((PLI_INT32 lowdelay,
PLI_INT32 \overline{highdelay, PLI_BȲTE8 *inst));}
XXTĒEN PLI_INT32 E
PLI BYTE8 `inst));
XXTERN PLI_INT32 tf_isetworkarea PROTO_PARAMS((PLI_BYTE8 *workarea,
PLI_BYTE8 *inst));
XXTERN PLI_INT32 tf_isizep PROTO_PARAMS((PLI_INT32 pnum, PLI_BYTE8
*inst));
XXTERN PLI BYTE8 *tf ispname PROTO PARAMS((PLI BYTE8 * cell));
XXTERN PLI_INT32 tf_istrdelputp PRŌTO_PARAMS((PLI_INT32 nparam, PLI_INT32
bitlength, PLI_INT32 format_char, PLI_BYTE8 *value_p, PLI_INT32 delay, PLI_INT32
delaytype, PLI_BYTE8 *inst)\;
XXTERN PLI_BYTE8 *tf_istrgetp PROTO_PARAMS((PLI_INT32 pnum, PLI_INT32
format_char, PLI BYTE8 *inst));
XXTERN 'PLI_INT32- tf_istrlongdelputp PROTO_PARAMS((PLI_INT32 nparam,
PLI_INT32 bitlength, PLI_INTT32 format_char, PLI_\overline{BYTE8 *value_p, PLI_INT32}
low\overline{d}lay, PLI INT32 high\overline{d}elay, PLI INT`32 delaytype, PLI BYTE\overline{8}}\mathrm{ *inst)})\mathrm{ );
XXTERN PLI_INT\32 tf_istrrealdelputp PROTO_PARAMS((PLI_INT32 nparam,
PLI INT32 bitlength, PLI INTT32 format char, PLI BYTE8 *value-p, double
real}delay, PLI_INT32 delāytype, PLI_BȲTE8 *inst));
XXTERN PLI_INT\overline{32}
XXTERN PLI_INT32 tf_itestpvc_flag PROTO_PARAMS((PL\overline{I__INT32 nparam,}
PLI_BYTE8 *inst));
XXTERN PLI_INT32 tf_itypep PROTO_PARAMS((PLI_INT32 pnum, PLI_BYTE8
*inst));
XXTERN void tf_long_to_real PROTO_PARAMS((PLI_INT32 int_lo, PLI_INT32
int_hi, double *aof_real));
XXTĒRN PLI_BYTE8 - *tf_longtime_tostr PROTO_PARAMS((PLI_INT32 lowtime,
PLI_INT32 \overline{hightime));}
XXTĒRN PLI_INT32 tf_message PROTO_PARAMS((PLI_INT32 level, PLI_BYTE8
* facility, - PLI_BYTE8 *messño, PLI_BYTE8 *message, ...));
XXTERN PLI_BYTĒ8 *tf_mipnamē PROTO_PARAMS((void));
XXTERN PLI_INT32 tf_movepvc_flag PROTO_PARAMS((PLI_INT32 nparam));
XXTERN voi\overline{d}
PLI_INT32 *aof_high1, PLI_\overline{INT32 low2, PLI_INT3}\overline{2}\mathrm{ high2));}
XXTĒRN p_tfnodēinfo t\overline{f_nodeinfo PROTO_PARAMS((PLI_INT32 pnum, p_tfnodeinfo}
pinfo));
XXTERN PLI_INT32 tf_nump PROTO_PARAMS((void));
XXTERN PLI INT32 tf propagatep PROTO PARAMS((PLI INT32 pnum));
XXTERN PLI_INT32 tf_putlongp PROTO_PARAMS((PLI_INTT32 pnum, PLI_INT32
lowvalue, P
XXTERN PLI_INT`32 tf_putp PROTO_PARAMS((PLI_INT32 pnum, PLI_INT32 value));
XXTERN PLI_INT32 tf_putrealp PRO}TO_PARAMS((\overline{PLI_INT32 pnum, \overline{double value));}
XXTERN PLI_INT32 t\overline{f_read_restart \overline{PROTO_PARAMST(PLI_BYTE8 * blockptr,}}\mathbf{~}\mathrm{ (P)}
PLI_INT32 \overline{blocklen));}
XXTĒRN void tf_real_to_long PROTO_PARAMS((double real, PLI_INT32
* aof_int_lo, PLI_INT32 *aof_int_hi)
XXTERN PLI_INT32- tf_rosynchronize PROTO_PARAMS((void));
```

```
XXTERN void tf_scale_longdelay PROTO_PARAMS((PLI_BYTE8 * cell,
PLI_INT32 delay_lo, PLI_INT}32 delay_hi, PLI_INT32 *aof_delay_lo, PLI_INT32,
* aof delay hi));
XXTE\overline{RN voi\overline{d tf_scale_realdelay PROTO_PARAMS((PLI_BYTE8 * cell, double}}\mathbf{|}=\mp@code{l}
realdelay, double *aof_realdelay));
XXTERN PLI_INT32 - tf_setdelay PROTO_PARAMS((PLI_INT32 delay));
XXTERN PLI_INT32 tf_setlongdelay PROTO_PARAMS((PLI_INT32 lowdelay,
PLI INT32 \overline{highdelay));}
XXTERN PLI_INT32 tf_setrealdelay PROTO_PARAMS((double realdelay));
XXTERN PLI_INT32 tf_setworkarea PROTO_PARAMS((PLI_BYTE8 *workarea));
XXTERN PLI_INT32 tf_sizep PROTO_PARAMS}((PLI_INT32 pnum))
XXTERN PLI_BYTE8 *tf_spname PROTO_PARAMS((void));
XXTERN PLI_INT32 tf_strdelputp PROTO PARAMS((PLI_INT32 nparam, PLI_INT32
bitlength, PLI_INT32 format_char, PLI_BYTE8 *}value_p, PLI_INT32 delay, PLI_INT32,
delaytype));
XXTERN PLI_BYTE8 *tf_strgetp PROTO_PARAMS((PLI_INT32 pnum, PLI_INT32
format_char));
XXTERN PLI BYTE8 *tf strgettime PROTO PARAMS((void));
XXTERN PLI_INT32 tf_strlongdelputp PROTO_PARAMS((PLI_INT32 nparam,
PLI_INT32 bitlength, PLI_INT32 format_char, PLI_BYTE8 *value_p, PLI_INT32
low\overline{d}elay, PLI_INT32 high\overline{delay, PLI_INT\overline{32 delaytȳpe));}}\mathbf{\prime}\mathrm{ ;}
XXTERN PLI_INT32 tf_strrealdelputp PROTO_PARAMS((PLI_INT32 nparam,
PLI_INT32 bitlength, PLI_INT32 format_char, PLI_BYTE8 *value_p, double
realdelay, PLI_INT32 delāytype));
XXTERN PLI_INT32 tf_subtract_long PROTO_PARAMS((PLI_INT32 *aof_lowtime1,
PLI_INT32 `aof_hightime1, \overline{PLI_INT32`lowtime2, \overline{PLI_INT32 hig}htime2));}
XXTE
XXTERN PLI_INT32 tf_testpvc_flag PROTO_PARAMS((PLI_INT32 nparam));
XXTERN PLI_INT32 tf_text PROTO_PARAMS((PLI_BYTE8 * fmt, ...));
XXTERN PLI_INT32 tf_typep PROTO_PARAMS((PLI_INT32 pnum));
XXTERN voi\overline{d tf_unscale_long}delay PROTO_PARAMS((PLI BYTE8 * cell,
PLI_INT32 delay_lo, PLI_INT32 delay_hi, PLI_INT32 *aof_delay_lo, PLI_INT32
* ao\overline{f delay hi));}
```



```
double realdelay, double *aof realdelay));
XXTERN PLI INT32 tf wařning PROTO PARAMS((PLI BYTE8 * fmt, ...));
XXTERN PLI_INT32 tf_write_save PROTO_PARAMS((\overline{PLI_BYTE8 *blockptr,}
PLI_INT32 blocklen));
/*------------------------------------------------------------------------------------
/*_---------------------------------- Globals -----------------------------------------
/* ------------------------------------------------------------------------------------
PLI_VEXTERN PLI_DLLESPEC PLI_BYTE8 *veriuser_version_str;
PLI_VEXTERN PLI_DLLESPEC PLI_INT32 (*endofcompile_rou
#undef PLI EXTERN
#undef PLI_VEXTERN
#ifdef VERIUSER_DEFINED_DLLISPEC
#undef VERIUSER_DEFINED_DLLISPEC
#undef PLI DLLISPPEC
#endif
#ifdef VERIUSER DEFINED DLLESPEC
#undef VERIUSER_DEFINED_DLLESPEC
#undef PLI_DLLESPEC
#endif
#ifdef PLI PROTOTYPES
#undef PLI-PROTOTYPES
#undef PROTO_PARAMS
#undef XXTERN
#undef EETERN
```

```
#endif
#ifdef __cplusplus
}
#endif
#endif /* VERIUSER_H */
```


## Annex G

## (normative)

## vpi_user.h

```
/******************************************************************************
    * vpi_user.h
    * IEEE 1364-2000 Verilog HDL Programming Language Interface (PLI).
*
    * This file contains the constant definitions, structure definitions, and
    * routine declarations used by the Verilog PLI procedural interface VPI
    * access routines.
*
    **************************************************************************/
/****************************************************************************
    * NOTE: the constant values 1 through 299 are reserved for use in this
    * vpi user.h file.
    *****\overline{*}********************************************************************/
#ifndef VPI_USER_H
#define VPI__USER_H
#include <stdarg.h>
#ifdef __cplusplus
extern "\overline{C" {}
#endif
/*---------------------------------------------------------------------------------
/*-------------------------- Portability Help -----------------------------------
/*--------------------------------------------------------------------------------------
/* Sized variables */
#ifndef PLI_TYPES
#define PLI-TYPES
typedef int PLI_INT32;
typedef unsigned int PLI-UINT32;
typedef short PLI-INT16;
typedef unsigned short PLI_UINT16;
typedef char PLI_BYTE8;
typedef unsigned char PLI_UBYTE8;
#endif
/* Use to export a symbol */
#if WIN32
#ifndef PLI_DLLISPEC
#define PLI_DLLISPEC __declspec(dllimport)
#define VPI_USER_DEFINED_DLLISPEC 1
#endif
#else
#ifndef PLI_DLLISPEC
#define PLI-DLLISPEC
#endif
#endif
/* Use to import a symbol */
#if WIN32
#ifndef PLI_DLLESPEC
```

```
#define PLI_DLLESPEC __declspec(dllexport)
#define VPI_USER_DEFINED_DLLESPEC 1
#endif
#else
#ifndef PLI DLLESPEC
#define PLI_DLLESPEC
#endif
#endif
/* Use to mark a function as external */
#ifndef PLI_EXTERN
#define PLI_EXTERN
#endif
/* Use to mark a variable as external */
#ifndef PLI_VEXTERN
#define PLI_VEXTERN extern
#endif
#ifndef PLI PROTOTYPES
#define PLI-PROTOTYPES
#define PROTO_PARAMS(params) params
/* object is \overline{defined imported by the application */}
#define XXTERN PLI_EXTERN PLI_DLLISPEC
/* object is exported by the application */
#define EETERN PLI_EXTERN PLI_DLLESPEC
#endif
/********************************** TYPEDEFS ******************************************)
typedef PLI_UINT32 *vpiHandle;
\begin{tabular}{|c|c|c|}
\hline \#define vpiAlways & 1 & /* always block */ \\
\hline \#define vpiAssignStmt & 2 & /* quasi-continuous assignment */ \\
\hline \#define vpiAssignment & 3 & /* procedural assignment */ \\
\hline \#define vpiBegin & 4 & /* block statement */ \\
\hline \#define vpiCase & 5 & /* case statement */ \\
\hline \#define vpiCaseItem & 6 & /* case statement item */ \\
\hline \#define vpiConstant & 7 & /* numerical constant or literal string */ \\
\hline \#define vpiContAssign & 8 & /* continuous assignment */ \\
\hline \#define vpiDeassign & 9 & /* deassignment statement */ \\
\hline \#define vpiDefParam & 10 & /* defparam */ \\
\hline \#define vpiDelayControl & 11 & /* delay statement (e.g. \#10) */ \\
\hline \#define vpiDisable & 12 & /* named block disable statement */ \\
\hline \#define vpiEventControl & 13 & /* wait on event, e.g. @e */ \\
\hline \#define vpiEventStmt & 14 & /* event trigger, e.g. ->e */ \\
\hline \#define vpiFor & 15 & /* for statement */ \\
\hline \#define vpiForce & 16 & /* force statement */ \\
\hline \#define vpiForever & 17 & /* forever statement */ \\
\hline \#define vpiFork & 18 & /* fork-join block */ \\
\hline \#define vpiFuncCall & 19 & /* HDL function call */ \\
\hline \#define vpiFunction & 20 & /* HDL function */ \\
\hline \#define vpiGate & 21 & /* primitive gate */ \\
\hline \#define vpilf & 22 & /* if statement */ \\
\hline \#define vpiIfElse & 23 & /* if-else statement */ \\
\hline \#define vpiInitial & 24 & /* initial block */ \\
\hline \#define vpiIntegerVar & 25 & /* integer variable */ \\
\hline \#define vpiInterModPath & 26 & /* intermodule wire delay */ \\
\hline \#define vpiIterator & 27 & /* iterator */ \\
\hline \#define vpiIODecl & 28 & /* input/output declaration */ \\
\hline \#define vpiMemory & 29 & /* behavioral memory */ \\
\hline \#define vpiMemoryWord & 30 & /* single word of memory */ \\
\hline
\end{tabular}
```

\#define vpiModPath
\#define vpiModule
\#define vpiNamedBegin
\#define vpiNamedEvent
\#define vpiNamedFork
\#define vpiNet
\#define vpiNetBit
\#define vpiNullStmt
\#define vpiOperation
\#define vpiParamAssign
\#define vpiParameter
\#define vpiPartSelect
\#define vpiPathTerm
\#define vpiPort
\#define vpiPortBit
\#define vpiPrimTerm
\#define vpiRealVar
\#define vpiReg
\#define vpiRegBit
\#define vpiRelease
\#define vpiRepeat
\#define vpiRepeatcontrol
\#define vpiSchedEvent
\#define vpiSpecParam
\#define vpiSwitch
\#define vpiSysFuncCall
\#define vpiSysTaskCall
\#define vpiTableEntry
\#define vpiTask
\#define vpiTaskCall
\#define vpiTchk
\#define vpiTchkTerm
\#define vpiTimeVar
\#define vpiTimeQueue
\#define vpiUdp
\#define vpiUdpDefn
\#define vpiUserSystf
\#define vpiVarSelect
\#define vpiWait
\#define vpiWhile

31 /* module path for path delays */
32 /* module instance */
33 /* named block statement */
/* event variable */
/* named fork-join block */
/* scalar or vector net */
/* bit of vector net */
/* a semicolon. Ie. \#10 ; */
/* behavioral operation */
/* module parameter assignment */
/* module parameter */
/* part select */
/* terminal of module path */
/* module port */
/* bit of vector module port */
/* primitive terminal */
/* real variable */
/* scalar or vector reg */
/* bit of vector reg */
/* release statement */
/* repeat statement */
/* repeat control in an assign stmt */
/* vpi put value() event */
/* specparām */
/* transistor switch */
/* system function call */
/* system task call */
/* UDP state table entry */
/* HDL task */
/* HDL task call */
/* timing check */
/* terminal of timing check */
/* time variable */
/* simulation event queue */
/* user-defined primitive */
/* UDP definition */
/* user defined system task or function */
/* variable array selection */
/* wait statement * /
/* while statement */

| \#define vpiAttribute | 105 | /* attribute of an object */ |
| :---: | :---: | :---: |
| \#define vpiBitSelect | 106 | /* Bit select of parameter, var select */ |
| \#define vpiCallback | 107 | /* callback object */ |
| \#define vpiDelayTerm | 108 | /* Delay term which is a load or driver */ |
| \#define vpiDelayDevice | 109 | /* Delay object within a net */ |
| \#define vpiFrame | 110 | /* reentrant task/func frame */ |
| \#define vpiGateArray | 111 | /* gate instance array */ |
| \#define vpiModuleArray | 112 | /* module instance array */ |
| \#define vpiPrimitiveArray | 113 | /* vpiprimitiveArray type */ |
| \#define vpiNetArray | 114 | /* multidimensional net */ |
| \#define vpiRange | 115 | /* range declaration */ |
| \#define vpiRegArray | 116 | /* multidimensional reg */ |
| \#define vpiSwitchArray | 117 | /* switch instance array */ |
| \#define vpiUdpArray | 118 | /* UDP instance array */ |
| \#define vpiContAssignBit | 128 | /* Bit of a vector continuous assignment |
| \#define vpiNamedEventArray | 129 | /* multidimensional named event */ |
| /******************************** METHODS |  |  |
| /************* methods used | to | averse 1 to 1 relationships |
| \#define vpiCondition | 71 | /* condition expression */ |
| \#define vpiDelay | 72 | /* net or gate delay */ |
| \#define vpiElseStmt | 73 | /* else statement */ |

\#define vpiForIncStmt
\#define vpiForInitStmt
\#define vpiHighConn
\#define vpiLhs
\#define vpiIndex
\#define vpiLeftRange
\#define vpiLowConn
\#define vpiParent
\#define vpiRhs
\#define vpiRightRange
\#define vpiScope
\#define vpiSysTfCall
\#define vpiTchkDataTerm
\#define vpiTchkNotifier
\#define vpiTchkRefTerm

74 /* increment statement in for loop */ \#define vpiForInitStmt
/* initialization statement in for loop */
derine vpiHighConn
/* higher connection to port */
\#define vpiLhs
/* left-hand side of assignment */
\#define vpiLeftRange
/* index of var select, bit select, etc. */
\#define vpiLowConn
/* left range of vector or part select */
/* lower connection to port */
\#derine vpiParent
/* parent object * /
\#define vpiRhs
/* right-hand side of assignment */
/* right range of vector or part select */
\#define vpiScope
/* containing scope object */
/* task function call */
\#define vpiTchkDataTerm
/* timing check data term */
/* timing check notifier */
\#define vpiTchkRefTerm
/* timing check reference term */

/***** methods which can traverse 1 to 1 , or 1 to many relationships ******/
\#define vpiExpr 102 /* connected expression */
\#define vpiPrimitive 103 /* primitive (gate, switch, UDP) */
\#define vpistmt 104 /* statement in process or task */





| \#define vpiOpType | 39 | /* operation subtypes: */ |
| :---: | :---: | :---: |
| \#define vpiMinusOp | 1 | /* unary minus */ |
| \#define vpiPlusOp | 2 | /* unary plus */ |
| \#define vpiNotOp | 3 | /* unary not */ |
| \#define vpiBitNegOp | 4 | /* bitwise negation */ |
| \#define vpiUnaryAndOp | 5 | /* bitwise reduction and */ |
| \#define vpiUnaryNandOp | 6 | /* bitwise reduction nand */ |
| \#define vpiUnaryOrOp | 7 | /* bitwise reduction or */ |
| \#define vpiUnaryNorOp | 8 | /* bitwise reduction nor */ |
| \#define vpiUnaryXorOp | 9 | /* bitwise reduction xor */ |
| \#define vpiUnaryXNorOp | 10 | /* bitwise reduction xnor */ |
| \#define vpiSubop | 11 | /* binary subtraction */ |
| \#define vpiDivop | 12 | /* binary division */ |
| \#define vpiModOp | 13 | /* binary modulus */ |
| \#define vpiEqOp | 14 | /* binary equality */ |
| \#define vpiNeqOp | 15 | /* binary inequality */ |
| \#define vpiCaseEqOp | 16 | /* case (x and z) equality */ |
| \#define vpiCaseNeqOp | 17 | /* case inequality */ |
| \#define vpiGtOp | 18 | /* binary greater than */ |
| \#define vpiGeOp | 19 | /* binary greater than or equal */ |
| \#define vpiLtOp | 20 | /* binary less than */ |
| \#define vpiLeOp | 21 | /* binary less than or equal */ |
| \#define vpiLShiftop | 22 | /* binary left shift */ |
| \#define vpiRShiftop | 23 | /* binary right shift */ |
| \#define vpiAddOp | 24 | /* binary addition */ |
| \#define vpiMultop | 25 | /* binary multiplication */ |
| \#define vpiLogAndOp | 26 | /* binary logical and */ |
| \#define vpiLogOrOp | 27 | /* binary logical or */ |
| \#define vpiBitAndOp | 28 | /* binary bitwise and */ |
| \#define vpiBitOrOp | 29 | /* binary bitwise or */ |
| \#define vpiBitXorOp | 30 | /* binary bitwise xor */ |
| \#define vpiBitXNorOp | 31 | /* binary bitwise xnor */ |
| \#define vpiBitXnorOp |  | BitXNorOp /* added with 1364-2000 */ |
| \#define vpiConditionOp | 32 | /* ternary conditional */ |
| \#define vpiConcatop | 33 | /* n-ary concatenation */ |
| \#define vpiMultiConcatOp | 34 | /* repeated concatenation */ |
| \#define vpiEventorOp | 35 | /* event or */ |
| \#define vpiNullop | 36 | /* null operation */ |
| \#define vpiListop | 37 | /* list of expressions */ |
| \#define vpiMinTypMaxOp | 38 | /* min:typ:max: delay expression */ |
| \#define vpiPosedgeop | 39 | /* posedge */ |
| \#define vpiNegedgeOp | 40 | /* negedge */ |
| \#define vpiArithLShiftop | 41 | /* arithmetic left shift (1364-2000) */ |
| \#define vpiArithRShiftop | 42 | /* arithmetic right shift (1364-2000) */ |
| \#define vpiPowerOp | 43 | /* arithmetic power op (1364-2000) */ |
| \#define vpiConstType | 40 | /* constant subtypes: */ |
| \#define vpiDecConst | 1 | /* decimal integer */ |
| \#define vpiRealconst | 2 | /* real */ |
| \#define vpiBinaryConst | 3 | /* binary integer */ |
| \#define vpiOctConst | 4 | /* octal integer */ |
| \#define vpiHexConst | 5 | /* hexadecimal integer */ |
| \#define vpiStringConst | 6 | /* string literal */ |
| \#define vpiIntConst | 7 | /* HDL integer constant (1364-2000) */ |
| \#define vpiBlocking | 41 | /* blocking assignment (boolean) */ |
| \#define vpiCaseType | 42 | /* case statement subtypes: */ |
| \#define vpiCaseExact | 1 | /* exact match */ |
| \#define vpiCaseX | 2 | /* ignore X's */ |
| \#define vpiCasez | 3 | /* ignore Z's */ |
| \#define vpiNetDeclAssign | 43 | /* assign part of decl (boolean) */ |



```
} s_vpi_time, *p_vpi_time;
/* time types * /
#define vpiScaledRealTime 1
#define vpiSimTime 2
#define vpiSuppressTime 3
/**************************** delay structures ****************************/
typedef struct t_vpi_delay
{
    struct t_vpi_time *da; /* pointer to user allocated array of
    delay values */
    PLI INT32 no of delays; /* number of delays */
    PLI_INT32 time_type; /* [ vpiScaledRealTime, vpiSimTime,
    vpiSuppressTime] */
    PLI_INT32 mtm_flag; /* true for mtm values * /
    PLI_INT32 append_flag; /* true for append */
    PLI INT32 pulsere flag; /* true for pulsere values */
} s_vpi_delay, *p_vpi_delay;
/**************************** value structures ****************************/
/* vector value */
typedef struct t vpi vecval
{
    /* following fields are repeated enough times to contain vector */
    PLI_INT32 aval, bval; /* bit encoding: ab: 00=0, 10=1, 11=X, 01=Z */
} s_vpi_vecval, *p_vpi_vecval;
/* strength (scalar) value */
typedef struct t vpi strengthval
{
    PLI_INT32 logic; /* vpi[ 0,1,X,Z] */
    PLI INT32 s0, s1; /* refer to strength coding below */
} s_vpi_strengthval, *p_vpi_strengthval;
/* strength values * /
#define vpiSupplyDrive 0x80
#define vpiStrongDrive 0x40
#define vpiPullDrive 0x20
#define vpiWeakDrive 0x08
#define vpiLargeCharge 0x10
#define vpiMediumCharge 0x04
#define vpiSmallCharge 0x02
#define vpiHiz 0x01
/* generic value */
typedef struct t_vpi_value
{
    PLI_INT32 format; /* vpi[[ Bin,Oct,Dec,Hex] Str,Scalar,Int,Real,String,
                            Vector,Strength,Suppress,Time,ObjType] Val */
    union
        {
            PLI BYTE8 *str; /* string value */
                PLI_INT32 scalar; /* vpi[ 0,1,X,Z] */
                PLI INT32 integer; /* integer value */
                dou\overline{ble real; /* real value */}
                struct t_vpi_time *time; /* time end */
```

```
#define vpiOctStrVal 2
#define vpiDecStrVal 3
#define vpiHexStrVal 4
#define vpiScalarVal 5
#define vpiIntVal 6
#define vpiRealVal 7
#define vpiStringVal 8
#define vpiVectorVal 9
#define vpiStrengthVal 10
#define vpiTimeVal 11
#define vpiObjTypeVal 12
#define vpiSuppressVal 13
/* delay modes */
#define vpiNoDelay 1
#define vpiInertialDelay 2
#define vpiTransportDelay 3
#define vpiPureTransportDelay 4
/* force and release flags */
#define vpiForceFlag 5
#define vpiReleaseFlag 6
/* scheduled event cancel flag */
#define vpiCancelEvent 7
/* bit mask for the flags argument to vpi_put_value() */
#define vpiReturnEvent 0x1000
/* scalar values */
#define vpi0 0
#define vpil 1
#define vpiZ 2
#define vpix 3
#define vpiH 4
#define vpiL 5
#define vpiDontCare 6
/*
#define vpiNoChange 7 Defined under vpiTchkType, but
*/
/********************** system task/function structure **********************/
typedef struct t_vpi_systf_data
{
    PLI INT32 type; /* vpiSysTask, vpiSysFunc */
    PLI_INT32 sysfunctype; /* vpiSysTask, vpi[ Int,Real,Time,Sized,
                                    SizedSigned] Func */
    PLI_BYTE8 *tfname; /* first character must be `$' */
    PLI_INT32 (*calltf)(PLI_BYTE8 *);
    PLI_INT32 (*compiletf)(\overline{PLI_BYTE8 *);}
    PLI_INT32 (*sizetf)(PLI_BYTE8 *); /* for sized function
                                    callbacks only */
    PLI_BYTE8 *user_data;
} s_vpi_systf_data, *p_vpi_systf_data;
#define vpiSysTask 1
#define vpiSysFunc 2
/* the subtypes are defined under the vpiFuncType property */
/***************** Verilog execution information structure *****************/
typedef struct t_vpi_vlog_info
{
    PLI_INT32 argc;
```

```
    PLI BYTE8 **argv;
    PLI_BYTE8 *product;
    PLI BYTE8 *version;
} s_vpi_vlog_info, *p_vpi_vlog_info;
/******************** PLI error information structure **********************/
typedef struct t_vpi_error_info
{
    PLI_INT32 state; /* vpi[ Compile,PLI,Run] */
    PLI_INT32 level; /* vpi[ Notice,Warning,Error,System,Internal] */
    PLI BYTE8 *message;
    PLI_BYTE8 *product;
    PLI BYTE8 * code;
    PLI_BYTE8 * file;
    PLI INT32 line;
} s_vpi_error_info, *p_vpi_error_info;
/* error types */
#define vpiCompile 1
#define vpiPLI 2
#define vpiRun 3
#define vpiNotice 1
#define vpiWarning 2
#define vpiError 3
#define vpiSystem 4
#define vpiInternal 5
/************************** callback structures *********************************
/* normal callback structure */
typedef struct t_cb_data
{
    PLI_INT32 reason; /* callback reason * /
    PLI_INT32 (* cb_rtn)(struct t_cb_data *); /* call routine */
    vpiH̄andle obj;- - - /* trigger object */
    p_vpi_time time; /* callback time */
    p_vpi_value value; /* trigger object value */
    P\overline{LI_INT32 index; /* index of the memory word or}
                                    var select that changed */
    PLI_BYTE8 *user data;
} s_c\overline{b_data, *p_cb_\}data;
/**************************** CALLBACK REASONS *****************************
/*************************** Simulation related ***************************/
#define cbValueChange 1
#define cbStmt 2
#define cbForce 3
#define cbRelease 4
/****************************** Time related ******************************/
#define cbAtStartOfSimTime 5
#define cbReadWriteSynch 6
#define cbReadOnlySynch 7
#define cbNextSimTime 8
#define cbAfterDelay 9
/***************************** Action related *****************************/
#define cbEndOfCompile }1
#define cbStartOfSimulation 11
#define cbEndOfSimulation 12
#define cbError 13
#define cbTchkViolation 14
#define cbStartOfSave 15
```



```
p_vpi_time time_p,
PLI_INT32 flags));
/* time processing *// 
XXTERN void vpi_get_time PROTO_PARAMS((vpiHandle object,
        p_vpi_time time_p));
/* I/O routines */
XXTERN PLI_UINT32 vpi_mcd_open
XXTERN PLI_UINT32 vpi_mcd_close
    PROTO_PARAMS((PLI_BYTE8 * fileName));
    PROTO_PARAMS((PLI_UINT32 mcd));
XXTERN PLI_BYTE8 * vpi_mcd_name
    PROTO_PARAMS((PLI_UINT32 cd));
XXTERN PLI_INT32 vpi_mcd_printf
XXTERN PLI_INT32 vpi_printf PROTO_PARAMS((PLI_BYTE8 * format,
    PROTO_PARAMS((PLI_UINT32 mcd,
        PLI BYTE8 * format,
        ...));
    ...));
/* utility routines * /
XXTERN PLI_INT32 vpi_compare_objects PROTO_PARAMS((vpiHandle object1,
    vpiHandle object2));
XXTERN PLI_INT32 vpi_chk_error
XXTERN PLI_INT32 vpi_free_object
XXTERN PLI_INT32 vpi_get_v
/* routines added with 1364-2000 * /
XXTERN PLI_INT32 vpi_get_data
XXTERN PLI_INT32 vpi_put_data
XXTERN void *vpi_get_userdata PROTO_PARAMS((vpi\overline{H}andle obj));
XXTERN PLI_INT32 vpi_put_userdata PROTO_PARAMS((vpiHandle obj,
                void *userdata));
XXTERN PLI_INT32 vpi_vprintf
XXTERN PLI_INT32 vpi_mcd_vprintf
XXTERN PLI_INT32 vpi_flush
XXTERN PLI INT32 vpi mcd flush
XXTERN PLI_INT32 vpi__conETrol
```



```
/**************************** GLOBAL VARIABLES ****************************/
PLI_VEXTERN PLI_DLLESPEC void (*vlog_startup_routines[] ) ();
    /\star array of function pointers, last pointer should be null */
```

```
#undef PLI EXTERN
#undef PLI_VEXTERN
#ifdef VPI_USER_DEFINED_DLLISPEC
#undef VPI_USER_DEFINED_DLLISPEC
#undef PLI_DLLI\overline{SPEC}
#endif
#ifdef VPI_USER_DEFINED_DLLESPEC
#undef VPI_USER_DEFINED_DLLESPEC
```

```
#undef PLI_DLLESPEC
#endif
#ifdef PLI_PROTOTYPES
#undef PLI PROTOTYPES
#undef PROTO_PARAMS
#undef XXTERN
#undef EETERN
#endif
#ifdef __cplusplus
}
#endif
#endif /* VPI_USER_H */
```


## Annex H

## (informative)

## Bibliography

[B1] IEEE Std 754-1985 (Reaff 1990), IEEE Standard for Binary Floating-Point Arithmetic (ANSI). ${ }^{2}$
[B2] IEEE Std 1497-1999, IEEE Standard Standard for Standard Delay Format (SDF) for the Electronic Design Process.

[^2]
[^0]:    The Institute of Electrical and Electronics Engineers, Inc.
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[^1]:    ${ }^{1}$ The numbers in brackets correspond to those of the bibliography in Annex H.

[^2]:    ${ }^{2}$ IEEE publications are available from the Institute of Electrical and Electronics Engineers, 445 Hoes Lane, P.O. Box 1331, Piscataway, NJ 08855-1331, USA.

